

Standard Z-Transform Pair

$$x(n)$$

$$\leftrightarrow X(z)$$

ROC

$$\delta[n]$$

$$1$$

All z

$$\delta[n-m]$$

$$z^{-m}$$

All z except 0 if $m > 0$

$$n[n]$$

$$\frac{1}{1-z^{-1}}, \frac{z}{z-1}$$

$$|z| > 1$$

$$-n[-n-1]$$

$$\frac{1}{1-z^{-1}}, \frac{z}{z-1}$$

$$|z| < 1$$

$$2n\,n[n]$$

$$\frac{1}{1-2z^{-1}}, \frac{z}{z-2}$$

$$|z| > |2|$$

$$-2n\,0[-n-1]$$

$$\frac{1}{1-2z^{-1}}, \frac{z}{z-2}$$

$$|z| < |2|$$

$$\cos \omega_0 n\,n[n]$$

$$\frac{z^2 - \cos \omega_0 z}{z^2 - 2 \cos \omega_0 z + 1}$$

$$|z| > 1$$

$$\sin \omega_0 n\,n[n]$$

$$\frac{\sin \omega_0 z}{z^2 - 2 \cos \omega_0 z + 1}$$

$$|z| > 1$$

$$\pi n \cos \omega_0 n\,n[n]$$

$$\frac{z^2 - 2 \cos \omega_0 z}{z^2 - 2 \cos \omega_0 z + r^2}$$

$$|z| > |r|$$

$y^n \sin(n\pi)$ $\sin(\omega z)$ $|z| > |r|$

$$\frac{z}{z^2 - 2r\cos(\omega z) + r^2}$$

$$x(n) = \begin{cases} ar^n, & \text{for } 0 \leq n \leq N-1 \\ 0, & \text{else} \end{cases} \quad \frac{1 - ar^N z^{-N}}{1 - arz^{-1}}$$

~~no~~ $n a^n v[n]$

$$\frac{z^2}{(z-2)^2}$$

 $|z| > |r|$ ~~no~~ $-n a^n v[-n-1]$

$$\frac{z^2}{(z+2)^2}$$

 $|z| < |r|$

formulae from property.

$$y[n] \leftrightarrow y(z)$$

$$x[n] \leftrightarrow X(z)$$

$$x[n-n_0] \leftrightarrow z^{-n_0} X(z)$$

$$x(-n) \leftrightarrow X(z^{-1})$$

$$a^n x(n) \leftrightarrow X(\frac{z}{a})$$

$$n x(n) \leftrightarrow -\frac{d}{dz} [x(z)]$$

$$x^*[n] \leftrightarrow X^*(z^*)$$

$$X(z) = \sum_{n=-\infty}^{\infty} x(n) \cdot z^{-n}$$

$$x(n) = z^{-1}(X(z))$$



Inverse \leftrightarrow Transform

$$x[n] = \mathcal{Z}^{-1}\{X(z)\}$$

Finding the seq. $x[n]$ from its \mathcal{Z} transform $X(z)$ is called inverse \mathcal{Z} -transform

① Inversion formula

$$x[n] = \frac{1}{2\pi j} \int_C X(z) z^{-n-1} dz$$

Where C is anticlockwise contour of integration enclosing origin.

② Using tables

Given $X(z)$,

$$X(z) = X_1(z) + X_2(z) + \dots + X_K(z)$$

where

$X_1(z), X_2(z), \dots, X_K(z)$ are known \mathcal{Z} transform pair of seq $x_1[n], x_2[n], \dots, x_K[n]$ respectively

so that,

$$x[n] = x_1[n] + x_2[n] + \dots + x_K[n]$$

(iii)

Division method (Power Series expansion)

$$X(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$$

$$= \dots + x[-2] z^2 + x[-1] z + x[0] + x[1] z^{-1} \\ + x[2] z^{-2} + \dots$$

In this method Coeff. of seq. $x[n]$ is determined by comparing the Coeff. of appropriate power of z^{-1} in $x(z)$. This method is useful for finite length Seq. This method may not provide closed form soln, in such a case long division method can be applied.

(iv)

Partial fraction expansion

Given $X(z)$,

$$X(z) = \frac{N(z)}{D(z)}$$

$$X(z) = \frac{(z - z_1)(z - z_2) \dots (z - z_m)}{(z - p_1)(z - p_2) \dots (z - p_n)}$$

$m \neq$ zeros

$n \neq$ poles

② if $m < n$ and all-poles are simple then partial fraction has the form

$$\frac{X(z)}{z} = \frac{c_0}{z} + \frac{c_1}{z-p_1} + \frac{c_2}{z-p_2} + \dots + \frac{c_n}{z-p_n}$$

where,

$$c_0 = X(z) \Big|_{z=0}$$

$$c_k = (z - p_k) X(z) \Big|_{z=p_k}$$

③ if $m > n$, improper rational function

first make $X(z)$ proper rational function then go for partial fraction expansion

$$X(z) = \sum_{q=0}^{m-n} b_q z^{-q} + \sum_{k=1}^n c_k \cdot \frac{1}{1 - p_k z^{-1}}$$

④ multiple poles at $(z - p_i)^r$

r poles at $z = p_i$

then term containing multiple poles can be expressed by partial function as

$$\frac{d_1}{z - p_i} + \frac{d_2}{(z - p_i)^2} + \dots + \frac{d_r}{(z - p_i)^r}$$

Where,

$$d_{r-k} = \frac{1}{k!} \frac{d^k}{dz^k} \Big|_{z=p_i} (z - p_i)^r X(z) \Big|_{z=p_i}$$

Properties of \bar{z} -Transform

\bar{z} transform of $x[n]$

$$X(\bar{z}) = \sum_{n=-\infty}^{\infty} x[n] \bar{z}^{-n}$$

and

inverse \bar{z} - transform of $X(\bar{z})$ is given by

$$x[n] = \frac{1}{2\pi j} \oint_C X(\bar{z}) \bar{z}^{n-1} d\bar{z}$$

Notation: $x[n] \leftrightarrow \bar{z} X(\bar{z}), \text{ ROC: } R$

i) linearity

$$\text{if } x_1[n] \leftrightarrow \bar{z} X_1(\bar{z}), \text{ ROC: } R_1 \\ x_2[n] \leftrightarrow \bar{z} X_2(\bar{z}), \text{ ROC: } R_2$$

then,

$$\alpha_1 x_1[n] + \alpha_2 x_2[n] \leftrightarrow \bar{z} \alpha_1 X_1(\bar{z}) + \alpha_2 X_2(\bar{z}),$$

ROC: $R_1 \cap R_2$

(ii) Time Reversal

$$\text{if } x[n] \leftrightarrow \bar{z} X(\bar{z}), \text{ ROC: } R$$

then

$$x[-n] \leftrightarrow \bar{z} X(\bar{z}^{-1}) = X\left(\frac{1}{\bar{z}}\right), \text{ ROC: } \frac{1}{R}$$

After time reversal, a pole or zero in $X(\frac{z}{\bar{z}})$ at $\bar{z} = \frac{1}{z} K$ moves to $\bar{z} = \frac{1}{\frac{1}{z} K}$

Proof

$$\bar{z} \{x[-n]\} = \sum_{n=-\infty}^{\infty} x[-n] \bar{z}^{-n}$$

put

$$-n = k$$

$$= \sum_{k=-\infty}^{\infty} x[k] \bar{z}^{-k}$$

$$= \sum_{k=-\infty}^{\infty} x[k] \left(\frac{1}{z}\right)^{-k}$$

$$= X\left(\frac{1}{z}\right), \text{ ROC: } \frac{1}{R}$$

(iii)

Time shifting

$$\text{if } x[n] \leftrightarrow X\left(\frac{z}{\bar{z}}\right), \text{ ROC: } R$$

then

$$x[n-n_0] \leftrightarrow \bar{z}^{-n_0} X\left(\frac{z}{\bar{z}}\right),$$

$$\text{ROC: } R n \left| \frac{z}{\bar{z}} \right| > 0 \text{ if } n_0 > 0$$

$$x[n-1] \leftrightarrow \bar{z}^{-1} X\left(\frac{z}{\bar{z}}\right), \text{ ROC: } R n \left| \frac{z}{\bar{z}} \right| < \infty \text{ if } n_0 < 0$$

unit delay

$n[n+1] \leftrightarrow z \rightarrow zX(z)$, ROC: $R > |z| < \infty$

↓
Unit Advance.

Proof Step

$$\sum n[n-n_0] = \sum_{n=-\infty}^{\infty} n[n-n_0] z^{-n}$$

$$\text{put } n-n_0=k$$

$$= \sum_{k=-\infty}^{\infty} n[k] z^{-k-n_0}$$

$$= z^{-n_0} \sum_{k=-\infty}^{\infty} n[k] z^{-k}$$

$$= z^{-n_0} X(z),$$

(iv)

multiplication by z^m

if $n[n_0] \leftrightarrow X(z)$, ROC: R

then,

$z^m n[n] \leftrightarrow X(\frac{z}{z^m})$, ROC: $|z^m|/R$

After multiplication of sequence $n[n]$ by z^m , a pole or zero in $n(z)$ at $z = zK$ moves to $z = z_0 \cdot zK$ if region of convergence will expand or contract by a factor of $\text{abs}(z_0)$

PROOF :-

$$\begin{aligned} \sum_{n=-\infty}^{\infty} n x[n] &= \sum_{n=-\infty}^{\infty} x[n] n z^{-n} \\ &= \sum_{n=-\infty}^{\infty} n x[n] \left(\frac{z}{z_0}\right)^{-n} \\ &= X\left(\frac{z}{z_0}\right) \end{aligned}$$

$$e^{j\omega_0 n} x[n] \leftrightarrow X\left(\frac{z}{e^{j\omega_0}}\right), \text{ ROC: } R$$

⑤ multiplication by n (differentiation in $\frac{d}{dz}$)

$$\text{if } x[n] \leftrightarrow X(z), \text{ ROC: } R$$

then,

$$n x[n] \leftrightarrow -\frac{d}{dz} X(z) \quad \text{ROC: } R$$

PROOF :-

We have

$$X(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$$

Differentiating on both sides wrt z ,

$$\frac{d}{dz} X(z) = \sum_{n=-\infty}^{\infty} x[n] (-n z^{-n-1})$$

$$-\bar{z} \frac{d}{dz} X(z) = \sum_{n=-\infty}^{\infty} n x[n] z^{-n}$$

$$\therefore -\bar{z} \left\{ n x[n] \right\} = -\bar{z} \frac{d}{dz} X(z)$$

Point 2:

(vi) Complex Conjugate:

(contd. from Red Copy:-)

$X(z) = \frac{z}{2z^2 - 3z + 1}, |z| > 1$

Using division method.

Since, ROC is $|z| > 1$ Seq. $a[n]$ must be Right-sided
 So we must divide ~~so we~~ to obtain the series
 in power of z^{-1}

$$\begin{array}{r}
 \frac{1}{2} z^{-1} \\
 \hline
 2z^2 - 3z + 1 \overline{)z} \\
 -z - \frac{3}{2} + \frac{1}{2} z^{-1} \\
 \hline
 \frac{3}{2} - \frac{1}{2} z^{-1} \\
 \hline
 \frac{3}{2} - \frac{9}{4} z^{-1} + \frac{3}{4} z^{-2} \\
 - \quad + \quad - \\
 \hline
 \frac{7}{4} z^{-1} - \frac{9}{4} z^{-2} \\
 \hline
 \frac{7}{4} z^{-1} - \frac{21}{8} z^{-2} + \frac{7}{8} z^{-1} \\
 + \quad - \quad \\
 \hline
 \frac{15}{8} z^{-2} - \frac{7}{8} z^{-3}
 \end{array}$$

$$X(z) = \frac{1}{2}z^{-1} + \frac{3}{4}z^{-2} + \frac{7}{8}z^{-3} + \frac{15}{16}z^{-4} + \dots$$

Computing with $X(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$

We get

$$x[1] = \frac{1}{2}, x[2] = \frac{3}{4}, x[3] = \frac{7}{8}, x[4] = \frac{15}{16}$$

$x[n] = 0$, for $n \leq 0$

$$x[n] = \left\{ 0, \frac{1}{2}, \frac{3}{4}, \frac{7}{8}, \frac{15}{16}, \dots \right\}$$

or
 $x[n] = \left(1 - \left(\frac{1}{2} \right)^n \right) u(n)$

④ $X(z) = \frac{z}{2z^2 - 3z + 1}$

① $|z| > 1$

② $|z| < \frac{1}{2}$

③ $\frac{1}{2} < |z| < 1$

so/n.

$$\frac{x(z)}{z} = \frac{1}{2z^2 - 3z + 1}$$

$$\frac{x(z)}{z} = \frac{1}{2(z - \frac{1}{2})(z - 1)} = \frac{c_1}{z - \frac{1}{2}} + \frac{c_2}{z - 1}$$

Where

$$C_1 = \left(\frac{z}{2} - \frac{1}{2} \right) \frac{x(z)}{z} \Bigg|_{z=1} \quad z \neq \frac{1}{2}$$

$$= \left(\frac{z}{2} - \frac{1}{2} \right) \frac{1}{2 \left(\frac{z}{2} - \frac{1}{2} \right) \left(\frac{z}{2} - 1 \right)} \Bigg|_{z=\frac{1}{2}}$$

$$= -1$$

$$C_2 = (z-1) \frac{x(z)}{z} \Bigg|_{z=1}$$

$$= (z-1) \frac{1}{2 \left(z - \frac{1}{2} \right) \left(z - 1 \right)} \Bigg|_{z=1}$$

$$= 1$$

$$X(z) = -\frac{z}{z - \frac{1}{2}} + \frac{z}{z - 1}$$

② $\text{ROC: } |z| > 1$,

We have

$$z \in \text{dom} \quad \xleftrightarrow{\frac{z}{z-2}} \quad |z| > |2|$$

Taking Inverse z -transform Using table,
we get

$$x[n] = \left(\frac{1}{2}\right)^n u[n] + v[n]$$

$$= \left(1 - \left(\frac{1}{2}\right)^n\right) v[n] //$$

(b) ROC: $|z| < \frac{1}{2}$

We have

$$-2^n v[-n-1] \leftrightarrow \frac{z}{z-2}, |z| > |2|$$

Taking inverse Z-transform Using table,
we get

$$x[n] = \left(\frac{1}{2}\right)^n v[-n-1] - v[-n-1]$$

$$= \left(\left(\frac{1}{2}\right)^n - 1\right) v[-n-1]$$

(c) ROC: $\frac{1}{2} < |z| < 1$

We have

$$2^n v[n] \leftrightarrow \frac{z}{z-2}, |z| > |2|$$

$$-2^n v[-n-1] \leftrightarrow \frac{z}{z-2}, |z| < |2|$$

Taking inverse Z-transform Using tables,
we get

$$x[n] = -\left(\frac{1}{2}\right)^n v[n] - v[-n-1]$$

Find $n \in \mathbb{N} \}$

+10

$$\text{if } X\left(\frac{z}{2}\right) = \frac{\frac{z^2}{2}}{z^2 - z - 6}$$

- (2) $|z| > 3$
- (3) $|z| < 2$
- (c) $2 < |z| < 3$

eg: find $n \in \mathbb{N} \}$

$$X\left(\frac{z}{2}\right) = \frac{2z^3 + 5z^2 + z - 1}{2z^2 - 3z + 1}, \left|\frac{z}{2}\right| < \frac{1}{2}$$

Since $X\left(\frac{z}{2}\right)$ is improportion of function first make it proportional function then go for partial fraction expansion.

so/n,

$$\begin{array}{r}
 \frac{z}{2} \\
 \hline
 2z^2 - 3z + 1) \overbrace{2z^3 + 5z^2 + z - 1} \\
 \quad \quad \quad \overbrace{2z^3 - 3z^2 + z} \\
 \quad \quad \quad - \quad + \quad - \\
 \hline
 \quad \quad \quad \overbrace{8z^2 + 0 - 1} \\
 \quad \quad \quad \overbrace{8z^2 - 12z + 4} \\
 \quad \quad \quad + \quad - \\
 \hline
 \quad \quad \quad 12z - 5
 \end{array}$$

$$X\left(\frac{z}{2}\right) = z + 4 + \frac{12z - 5}{2z^2 - 3z + 1}$$

Let

$$X_1(z) = \frac{12z-5}{2z^2-3z+1}$$

$$\begin{aligned} X_1(z) &= \frac{12z-5}{2z(z-\frac{1}{2})(z-1)} \\ &= \frac{C_0}{2z} + \frac{C_1}{z-\frac{1}{2}} + \frac{C_2}{z-1} \end{aligned}$$

(where)

$$C_0 = X_1(z) \Big|_{z=0}$$

$$= \frac{12z-5}{2z^2-3z+1} \Big|_{z=0}$$

$$= -5$$

$$C_1 = \left(z - \frac{1}{2} \right) \frac{X_1(z)}{z} \Big|_{z=\frac{1}{2}}$$

$$= \left(z - \frac{1}{2} \right) \times \frac{12z-5}{2z(z-\frac{1}{2})(z-1)}$$

$$= \frac{12z-5}{2(z-\frac{1}{2})} = \frac{6-5}{-\frac{1}{2}} = -2$$

$$C_2 = \left(\frac{z-1}{z} \right) \frac{x_1(z)}{z} \Bigg|_{z=1}$$

$$= \frac{12z - 5}{2z(z-1)} \Bigg|_{z=1}$$

$$= 7$$

$$x_1(z) = -5 - \underbrace{\frac{2z}{z-\frac{1}{2}}}_{\frac{2z}{z-\frac{1}{2}}} + \underbrace{\frac{7z}{z-1}}_{\frac{7z}{z-1}}$$

$$x(z) = z + \underbrace{4 - 5 - \frac{2z}{z-\frac{1}{2}}}_{\frac{2z}{z-\frac{1}{2}}} + \underbrace{7z}_{z-1}$$

$$= z - 1 - \underbrace{\frac{2z}{z-\frac{1}{2}}}_{\frac{2z}{z-\frac{1}{2}}} + \underbrace{\frac{7z}{z-1}}_{\frac{7z}{z-1}}$$

$$\text{ROC: } |z| < \underline{\frac{1}{2}}$$

We have,

$$\delta[n] \xleftarrow[z]{z} 1, \text{ All } z$$

$$\delta[n-m] \xleftarrow[z]{z} z^{-m}, \text{ All } z \text{ except 0 if } m > 0 \\ \infty \text{ if } m < 0$$

$$-a^n \delta[-n-1] \xleftarrow[z]{z} \frac{z}{z-a}, |z| < |a|$$

Taking inverse Z-transform Using Tables

We get

$$x(n) = \delta(n+1) - \delta(n) + 2\left(\frac{1}{2}\right)^n u(n-1) - 7u(-n-1)$$

H/W

$$X(z) = \frac{2z^4 + z^2 - 6}{z^2 - 5z + 6}, |z| < 2$$

Now find $x(n)$

$$\text{if } X(z) = \frac{z}{(z-1)(z-2)^2}, |z| > 2$$

so/n

$$\frac{X(z)}{z} = \frac{1}{(z-1)(z-2)^2}$$

$$\frac{X(z)}{z} = \frac{c_1}{z-1} + \frac{c_1}{z-2} + \frac{c_2}{(z-2)^2}$$

where,

$$C_1 = \left(\frac{z-1}{z} \right) \frac{x(z)}{z} \Big|_{z=1}$$

$$= (z-1) \frac{1}{(z-1)(z-2)^2} \Big|_{z=1}$$

~~z=1~~

$$C_2 = \frac{1}{0!} \frac{d^0}{dz^0} \left(\frac{(z-2)^2 x(z)}{z} \right) \Big|_{z=2}$$

$$= (z-2)^2 + \frac{1}{(z-1)(z-2)^2} \Big|_{z=2}$$

= 1

$$C_1 = \frac{1}{1!} \frac{d}{dz} \left(\frac{(z-2)^2 x(z)}{z} \right) \Big|_{z=2}$$

$$= \frac{d}{dz} (z-2)^2 \frac{1}{(z-1)(z-2)^2} \Big|_{z=2}$$

$$= -1 (z-1)^{-2} \frac{d}{dz} (z-1) \Big|_{z=2}$$

$$= - (z-1)^{-2} \Big|_{z=2}$$

= -1

$$X(z) = \frac{z}{z-1} - \frac{z}{z-2} \left(\frac{z}{z+2} \right)^2 \times \frac{1}{2} z^2$$

ROC: $|z| > 2$

We have

$$e^{n\ln z} \xleftarrow[z \rightarrow \infty]{z} z^n, |z| > |z|$$

$$\eta 2^n e^{n\ln z} \xleftarrow[z \rightarrow \infty]{z} \frac{2^n}{(z-2)^2}, |z| > |z|$$

Taking inverse Z-transform apply tables

We get

$$\begin{aligned} x[n] &= v[n] - 2^n v[n] + \frac{1}{2} \eta 2^n v[n] \\ &= (1 - 2^n + n 2^{n-1}) v[n] \end{aligned}$$

H/w.

Find $x[n]$

$$\text{If } X(z) = \frac{z}{(z+2)(z-3)^2}, |z| < 2$$

Find O/P of LTI system having I/P sequence $x(n) = \{1, -2, 3, 5\}$
and impulse Response $h(n) = \{2, 1, 6, 3\}$ Using Z-transform.

80/n.

Z-transform of $x(n)$

$$X(z) = \sum_{n=-\infty}^{\infty} x(n) z^{-n}$$

$$= \dots + x(-2)z^2 + x(-1)z + x(0) + x(1)z^{-1} + \dots$$

$$= 4z^2 - 2z + 3 + 5z^{-1} \quad \text{ROC: } 0 < |z| < \infty$$

(∞ तक सम्पूर्ण जगह सुनिश्चित है)

Similarly Z-transform of $h(n)$

$$H(z) = \sum_{n=-\infty}^{\infty} h(n) z^{-n}$$

$$= \dots + h(0) + h(1)z^{-1} + h(2)z^{-2} + h(3)z^{-3} + \dots$$

$$= 2 + z^{-1} + 6z^{-2} + 3z^{-3}, \quad \text{ROC: } |z| > 0$$

बास इसीलिए 0 का वाला वर्ग वाला है।

O/P of LTI system is given by

$$y(n) = x(n) * h(n)$$

In Z-domain,

$$\begin{aligned} y(z) &= X(z) \cdot H(z), \quad \text{ROC: } R_1 \cap R_2 \\ &= (4z^2 - 2z + 3 + 5z^{-1})(2 + z^{-1} + 6z^{-2} + 3z^{-3}) \\ &= 8z^2 - 4z + 6 + 10z^{-1} + 4z^{-2} - 2 + 3z^{-3} + 5z^{-4} \\ &\quad + 24 - 12z^{-1} + 18z^{-2} + 30z^{-3} + 12z^{-4} \\ &\quad - 6z^{-2} + 9z^{-3} + 15z^{-4} \\ &= 8z^2 + 28 + 17z^{-1} + 17z^{-2} + 39z^{-3} + 15z^{-4} \end{aligned}$$

ROC: $0 < |z| < \infty$

$$C_2 = \left(\frac{z-\beta}{z-\alpha} \right) \cdot \frac{Y(z)}{z} \Bigg|_{z=\beta}$$

$$= \left(\frac{z-\beta}{z-\alpha} \right) \frac{z}{(z-\alpha)(z-\beta)} \Bigg|_{z=\beta}$$

$$= \frac{\beta}{\beta-\alpha} = -\frac{\beta}{\alpha-\beta}$$

$$Y(z) = \frac{\alpha}{\alpha-\beta} \cdot \frac{z}{z-\alpha} - \frac{\beta}{\alpha-\beta} \cdot \frac{z}{z-\beta}, |z| > \max\{|\alpha|, |\beta|\}$$

Taking inverse \bar{z} -transform using table,

$$Y(n) = \frac{\alpha}{\alpha-\beta} \alpha^n u(n) - \frac{\beta}{\alpha-\beta} \beta^n u(n)$$

$$= \frac{\alpha^{n+1} - \beta^{n+1}}{\alpha-\beta} u(n)$$

When $\alpha = \beta$,

$$Y(z) = \frac{z}{z-\alpha} \cdot \frac{z}{z-\alpha}, |z| > |\alpha|$$

$$\frac{Y(z)}{z} = \frac{z}{(z-\alpha)^2} = \frac{d_1}{z-\alpha} + \frac{d_2}{(z-\alpha)^2}$$

where,

$$d_2 = \frac{1}{0!} \frac{d^0}{dz^0} \frac{(z-\alpha)^2 Y(z)}{z} \Bigg|_{z=\alpha}$$

$$= (z-\alpha)^2 \cdot \frac{z}{(z-\alpha)^2} \Bigg|_{z=\alpha}$$

$$= \alpha$$

$$d_1 = \frac{1}{1!} \left. \frac{d}{dz} (z-\alpha)^2 y(z) \right|_{z=\alpha}$$

$$= \frac{d}{dz} (z-\alpha)^2 \left. \frac{z}{(z-\alpha)^2} \right|_{z=\alpha}$$

$$= 1$$

$$y(z) = \frac{z}{z-\alpha} + \frac{\alpha z}{(z-\alpha)^2}, |z| > |\alpha|$$

We have

$$n(n) \leftrightarrow \frac{z}{(z-a)} , |z| > |a|$$

Taking inverse Z transform using table

We get,

$$\begin{aligned} y(n) &= \alpha^n n(n) + n \alpha^n u(n) \\ &= (1+n) \alpha^n u(n) \end{aligned}$$

H/w find $y(n) = n(n) * h(n)$

$$\begin{aligned} ① \quad n(n) &= \left(\frac{1}{2}\right)^n u(n) \\ h(n) &= \left(\frac{1}{3}\right)^n u(n) \end{aligned}$$

$$② \quad n(n) = \left(\frac{1}{2}\right)^n u(n)$$

$$h(n) = 2^n u(n-1)$$

→ Complexity of LTI System in \mathbb{Z} -domain

To be an LTI System (causal)

$$h(n) = 0, \text{ for } n < 0$$

i.e. $h(n)$ is Right-sided sequence

If $H(z)$ exists,

then,

Roc has the form at least

$$|z| > 0, |z| > r_{\min}$$

→ Stability of LTI System in \mathbb{Z} -domain

To be an LTI System stable,

$$\left| \sum_{n=-\infty}^{\infty} h(n) \right| < \infty$$

If $H(z)$ exists,

$$H(z) = \left| \sum_{n=-\infty}^{\infty} h(n) z^{-n} \right|$$

To get $H(z)$ finite,

Roc must include $|z| = 1$ i.e. Unit Circle in z -plane

→ If LTI System is both stable & causal, then

poles of LTI System lies inside the Unit Circle
in z -plane

$$\textcircled{a} \quad h(n) = 3^n u(-n+1) \quad \xrightarrow{\text{Z}} H(z) = \frac{z}{z-3}, |z| < 3$$

non causal
(as $z > 3$ for causal),

$$\textcircled{b} \quad h(n) = 3^n u(n) \quad \leftrightarrow \quad \cancel{H(z) = \frac{z}{z-3}, |z| > 3}$$

stable (included)
causal,

as $z > 3$ and 1 is not included
so unstable

$$\textcircled{c} \quad h(n) = \left(\frac{1}{3}\right)^n u(-n-1) \quad \xrightarrow{\text{Z}} H(z) = \frac{z}{z-\frac{1}{3}}, |z| < \frac{1}{3}$$

non causal, unstable

$$\textcircled{d} \quad h(n) = \left(\frac{1}{3}\right)^n u(n) \quad \xrightarrow{\text{Z}} H(z) = \frac{z}{z-\frac{1}{3}}, |z| > \frac{1}{3}$$

causal, stable.

→ Recursive System:

There are many system whose present O/P depends upon a no. of past O/P. Such a system is called Recursive System & its O/P can be represented as

$$y[n] = f\{y[n-1], y[n-2], \dots, y[n-N], n[n], n[n-1], \dots, n[n-M]\}$$

where,

$f\{\cdot\} \rightarrow$ function

i.e: Feedback System.

→ Non-Recursive System:

A System is said to be non-recursive if its present O/P only depends upon present I/O & past I/P and represented as

$$y[n] = f\{n[n], n[n-1], \dots, n[n-M]\}$$

where

$f\{\cdot\} \rightarrow$ function

i.e: non-feedback System.

* Infinite Impulse Response (IIR) System

If length of impulse Response is ∞ then it is called IIR.

$$\text{eg: } h[n] = \left(\frac{1}{2}\right)^n u(n)$$

↳ Feedback System

↳ Recursive System

* Finite impulse Response (FIR) System

If length of impulse Response is finite then it is called FIR System.

$$\text{eg: } h[n] = \{3, 2, 5, 6\}$$

→ It is non-feedback system

→ It is non-recursive System.

moving avg. System

$$y[n] = \frac{1}{n+1} \sum_{k=0}^n x[k] \rightarrow \text{non recursive system}$$

$$y[n-1] = \frac{1}{n-1+1} \sum_{k=0}^{n-1} x[k]$$

$$ny[n-1] = \sum_{k=0}^{n-1} x[k]$$

$$y[n] = \frac{1}{n+1} \left[\sum_{k=0}^{n-1} n[k] + n[n] \right]$$

$$y[n] = \frac{1}{n+1} [n y[n-1] + n[n]]$$

$$y[n] = \frac{n}{n+1} y[n-1] + \frac{1}{n+1} n[n] \rightarrow \text{Recursive.}$$

→ Recursivity depends upon implementation of system.

Difference Equations:-

↑ The General form of Const. Coeff. difference equation for LTI System can be represented as:

$$\sum_{k=0}^N a_k y[n-k] = \sum_{k=0}^M b_k x[n-k]$$

$a_k, b_k \rightarrow \text{Coeff.}$

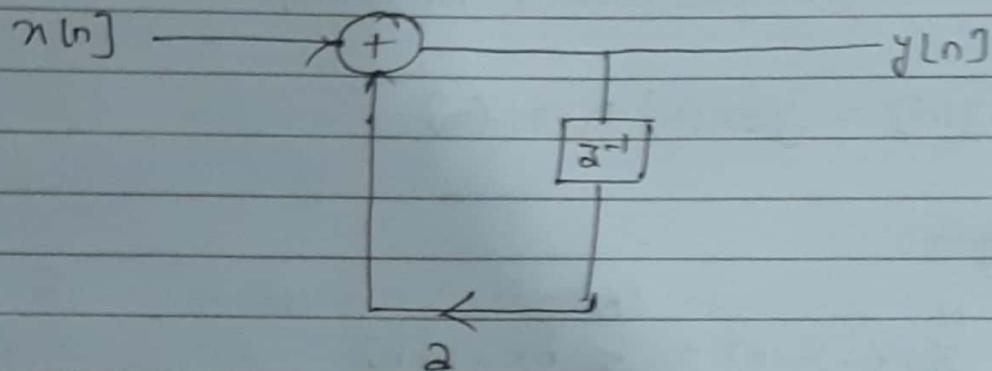
$M, N \rightarrow \text{no. of delay in all I/P \& O/P respectively.}$

$$a_0 y[n] + \sum_{k=1}^N a_k y[n-k] = \sum_{k=0}^M b_k x[n-k]$$

$$y[n] = \frac{1}{a_0} \left[\sum_{k=0}^M b_k x[n-k] - \sum_{k=1}^N a_k y[n-k] \right]$$

First order difference Equation

$$y[n] = ay[n-1] + x[n]$$



for $n=0$,

$$y[0] = ay[-1] + x[0]$$

for $n=1$,

$$y[1] = ay[0] + x[1] = a^2y[-1] + ax[0] + x[1]$$

for $n=2$,

$$y[2] = ay[1] + x[2] = a^3y[-1] + a^2x[0] + ax[1] + x[2]$$

:

:

:

$$y[n] = a^{n+1}y[-1] + a^n x[0] + a^{n-1}x[1] + \dots + x[n]$$

$$y[n] = a^{n+1}y[-1] + \sum_{k=0}^n a^k x[n-k]$$

I/P

When $x[n]=0$, for $n \geq 0$, ie zero initial condn then

$$y_{sr}(n) = a^{n+1}y[-1]$$

↗ Natural response

↗ zero-I/P response

When initial cond) zero ie $y(1-)=0$

then,

$$y[n] = \sum_{k=0}^n a_k x[n-k]$$

(\rightarrow Zero-state or forced response.)

$$y[n] = y_{ST}(n) + y_{IS}(n)$$

Again,

$$\sum_{k=0}^N a_k y[n-k] = \sum_{k=0}^N b_k x[n-k]$$

Taking \mathcal{Z} transform Using time shifting property of \mathcal{Z} -transform

$$\sum_{k=0}^N a_k z^{-k} y(z) = \sum_{k=0}^N b_k z^{-k} x(z)$$

$$\frac{y(z)}{x(z)} = \frac{\sum_{k=0}^M b_k z^{-k}}{\sum_{k=0}^N a_k z^{-k}}$$

$$H(z) = \frac{\sum_{k=0}^M b_k z^{-k}}{\sum_{k=0}^N a_k z^{-k}}$$

$$\leftarrow \sum_{k=0}^N a_k z^{-k}$$

Transfer function / System function :- Ratio of O/P to I/O in frequency domain, laplace domain or in \mathcal{Z} -domain.

$$H(z) = \frac{b_0 + b_1 z^{-1} + b_2 z^{-2} + \dots + b_N z^{-N}}{a_0 + a_1 z^{-1} + a_2 z^{-2} + \dots + a_N z^{-N}}$$

$H(z)$ contains M zeros
 N poles

If $N=0$, no poles (non-zero poles)
then FIR System

If $N \geq 1$, at least one non-zero pole then
IIR System