

1. (30%) For the filter network shown in Fig. A

(a) (15%) Derive the voltage transfer function $H(S) = V_o(S) / V_s(S)$. What type of filter is it? Verify your answer.

(b) (15%) If $R_s = 1 \text{ k}\Omega$, $R_L = 9 \text{ k}\Omega$, $L = 1 \text{ mH}$, what is the cutoff frequency?

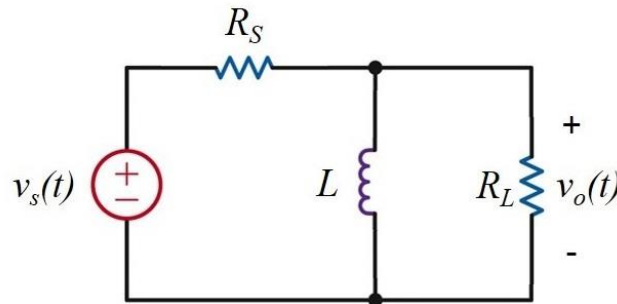


Fig. A

(a)

Derive the voltage transfer function (10%)

To derive the voltage transfer function $H(s) = \frac{V_o(s)}{V_s(s)}$ for the given circuit, we need to analyze the impedance of each component and use voltage division.

The components are:

R_s : series resistance

L : inductance

R_L : parallel resistance

First, let's write down the impedances:

Impedance of R_s : R_s (1%)

Impedance of L : sL (where s is the complex frequency) (1%)

Impedance of R_L : R_L (1%)

Since L and R_L are in parallel, their combined impedance $Z_{parallel}$ is: (2%)

$$Z_{parallel} = \frac{R_L sL}{R_L + sL}$$

The total impedance Z_{total} of the circuit is the sum of R_s and $Z_{parallel}$: (2%)

$$Z_{total} = R_s + \frac{R_L sL}{R_L + sL}$$

Now, the transfer function $H(s)$ can be found using the voltage division rule: (3%)

$$H(s) = \frac{Z_{parallel}}{Z_{total}} = \frac{\frac{R_L sL}{R_L + sL}}{R_s + \frac{R_L sL}{R_L + sL}} = \frac{sLR_L}{R_s R_L + sL(R_L + R_s)} = \frac{j\omega LR_L}{R_s R_L + j\omega L(R_L + R_s)}$$

What type of filter is it? (5%)

To determine the type of filter, analyze the transfer function at different frequency limits:

At low frequencies $s \rightarrow 0$, $H(s) \rightarrow 0$ (2%)

At high frequencies $s \rightarrow \infty$, $H(s) \rightarrow 1$ (2%)

This suggests that the circuit is a **high-pass filter**. (3%)

(b)

Cutoff frequency (15%)

The cutoff frequency ω_c is where the magnitude of the transfer function is $\frac{1}{\sqrt{2}}$ of its maximum value.

This occurs when the imaginary part of the denominator equals the real part:

$$\omega_c L(R_L + R_S) = R_L R_S \quad (7\%)$$

Solving for ω_c :

$$\omega_c = \frac{R_L R_S}{L(R_L + R_S)} \quad (8\%)$$

Or the cutoff frequency f_c in hertz is:

$$f_c = \frac{\omega_c}{2\pi} = \frac{R_L R_S}{2\pi L(R_L + R_S)}$$

2. (30%) For the following transfer function

$$H(j\omega) = \frac{-1000(j\omega + 10)}{(j\omega + 100)(j\omega + 200)}$$

- (a) (15%) Construct its Bode plot of the magnitude (label the important values and features in the plot, such as break frequencies, slopes, and actual curve).
- (b) (15%) Construct its asymptotic Bode plot of phase (label the important values and features in the plot, such as break frequencies and slopes).

1. Identify Zeros and Poles

Zero:

$$j\omega = -10$$

Poles:

$$j\omega = -100$$

$$j\omega = -200$$

2. Calculate Initial Gain and Phase

Initial Gain (Magnitude)

For very low frequencies ($\omega \rightarrow 0$):

$$H(j\omega) \approx -1000(0 + 10) / ((0 + 100)(0 + 200)) = -10000 / 20000 = -0.5$$

Convert to dB:

$$20 \log_{10}(0.5) = 20 \log_{10}(1/2) = 20 \times (-0.3010) = -6.02 \text{ dB}$$

Initial Phase

The constant term -1000 contributes -180° to the phase.

The zero and poles at low frequencies contribute negligible phase shift.

So, the initial phase is:

$$-180^\circ + 0^\circ - 0^\circ - 0^\circ = -180^\circ$$

3. Bode Plot: Magnitude

Frequency Range $\omega < 10$:

- Initial magnitude is -6.02 dB.
- Slope: 0 dB/decade.

Frequency Range $10 < \omega < 100$:

- Zero at $\omega = 10$ adds +20 dB/decade.
- At $\omega = 100$:
 $-6.02 \text{ dB} + 20 \text{ dB/decade} \times 1 = 13.98 \text{ dB}$

Frequency Range $100 < \omega < 200$:

- Pole at $\omega = 100$ adds -20 dB/decade.
- Slope: 0 dB/decade.

- Magnitude at $\omega = 200$:

13.98 dB

Frequency Range $\omega > 200$:

- Pole at $\omega = 200$ adds -20 dB/decade.
- Slope: -20 dB/decade.

4. Bode Plot: Phase

Frequency Range $\omega < 10$:

- Initial phase is -180° .

Frequency Range $10 < \omega < 100$:

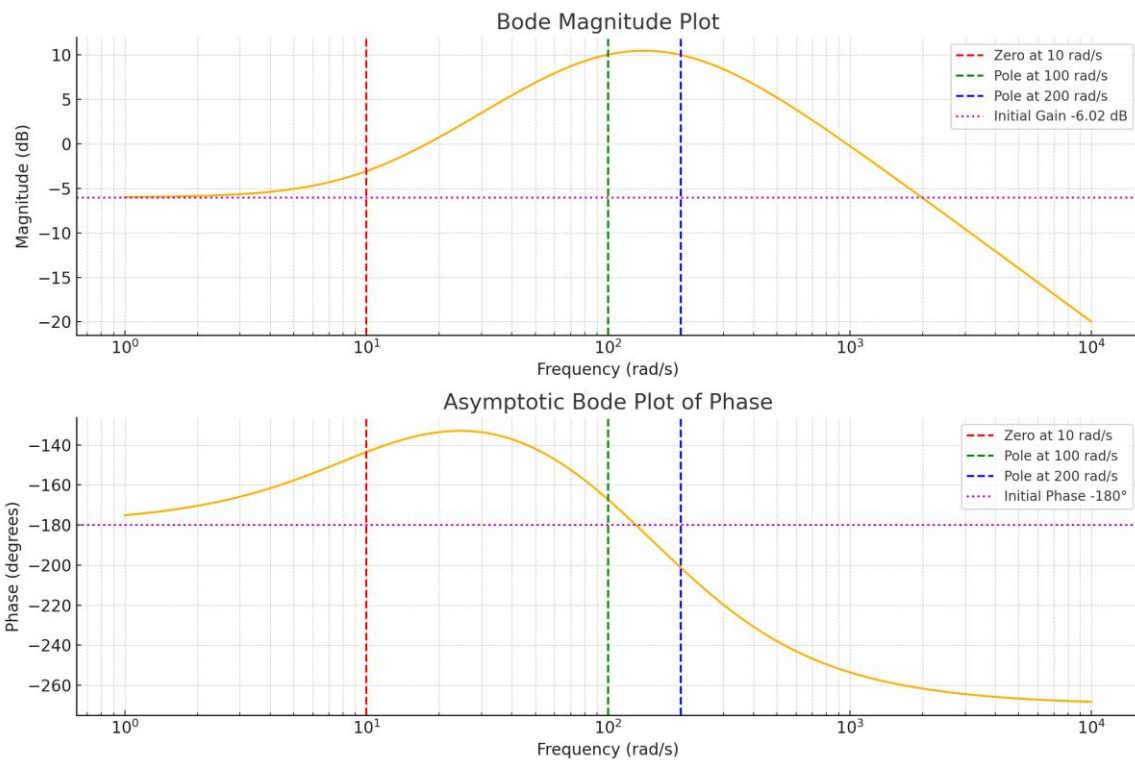
- Zero at $\omega = 10$ adds $+45^\circ$.
- Phase at $\omega = 100$:
 $-180^\circ + 45^\circ = -135^\circ$

Frequency Range $100 < \omega < 200$:

- Pole at $\omega = 100$ adds -45° .
- Phase at $\omega = 200$:
 $-135^\circ - 45^\circ = -180^\circ$

Frequency Range $\omega > 200$:

- Pole at $\omega = 200$ adds -45° .
- Phase:
 $-180^\circ - 45^\circ = -225^\circ$



配分

(a) Bode Magnitude Plot (15%)

1. Initial Gain Marking (3%): Correctly mark the initial gain as -6.02 dB.
2. Zero Marking (3%): Correctly identify and mark the zero at $\omega = 10$.
3. First Pole Marking (3%): Correctly identify and mark the first pole at $\omega = 100$.
4. Second Pole Marking (3%): Correctly identify and mark the second pole at $\omega = 200$.
5. Plot Accuracy (3%): Correctly plot the magnitude response curve with appropriate slopes and transitions.

b) Asymptotic Bode Phase Plot (15%)

1. Initial Phase Marking (3%): Correctly mark the initial phase as -180° .
2. Zero Phase Shift (3%): Correctly identify and mark the phase shift due to the zero at $\omega = 10$.
3. First Pole Phase Shift (3%): Correctly identify and mark the phase shift due to the first pole at $\omega = 100$.
4. Second Pole Phase Shift (3%): Correctly identify and mark the phase shift due to the second pole at $\omega = 200$.
5. Plot Accuracy (3%): Correctly plot the phase response curve with appropriate phase shifts and transitions.

3. (40%) Consider the circuit in Fig. B.

- (a) (10%) Derive the frequency response of the gain $G_v(j\omega) = V_o(j\omega)/V_s(j\omega)$ in terms of R_S (source resistance), R_L (load resistance), L (inductor) and C (capacitor), where $V_s(j\omega)$ and $V_o(j\omega)$ are the phasors of $v_s(t)$ and $v_o(t)$, respectively.
- (b) (10%) Determine the quality factor (Q) in terms of R_S, R_L, L and C .
- (c) (10%) Determine the bandwidth (BW) in terms of R_S, R_L, L and C .
- (d) (10%) If $R_S = 30 \Omega, R_L = 150 \Omega, L = 2 \times 10^{-4} \text{ H}$ and $C = 2 \times 10^{-6} \text{ F}$, calculate the half-power frequencies.

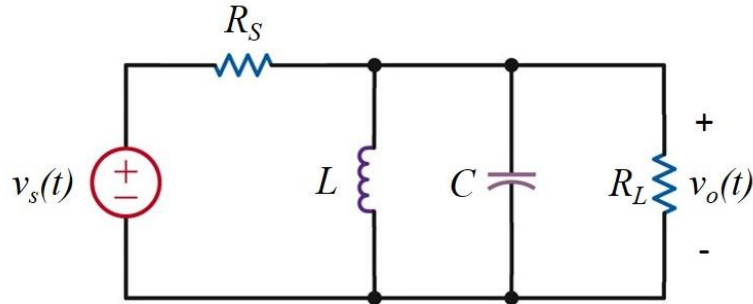


Fig. B

Sol:

(a)

$$G_v(j\omega) = G_v(S) = \frac{v_o(S)}{v_s(S)} = \frac{SL // \frac{1}{SC} // R_L}{R_S + (SL // \frac{1}{SC} // R_L)} \quad (2\%)$$

$$= \frac{S \frac{1}{R_S C}}{S^2 + S \left(\frac{R_S + R_L}{R_S R_L C} \right) + \frac{1}{LC}} \quad (2\%)$$

Let $S = j\omega$

$$= \frac{j\omega \frac{1}{R_S C}}{(j\omega)^2 + j\omega \left(\frac{R_S + R_L}{R_S R_L C} \right) + \frac{1}{LC}}$$

$$= \frac{j}{\left(\frac{R_S}{\omega L} - \omega R_S C \right) + j \frac{R_S + R_L}{R_L}} \quad (6\%)$$

(c)

At resonant, $\omega = \omega_0$, the network become pure resistive: (2%)

$$\frac{R_s}{\omega_0 L} - \omega_0 R_s C = 0$$

Yield:

$$\omega_0 = \frac{1}{\sqrt{LC}}$$

Next, find ω_{Lo} and ω_{Hi} by letting:

$$\begin{aligned} \left(\frac{R_s}{\omega L} - \omega R_s C\right)^2 &= \left(\frac{R_s + R_L}{R_L}\right)^2 \\ \Rightarrow \omega^2 \pm \omega \frac{R_s + R_L}{R_s R_L C} - \frac{1}{LC} &= 0 \quad (2\%) \end{aligned}$$

Solve for ω_{Hi} and ω_{Lo}

$$\begin{aligned} \omega_{Hi} &= \frac{\frac{R_s + R_L}{R_s R_L C} \pm \sqrt{\left(\frac{R_s + R_L}{R_s R_L C}\right)^2 + \frac{4}{LC}}}{2} \quad (2\%) \\ \omega_{Lo} &= \frac{-\frac{R_s + R_L}{R_s R_L C} \pm \sqrt{\left(\frac{R_s + R_L}{R_s R_L C}\right)^2 + \frac{4}{LC}}}{2} \quad (2\%) \end{aligned}$$

The bandwidth is defined as $\omega_{Hi} - \omega_{Lo}$:

$$BW = \frac{R_s + R_L}{R_s R_L C} \quad (2\%)$$

(b) Q is defined as:

$$\begin{aligned} Q &= \frac{\omega_0}{BW} \quad (3\%) \\ &= \frac{R_s R_L C}{(R_s + R_L)\sqrt{LC}} \quad (7\%) \end{aligned}$$

(d)

Half-power frequencies:

$$\omega_{Hi} = 9703\text{Hz}(60966\text{rad/s}) \quad (5\%)$$

$$\omega_{Lo} = 6527\text{Hz}(41010\text{rad/s}) \quad (5\%)$$

$$BW = 3176\text{Hz}(19956\text{rad/s})$$

$$\omega_0 = 7960\text{Hz}(50014\text{rad/s})$$

$$Q = \frac{\omega_0}{BW} = 2.5$$