

Step 2 - Filter the tones

Input data file name

DataFileName := "Data3.NoSpin.NoPhNoise.byte"

A priori Doppler Polynomial model file name

DpcFileName0 := "Data3.Det.Pass0.tone0.fitcoeffs.txt"

Pass1 Doppler Polynomial model file name

DpcFileName1 := "Data3.Det.Pass1.tone0.fitcoeffs.txt"

Pass2 Doppler Polynomial model file name

Data set and processing essential parameters

Band Width

BW := 250 · kHz

Filter ratio

FR := 100

Scan length (in samples)

Nt := 8 · 1000 · 1000

Output bandwidth

$BW_o := BW \cdot FR^{-1}$ $BW_o = 2500 \text{ s}^{-1}$

N of points to input FFTsegment

Npsi := 10 · 1000

Number of points in output segment

$Npso := Npsi \cdot FR^{-1}$ Npso = 100

Overlapping of FFT segments

Ovlp := 2

Number of samples in the output signal

$Nto := \text{floor}(Nt \cdot FR^{-1})$ Nto = 80000

Note the decimal number for FFT length !

And input length is abbreviated to make
a nice decimal number of output samples

Number of spectral points to extract to output FFT

$Npfo := Npso \cdot 0.5 + 1$ Npfo = 51

Zero padding of output FFT

$Npfz := Npso \cdot 0.5 - 1$ jpfz := 0 .. Npfz - 1

padding array for neg. frequencies

dpadd_{jpfz} := 0 + i · 0

Derived parameters (some of them are just for illustration)

Sampling rate

Sr := 2 · BW

And sampling interval

~~xxx~~ dt := Sr⁻¹

Dimensionless

dt_n := dt · s⁻¹

Total time span

Tspan := Nt · dt Tspan = 16 s

Input Time grid

jt := 0 .. Nt - 1 tt_{jt} := jt · dt

Output Time grid

jto := 0 .. Nto - 1 tto_{jto} := jto · dt · FR

Binning within FFT input and output segments

jpsi := 0 .. Npsi - 1 jpso := 0 .. Npso - 1

Input Window function in time domain

$Wini_{jpsi} := \cos\left[\frac{\pi}{Npsi} \cdot (jpsi - 0.5 \cdot Npsi + 0.5)\right]$

Output Window function in time domain

$Wino_{jpso} := \cos\left[\frac{\pi}{Npso} \cdot (jpso - 0.5 \cdot Npso + 0.5)\right]$

Derived parameters , Continue

Number of FFT segments to process (accounting for overlap)

$$N_{\text{segm}} := \text{floor}\left(\frac{N_t}{N_{\text{psi}}}\right) \cdot \text{Ovlp} - (\text{Ovlp} - 1) \quad N_{\text{segm}} = 1599$$

shift (in samples) between overlapping segments

$$O_{\text{shifti}} := \frac{N_{\text{psi}}}{\text{Ovlp}}$$

$$O_{\text{shifto}} := \frac{N_{\text{pso}}}{\text{Ovlp}}$$

Frequency resolution of the input FFT

$$df_i := (N_{\text{psi}} \cdot dt)^{-1}$$

$$df_i = 50 \text{ s}^{-1}$$

Read Dorrpler Frequency polynomials

Fcd0 := READPRN(DpcFileName0)

Fcd1 := READPRN(DpcFileName1)

Npf := Fcd0₀ Npf = 3 jpf := 0 .. Npf

Cf0_{jpf} := Fcd0_{jpf+2}

Cf1_{jpf} := Fcd1_{jpf+2} Tspanp := Fcd1₁ Tspanp = 16.777216 Tspanps := Tspanp · s

Combine all data sets Cf := Cf0 + Cf1

~~Cf~~₀ := Cf1₀

Select the Doppler constant offset from the last pass data

Make Phase polynomials

Cpp₀ := 0 ~~Cpp~~_{jpf+1} := $2\pi \cdot \frac{Cf_{jpf}}{jpf + 1}$ ~~Cpp~~₁ := 0 Npp := Npf + 1

Frequency polynomial

Phase polynomial

$$Cf = \begin{pmatrix} 70048.91028919304 \\ 7052.868226490915 \\ -7888.813451185823 \\ 1.104737497866154 \end{pmatrix}$$

BWoh := BWo · 0.5

$$Cpp = \begin{pmatrix} 0 \\ 0 \\ 22157.23900708073 \\ -16522.292255857148 \\ 1.735317603720739 \end{pmatrix}$$

We know, that several tones have certain offsets from the carrier line

Defining the start/end bin of the filter, which will put tone lines in the center of the output band

carrier

Fcc := Cf₀ · Hz

Bsc := floor[(Fcc - BWoh) · dfi⁻¹]

Bsc = 1375

Bec := Bsc + Npfo - 1

Fstartc := Bsc · dfi

Fstartc = 68750 s⁻¹

tones

Fc1 := Fcc - 10000 · Hz

Bs1 := floor[(Fc1 - BWoh) · dfi⁻¹]

Bs1 = 1175

Be1 := Bs1 + Npfo - 1

Fstart1 := Bs1 · dfi

Fstart1 = 58750 s⁻¹

Fc2 := Fcc - 50000 · Hz

Bs2 := floor[(Fc2 - BWoh) · dfi⁻¹]

Bs2 = 375

Be2 := Bs2 + Npfo - 1

Fstart2 := Bs2 · dfi

Fstart2 = 18750 s⁻¹

Fc3 := Fcc + 20000 · Hz

Bs3 := floor[(Fc3 - BWoh) · dfi⁻¹]

Bs3 = 1775

Be3 := Bs3 + Npfo - 1

Fstart3 := Bs3 · dfi

Fstart3 = 88750 s⁻¹

Integrate the phase

$$\text{Phdopp}_{jt} := \text{Cpp}_0 + \text{Tspanp} \cdot \sum_{jip=2}^{\text{Npf}} \left[\text{Cpp}_{jip} \cdot \left(\frac{tt_{jt}}{\text{Tspanps}} \right)^{jip} \right]$$

Make a segment time shift phase correction coefficient, actually a start bin of the filter can be selected in such way, that this coeff will be +1, although it can be even complex

$\text{Fstartc} \cdot \text{Oshifti} \cdot \text{dt} = 687.5$	$\text{Pssc} := \text{Fstartc} \cdot \text{Oshifti} \cdot \text{dt} - \text{floor}(\text{Fstartc} \cdot \text{Oshifti} \cdot \text{dt})$	$\text{Esc} := \exp(i \cdot 2 \cdot \pi \cdot \text{Pssc})$	$\text{Esc} = -1$
$\text{Fstart1} \cdot \text{Oshifti} \cdot \text{dt} = 587.5$	$\text{Pss1} := \text{Fstart1} \cdot \text{Oshifti} \cdot \text{dt} - \text{floor}(\text{Fstart1} \cdot \text{Oshifti} \cdot \text{dt})$	$\text{Es1} := \exp(i \cdot 2 \cdot \pi \cdot \text{Pss1})$	$\text{Es1} = -1$
$\text{Fstart2} \cdot \text{Oshifti} \cdot \text{dt} = 187.5$	$\text{Pss2} := \text{Fstart2} \cdot \text{Oshifti} \cdot \text{dt} - \text{floor}(\text{Fstart2} \cdot \text{Oshifti} \cdot \text{dt})$	$\text{Es2} := \exp(i \cdot 2 \cdot \pi \cdot \text{Pss2})$	$\text{Es2} = -1$
$\text{Fstart3} \cdot \text{Oshifti} \cdot \text{dt} = 887.5$	$\text{Pss3} := \text{Fstart3} \cdot \text{Oshifti} \cdot \text{dt} - \text{floor}(\text{Fstart3} \cdot \text{Oshifti} \cdot \text{dt})$	$\text{Es3} := \exp(i \cdot 2 \cdot \pi \cdot \text{Pss3})$	$\text{Es3} = -1$

```

MakeFiltX(Phcorr, Fbinstart, Fbinend, Es) :=
  for jjo ∈ 0 .. Nto - 1
    foutjjo ← 0
    for jsegm ∈ 0 .. Nsegm - 1
      skip ← jsegm · Oshifti
      din ← READBIN(DataFileName, "byte", 0, 1, skip, Npsi)
      din ← din - 127
      din ← (din · Wini)
      phc ← submatrix(Phcorr, skip, skip + Npsi - 1, 0, 0)
      ephc ← exp(i · phc)
      din ← (din · ephc)
      sp ← cfft(din)
      spo ← submatrix(sp, Fbinstart, Fbinend, 0, 0)
      spo0 ← 1 · Re(spo0)
      spoNpfo-1 ← 1 · Re(spoNpfo-1)
      spop ← stack(spo, dpadd)
      dout ← icfft(spop)
      dout ← (dout · Wino)
      for jjso ∈ 0 .. Npso - 1
        foutjjso+jsegm·Oshifto ← foutjjso+jsegm·Oshifto + doutjjso · Esjsegm
    return fout

```

Major function to do Phase Tracking,
Filtering and Hilbert transform

Phase integration should be done differently in C

Multiple tones can be extracted in parallell,
using only 1 "big" input Fourier transform
and several "small" output FFTs.

Mathematically this filter is equivalent to PFB,
but better, because it allows arbitrary positioning
of output channels, with different width.

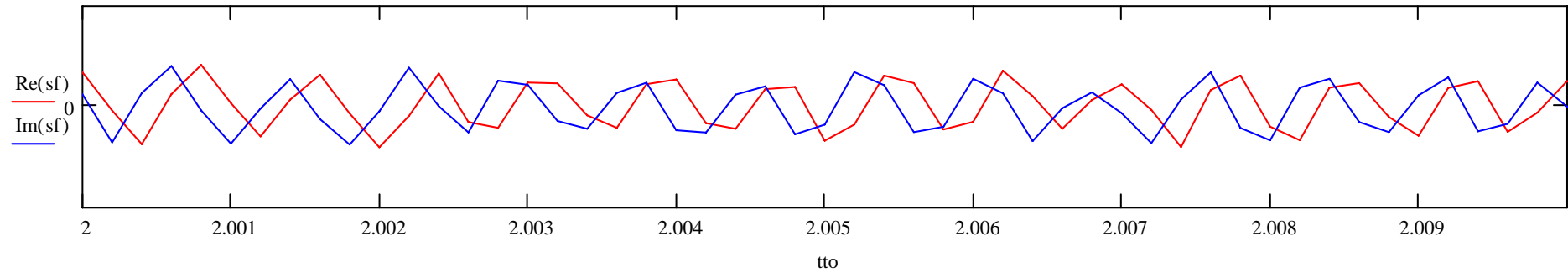
This is a tricky part to make in C

Make a filtered signal for major tone

$\text{sf} := \text{MakeFiltX}(\text{Phdopp}, \text{Bsc}, \text{Bec}, \text{Esc})$

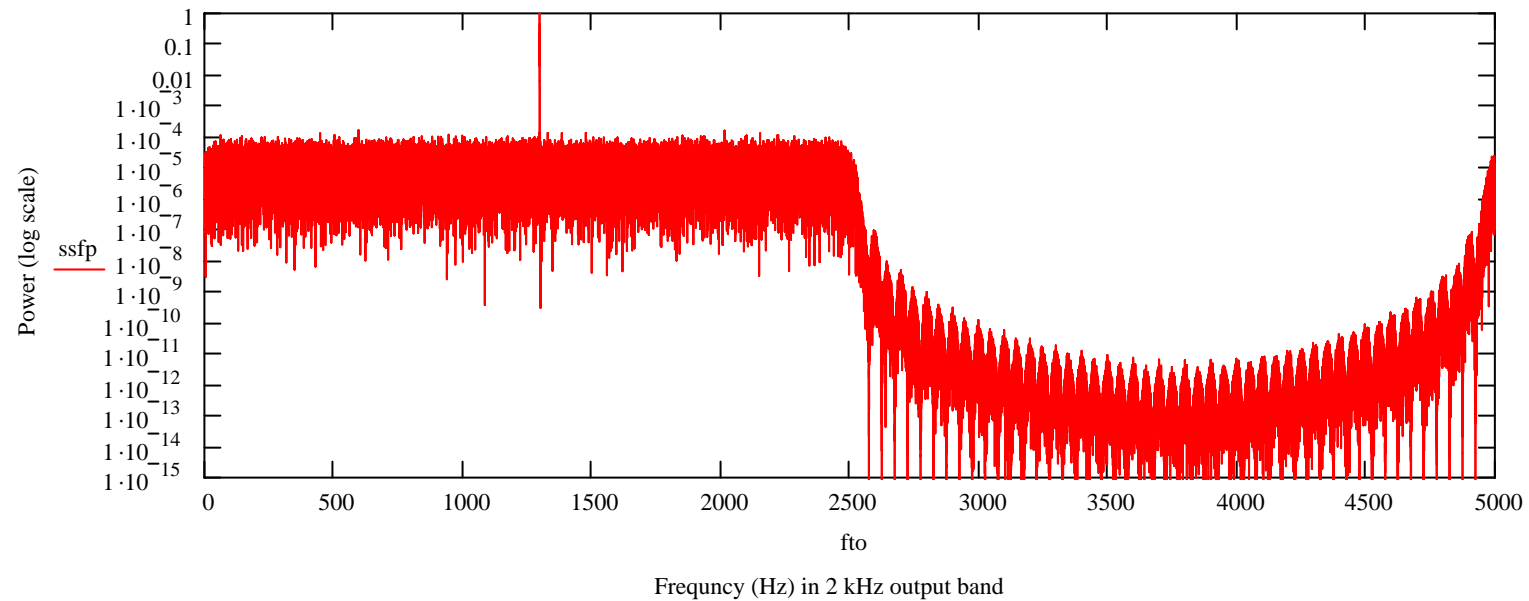
$\text{length}(\text{sf}) = 80000$

fragment of the filtered complex signal in a time domain

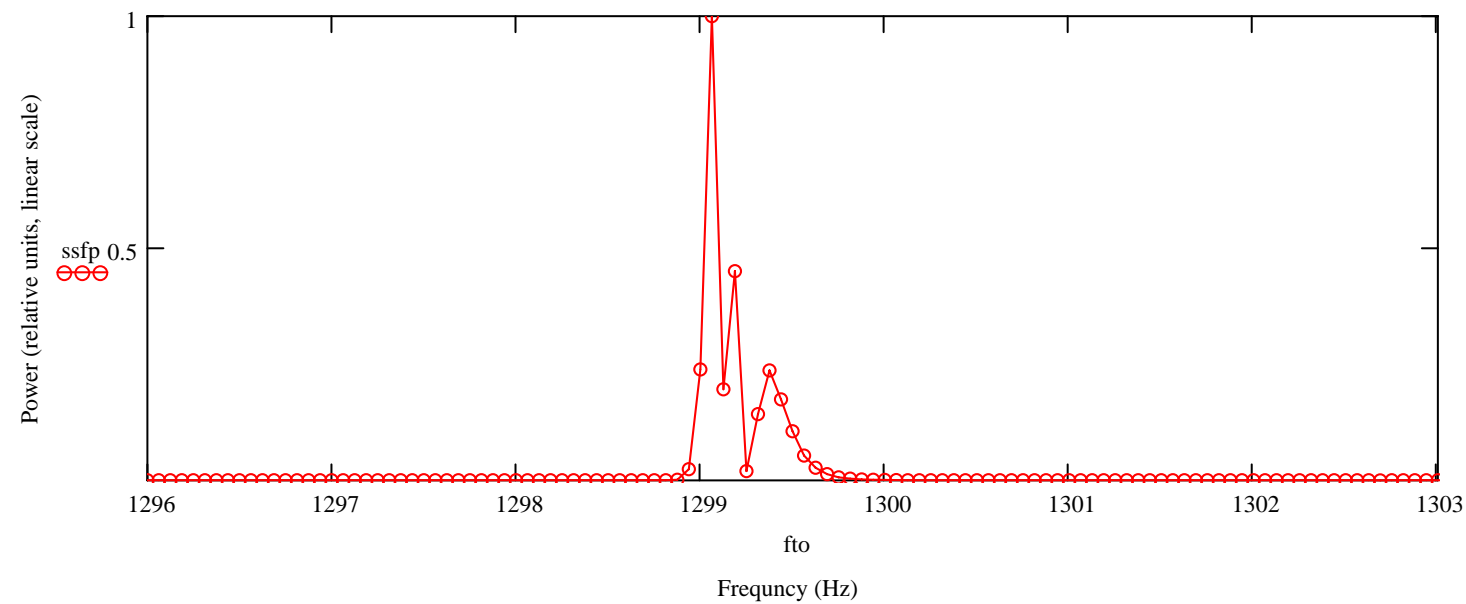


Full length spectrum (two-sided FFT), shows good suppression of negative frequencies

$\text{ssf} := \text{cfft}(\text{sf})$
 $\xrightarrow{\quad} \text{ssfp} := (|\text{ssf}|)^2$
 $\text{xssf} := \max(\text{ssfp})$
 $\text{ssfp} := \text{ssfp} \cdot \text{xssf}^{-1}$
 $\text{fto}_{\text{jto}} := \frac{1}{\text{Tspan}} \cdot \text{jto}$



Zoom into central line, all the power is concentrated in a sub-Hz wide line, SNR at 50 dB level.



So, just write this narrow band complex signal into file, PLL it later.