

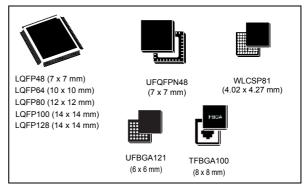
STM32G473xB STM32G473xC STM32G473xE

Arm® Cortex®-M4 32-bit MCU+FPU, up to 512 KB Flash, 170 MHz / 213 DMIPS, 128 KB SRAM, rich analog, math accelerator

Datasheet - production data

Features

- Includes ST state-of-the-art patented technology
- Core: Arm[®] 32-bit Cortex[®]-M4 CPU with FPU, Adaptive real-time accelerator (ART Accelerator) allowing 0-wait-state execution from Flash memory, frequency up to 170 MHz with 213 DMIPS, MPU, DSP instructions
- · Operating conditions:
 - V_{DD}, V_{DDA} voltage range: 1.71 V to 3.6 V
- · Mathematical hardware accelerators
 - CORDIC for trigonometric functions acceleration
 - FMAC: filter mathematical accelerator
- Memories
 - 512 Kbytes of Flash memory with ECC support, two banks read-while-write, proprietary code readout protection (PCROP), securable memory area, 1 Kbyte OTP
 - 96 Kbytes of SRAM, with hardware parity check implemented on the first 32 Kbytes
 - Routine booster: 32 Kbytes of SRAM on instruction and data bus, with hardware parity check (CCM SRAM)
 - External memory interface for static memories FSMC supporting SRAM, PSRAM, NOR and NAND memories
 - Quad-SPI memory interface
- Reset and supply management
 - Power-on/power-down reset (POR/PDR/BOR)
 - Programmable voltage detector (PVD)
 - Low-power modes: sleep, stop, standby and shutdown
 - V_{BAT} supply for RTC and backup registers



- · Clock management
 - 4 to 48 MHz crystal oscillator
 - 32 kHz oscillator with calibration
 - Internal 16 MHz RC with PLL option (± 1%)
 - Internal 32 kHz RC oscillator (± 5%)
- Up to 107 fast I/Os
 - All mappable on external interrupt vectors
 - Several I/Os with 5 V tolerant capability
- Interconnect matrix
- 16-channel DMA controller
- 5 x 12-bit ADCs 0.25 μs, up to 42 channels.
 Resolution up to 16-bit with hardware oversampling, 0 to 3.6 V conversion range
- 7 x 12-bit DAC channels
 - 3 x buffered external channels 1 MSPS
 - 4 x unbuffered internal channels 15 MSPS
- 7 x ultra-fast rail-to-rail analog comparators
- 6 x operational amplifiers that can be used in PGA mode, all terminals accessible
- Internal voltage reference buffer (VREFBUF) supporting three output voltages (2.048 V, 2.5 V, 2.9 V)
- 14 timers:
 - 2 x 32-bit timer and 2 x 16-bit timers with up to four IC/OC/PWM or pulse counter and quadrature (incremental) encoder input
 - 3 x 16-bit 8-channel advanced motor control timers, with up to 8 x PWM

- channels, dead time generation and emergency stop
- 1 x 16-bit timer with 2 x IC/OCs, one OCN/PWM, dead time generation and emergency stop
- 2 x 16-bit timers with IC/OC/OCN/PWM, dead time generation and emergency stop
- 2 x watchdog timers (independent, window)
- 1 x SysTick timer: 24-bit downcounter
- 2 x 16-bit basic timers
- 1 x low-power timer
- Calendar RTC with alarm, periodic wakeup from stop/standby
- · Communication interfaces
 - 3 x FDCAN controller supporting flexible data rate
 - 4 x I²C Fast mode plus (1 Mbit/s) with 20 mA current sink, SMBus/PMBus, wakeup from stop

- 5 x USART/UARTs (ISO 7816 interface, LIN, IrDA, modem control)
- 1x LPUART
- 4 x SPIs, 4 to 16 programmable bit frames, 2 x with multiplexed half duplex I²S interface
- 1 x SAI (serial audio interface)
- USB 2.0 full-speed interface with LPM and BCD support
- IRTIM (infrared interface)
- USB Type-C[™] /USB power delivery controller (UCPD)
- True random number generator (RNG)
- CRC calculation unit, 96-bit unique ID
- Development support: serial wire debug (SWD), JTAG, Embedded Trace Macrocell™

Table 1. Device summary

Reference	Part number
STM32G473xB	STM32G473CB, STM32G473MB, STM32G473PB, STM32G473RB, STM32G473VB, STM32G473QB
STM32G473xC	STM32G473CC, STM32G473MC, STM32G473PC, STM32G473RC, STM32G473VC, STM32G473QC
STM32G473xE	STM32G473CE, STM32G473ME, STM32G473PE, STM32G473RE, STM32G473VE, STM32G473QE

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1 Introduction

This datasheet provides the ordering information and mechanical device characteristics of the STM32G473xB/xC/xE microcontrollers.

This document should be read in conjunction with the reference manual RM0440 "STM32G4 Series advanced Arm[®] 32-bit MCUs". The reference manual is available from the STMicroelectronics website *www.st.com*.

For information on the Arm^{®(a)} Cortex[®]-M4 core, refer to the Cortex[®]-M4 technical reference manual, available from the www.arm.com website.

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2 Description

The STM32G473xB/xC/xE devices are based on the high-performance Arm[®] Cortex[®]-M4 32-bit RISC core. They operate at a frequency of up to 170 MHz.

The Cortex-M4 core features a single-precision floating-point unit (FPU), which supports all the Arm single-precision data-processing instructions and all the data types. It also implements a full set of DSP (digital signal processing) instructions and a memory protection unit (MPU) which enhances the application's security.

These devices embed high-speed memories (up to 512 Kbytes of flash memory, and 128 Kbytes of SRAM), a flexible external memory controller (FSMC) for static memories (for devices with packages of 100 pins and more), a Quad-SPI flash memory interface, and an extensive range of enhanced I/Os and peripherals connected to two APB buses, two AHB buses and a 32-bit multi-AHB bus matrix.

The devices also embed several protection mechanisms for embedded flash memory and SRAM: readout protection, write protection, securable memory area and proprietary code readout protection.

The devices embed peripherals allowing mathematical/arithmetic function acceleration (CORDIC for trigonometric functions and FMAC unit for filter functions).

They offer five fast 12-bit ADCs (4 Msps), seven comparators, six operational amplifiers, seven DAC channels (3 external and 4 internal), an internal voltage reference buffer, a low-power RTC, twothree general-purpose 32-bit timers, three 16-bit PWM timers dedicated to motor control, seven general-purpose 16-bit timers, and one 16-bit low-power timer.

They also feature standard and advanced communication interfaces such as:

- Four I2Cs
- Four SPIs multiplexed with two half duplex I2Ss
- Three USARTs, two UARTs and one low-power UART.
- Three FDCANs
- One SAI
- USB device
- UCPD

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The devices operate in the -40 to +85 °C (+105 °C junction) and -40 to +125 °C (+130 °C junction) temperature ranges from a 1.71 to 3.6 V power supply. A comprehensive set of power-saving modes allows the design of low-power applications.

Some independent power supplies are supported including an analog independent supply input for ADC, DAC, OPAMPs and comparators. A V_{BAT} input allows backup of the RTC and the registers.

The STM32G473xB/xC/xE family offers 8 packages from 48-pin to 128-pin.

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Table 2. STM32G473xB/xC/xE features and peripheral counts

	Peripheral	STM32G473 Cx			STM32G473 Rx			STM32G473 Mx			STM32G473 Vx			STM32G473 Px			STM32G473 Qx		
Fla	Flash memory		256 KB	512 KB	128 KB	256 KB	512 KB	128 KB	256 KB	512 KB	128 KB	256 KB	512 KB	128 KB	256 KB	512 KB	128 KB	256 KB	512 KB
SF	RAM							,	128 (80	+ 16-	+ 32)	KB							
External memory controller for static memories (FSMC)		No				Yes Yes ⁽¹⁾			Yes Yes										
Ql	JADSPI	1																	
	Advanced motor control	3 (16-bit)																	
	General purpose		5 (16-bit) 2 (32-bit)																
rs.	Basic		2 (16-bit)																
Timers	Low power	1 (16-bit)																	
	SysTick timer		1																
	Watchdog timers (independent, window)	2																	
	SPI(I2S) ⁽²⁾	3 (2)						4 (2)											
	I ² C									4									
SS	USART									3									
Common interfaces	UART		ı LQFF JFQFF		2														
i uoi	LPUART									1									
nmc	FDCANs									3									
ŏ	USB device									Yes									
	UCPD									Yes									
	SAI	Yes																	
RI	С									Yes									
Та	mper pins			2									3						
	ndom number nerator	Yes																	
CC	ORDIC	Yes																	
F۱	1AC									Yes									
GF	PIOs		n LQF UFQF			52			WLCS n LQF			86			102			107	
Wa	akeup pins		3			4			4			5			5			5	



Table 2. STM32G473xB/xC/xE features and peripheral counts (continued)

			•	·	<u> </u>	<u>, </u>		
Peripheral	STM32G473 Cx	STM32G473 Rx	STM32G473 Mx	STM32G473 Vx	STM32G473 Px	STM32G473 Qx		
12-bit ADCs	5							
Number of channels	20 in LQFP48 21 in UFQFPN48	26	42 in WLCSP81 41 in LQFP80	42	42	42		
12-bit DAC Number of channels			7 (3 external +	4 internal)				
Internal voltage reference buffer	Yes							
Analog comparator		7						
Operational amplifiers		6						
Max. CPU frequency			170 Mi	Hz				
Operating voltage		1.71 V to 3.6 V						
Operating temperature		Ambient opera	ating temperature:	re: -40 to 85 °C / -40 to 125 °C				
Packages	LQFP48/ UFQFPN48	LQFP64	WLCSP81 LQFP80	LQFP100/ TFBGA100	UFBGA121	LQFP128		

For the LQFP100 package, only FMC bank1 and NAND bank are available. Bank1 can only support a multiplexed NOR/PSRAM memory using the NE1 chip select.

^{2.} The SPI2/3 interfaces can work in an exclusive way in either the SPI mode or the I2S audio mode.

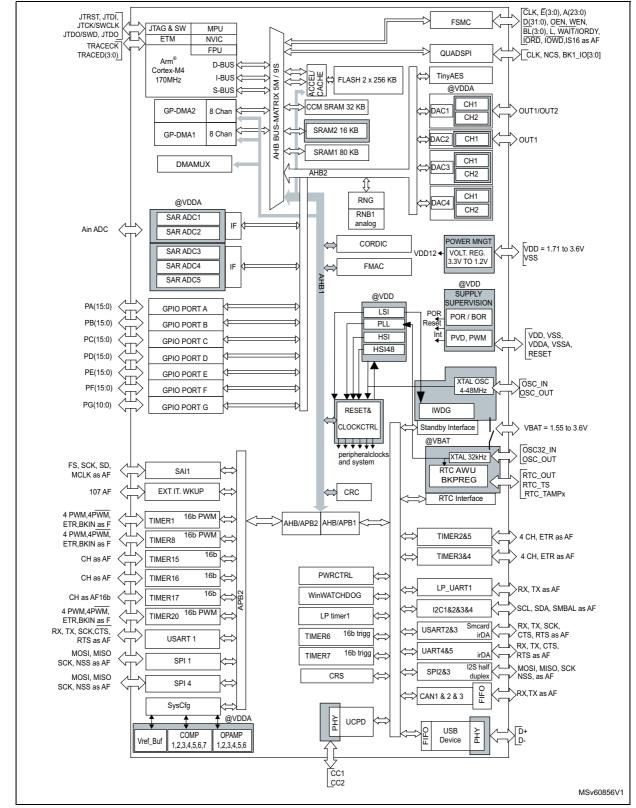


Figure 1. STM32G473xB/xC/xE block diagram

1. AF: alternate function on I/O pins.



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3 Functional overview

3.1 Arm[®] Cortex[®]-M4 core with FPU

The Arm[®] Cortex[®]-M4 with FPU processor is the latest generation of Arm processors for embedded systems. It was developed to provide a low-cost platform that meets the needs of the MCU implementation, with a reduced pin count and with low-power consumption, while delivering outstanding computational performance and an advanced response to interrupts.

The Arm[®] Cortex[®]-M4 with FPU 32-bit RISC processor features an exceptional code-efficiency, delivering the expected high-performance from an Arm core in a memory size usually associated with 8-bit and 16-bit devices.

The processor supports a set of DSP instructions which allows an efficient signal processing and a complex algorithm execution. Its single precision FPU speeds up the software development by using metalanguage development tools to avoid saturation.

With its embedded Arm core, the STM32G473xB/xC/xE family is compatible with all Arm tools and software.

Figure 1 shows the general block diagram of the STM32G473xB/xC/xE devices.

3.2 Adaptive real-time memory accelerator (ART accelerator)

The ART accelerator is a memory accelerator that is optimized for the STM32 industry-standard Arm[®] Cortex[®]-M4 processors. It balances the inherent performance advantage of the Arm[®] Cortex[®]-M4 over flash memory technologies, which normally requires the processor to wait for the flash memory at higher frequencies.

3.3 Memory protection unit

The memory protection unit (MPU) is used to manage the CPU accesses to the memory and to prevent one task to accidentally corrupt the memory or the resources used by any other active task. This memory area is organized into up to 8 protected areas, which can be divided in up into 8 subareas each. The protection area sizes range between 32 bytes and the whole 4 gigabytes of addressable memory.

The MPU is especially helpful for applications where some critical or certified code has to be protected against the misbehavior of other tasks. It is usually managed by an RTOS (real-time operating system). If a program accesses a memory location that is prohibited by the MPU, the RTOS can detect it and take action. In an RTOS environment, the kernel can dynamically update the MPU area setting based on the process to be executed.

The MPU is optional and can be bypassed for applications that do not need it.

3.4 Embedded flash memory

The STM32G473xB/xC/xE devices feature up to 512 Kbytes of embedded flash memory, which is available for storing programs and data.

The flash memory interface features:

- Single or dual bank operating modes
- Read-while-write (RWW) in dual bank mode

This feature allows to perform a read operation from one bank while an erase or program operation is performed to the other bank. The dual bank boot is also supported.

Flexible protections can be configured thanks to the option bytes:

- Readout protection (RDP) to protect the whole memory. Three levels of protection are available:
 - Level 0: no readout protection
 - Level 1: memory readout protection; the flash memory cannot be read from or written to if either the debug features are connected or the boot in RAM or bootloader are selected
 - Level 2: chip readout protection; the debug features (Cortex-M4 JTAG and serial wire), the boot in RAM and the bootloader selection are disabled (JTAG fuse). This selection is irreversible.
- Write protection (WRP): the protected area is protected against erasing and programming.
- Proprietary code readout protection (PCROP): a part of the flash memory can be
 protected against read and write from third parties. The protected area is execute-only
 and it can only be reached by the STM32 CPU as an instruction code, while all other
 accesses (DMA, debug and CPU data read, write and erase) are strictly prohibited. An
 additional option bit (PCROP_RDP) allows to select if the PCROP area is erased or not
 when the RDP protection is changed from Level 1 to Level 0.
- Securable memory area: a part of flash memory can be configured by option bytes to be securable. After reset this securable memory area is not secured and it behaves like the remainder of main flash memory (execute, read, write access). When secured, any access to this securable memory area generates corresponding read/write error. Purpose of the Securable memory area is to protect sensitive code and data (secure keys storage) which can be executed only once at boot, and never again unless a new reset occurs.

The flash memory embeds the error correction code (ECC) feature supporting:

- Single error detection and correction
- Double error detection
- The address of the ECC fail can be read in the ECC register
- 1 Kbyte (128 double word) OTP (one-time programmable) for user data. The OTP area is available in Bank 1 only. The OTP data cannot be erased and can be written only once.

 $\overline{\mathbf{A}}$

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3.5 Embedded SRAM

STM32G473xB/xC/xE devices feature 128 Kbytes of embedded SRAM. This SRAM is split into three blocks:

- 80 Kbytes mapped at address 0x2000 0000 (SRAM1). The CM4 can access the SRAM1 through the System Bus. The first 32 Kbytes of SRAM1 support hardware parity check.
- 16 Kbytes mapped at address 0x2001 4000 (SRAM2). The CM4 can access the SRAM2 through the System bus. SRAM2 can be retained in standby modes.
- 32 Kbytes mapped at address 0x1000 0000 (CCM SRAM). It is accessed by the CPU through I-Code/D-Code bus for maximum performance.
 It is also aliased at 0x2001 8000 address to be accessed by all masters (CPU, DMA1, DMA2) through SBUS contiguously to SRAM1 and SRAM2. The CCM SRAM supports hardware parity check and can be write-protected with 1-Kbyte granularity.
- The memory can be accessed in read/write at max CPU clock speed with 0 wait states.

3.6 Multi-AHB bus matrix

The 32-bit multi-AHB bus matrix interconnects all the masters (CPU, DMAs) and the slaves (flash memory, RAM, FSMC, QUADSPI, AHB and APB peripherals). It also ensures a seamless and efficient operation even when several high-speed peripherals work simultaneously.

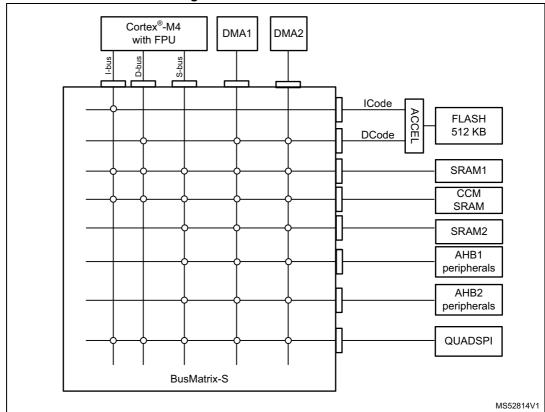


Figure 2. Multi-AHB bus matrix

3.7 Boot modes

At startup, a BOOT0 pin (or nBOOT0 option bit) and an nBOOT1 option bit are used to select one of three boot options:

- Boot from user flash memory
- Boot from system memory
- Boot from embedded SRAM

The BOOT0 value may come from the PB8-BOOT0 pin or from an nBOOT0 option bit depending on the value of a user nBOOT_SEL option bit to free the GPIO pad if needed.

The boot loader is located in the system memory. It is used to reprogram the flash memory by using USART, I2C, SPI, and USB through the DFU (device firmware upgrade).

3.8 CORDIC

The CORDIC provides hardware acceleration of certain mathematical functions, notably trigonometric, commonly used in motor control, metering, signal processing and many other applications.

It speeds up the calculation of these functions compared to a software implementation, allowing a lower operating frequency, or freeing up processor cycles in order to perform other tasks.

Cordic features

- 24-bit CORDIC rotation engine
- Circular and Hyperbolic modes
- Rotation and Vectoring modes
- Functions: Sine, Cosine, Sinh, Cosh, Atan, Atan2, Atanh, Modulus, Square root, Natural logarithm
- Programmable precision up to 20-bit
- Fast convergence: 4 bits per clock cycle
- Supports 16-bit and 32-bit fixed point input and output formats
- Low latency AHB slave interface
- Results can be read as soon as ready without polling or interrupt
- DMA read and write channels

3.9 Filter mathematical accelerator (FMAC)

The filter mathematical accelerator unit performs arithmetic operations on vectors. It comprises a multiplier/accumulator (MAC) unit, together with address generation logic, which allows it to index vector elements held in local memory.

The unit includes support for circular buffers on input and output, which allows digital filters to be implemented. Both finite and infinite impulse response filters can be realized.

The unit allows frequent or lengthy filtering operations to be offloaded from the CPU, freeing up the processor for other tasks. In many cases it can accelerate such calculations compared to a software implementation, resulting in a speed-up of time critical tasks.

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FMAC features

- 16 x 16-bit multiplier
- 24+2-bit accumulator with addition and subtraction
- 16-bit input and output data
- 256 x 16-bit local memory
- Up to three areas can be defined in memory for data buffers (two input, one output), defined by programmable base address pointers and associated size registers
- Input and output sample buffers can be circular
- Buffer "watermark" feature reduces overhead in interrupt mode
- Filter functions: FIR, IIR (direct form 1)
- AHB slave interface
- DMA read and write data channels

3.10 Cyclic redundancy check calculation unit (CRC)

The CRC (cyclic redundancy check) calculation unit is used to get a CRC code using a configurable generator with polynomial value and size.

Among other applications, the CRC-based techniques are used to verify data transmission or storage integrity. In the scope of the EN/IEC 60335-1 standard, they offer a mean to verify the flash memory integrity.

The CRC calculation unit helps to compute a signature of the software during runtime, which can be ulteriorly compared with a reference signature generated at link-time and which can be stored at a given memory location.

3.11 Power supply management

3.11.1 Power supply schemes

The STM32G473xB/xC/xE devices require a 1.71 V to 3.6 V V_{DD} operating voltage supply. Several independent supplies, can be provided for specific peripherals:

- V_{DD} = 1.71 V to 3.6 V
 - V_{DD} is the external power supply for the I/Os, the internal regulator and the system analog such as reset, power management and internal clocks. It is provided externally through the VDD pins.
- V_{DDA} = 1.62 V to 3.6 V (see Section 5: Electrical characteristics for the minimum V_{DDA} voltage required for ADC, DAC, COMP, OPAMP, VREFBUF operation).
 V_{DDA} is the external analog power supply for A/D converters, D/A converters, voltage reference buffer, operational amplifiers and comparators. The V_{DDA} voltage level is independent from the V_{DD} voltage and should preferably be connected to V_{DD} when these peripherals are not used.
- V_{BAT} = 1.55 V to 3.6 V
 - V_{BAT} is the power supply for RTC, external clock 32 kHz oscillator and backup registers (through power switch) when V_{DD} is not present.

VREF-, VREF+

 $V_{\mathsf{REF+}}$ is the input reference voltage for ADCs and DACs. It is also the output of the internal voltage reference buffer when enabled.

When V_{DDA} < 2 V V_{REF+} must be equal to V_{DDA} .

When $V_{DDA} \ge 2 \text{ V } V_{REF+}$ must be between 2 V and V_{DDA} .

The internal voltage reference buffer supports three output voltages, which are configured with VRS bits in the VREFBUF_CSR register:

- $V_{RFF+} = 2.048 V$
- $V_{RFF+} = 2.5 V$
- $V_{RFF+} = 2.9 V$

 $V_{\text{REF-}}$ is double bonded with V_{SSA} .

During power up and power down, the following power sequence is required:

- When V_{DD} is below 1 V, then V_{DDA} supply must remain below V_{DD} + 300 mV
- When V_{DD} is above 1 V, all power supplies became independent.
 During power down phase, V_{DD} can temporarily become lower then other supplies only if the energy provided to the MCU remains below 1 mJ. This allows external decoupling capacitors to be discharged with different time constants during the power down transient phase.

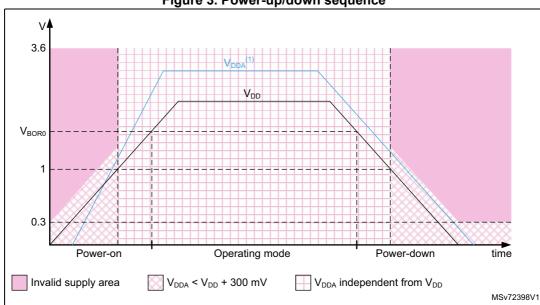


Figure 3. Power-up/down sequence

3.11.2 Power supply supervisor

The device has an integrated ultra-low-power brown-out reset (BOR) active in all modes (except for Shutdown mode). The BOR ensures proper operation of the device after power-on and during power down. The device remains in reset mode when the monitored supply voltage V_{DD} is below a specified threshold, without the need for an external reset circuit.

The lowest BOR level is 1.71 V at power on, and other higher thresholds can be selected through option bytes. The device features an embedded programmable voltage detector (PVD) that monitors the V_{DD} power supply and compares it to the VPVD threshold. An interrupt can be generated when V_{DD} drops below the VPVD threshold and/or when V_{DD} is higher than the VPVD threshold. The interrupt service routine can then generate a warning message and/or put the MCU into a safe state. The PVD is enabled by software.

In addition, the device embeds a peripheral voltage monitor which compares the independent supply voltages V_{DDA} , with a fixed threshold in order to ensure that the peripheral is in its functional supply range.

3.11.3 Voltage regulator

Two embedded linear voltage regulators, main regulator (MR) and low-power regulator (LPR), supply most of digital circuitry in the device. The MR is used in Run and Sleep modes. The LPR is used in Low-power run, Low-power sleep and Stop modes. In Standby and Shutdown modes, both regulators are powered down and their outputs set in high-impedance state, such as to bring their current consumption close to zero.

The device supports dynamic voltage scaling to optimize its power consumption in Run mode. the voltage from the main regulator that supplies the logic (VCORE) can be adjusted according to the system's maximum operating frequency.

The main regulator (MR) operates in the following ranges:

- Range 1 boost mode with the CPU running at up to 170 MHz.
- Range 1 normal mode with CPU running at up to 150 MHz.
- Range 2 with a maximum CPU frequency of 26 MHz.

3.11.4 Low-power modes

By default, the microcontroller is in Run mode after system or power Reset. It is up to the user to select one of the low-power modes described below:

- Sleep mode: In Sleep mode, only the CPU is stopped. All peripherals continue to operate and can wake up the CPU when an interrupt/event occurs.
- Low-power run mode: This mode is achieved with VCORE supplied by the low-power regulator to minimize the regulator's operating current. The code can be executed from SRAM or from flash, and the CPU frequency is limited to 2 MHz. The peripherals with independent clock can be clocked by HSI16.
- Low-power sleep mode: This mode is entered from the low-power run mode. Only the CPU clock is stopped. When wakeup is triggered by an event or an interrupt, the system reverts to the Low power run mode.
- Stop mode: In Stop mode, the device achieves the lowest power consumption while
 retaining the SRAM and register contents. All clocks in the VCORE domain are
 stopped. The PLL, as well as the HSI16 RC oscillator and the HSE crystal oscillator are
 disabled. The LSE or LSI keep running. The RTC can remain active (Stop mode with



- RTC, Stop mode without RTC). Some peripherals with wakeup capability can enable the HSI16 RC during Stop mode, so as to get clock for processing the wakeup event.
- Standby mode: The Standby mode is used to achieve the lowest power consumption with brown-out reset, BOR. The internal regulator is switched off to power down the VCORE domain. The PLL, as well as the HSI16 RC oscillator and the HSE crystal oscillator are also powered down. The RTC can remain active (Standby mode with RTC, Standby mode without RTC). The BOR always remains active in Standby mode. For each I/O, the software can determine whether a pull-up, a pull-down or no resistor shall be applied to that I/O during Standby mode. Upon entering Standby mode, SRAM and register contents are lost except for registers in the RTC domain and standby circuitry. The device exits Standby mode upon external reset event (NRST pin), IWDG reset event, wakeup event (WKUP pin, configurable rising or falling edge) or RTC event (alarm, periodic wakeup, timestamp, tamper), or when a failure is detected on LSE (CSS on LSE).
- Shutdown mode: The Shutdown mode allows to achieve the lowest power consumption. The internal regulator is switched off to power down the VCORE domain. The PLL, as well as the HSI16 and LSI RC-oscillators and HSE crystal oscillator are also powered down. The RTC can remain active (Shutdown mode with RTC, Shutdown mode without RTC). The BOR is not available in Shutdown mode. No power voltage monitoring is possible in this mode. Therefore, switching to RTC domain is not supported. SRAM and register contents are lost except for registers in the RTC domain. The device exits Shutdown mode upon external reset event (NRST pin), IWDG reset event, wakeup event (WKUP pin, configurable rising or falling edge) or RTC event (alarm, periodic wakeup, timestamp, tamper).

3.11.5 Reset mode

In order to improve the consumption under reset, the I/Os state under and after reset is "analog state" (the I/O schmitt trigger is disabled). In addition, the internal reset pull-up is deactivated when the reset source is internal.

3.11.6 V_{BAT} operation

The V_{BAT} pin allows to power the device V_{BAT} domain from an external battery, an external supercapacitor, or from V_{DD} when there is no external battery and when an external supercapacitor is present. The V_{BAT} pin supplies the RTC with LSE and the backup registers. Three anti-tamper detection pins are available in V_{BAT} mode.

The V_{BAT} operation is automatically activated when V_{DD} is not present. An internal V_{BAT} battery charging circuit is embedded and can be activated when V_{DD} is present.

Note: When the microcontroller is supplied from V_{BAT} , neither external interrupts nor RTC alarm/events exit the microcontroller from the V_{BAT} operation.

4

3.12 Interconnect matrix

Several peripherals have direct connections between them. This allows autonomous communication between peripherals, saving CPU resources thus power supply consumption. In addition, these hardware connections allow fast and predictable latency.

Depending on peripherals, these interconnections can operate in Run, Sleep and Stop modes.

Table 3. STM32G473xB/xC/xE peripherals interconnect matrix

Interconnect source	Interconnect destination	Interconnect action		Sleep	Low-power run	Low-power sleep	Stop
	TIMx	Timers synchronization or chaining	Υ	Υ	Υ	Υ	-
TIMx	ADCx DACx	Conversion triggers	Υ	Υ	Υ	Υ	-
	DMA	Memory to memory transfer trigger	Υ	Υ	Υ	Υ	-
	COMPx	Comparator output blanking	Υ	Υ	Υ	Υ	-
TIM16/TIM17	IRTIM	Infrared interface output generation	Υ	Υ	Υ	Υ	-
COMP	TIM1, 8, 20 TIM2, 3, 4, 5	Timer input channel, trigger, break from analog signals comparison		Υ	Υ	Υ	-
COMPx	LPTIMER1	Low-power timer triggered by analog signals comparison		Υ	Υ	Υ	Υ
ADCx	TIM1, 8, 20 Timer triggered by analog watchdog		Υ	Υ	Υ	Υ	-
	TIM16	Timer input channel from RTC events	Υ	Υ	Υ	Υ	-
RTC	LPTIMER1	Low-power timer triggered by RTC alarms or tampers		Υ	Υ	Υ	Υ
All clocks sources (internal and external)			Υ	Υ	Υ	Υ	-
USB	TIM2	Timer triggered by USB SOF	Υ	Υ	-	-	-
CSS RAM (parity error) Flash memory (ECC error) COMPx PVD	TIM1,8, 20 TIM15,16,17			Υ	Υ	Y	-
CPU (hard fault)	d fault) TIM1,8,20 TIM15/16/17 Timer break		Υ	Υ	Υ	Υ	-

Table 3. STM32G473xB/xC/xE peripherals interconnect matrix (continued)

Interconnect source Interconnect destination		Interconnect action		Sleep	Low-power run	Low-power sleep	Stop
	TIMx	External trigger	Υ	Υ	Υ	Υ	-
GPIO	LPTIMER1	External trigger		Υ	Υ	Υ	Υ
	ADCx DACx	Conversion external trigger		Υ	Υ	Υ	-

3.13 Clocks and startup

The clock controller distributes the clocks coming from different oscillators to the core and the peripherals. It also manages clock gating for low-power modes and ensures clock robustness. It features:

- Clock prescaler: to get the best trade-off between speed and current consumption, the clock frequency to the CPU and peripherals can be adjusted by a programmable prescaler
- **Safe clock switching:** clock sources can be changed safely on the fly in run mode through a configuration register.
- **Clock management:** to reduce power consumption, the clock controller can stop the clock to the core, individual peripherals or memory.
- System clock source: three different sources can deliver SYSCLK system clock:
 - 4 48 MHz high-speed oscillator with external crystal or ceramic resonator (HSE).
 It can supply clock to system PLL. The HSE can also be configured in bypass mode for an external clock.
 - 16 MHz high-speed internal RC oscillator (HSI16), trimmable by software. It can supply clock to system PLL.
 - System PLL with maximum output frequency of 170 MHz. It can be fed with HSE or HSI16 clocks.
- Auxiliary clock source: two ultra-low-power clock sources for the real-time clock (RTC):
 - 32.768 kHz low-speed oscillator with external crystal (LSE), supporting four drive capability modes. The LSE can also be configured in bypass mode for using an external clock.
 - 32 kHz low-speed internal RC oscillator (LSI) with ±5% accuracy, also used to clock an independent watchdog.
- **Peripheral clock sources:** several peripherals (I2S, USART, I2C, LPTimer, ADC, SAI, RNG) have their own clock independent of the system clock.
- Clock security system (CSS): in the event of HSE clock failure, the system clock is automatically switched to HSI16 and, if enabled, a software interrupt is generated. LSE clock failure can also be detected and generate an interrupt.
- Clock-out capability:
 - MCO: microcontroller clock output: it outputs one of the internal clocks for external use by the application
 - LSCO: low speed clock output: it outputs LSI or LSE in all low-power modes.

Several prescalers allow to configure the AHB frequency, the High-speed APB (APB2) and the low speed APB (APB1) domains. The maximum frequency of the AHB and the APB domains is 170 MHz.

3.14 General-purpose inputs/outputs (GPIOs)

Each of the GPIO pins can be configured by software as output (push-pull or open-drain), as input (with or without pull-up or pull-down) or as peripheral alternate function. Most of the GPIO pins are shared with digital or analog alternate functions. Fast I/O toggling can be achieved thanks to their mapping on the AHB2 bus.

The I/Os alternate function configuration can be locked if needed following a specific sequence in order to avoid spurious writing to the I/Os registers.

3.15 Direct memory access controller (DMA)

The device embeds 2 DMAs. Refer to *Table 4: DMA implementation* for the features implementation.

Direct memory access (DMA) is used in order to provide a high-speed data transfer between peripherals and memory as well as from memory to memory. Data can be quickly moved by DMA without any CPU actions. This keeps the CPU resources free for other operations.

The two DMA controllers have 16 channels in total, each one dedicated to manage memory access requests from one or more peripherals. Each controller has an arbiter for handling the priority between DMA requests.

The DMA supports:

- 16 independently configurable channels (requests)
 - Each channel is connected to a dedicated hardware DMA request, a software trigger is also supported on each channel. This configuration is done by software.
- Priorities between requests from channels of one DMA are both software programmable (4 levels: very high, high, medium, low) or hardware programmable in case of equality (request 1 has priority over request 2, etc.)
- Independent source and destination transfer size (byte, half word, word), emulating packing and unpacking. Source/destination addresses must be aligned on the data size.
- Support for circular buffer management
- 3 event flags (DMA half transfer, DMA transfer complete and DMA transfer error) logically ORed together in a single interrupt request for each channel
- Memory-to-memory transfer
- Peripheral-to-memory, memory-to-peripheral, and peripheral-to-peripheral transfers
- Access to flash memory, SRAM, APB and AHB peripherals as source and destination
- Programmable number of data to be transferred: up to 65536.

Table 4. DMA implementation

DMA features	DMA1	DMA2
Number of regular channels	8	8



3.16 DMA request router (DMAMUX)

When a peripheral indicates a request for DMA transfer by setting its DMA request line, the DMA request is pending until it is served and the corresponding DMA request line is reset. The DMA request router allows to route the DMA control lines between the peripherals and the DMA controllers of the product.

An embedded multi-channel DMA request generator can be considered as one of such peripherals. The routing function is ensured by a multi-channel DMA request line multiplexer. Each channel selects a unique set of DMA control lines, unconditionally or synchronously with events on synchronization inputs.

For simplicity, the functional description is limited to DMA request lines. The other DMA control lines are not shown in figures or described in the text. The DMA request generator produces DMA requests following events on DMA request trigger inputs.

3.17 Interrupts and events

3.17.1 Nested vectored interrupt controller (NVIC)

The STM32G473xB/xC/xE devices embed a nested vectored interrupt controller which is able to manage 16 priority levels, and to handle up to 102 maskable interrupt channels plus the 16 interrupt lines of the Cortex[®]-M4.

The NVIC benefits are the following:

- Closely coupled NVIC gives low latency interrupt processing
- Interrupt entry vector table address passed directly to the core
- Allows early processing of interrupts
- Processing of late arriving higher priority interrupts
- Support for tail chaining
- Processor state automatically saved
- Interrupt entry restored on interrupt exit with no instruction overhead

The NVIC hardware block provides flexible interrupt management features with minimal interrupt latency.

3.17.2 Extended interrupt/event controller (EXTI)

The extended interrupt/event controller consists of 44 edge detector lines used to generate interrupt/event requests and to wake-up the system from the Stop mode. Each external line can be independently configured to select the trigger event (rising edge, falling edge, both) and can be masked independently.

A pending register maintains the status of the interrupt requests. The internal lines are connected to peripherals with wakeup from Stop mode capability. The EXTI can detect an external line with a pulse width shorter than the internal clock period. Up to 107 GPIOs can be connected to the 16 external interrupt lines.

3.18 Analog-to-digital converter (ADC)

The device embeds five successive approximation analog-to-digital converters with the following features:

- 12-bit native resolution, with built-in calibration
- 4 Msps maximum conversion rate with full resolution
 - Down to 41.67 ns sampling time
 - Increased conversion rate for lower resolution (up to 6.66 Msps for 6-bit resolution)
- One external reference pin is available on all packages, allowing the input voltage range to be independent from the power supply
- Single-ended and differential mode inputs
- Low-power design
 - Capable of low-current operation at low conversion rate (consumption decreases linearly with speed)
 - Dual clock domain architecture: ADC speed independent from CPU frequency
- Highly versatile digital interface
 - Single-shot or continuous/discontinuous sequencer-based scan mode: 2 groups of analog signals conversions can be programmed to differentiate background and high-priority real-time conversions
 - Each ADC support multiple trigger inputs for synchronization with on-chip timers and external signals
 - Results stored into a data register or in RAM with DMA controller support
 - Data pre-processing: left/right alignment and per channel offset compensation
 - Built-in oversampling unit for enhanced SNR
 - Channel-wise programmable sampling time
 - Analog watchdog for automatic voltage monitoring, generating interrupts and trigger for selected timers
 - Hardware assistant to prepare the context of the injected channels to allow fast context switching
 - Flexible sample time control
 - Hardware gain and offset compensation

3.18.1 Temperature sensor

The temperature sensor (TS) generates a voltage V_{TS} that varies linearly with temperature. The temperature sensor is internally connected to the ADCs input channels which is used to convert the sensor output voltage into a digital value.

The sensor provides good linearity but it has to be calibrated to obtain good overall accuracy of the temperature measurement. As the offset of the temperature sensor varies from chip to chip due to process variation, the uncalibrated internal temperature sensor is suitable for applications that detect temperature changes only.

To improve the accuracy of the temperature sensor measurement, each device is individually factory-calibrated by ST. The temperature sensor factory calibration data are stored by ST in the system memory area, accessible in read-only mode.

4

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Calibration value name	Description	Memory address
TS_CAL1	TS ADC raw data acquired at a temperature of 30 °C (± 5 °C), V _{DDA} = V _{REF+} = 3.0 V (± 10 mV)	0x1FFF 75A8 - 0x1FFF 75A9
TS_CAL2	TS ADC raw data acquired at a temperature of 130 °C (± 5 °C), V _{DDA} = V _{REF+} = 3.0 V (± 10 mV)	0x1FFF 75CA - 0x1FFF 75CB

Table 5. Temperature sensor calibration values

3.18.2 Internal voltage reference (V_{REFINT})

The internal voltage reference (VREFINT) provides a stable (bandgap) voltage output for the ADC and the comparators. The VREFINT is internally connected to the ADCx_IN18, x = 1,3,4,5 input channel. The precise voltage of VREFINT is individually measured for each part by ST during production test and stored in the system memory area. It is accessible in read-only mode.

Table 6. Internal voltage reference calibration values

Calibration value name	Description	Memory address
VREFINT	Raw data acquired at a temperature of 30 °C (± 5 °C), V _{DDA} = V _{REF+} = 3.0 V (± 10 mV)	0x1FFF 75AA - 0x1FFF 75AB

3.18.3 V_{BAT} battery voltage monitoring

This embedded hardware enables the application to measure the V_{BAT} battery voltage using the internal ADC1_IN17 channel. As the V_{BAT} voltage may be higher than the V_{DDA} , and thus outside the ADC input range, the VBAT pin is internally connected to a bridge divider by 3. As a consequence, the converted digital value is one third of the V_{BAT} voltage.

3.18.4 Operational amplifier internal output (OPAMPxINT):

The OPAMPx (x = 1...6) output OPAMPxINT can be sampled using an ADCx (x = 1...5) internal input channel. In this case, the I/O on which the OPAMPx output is mapped can be used as GPIO.

3.19 Digital to analog converter (DAC)

Seven 12 bit DAC channels (3 external buffered and 4 internal unbuffered) can be used to convert digital signals into analog voltage signal outputs. The chosen design structure is composed of integrated resistor strings and an amplifier in inverting configuration.

This digital interface supports the following features:

- Up to two DAC output channels
- 8-bit or 12-bit output mode
- Buffer offset calibration (factory and user trimming)
- Left or right data alignment in 12-bit mode
- Synchronized update capability
- Noise-wave generation
- Triangular-wave generation
- Saw tooth wave generation
- Dual DAC channel independent or simultaneous conversions
- DMA capability for each channel
- External triggers for conversion
- Sample and hold low-power mode, with internal or external capacitor
- Up to 1 Msps for external output and 15 Msps for internal output

The DAC channels are triggered through the timer update outputs that are also connected to different DMA channels.

3.20 Voltage reference buffer (V_{REFBUF})

The STM32G473xB/xC/xE devices embed a voltage reference buffer which can be used as voltage reference for ADC, DACs and also as voltage reference for external components through the VREF+ pin.

The internal voltage reference buffer supports three voltages:

- 2.048 V
- 2.5 V
- 2.9 V

An external voltage reference can be provided through the VREF+ pin when the internal voltage reference buffer is off.

The VREF+ pin is double-bonded with V_{DDA} on some packages. In these packages the internal voltage reference buffer is not available.

VREFBUF
VDDA DAC, ADC
Bandgap
Low frequency cut-off capacitor

MSv40197V1

Figure 4. Voltage reference buffer

4

3.21 Comparators (COMP)

The STM32G473xB/xC/xE devices embed seven rail-to-rail comparators with programmable reference voltage (internal or external), hysteresis.

The reference voltage can be one of the following:

- External I/O
- DAC output channels
- Internal reference voltage or submultiple (1/4, 1/2, 3/4).

All comparators can wake up from Stop mode, generate interrupts and breaks for the timers.

3.22 Operational amplifier (OPAMP)

The STM32G473xB/xC/xE devices embed six operational amplifiers with external or internal follower routing and PGA capability.

The operational amplifier features:

- 13 MHz bandwidth
- Rail-to-rail input/output
- PGA with a non-inverting gain ranging of 2, 4, 8, 16, 32 or 64 or inverting gain ranging of -1, -3, -7, -15, -31 or -63

3.23 Random number generator (RNG)

All devices embed an RNG that delivers 32-bit random numbers generated by an integrated analog circuit.

3.24 Timers and watchdogs

The STM32G473xB/xC/xE devices include three advanced motor control timers, up to nine general-purpose timers, two basic timers, one low-power timer, two watchdog timers and a SysTick timer. The table below compares the features of the advanced motor control, general purpose and basic timers.

Table 7. Timer feature comparison

Timer type	Timer	Counter resolution	Counter type	Prescaler factor	DMA request generation	Capture/ compare channels	Complementary outputs
Advanced motor control	TIM1, TIM8, TIM20	16-bit	Up, down, Up/down	Any integer between 1 and 65536	Yes	4	4
General- purpose	TIM2, TIM5	32-bit	Up, down, Up/down	Any integer between 1 and 65536	Yes	4	No
General- purpose	TIM3, TIM4	16-bit	Up, down, Up/down	Any integer between 1 and 65536	Yes	4	No

Timer type	Timer	Counter resolution	Counter type	Prescaler factor	DMA request generation	Capture/ compare channels	Complementary outputs
General- purpose	TIM15	16-bit	Up	Any integer between 1 and 65536	Yes	2	1
General- purpose	TIM16, TIM17	16-bit	Up	Any integer between 1 and 65536	Yes	1	1
Basic	TIM6, TIM7	16-bit	Up	Any integer between 1 and 65536	Yes	0	No

Table 7. Timer feature comparison (continued)

3.24.1 Advanced motor control timer (TIM1, TIM8, TIM20)

The advanced motor control timers can each be seen as a four-phase PWM multiplexed on 8 channels. They have complementary PWM outputs with programmable inserted dead-times. They can also be seen as complete general-purpose timers.

The 4 independent channels can be used for:

- Input capture
- Output compare
- PWM generation (edge or center-aligned modes) with full modulation capability (0-100%)
- One-pulse mode output

In debug mode, the advanced motor control timer counter can be frozen and the PWM outputs disabled in order to turn off any power switches driven by these outputs.

Many features are shared with the general-purpose TIMx timers (described in Section 3.24.2) using the same architecture, so the advanced motor control timers can work together with the TIMx timers via the Timer Link feature for synchronization or event chaining.

3.24.2 General-purpose timers (TIM2, TIM3, TIM4, TIM5, TIM15, TIM16, TIM17)

There are up to seven synchronizable general-purpose timers embedded in the STM32G473xB/xC/xE devices (see *Table 7* for differences). Each general-purpose timer can be used to generate PWM outputs, or act as a simple time base.

TIM2, TIM3, TIM4 and TIM5

They are full-featured general-purpose timers:

- TIM2 and TIM5 have a 32-bit auto-reload up/downcounter and 32-bit prescaler
- TIM3 and TIM4 have 16-bit auto-reload up/downcounter and 16-bit prescaler.

These timers feature 4 independent channels for input capture/output compare, PWM or one-pulse mode output. They can work together, or with the other general-purpose timers via the Timer Link feature for synchronization or event chaining.

The counters can be frozen in debug mode.

All have independent DMA request generation and support quadrature encoders.

TIM15, 16 and 17

They are general-purpose timers with mid-range features:

They have 16-bit auto-reload upcounters and 16-bit prescalers.

- TIM15 has 2 channels and 1 complementary channel
- TIM16 and TIM17 have 1 channel and 1 complementary channel

All channels can be used for input capture/output compare, PWM or one-pulse mode output.

The timers can work together via the Timer Link feature for synchronization or event chaining. The timers have independent DMA request generation.

The counters can be frozen in debug mode.

3.24.3 Basic timers (TIM6 and TIM7)

The basic timers are mainly used for DAC trigger generation. They can also be used as generic 16-bit timebases.

3.24.4 Low-power timer (LPTIM1)

The devices embed a low-power timer. This timer has an independent clock and are running in Stop mode if it is clocked by LSE, LSI or an external clock. It is able to wakeup the system from Stop mode.

LPTIM1 is active in Stop mode.

This low-power timer supports the following features:

- 16-bit up counter with 16-bit autoreload register
- 16-bit compare register
- Configurable output: pulse, PWM
- Continuous/ one shot mode
- Selectable software/hardware input trigger
- Selectable clock source
 - Internal clock sources: LSE, LSI, HSI16 or APB clock
 - External clock source over LPTIM input (working even with no internal clock source running, used by pulse counter application).
- Programmable digital glitch filter
- Encoder mode

3.24.5 Independent watchdog (IWDG)

The independent watchdog is based on a 12-bit downcounter and an 8-bit prescaler. It is clocked from an independent 32 kHz internal RC (LSI) and as it operates independently from the main clock, it can operate in Stop and Standby modes. It can be used either as a watchdog to reset the device when a problem occurs, or as a free running timer for application timeout management. It is hardware or software configurable through the option bytes. The counter can be frozen in debug mode.

3.24.6 System window watchdog (WWDG)

The window watchdog is based on a 7-bit downcounter that can be set as free running. It can be used as a watchdog to reset the device when a problem occurs. It is clocked from the main clock. It has an early warning interrupt capability and the counter can be frozen in debug mode.

3.24.7 SysTick timer

This timer is dedicated to real-time operating systems, but could also be used as a standard down counter. It features:

- A 24-bit down counter
- Autoreload capability
- Maskable system interrupt generation when the counter reaches 0.
- Programmable clock source

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3.25 Real-time clock (RTC) and backup registers

The RTC supports the following features:

- Calendar with subsecond, seconds, minutes, hours (12 or 24 format), week day, date, month, year, in BCD (binary-coded decimal) format.
- Automatic correction for 28, 29 (leap year), 30, and 31 days of the month.
- Two programmable alarms.
- On-the-fly correction from 1 to 32767 RTC clock pulses. This can be used to synchronize it with a master clock.
- Reference clock detection: a more precise second source clock (50 or 60 Hz) can be used to enhance the calendar precision.
- Digital calibration circuit with 0.95 ppm resolution, to compensate for quartz crystal inaccuracy.
- Timestamp feature which can be used to save the calendar content. This function can be triggered by an event on the timestamp pin, or by a tamper event, or by a switch to V_{BAT} mode.
- 17-bit auto-reload wakeup timer (WUT) for periodic events with programmable resolution and period.

The RTC is supplied through a switch that takes power either from the V_{DD} supply when present or from the VBAT pin.

The RTC clock sources can be:

- A 32.768 kHz external crystal (LSE)
- An external resonator or oscillator (LSE)
- The internal low power RC oscillator (LSI, with typical frequency of 32 kHz)
- The high-speed external clock (HSE) divided by 32.

The RTC is functional in V_{BAT} mode and in all low-power modes when it is clocked by the LSE. When clocked by the LSI, the RTC is not functional in V_{BAT} mode, but is functional in all low-power modes except Shutdown mode.

All RTC events (Alarm, WakeUp Timer, Timestamp) can generate an interrupt and wakeup the device from the low-power modes.

3.26 Tamper and backup registers (TAMP)

- 32 32-bit backup registers, retained in all low-power modes and also in V_{BAT} mode.
 They can be used to store sensitive data as their content is protected by an tamper detection circuit. They are not reset by a system or power reset, or when the device wakes up from Standby or Shutdown mode.
- Up to three tamper pins for external tamper detection events. The external tamper pins can be configured for edge detection, edge and level, level detection with filtering.
- Five internal tampers events.
- Any tamper detection can generate a RTC timestamp event.
- Any tamper detection erases the backup registers.
- Any tamper detection can generate an interrupt and wake-up the device from all lowpower modes.



3.27 Infrared transmitter

The STM32G473xB/xC/xE devices provide an infrared transmitter solution. The solution is based on internal connections between TIM16 and TIM17 as shown in the figure below.

TIM17 is used to provide the carrier frequency and TIM16 provides the main signal to be sent. The infrared output signal is available on PB9 or PA13.

To generate the infrared remote control signals, TIM16 channel 1 and TIM17 channel 1 must be properly configured to generate correct waveforms. All standard IR pulse modulation modes can be obtained by programming the two timers output compare channels.

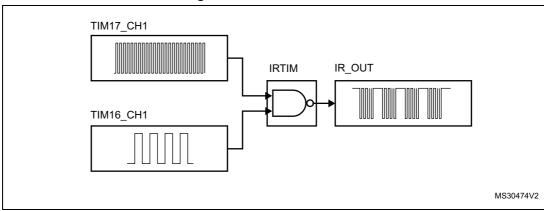


Figure 5. Infrared transmitter

Inter-integrated circuit interface (I²C) 3.28

The device embeds four I2Cs. Refer to Table 8: I2C implementation for the features implementation.

The I²C bus interface handles communications between the microcontroller and the serial I²C bus. It controls all I²C bus-specific sequencing, protocol, arbitration and timing.

The I2C peripheral supports:

- I²C-bus specification and user manual rev. 5 compatibility:
 - Slave and master modes, multimaster capability
 - Standard-mode (Sm), with a bitrate up to 100 kbit/s
 - Fast-mode (Fm), with a bitrate up to 400 kbit/s
 - Fast-mode Plus (Fm+), with a bitrate up to 1 Mbit/s and 20 mA output drive I/Os
 - 7-bit and 10-bit addressing mode, multiple 7-bit slave addresses
 - Programmable setup and hold times
 - Optional clock stretching
- System management bus (SMBus) specification rev 2.0 compatibility:
 - Hardware PEC (packet error checking) generation and verification with ACK control
 - Address resolution protocol (ARP) support
 - SMBus alert
- Power system management protocol (PMBusTM) specification rev 1.1 compatibility
- Independent clock: a choice of independent clock sources allowing the I2C communication speed to be independent from the PCLK reprogramming.
- Wakeup from Stop mode on address match
- Programmable analog and digital noise filters
- 1-byte buffer with DMA capability

Table 8. I2C implementation

I2C features ⁽¹⁾	I2C1	I2C2	I2C3	I2C4
Standard-mode (up to 100 kbit/s)	X	X	X	Х
Fast-mode (up to 400 kbit/s)	Х	Х	Х	Х
Fast-mode Plus with 20mA output drive I/Os (up to 1 Mbit/s)	Х	Х	Х	Х
Programmable analog and digital noise filters	Х	Х	Х	Х
SMBus/PMBus hardware support	Х	Х	Х	Х
Independent clock	Х	Х	Х	Х
Wakeup from Stop mode on address match	Х	Х	Х	Х

^{1.} X: supported



3.29 Universal synchronous/asynchronous receiver transmitter (USART)

The STM32G473xB/xC/xE devices have three embedded universal synchronous receiver transmitters (USART1, USART2 and USART3) and two universal asynchronous receiver transmitters (UART4, USART5).

These interfaces provide asynchronous communication, IrDA SIR ENDEC support, multiprocessor communication mode, single-wire half-duplex communication mode and have LIN master/slave capability. They provide hardware management of the CTS and RTS signals, and RS485 driver enable.

The USART1, USART2 and USART3 also provide a Smartcard mode (ISO 7816 compliant) and an SPI-like communication capability.

The USART comes with a Transmit FIFO (TXFIFO) and a Receive FIFO (RXFIFO). FIFO mode is enabled by software and is disabled by default.

All USARTs have a clock domain independent from the CPU clock, allowing the U(S)ARTx (x=1,2,3,4,5) to wake up the MCU from Stop mode. The wakeup from Stop mode can be done on:

- Start bit detection
- Any received data frame
- A specific programmed data frame
- Some specific TXFIFO/RXFIFO status interrupts when FIFO mode is enabled

All USART interfaces can be served by the DMA controller.

USART modes/features⁽¹⁾ **USART1** USART2 **USART3 UART4 UART5** LPUART1 Hardware flow control for modem Х Х Χ Χ Χ Χ Χ Χ Continuous communication using DMA Χ Χ Х Χ Х Χ Χ Χ Multiprocessor communication Х Х Χ Χ Х Synchronous mode Smartcard mode Χ Χ Χ Χ Χ Χ Single-wire half-duplex communication Χ Х Х IrDA SIR ENDEC block Χ Х Χ Χ Χ Χ LIN mode Χ Χ Χ Х Dual clock domain Χ Х Х Χ Χ Х Wakeup from Stop mode Χ Χ Х Χ Χ Χ Receiver timeout interrupt Х Χ Χ Χ Χ Modbus communication Χ Χ Χ Χ Χ Auto baud rate detection X (4 modes) **Driver Enable** Х Χ Х Χ Х LPUART/USART data length 7, 8 and 9 bits

Table 9. USART/UART/LPUART features



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USART modes/features ⁽¹⁾	USART1	USART2	USART3	UART4	UART5	LPUART1				
Tx/Rx FIFO	X									
Tx/Rx FIFO size	8									

Table 9. USART/UART/LPUART features (continued)

3.30 Low-power universal asynchronous receiver transmitter (LPUART)

The STM32G473xB/xC/xE devices embed one Low-Power UART. The LPUART supports asynchronous serial communication with minimum power consumption. It supports half-duplex single-wire communication and modem operations (CTS/RTS). It allows multiprocessor communication.

The LPUART comes with a Transmit FIFO (TXFIFO) and a Receive FIFO (RXFIFO). FIFO mode is enabled by software and is disabled by default. It has a clock domain independent from the CPU clock, and can wakeup the system from Stop mode. The wake up from Stop mode can be done on:

- Start bit detection
- Any received data frame
- A specific programmed data frame
- Some specific TXFIFO/RXFIFO status interrupts when FIFO mode is enabled

Only a 32.768 kHz clock (LSE) is needed to allow LPUART communication up to 9600 baud. Therefore, even in Stop mode, the LPUART can wait for an incoming frame while having an extremely low energy consumption. Higher speed clock can be used to reach higher baudrates.

The LPUART interface can be served by the DMA controller.

3.31 Serial peripheral interface (SPI)

Four SPI interfaces allow communication up to 75 Mbits/s in master and up to 41 Mbits/s in slave, half-duplex, full-duplex and simplex modes. The 3-bit prescaler gives 8 master mode frequencies and the frame size is configurable from 4 bits to 16 bits. The SPI interfaces support NSS pulse mode, TI mode and hardware CRC calculation.

Two standard I²S interfaces (multiplexed with SPI2 and SPI3) supporting four different audio standards can operate as master or slave at half-duplex communication modes. They can be configured to transfer 16 and 24 or 32 bits with 16-bit or 32-bit data resolution and synchronized by a specific signal. Audio sampling frequency from 8 kHz up to 192 kHz can be set by 8-bit programmable linear prescaler. When operating in master mode it can output a clock for an external audio component at 256 times the sampling frequency.

All SPI interfaces can be served by the DMA controller.

^{1.} X = supported.

3.32 Serial audio interfaces (SAI)

The device embeds 1 SAI. The SAI bus interface handles communications between the microcontroller and the serial audio protocol.

SAI peripheral supports:

- Two independent audio sub-blocks which can be transmitters or receivers with their respective FIFO.
- 8-word integrated FIFOs for each audio sub-block.
- Synchronous or asynchronous mode between the audio sub-blocks.
- Master or slave configuration independent for both audio sub-blocks.
- Clock generator for each audio block to target independent audio frequency sampling when both audio sub-blocks are configured in master mode.
- Data size configurable: 8-, 10-, 16-, 20-, 24-, 32-bit.
- Peripheral with large configurability and flexibility allowing to target as example the following audio protocol: I2S, LSB or MSB-justified, PCM/DSP, TDM, AC'97 and SPDIF out.
- Up to 16 slots available with configurable size and with the possibility to select which
 ones are active in the audio frame.
- Number of bits by frame may be configurable.
- Frame synchronization active level configurable (offset, bit length, level).
- First active bit position in the slot is configurable.
- LSB first or MSB first for data transfer.
- Mute mode.
- Stereo/Mono audio frame capability.
- Communication clock strobing edge configurable (SCK).
- Error flags with associated interrupts if enabled respectively.
 - Overrun and underrun detection.
 - Anticipated frame synchronization signal detection in slave mode.
 - Late frame synchronization signal detection in slave mode.
 - Codec not ready for the AC'97 mode in reception.
- Interruption sources when enabled:
 - Errors.
 - FIFO requests.
- DMA interface with 2 dedicated channels to handle access to the dedicated integrated FIFO of each SAI audio sub-block.

Table 10. SAI features implementation

SAI features	Support ⁽¹⁾
I2S, LSB or MSB-justified, PCM/DSP, TDM, AC'97	Х
Mute mode	Х
Stereo/Mono audio frame capability	Х
16 slots	X



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SAI features	Support ⁽¹⁾
Data size configurable: 8-, 10-, 16-, 20-, 24-, 32-bit	Х
FIFO size	X (8 word)
SPDIF	Х

Table 10. SAI features implementation (continued)

3.33 Controller area network (FDCAN1, FDCAN2, FDCAN3)

The controller area network (CAN) subsystem consists of three CAN modules and a shared message RAM memory.

The three CAN modules (FDCAN1, FDCAN2 and FDCAN3) are compliant with ISO 11898-1 (CAN protocol specification version 2.0 part A, B) and CAN FD protocol specification version 1.0.

A 3-Kbyte message RAM memory implements filters, receive FIFOs, receive buffers. transmit event FIFOs, transmit buffers. This message RAM is shared between the three FDCAN modules.

3.34 Universal serial bus (USB)

The STM32G473xB/xC/xE devices embed a full-speed USB device peripheral compliant with the USB specification version 2.0. The internal USB PHY supports USB FS signaling, embedded DP pull-up and also battery charging detection according to Battery Charging Specification Revision 1.2. The USB interface implements a full-speed (12 Mbit/s) function interface with added support for USB 2.0 Link Power Management. It has softwareconfigurable endpoint setting with packet memory up-to 1 Kbyte and suspend/resume support. It requires a precise 48 MHz clock which can be generated from the internal main PLL (the clock source must use a HSE crystal oscillator) or by the internal 48 MHz oscillator in automatic trimming mode. The synchronization for this oscillator can be taken from the USB data stream itself (SOF signalization) which allows crystal less operation.

3.35 USB Type-C[™] / USB Power Delivery controller (UCPD)

The device embeds one controller (UCPD) compliant with USB Type-C Rev. 1.2 and USB Power Delivery Rev. 3.0 specifications.

The controller uses specific I/Os supporting the USB Type-C and USB Power Delivery requirements, featuring:

- USB Type-C pull-up (Rp, all values) and pull-down (Rd) resistors
- "Dead battery" support
- USB Power Delivery message transmission and reception
- FRS (fast role swap) support



^{1.} X: supported.

The digital controller handles notably:

- USB Type-C level detection with de-bounce, generating interrupts
- FRS detection, generating an interrupt
- Byte-level interface for USB Power Delivery payload, generating interrupts (DMA compatible)
- USB Power Delivery timing dividers (including a clock pre-scaler)
- CRC generation/checking
- 4b5b encode/decode
- Ordered sets (with a programmable ordered set mask at receive)
- Frequency recovery in receiver during preamble

The interface offers low-power operation compatible with Stop mode, maintaining the capacity to detect incoming USB Power Delivery messages and FRS signaling.

3.36 Clock recovery system (CRS)

The devices embed a special block which allows automatic trimming of the internal 48 MHz oscillator to guarantee its optimal accuracy over the whole device operational range. This automatic trimming is based on the external synchronization signal, which could be either derived from USB SOF signalization, from LSE oscillator, from an external signal on CRS_SYNC pin or generated by user software. For faster lock-in during startup it is also possible to combine automatic trimming with manual trimming action.

3.37 Flexible static memory controller (FSMC)

The Flexible static memory controller (FSMC) includes two memory controllers:

- The NOR/PSRAM memory controller
- The NAND/memory controller

This memory controller is also named Flexible memory controller (FMC).

The main features of the FMC controller are the following:

- Interface with static-memory mapped devices including:
 - Static random access memory (SRAM)
 - NOR flash memory/OneNAND flash memory
 - PSRAM (4 memory banks)
 - NAND flash memory with ECC hardware to check up to 8 Kbytes of data
 - Ferroelectric RAM (FRAM)
- 8-.16- bit data bus width
- Independent Chip Select control for each memory bank
- Independent configuration for each memory bank
- Write FIFO
- The Maximum FMC_CLK frequency for synchronous accesses is HCLK/2.

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LCD parallel interface

The FMC can be configured to interface seamlessly with most graphic LCD controllers. It supports the Intel 8080 and Motorola 6800 modes, and is flexible enough to adapt to specific LCD interfaces. This LCD parallel interface capability makes it easy to build cost effective graphic applications using LCD modules with embedded controllers or high performance solutions using external controllers with dedicated acceleration.

3.38 Quad-SPI memory interface (QUADSPI)

The Quad-SPI is a specialized communication interface targeting single, dual or quad SPI flash memories. It can operate in any of the three following modes:

- Indirect mode: all the operations are performed using the QUADSPI registers
- Status polling mode: the external flash memory status register is periodically read and an interrupt can be generated in case of flag setting
- Memory-mapped mode: the external flash is memory mapped and is seen by the system as if it were an internal memory.

Both throughput and capacity can be increased two-fold using dual-flash mode, where two quad SPI flash memories are accessed simultaneously.

The Quad-SPI interface supports:

- Indirect mode: all the operations are performed using the QUADSPI registers
- Status polling mode: the external flash status register is periodically read and an interrupt can be generated in case of flag setting
- Memory-mapped mode: the external flash is memory mapped and is seen by the system as if it were an internal memory
- Three functional modes: indirect, status-polling, and memory-mapped
- SDR and DDR support
- Fully programmable opcode for both indirect and memory mapped mode
- Fully programmable frame format for both indirect and memory mapped mode
 - Each of the 5 following phases can be configured independently (enable, length, single/dual/quad communication)
 - Instruction phase
 - Address phase
 - Alternate bytes phase
 - Dummy cycles phase
 - Data phase
- Integrated FIFO for reception and transmission
- 8, 16, and 32-bit data accesses are allowed
- DMA channel for indirect mode operations
- Programmable masking for external flash flag management
- Timeout management
- Interrupt generation on FIFO threshold, timeout, status match, operation complete, and access error



3.39 Development support

3.39.1 Serial wire JTAG debug port (SWJ-DP)

The Arm SWJ-DP interface is embedded, and is a combined JTAG and serial wire debug port that enables either a serial wire debug or a JTAG probe to be connected to the target.

Debug is performed using 2 pins only instead of 5 required by the JTAG (JTAG pins could be re-use as GPIO with alternate function): the JTAG TMS and TCK pins are shared with SWDIO and SWCLK, respectively, and a specific sequence on the TMS pin is used to switch between JTAG-DP and SW-DP.

3.39.2 Embedded trace macrocell™

The Arm embedded trace macrocell provides a greater visibility of the instruction and data flow inside the CPU core by streaming compressed data at a very high rate from the STM32G473xB/xC/xE devices through a small number of ETM pins to an external hardware trace port analyzer (TPA) device. Real-time instruction and data flow activity be recorded and then formatted for display on the host computer that runs the debugger software. TPA hardware is commercially available from common development tool vendors.

The Embedded trace macrocell operates with third party debugger software tools.



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Pinouts and pin description 4

UFQFPN48 pinout description 4.1

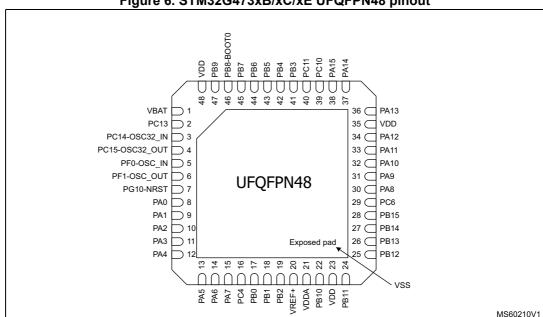


Figure 6. STM32G473xB/xC/xE UFQFPN48 pinout

- 1. The above figure shows the package top view.
- 2. VSS pads are connected to the exposed pad.

MS60210V1

4.2 LQFP48 pinout description

| VDD | VSS | VSS | PB9 | PB8-BOOT0 | PB6 | PB6 | PB7 | PB1 | PB3 | PB15 | PB15 | PB17 | PB18 ______ VBAT ☐ 1 36 🗖 VDD □ vss PC13 2 35 ☐ PA12 PC14 - OSC32_IN [PC15 - OSC32 OUT PA11 PF0 - OSC_IN 🔲 5 32 🗖 PA10 PA9 PF1 - OSC_OUT 6 31 LQFP48 PA8 PG10 - NRST 🔲 7 30 PA0 🗆 8 29 PB15 PA1 🔲 9 28 PB14 27 PB13 26 PB12 PA2 🛮 10 PA3 🔲 11 PA4 🔲 12 25 PB11 13 14 14 16 17 17 17 18 19 20 22 22 22 23 23 MSv42659V2

Figure 7. STM32G473xB/xC/xE LQFP48 pinout

1. The above figure shows the package top view.

4.3 LQFP64 pinout description

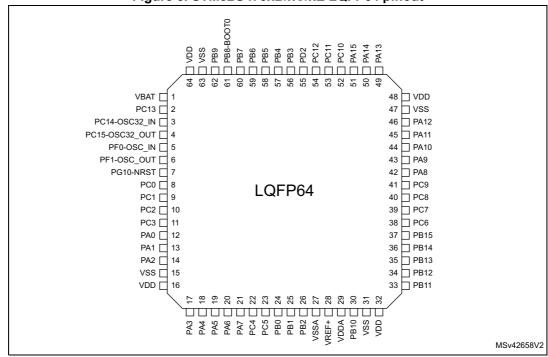


Figure 8. STM32G473xB/xC/xE LQFP64 pinout

1. The above figure shows the package top view.

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4.4 LQFP80 pinout description

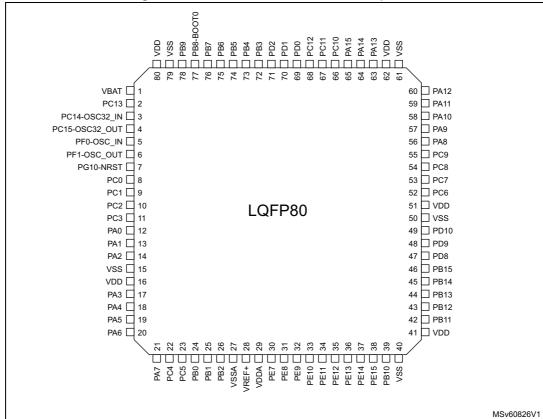


Figure 9. STM32G473xB/xC/xE LQFP80 pinout

1. The above figure shows the package top view.

4.5 LQFP100 pinout description

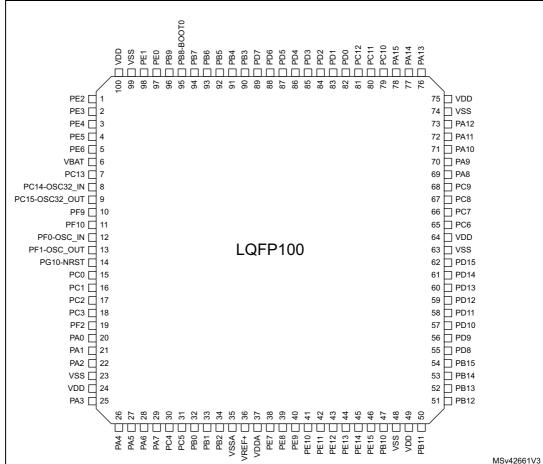


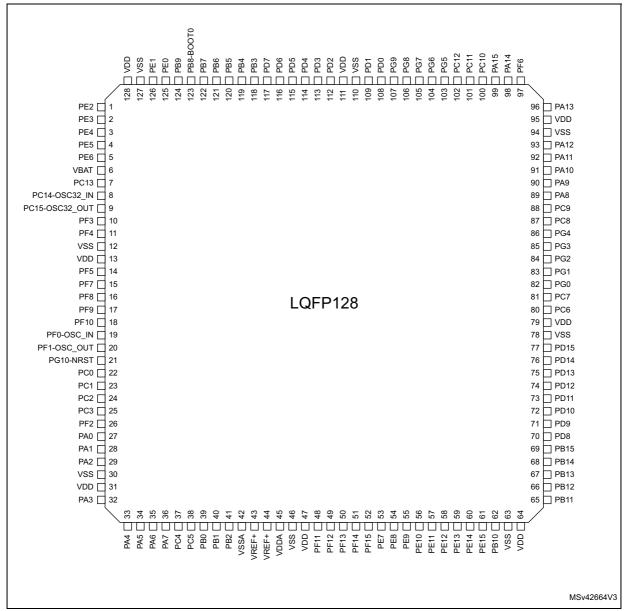
Figure 10. STM32G473xB/xC/xE LQFP100 pinout

1. The above figure shows the package top view.



4.6 LQFP128 pinout description

Figure 11. STM32G473xB/xC/xE LQFP128 pinout



1. The above figure shows the package top view.

4.7 WLCSP81 pinout description

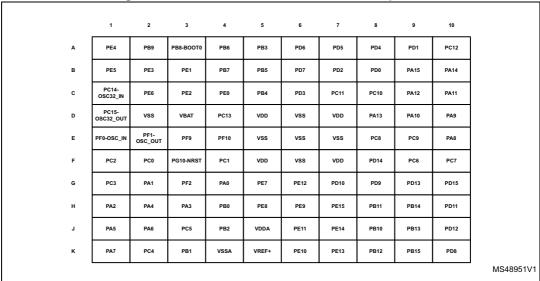
Figure 12. STM32G473xB/xC/xE WLCSP81 pinout

	1	2	3	4	5	6	7	8	9
A	VDD	PA15	PC12	PD1	PB3	PB5	PB9	vss	VDD
В	vss	PA13	PC10	PD0	PD2	PB6	PB8-BOOT0	PC13	VBAT
С	PA12	PA11	PA14	PC11	PC8	PB4	PB7	PC1	PC14- OSC32_IN
D	PA8	PC9	PA10	PA9	PC7	PA4	PA0	PG10-NRST	PC15- OSC32_OUT
E	VDD	PD11	PC6	PB15	PE12	PC4	PA1	PC0	PF0-OSC_IN
F	vss	PD10	PD9	PE15	PE9	PB0	PA5	PC2	PF1- OSC_OUT
G	PD8	PB14	PB12	PE13	PE8	PB1	PA6	PA2	PC3
н	PB13	PB11	PB10	PE11	PE7	VSSA	PC5	PA3	vss
J	VDD	vss	PE14	PE10	VDDA	VREF+	PB2	PA7	VDD

1. The above figure shows the package top view.

4.8 TFBGA100 pinout description

Figure 13. STM32G473xB/xC/xE TFBGA100 pinout



1. The above figure shows the package top view.

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4.9 UFBGA121 pinout description

Figure 14. STM32G473xB/xC/xE UFBGA121 pinout

			94.0	🔾			XO/XL	U. U	<i></i>	piiio	•
	1	2	3	4	5	6	7	8	9	10	11
A	PE4	PE2	VDD	PB9	PB6	PB3	PD4	VDD	PD1	PA15	PF6
В	PE5	PE3	vss	PE0	PB5	PD7	PD3	vss	PD0	PA14	PA13
С	PC13	VBAT	PE6	PE1	PB7	PB4	PD2	PC11	PC10	vss	VDD
D	PC14- OSC32_IN	PC15- OSC32_OUT	PF3	PF4	PB8-BOOT0	PD6	PC12	PA9	PA10	PA12	PA11
E	VDD	vss	PF5	PF7	PF8	PD5	PA8	PC9	PC8	PG4	PG3
F	PF0-OSC_IN	PF1- OSC_OUT	PF9	PF10	PG10-NRST	PD15	PG2	PG1	PG0	PC6	PC7
G	PC1	PC0	PC2	PA0	PB1	PF15	PD11	PD12	PD13	PD14	VDD
н	PC3	PF2	PA1	PC5	PF12	PF14	PE10	PB15	PD8	PD9	PD10
J	VDD	vss	PA2	PB0	PF11	PF13	PE9	PE13	PB12	PB14	PB13
к	PA3	PA5	PA7	PB2	VSSA	vss	PE8	PE12	PE14	vss	VDD
L	PA4	PA6	PC4	VREF+	VDDA	VDD	PE7	PE11	PE15	PB10	PB11
							•			•	

1. The above figure shows the package top view.



4.10 Pin definition

Table 11. Legend/abbreviations used in the pinout table

Name	Abbreviation	Definition								
Pin name		specified in brackets below the pin name, the pin function during and after as the actual pin name								
	S	Supply pin								
Pin type	I	Input only pin								
	I/O	Input / output pin								
	FT	5 V tolerant I/O								
	TT	3.6 V tolerant I/O								
	В	Dedicated BOOT0 pin								
	NRST	Bidirectional reset pin with embedded weak pull-up resistor								
I/O structure	Option for TT or FT I/Os									
i/O structure	_a ⁽¹⁾	I/O, with Analog switch function supplied by V _{DDA}								
	_c	I/O, USB Type-C PD capable								
	_d	I/O, USB Type-C PD Dead Battery function								
	_f ⁽²⁾	I/O, Fm+ capable								
	_u ⁽³⁾	I/O, with USB function								
Notes	Unless otherwise	specified by a note, all I/Os are set as floating inputs during and after reset								
	Alternate functions	Functions selected through GPIOx_AFR registers								
Pin functions	Additional functions	Functions directly selected/enabled through peripheral registers								

^{1.} The related I/O structures in *Table 12* are: FT_a, FT_fa, TT_a.

^{2.} The related I/O structures in *Table 12* are: FT_f, FT_fa.

^{3.} The related I/O structures in *Table 12* are FT_u.

Table 12. STM32G473xB/xC/xE pin definition

			Pir	Nur	nber				ter		စ်			
WLCSP81	UFQFPN48	LQFP48	LQFP64	LQFP80	TFBGA100	LPQF100	UFBGA121	LPQF128	Pin name (function after reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
-	1	-	-	-	C3	1	A2	1	PE2	I/O	FT	-	TRACECK, TIM3_CH1, SAI1_CK1, SPI4_SCK, TIM20_CH1, FMC_A23, SAI1_MCLK_A, EVENTOUT	-
-	1	1	1	-	B2	2	B2	2	PE3	I/O	FT	-	TRACED0, TIM3_CH2, SPI4_NSS, TIM20_CH2, FMC_A19, SAI1_SD_B, EVENTOUT	-
1	1	-	-	-	A1	3	A1	3	PE4	I/O	FT	-	TRACED1, TIM3_CH3, SAI1_D2, SPI4_NSS, TIM20_CH1N, FMC_A20, SAI1_FS_A, EVENTOUT	-
1	1	-	•	-	B1	4	B1	4	PE5	I/O	FT	-	TRACED2, TIM3_CH4, SAI1_CK2, SPI4_MISO, TIM20_CH2N, FMC_A21, SAI1_SCK_A, EVENTOUT	-
-	1	-	1	-	C2	5	C3	5	PE6	I/O	FT	-	TRACED3, SAI1_D1, SPI4_MOSI, TIM20_CH3N, FMC_A22, SAI1_SD_A, EVENTOUT	WKUP3, RTC_TAMP3
В9	1	1	1	1	D3	6	C2	6	VBAT	S	1	-	-	-
В8	2	2	2	2	D4	7	C1	7	PC13	I/O	FT	(2)	TIM1_BKIN, TIM1_CH1N, TIM8_CH4N, EVENTOUT	WKUP2, RTC_TAMP1, RTC_TS, RTC_OUT1
С9	3	3	3	3	C1	8	D1	8	PC14- OSC32_IN	I/O	FT	(2)	EVENTOUT	OSC32_IN
D9	4	4	4	4	D1	9	D2	9	PC15- OSC32_OUT	I/O	FT	(2) (3)	EVENTOUT	OSC32_OUT
-	-	-	-	-	-	-	D3	10	PF3	I/O	FT_f	-	TIM20_CH4, I2C3_SCL, FMC_A3, EVENTOUT	-

Table 12. STM32G473xB/xC/xE pin definition (continued)

	Pin Number												(continued)	
WLCSP81	UFQFPN48	LQFP48	LQFP64	LQFP80	TFBGA100	LPQF100	UFBGA121	LPQF128	Pin name (function after reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
-	ı	ı	-	-	-	1	D4	11	PF4	I/O	FT_f	ı	COMP1_OUT, TIM20_CH1N, I2C3_SDA, FMC_A4, EVENTOUT	-
F1	-	-	-	-	D2	-	E2	12	VSS	S	-	-	-	-
A9	-	-	-	-	D5	-	E1	13	VDD	S	-	-	-	-
-	-	ı	-	-	-	1	E3	14	PF5	I/O	FT	-	TIM20_CH2N, FMC_A5, EVENTOUT	-
-	1	1	-	-	-	-	E4	15	PF7	I/O	FT	1	TIM20_BKIN, TIM5_CH2, QUADSPI1_BK1_IO2, FMC_A1, SAI1_MCLK_B, EVENTOUT	-
-	1	1	-	-	-	-	E5	16	PF8	I/O	FT	-	TIM20_BKIN2, TIM5_CH3, QUADSPI1_BK1_IO0, FMC_A24, SAI1_SCK_B, EVENTOUT	-
-	1	-	-	-	E3	10	F3	17	PF9	I/O	FT	-	TIM20_BKIN, TIM15_CH1, SPI2_SCK, TIM5_CH4, QUADSPI1_BK1_IO1, FMC_A25, SAI1_FS_B, EVENTOUT	-
-	1	1	-	-	E4	11	F4	18	PF10	I/O	FT	1	TIM20_BKIN2, TIM15_CH2, SPI2_SCK, QUADSPI1_CLK, FMC_A0, SAI1_D3, EVENTOUT	-
E9	5	5	5	5	E1	12	F1	19	PF0-OSC_IN	I/O	FT_fa	-	I2C2_SDA, SPI2_NSS/I2S2_WS, TIM1_CH3N, EVENTOUT	ADC1_IN10, OSC_IN
F9	6	6	6	6	E2	13	F2	20	PF1- OSC_OUT	I/O	FT_a	-	SPI2_SCK/I2S2_CK, EVENTOUT	ADC2_IN10, COMP3_INM, OSC_OUT
D8	7	7	7	7	F3	14	F5	21	PG10-NRST	I/O	NRST (4)	-	MCO, EVENTOUT	NRST



Table 12. STM32G473xB/xC/xE pin definition (continued)

Pin Number													(Continued)	
WLCSP81	UFQFPN48	LQFP48	LQFP64	LQFP80	TFBGA100	LPQF100	UFBGA121	LPQF128	Pin name (function after reset) ⁽¹⁾ Pin type		I/O structure	Notes	Alternate functions	Additional functions
E8	1	-	8	8	F2	15	G2	22	PC0	I/O	FT_a	-	LPTIM1_IN1, TIM1_CH1, LPUART1_RX, EVENTOUT	ADC12_IN6, COMP3_INM
C8	1	1	9	9	F4	16	G1	23	PC1	I/O	TT_a	1	LPTIM1_OUT, TIM1_CH2, LPUART1_TX, QUADSPI1_BK2_IO0, SAI1_SD_A, EVENTOUT	ADC12_IN7, COMP3_INP
F8	1	1	10	10	F1	17	G3	24	PC2	I/O	FT_a	1	LPTIM1_IN2, TIM1_CH3, COMP3_OUT, TIM20_CH2, QUADSPI1_BK2_IO1, EVENTOUT	ADC12_IN8
G9	ı	1	11	11	G1	18	H1	25	PC3	I/O	TT_a	1	LPTIM1_ETR, TIM1_CH4, SAI1_D1, TIM1_BKIN2, QUADSPI1_BK2_IO2, SAI1_SD_A, EVENTOUT	ADC12_IN9, OPAMP5_VINP
-	1	-	-	-	G3	19	H2	26	PF2	I/O	FT	-	TIM20_CH3, I2C2_SMBA, FMC_A2, EVENTOUT	-
D7	8	8	12	12	G4	20	G4	27	PA0	I/O	TT_a	1	TIM2_CH1, TIM5_CH1, USART2_CTS, COMP1_OUT, TIM8_BKIN, TIM8_ETR, TIM2_ETR, EVENTOUT	ADC12_IN1, COMP1_INM, COMP3_INP, RTC_TAMP2,WK UP1
E7	0	9	13	13	G2	21	Н3	28	PA1	I/O	TT_a	-	RTC_REFIN, TIM2_CH2, TIM5_CH2, USART2_RTS_DE, TIM15_CH1N, EVENTOUT	ADC12_IN2, COMP1_INP, OPAMP1_VINP, OPAMP3_VINP, OPAMP6_VINM
G8	10	10	14	14	H1	22	J3	29	PA2	I/O	FT_a	-	TIM2_CH3, TIM5_CH3, USART2_TX, COMP2_OUT, TIM15_CH1, QUADSPI1_BK1_NCS , LPUART1_TX, UCPD1_FRSTX, EVENTOUT	ADC1_IN3, COMP2_INM, OPAMP1_VOUT, WKUP4/LSCO

Table 12. STM32G473xB/xC/xE pin definition (continued)

Pin Number									ter	-	ē		,	
WLCSP81	UFQFPN48	LQFP48	LQFP64	LQFP80	TFBGA100	LPQF100	UFBGA121	LPQF128	Pin name (function after reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
Н9	-	-	15	15	D6	23	J2	30	VSS	S	-	-	-	-
J9	-	-	16	16	D7	24	J1	31	VDD	S	1	-	-	-
Н8	11	11	17	17	НЗ	25	K1	32	PA3	I/O	TT_a	-	TIM2_CH4, TIM5_CH4, SAI1_CK1, USART2_RX, TIM15_CH2, QUADSPI1_CLK, LPUART1_RX, SAI1_MCLK_A, EVENTOUT	ADC1_IN4, COMP2_INP, OPAMP1_VINM/ OPAMP 1_VINP, OPAMP5_VINM
D6	12	12	18	18	H2	26	L1	33	PA4	I/O	TT_a	1	TIM3_CH2, SPI1_NSS, SPI3_NSS/I2S3_WS, USART2_CK, SAI1_FS_B, EVENTOUT	ADC2_IN17, DAC1_OUT1, COMP1_INM
F7	13	13	19	19	J1	27	K2	34	PA5	I/O	TT_a	-	TIM2_CH1, TIM2_ETR, SPI1_SCK, UCPD1_FRSTX, EVENTOUT	ADC2_IN13, DAC1_OUT2, COMP2_INM, OPAMP2_VINM
G7	14	14	20	20	J2	28	L2	35	PA6	I/O	TT_a	1	TIM16_CH1, TIM3_CH1, TIM8_BKIN, SPI1_MISO, TIM1_BKIN, COMP1_OUT, QUADSPI1_BK1_IO3, LPUART1_CTS, EVENTOUT	ADC2_IN3, DAC2_OUT1, OPAMP2_VOUT
J8	15	15	21	21	K1	29	K3	36	PA7	I/O	TT_a	-	TIM17_CH1, TIM3_CH2, TIM8_CH1N, SPI1_MOSI, TIM1_CH1N, COMP2_OUT, QUADSPI1_BK1_IO2, UCPD1_FRSTX, EVENTOUT	ADC2_IN4, COMP2_INP, OPAMP1_VINP, OPAMP2_VINP
E6	16	-	22	22	K2	30	L3	37	PC4	I/O	FT_fa	-	TIM1_ETR, I2C2_SCL, USART1_TX, QUADSPI1_BK2_IO3, EVENTOUT	ADC2_IN5



Table 12. STM32G473xB/xC/xE pin definition (continued)

			Pin	Nur	nber				<u>.</u>				(continuou)	
WLCSP81	UFQFPN48	LQFP48	LQFP64	LQFP80	TFBGA100	LPQF100	UFBGA121	LPQF128	Pin name (function after reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
H7	1	-	23	23	J3	31	H4	38	PC5	I/O	TT_a	-	TIM15_BKIN, SAI1_D3, TIM1_CH4N, USART1_RX, EVENTOUT	ADC2_IN11, OPAMP1_VINM, OPAMP2_VINM, WKUP5
F6	17	16	24	24	H4	32	J4	39	PB0	I/O	TT_a	-	TIM3_CH3, TIM8_CH2N, TIM1_CH2N, QUADSPI1_BK1_IO1, UCPD1_FRSTX, EVENTOUT	ADC3_IN12/ ADC1_IN15, COMP4_INP, OPAMP2_VINP, OPAMP3_VINP
G6	18	17	25	25	K3	33	G5	40	PB1	I/O	TT_a	-	TIM3_CH4, TIM8_CH3N, TIM1_CH3N, COMP4_OUT, QUADSPI1_BK1_IO0, EVENTOUT	ADC3_IN1/ ADC1_IN12, COMP1_INP, OPAMP3_VOUT, OPAMP6_VINM
J7	19	18	26	26	J4	34	K4	41	PB2	I/O	TT_a	-	RTC_OUT2, LPTIM1_OUT, TIM5_CH1, TIM20_CH1, I2C3_SMBA, QUADSPI1_BK2_IO1, EVENTOUT	ADC2_IN12, COMP4_INM, OPAMP3_VINM
Н6	-	19	27	27	K4	35	K5	42	VSSA	S	-	-	-	-
J6	20	20	28	28	K5	36	L4	43	VREF+	S	-	-	-	VREFBUF_OUT
-	1	1	-	-	-	1	-	44	VREF+	S	1	-	-	VREFBUF_OUT
J5	21	21	29	29	J5	37	L5	45	VDDA	S	-	-	-	-
Н9	1	-	-	-	E5	-	K6	46	VSS	S	-	-	-	-
J1	-	-	-	-	F5	-	L6	47	VDD	S	1	-	-	-
-	-	-	-	-	-	-	J5	48	PF11	I/O	FT	-	TIM20_ETR, FMC_NE4, EVENTOUT	-
-	1	-	-	-	-	-	H5	49	PF12	I/O	FT	-	TIM20_CH1, FMC_A6, EVENTOUT	-
-	-	1	-	-	-	-	J6	50	PF13	I/O	FT	-	TIM20_CH2, I2C4_SMBA, FMC_A7, EVENTOUT	-
-	-	-	-	-	-	-	Н6	51	PF14	I/O	FT_f	-	TIM20_CH3, I2C4_SCL, FMC_A8, EVENTOUT	-
-	-	-	-	-	-	-	G6	52	PF15	I/O	FT_f	-	TIM20_CH4, I2C4_SDA, FMC_A9, EVENTOUT	-

Table 12. STM32G473xB/xC/xE pin definition (continued)

			Pir	Nur	nber				ter	-	ē		,	
WLCSP81	UFQFPN48	LQFP48	LQFP64	LQFP80	TFBGA100	LPQF100	UFBGA121	LPQF128	Pin name (function after reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
H5	-	-	-	30	G5	38	L7	53	PE7	I/O	TT_a	-	TIM1_ETR, FMC_D4, SAI1_SD_B, EVENTOUT	ADC3_IN4, COMP4_INP
G5	-	-	-	31	H5	39	K7	54	PE8	I/O	FT_a	-	TIM5_CH3, TIM1_CH1N, FMC_D5, SAI1_SCK_B, EVENTOUT	ADC345_IN6, COMP4_INM
F5	1	ı	ı	32	H6	40	J7	55	PE9	I/O	FT_a	-	TIM5_CH4, TIM1_CH1, FMC_D6, SAI1_FS_B, EVENTOUT	ADC3_IN2
J4	1	1	-	33	K6	41	H7	56	PE10	I/O	FT_a	1	TIM1_CH2N, QUADSPI1_CLK, FMC_D7, SAI1_MCLK_B, EVENTOUT	ADC345_IN14
H4	-	-	-	34	J6	42	L8	57	PE11	I/O	FT_a	-	TIM1_CH2, SPI4_NSS, QUADSPI1_BK1_NCS , FMC_D8, EVENTOUT	ADC345_IN15
E5	1	1	-	35	G6	43	K8	58	PE12	I/O	FT_a	-	TIM1_CH3N, SPI4_SCK, QUADSPI1_BK1_IO0, FMC_D9, EVENTOUT	ADC345_IN16
G4	1	1	-	36	K7	44	J8	59	PE13	I/O	FT_a	-	TIM1_CH3, SPI4_MISO, QUADSPI1_BK1_IO1, FMC_D10, EVENTOUT	ADC3_IN3
J3	-	-	-	37	J7	45	K9	60	PE14	I/O	FT_a	-	TIM1_CH4, SPI4_MOSI, TIM1_BKIN2, QUADSPI1_BK1_IO2, FMC_D11, EVENTOUT	ADC4_IN1
F4	-	-	-	38	H7	46	L9	61	PE15	I/O	FT_a	-	TIM1_BKIN, TIM1_CH4N, USART3_RX, QUADSPI1_BK1_IO3, FMC_D12, EVENTOUT	ADC4_IN2



Table 12. STM32G473xB/xC/xE pin definition (continued)

			Pir	Nur	nber								(continuou)	
WLCSP81	UFQFPN48	LQFP48	LQFP64	LQFP80	TFBGA100	LPQF100	UFBGA121	LPQF128	Pin name (function after reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
НЗ	22	22	30	39	J8	47	L10	62	PB10	I/O	TT_a	-	TIM2_CH3, USART3_TX, LPUART1_RX, QUADSPI1_CLK, TIM1_BKIN, SAI1_SCK_A, EVENTOUT	COMP5_INM, OPAMP3_VINM, OPAMP4_VINM
J2	-	23	31	40	E6	48	K10	63	VSS	S	-	-	-	-
J1	23	24	32	41	F7	49	K11	64	VDD	S	-	-	-	-
H2	24	25	33	42	Н8	50	L11	65	PB11	I/O	TT_a	-	TIM2_CH4, USART3_RX, LPUART1_TX, QUADSPI1_BK1_NCS , EVENTOUT	ADC12_IN14, COMP6_INP, OPAMP4_VINP, OPAMP6_VOUT
G3	25	26	34	43	K8	51	J9	66	PB12	I/O	TT_a	-	TIM5_ETR, I2C2_SMBA, SPI2_NSS/I2S2_WS, TIM1_BKIN, USART3_CK, LPUART1_RTS_DE, FDCAN2_RX, EVENTOUT	ADC4_IN3/ ADC1_IN11, COMP7_INM, OPAMP4_VOUT, OPAMP6_VINP
H1	26	27	35	44	J9	52	J11	67	PB13	I/O	TT_a	-	SPI2_SCK/I2S2_CK, TIM1_CH1N, USART3_CTS, LPUART1_CTS, FDCAN2_TX, EVENTOUT	ADC3_IN5, COMP5_INP, OPAMP3_VINP, OPAMP4_VINP, OPAMP6_VINP
G2	27	28	36	45	Н9	53	J10	68	PB14	I/O	TT_a	-	TIM15_CH1, SPI2_MISO, TIM1_CH2N, USART3_RTS_DE, COMP4_OUT, EVENTOUT	ADC4_IN4/ ADC1_IN5, COMP7_INP, OPAMP2_VINP, OPAMP5_VINP
E4	28	29	37	46	K9	54	Н8	69	PB15	I/O	TT_a	-	RTC_REFIN, TIM15_CH2, TIM15_CH1N, COMP3_OUT, TIM1_CH3N, SPI2_MOSI/I2S2_SD, EVENTOUT	ADC4_IN5/ ADC2_IN15, COMP6_INM, OPAMP5_VINM
G1	-	-	-	47	K10	55	Н9	70	PD8	I/O	TT_a	-	USART3_TX, FMC_D13, EVENTOUT	ADC4_IN12/ ADC5_IN12, OPAMP4_VINM
F3	-	-	-	48	G8	56	H10	71	PD9	I/O	TT_a	-	USART3_RX, FMC_D14, EVENTOUT	ADC4_IN13/ ADC5_IN13, OPAMP6_VINP

Table 12. STM32G473xB/xC/xE pin definition (continued)

			Pin	Nur	nber				ter	-	ē		,	
WLCSP81	UFQFPN48	LQFP48	LQFP64	LQFP80	TFBGA100	LPQF100	UFBGA121	LPQF128	Pin name (function after reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
F2	1	-	-	49	G7	57	H11	72	PD10	I/O	FT_a	-	USART3_CK, FMC_D15, EVENTOUT	ADC345_IN7, COMP6_INM
E2	1	-	1	-	H10	58	G7	73	PD11	I/O	TT_a	1	TIM5_ETR, I2C4_SMBA, USART3_CTS, FMC_A16, EVENTOUT	ADC345_IN8, COMP6_INP, OPAMP4_VINP
-	1	1	1	-	J10	59	G8	74	PD12	I/O	TT_a	1	TIM4_CH1, USART3_RTS_DE, FMC_A17, EVENTOUT	ADC345_IN9, COMP5_INP, OPAMP5_VINP
-	-	-	-	-	G9	60	G9	75	PD13	I/O	FT_a	-	TIM4_CH2, FMC_A18, EVENTOUT	ADC345_IN10, COMP5_INM
-	1	-	-	-	F8	61	G10	76	PD14	I/O	TT_a	-	TIM4_CH3, FMC_D0, EVENTOUT	ADC345_IN11, COMP7_INP, OPAMP2_VINP
-	ı	ı	1	-	G10	62	F6	77	PD15	I/O	FT_a	-	TIM4_CH4, SPI2_NSS, FMC_D1, EVENTOUT	COMP7_INM
B1	-	-	-	50	E7	63	-	78	VSS	S	-	-	-	-
E1	-	-	-	51	-	64	G11	79	VDD	S	-	-	-	-
E3	29	1	38	52	F9	65	F10	80	PC6	I/O	FT_f	1	TIM3_CH1, TIM8_CH1, I2S2_MCK, COMP6_OUT, I2C4_SCL, EVENTOUT	-
D5	1	-	39	53	F10	66	F11	81	PC7	I/O	FT_f	-	TIM3_CH2, TIM8_CH2, I2S3_MCK, COMP5_OUT, I2C4_SDA, EVENTOUT	-
-	1	1	-	-	-	-	F9	82	PG0	I/O	FT	1	TIM20_CH1N, FMC_A10, EVENTOUT	-
-	-	1	-	-	-	-	F8	83	PG1	I/O	FT	- 1	TIM20_CH2N, FMC_A11, EVENTOUT	-
-	-	-	-	-	-	-	F7	84	PG2	I/O	FT	-	TIM20_CH3N, SPI1_SCK, FMC_A12, EVENTOUT	-



Table 12. STM32G473xB/xC/xE pin definition (continued)

			Pir	Nur	nber				ter		ē		•	
WLCSP81	UFQFPN48	LQFP48	LQFP64	LQFP80	TFBGA100	LPQF100	UFBGA121	LPQF128	Pin name (function after reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
-	1	-	-	-	1	-	E11	85	PG3	I/O	FT_f	1	TIM20_BKIN, I2C4_SCL, SPI1_MISO, TIM20_CH4N, FMC_A13, EVENTOUT	-
-	1	-	-	-	1	-	E10	86	PG4	I/O	FT_f	-	TIM20_BKIN2, I2C4_SDA, SPI1_MOSI, FMC_A14, EVENTOUT	-
C5	1	-	40	54	E8	67	E9	87	PC8	I/O	FT_f	-	TIM3_CH3, TIM8_CH3, TIM20_CH3, COMP7_OUT, I2C3_SCL, EVENTOUT	-
D2	1	-	41	55	E9	68	E8	88	PC9	I/O	FT_f	1	TIM3_CH4, TIM8_CH4, I2SCKIN, TIM8_BKIN2, I2C3_SDA, EVENTOUT	-
D1	30	30	42	56	E10	69	E7	89	PA8	I/O	FT_a	1	MCO, I2C3_SCL, I2C2_SDA, I2S2_MCK, TIM1_CH1, USART1_CK, COMP7_OUT, TIM4_ETR, FDCAN3_RX, SAI1_CK2, SAI1_SCK_A, EVENTOUT	ADC5_IN1, OPAMP5_VOUT
D4	31	31	43	57	D10	70	D8	90	PA9	I/O	FT_fda	-	I2C3_SMBA, I2C2_SCL,I2S3_MCK, TIM1_CH2, USART1_TX, OMP5_OUT, TIM15_BKIN, TIM2_CH3, SAI1_FS_A, EVENTOUT	ADC5_IN2, UCPD1_DBCC1

Table 12. STM32G473xB/xC/xE pin definition (continued)

			Pir	Nur	mber				ter				,	
WLCSP81	UFQFPN48	LQFP48	LQFP64	LQFP80	TFBGA100	LPQF100	UFBGA121	LPQF128	Pin name (function after reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
D3	32	32	44	58	D9	71	D9	91	PA10	I/O	FT_fda	-	TIM17_BKIN, USB_CRS_SYNC, I2C2_SMBA, SPI2_MISO, TIM1_CH3, USART1_RX, COMP6_OUT, TIM2_CH4, TIM8_BKIN, SAI1_D1, SAI1_SD_A, EVENTOUT	UCPD1_DBCC2, PVD_IN
C2	33	33	45	59	C10	72	D11	92	PA11	I/O	FT_u	-	SPI2_MOSI/I2S2_SD, TIM1_CH1N, USART1_CTS, COMP1_OUT, FDCAN1_RX, TIM4_CH1, TIM1_CH4, TIM1_BKIN2, EVENTOUT	USB_DM
C1	34	34	46	60	С9	73	D10	93	PA12	I/O	FT_u	-	TIM16_CH1, I2SCKIN, TIM1_CH2N, USART1_RTS_DE, COMP2_OUT, FDCAN1_TX, TIM4_CH2, TIM1_ETR, EVENTOUT	USB_DP
A8	-	35	47	61	F6	74	C10	94	VSS	S	-	-	-	-
A1	35	36	48	62	-	75	C11	95	VDD	S	-	-	-	-
B2	36	37	49	63	D8	76	B11	96	PA13	I/O	FT_f	(5)	SWDIO-JTMS, TIM16_CH1N, I2C4_SCL, I2C1_SCL, IR_OUT, USART3_CTS, TIM4_CH3, SAI1_SD_B, EVENTOUT	-
-	-	-	-	-	-	-	A11	97	PF6	I/O	FT_f	-	TIM5_ETR, TIM4_CH4, SAI1_SD_B, I2C2_SCL,TIM5_CH1, USART3_RTS, QUADSPI1_BK1_IO3, EVENTOUT	-



Table 12. STM32G473xB/xC/xE pin definition (continued)

			Pir	Nur	nber								(continued)	
WLCSP81	UFQFPN48	LQFP48	LQFP64	LQFP80	TFBGA100	LPQF100	UFBGA121	LPQF128	Pin name (function after reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
C3	37	38	50	64	B10	77	B10	98	PA14	I/O	FT_f	(5)	SWCLK-JTCK, LPTIM1_OUT, I2C4_SMBA, I2C1_SDA, TIM8_CH2, TIM1_BKIN, USART2_TX, SAI1_FS_B, EVENTOUT	-
A2	38	39	51	65	B9	78	A10	99	PA15	I/O	FT_f	(5)	JTDI, TIM2_CH1, TIM8_CH1, I2C1_SCL, SPI1_NSS, SPI3_NSS/I2S3_WS, USART2_RX, UART4_RTS_DE, TIM1_BKIN, FDCAN3_TX, TIM2_ETR, EVENTOUT	-
В3	39	-	52	66	C8	79	С9	100	PC10	I/O	FT	-	TIM8_CH1N, UART4_TX, SPI3_SCK/I2S3_CK, USART3_TX, EVENTOUT	-
C4	40	1	53	67	C7	80	C8	101	PC11	I/O	FT_f	-	TIM8_CH2N, UART4_RX, SPI3_MISO, USART3_RX, I2C3_SDA, EVENTOUT	-
A3	-	-	54	68	A10	81	D7	102	PC12	I/O	FT	-	TIM5_CH2, TIM8_CH3N, UART5_TX, SPI3_MOSI/I2S3_SD, USART3_CK, UCPD1_FRSTX, EVENTOUT	-
-	-	-	-	-	1	-	1	103	PG5	I/O	FT	-	TIM20_ETR, SPI1_NSS, LPUART1_CTS, FMC_A15, EVENTOUT	-
-	-	-	-	-	-	-	-	104	PG6	I/O	FT	-	TIM20_BKIN, I2C3_SMBA, LPUART1_RTS_DE, FMC_INT, EVENTOUT	-

Table 12. STM32G473xB/xC/xE pin definition (continued)

			Pir	Nur	nber								(continued)	
WLCSP81	UFQFPN48	LQFP48	LQFP64	LQFP80	TFBGA100	LPQF100	UFBGA121	LPQF128	Pin name (function after reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
-	-	-	-	-	-	-	-	105	PG7	I/O	FT_f	-	SAI1_CK1, I2C3_SCL, LPUART1_TX, FMC_INT, SAI1_MCLK_A, EVENTOUT	-
-	-	-	-	-	-	-	-	106	PG8	I/O	FT_f	-	I2C3_SDA, LPUART1_RX, FMC_NE3, EVENTOUT	-
-	-	-	-	-	-	-	-	107	PG9	I/O	FT	-	SPI3_SCK, USART1_TX, FMC_NCE/FMC_NE2, TIM15_CH1N, EVENTOUT	-
B4	-	-	-	69	В8	82	В9	108	PD0	I/O	FT	-	TIM8_CH4N, FDCAN1_RX, FMC_D2, EVENTOUT	-
A4	-	-	-	70	A9	83	A9	109	PD1	1/0	FT	-	TIM8_CH4, TIM8_BKIN2, FDCAN1_TX, FMC_D3, EVENTOUT	-
-	-	-	-	-	-	-	В8	110	VSS	S	-	-	-	-
A1	-	-	-	-	-	-	A8	111	VDD	S	-	-	-	-
B5	-	-	55	71	В7	84	C7	112	PD2	I/O	FT	-	TIM3_ETR, TIM8_BKIN, UART5_RX, EVENTOUT	-
-	1	1	-	-	C6	85	В7	113	PD3	I/O	FT	-	TIM2_CH1/ TIM2_ETR, USART2_CTS, QUADSPI1_BK2_NCS , FMC_CLK, EVENTOUT	-
-	-	-	-	-	A8	86	A7	114	PD4	I/O	FT	-	TIM2_CH2, USART2_RTS_DE, QUADSPI1_BK2_IO0, FMC_NOE, EVENTOUT	-
-	-	-	-	-	A7	87	E6	115	PD5	I/O	FT	-	USART2_TX, QUADSPI1_BK2_IO1, FMC_NWE, EVENTOUT	-



Table 12. STM32G473xB/xC/xE pin definition (continued)

			Pir	Nur	nber				ter					
WLCSP81	UFQFPN48	LQFP48	LQFP64	LQFP80	TFBGA100	LPQF100	UFBGA121	LPQF128	Pin name (function after reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
-	-	-	-	1	A6	88	D6	116	PD6	I/O	FT	-	TIM2_CH4, SAI1_D1, USART2_RX, QUADSPI1_BK2_IO2, FMC_NWAIT, SAI1_SD_A, EVENTOUT	-
-	-	-	-	-	В6	89	В6	117	PD7	I/O	FT	-	TIM2_CH3, USART2_CK, QUADSPI1_BK2_IO3, FMC_NCE/FMC_NE1, EVENTOUT	-
A5	41	40	56	72	A5	90	A6	118	PB3	I/O	FT	(5)	JTDO-TRACESWO, TIM2_CH2, TIM4_ETR, UCPD1_CRS_SYNC, TIM8_CH1N, SPI1_SCK, SPI3_SCK/I2S3_CK, USART2_TX, TIM3_ETR, FDCAN3_RX, SAI1_SCK_B, EVENTOUT	-
C6	42	41	57	73	C5	91	C6	119	PB4	I/O	FT_c	(5) (6)	JTRST, TIM16_CH1, TIM3_CH1, TIM8_CH2N, SPI1_MISO, SPI3_MISO, USART2_RX, UART5_RTS_DE, TIM17_BKIN, FDCAN3_TX, SAI1_MCLK_B, EVENTOUT	UCPD1_CC2
A6	43	42	58	74	B5	92	B5	120	PB5	I/O	FT_f	-	TIM16_BKIN, TIM3_CH2, TIM8_CH3N, I2C1_SMBA, SPI1_MOSI, SPI3_MOSI/I2S3_SD, USART2_CK, I2C3_SDA, FDCAN2_RX, TIM17_CH1, LPTIM1_IN1, SAI1_SD_B, UART5_CTS, EVENTOUT	-

Table 12. STM32G473xB/xC/xE pin definition (continued)

			Pir	Nur	nber				ter		é			
WLCSP81	UFQFPN48	LQFP48	LQFP64	LQFP80	TFBGA100	LPQF100	UFBGA121	LPQF128	Pin name (function after reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
В6	44	43	59	75	A4	93	A5	121	PB6	I/O	FT_c	(6)	TIM16_CH1N, TIM4_CH1, TIM8_CH1, TIM8_ETR, USART1_TX, COMP4_OUT, FDCAN2_TX, TIM8_BKIN2, LPTIM1_ETR, SAI1_FS_B, EVENTOUT	UCPD1_CC1
C7	45	44	60	76	B4	94	C5	122	PB7	I/O	FT_f	-	TIM17_CH1N, TIM4_CH2, I2C4_SDA,I2C1_SDA, TIM8_BKIN, USART1_RX, COMP3_OUT, TIM3_CH4, LPTIM1_IN2, FMC_NL, UART4_CTS, EVENTOUT	-
В7	46	45	61	77	А3	95	D5	123	PB8-BOOT0	I/O	FT_f	(7)	TIM16_CH1, TIM4_CH3, SAI1_CK1, I2C1_SCL, USART3_RX, COMP1_OUT, FDCAN1_RX, TIM8_CH2, TIM1_BKIN, SAI1_MCLK_A, EVENTOUT	-
A7	47	46	62	78	A2	96	A4	124	PB9	I/O	FT_f	1	TIM17_CH1, TIM4_CH4, SAI1_D2, I2C1_SDA, IR_OUT, USART3_TX, COMP2_OUT, FDCAN1_TX, TIM8_CH3, TIM1_CH3N, SAI1_FS_A, EVENTOUT	-
-	-	-	-	-	C4	97	B4	125	PE0	I/O	FT	-	TIM4_ETR, TIM20_CH4N, TIM16_CH1, TIM20_ETR, USART1_TX, FMC_NBL0, EVENTOUT	-



			Pir	Nur	nber				ter		re			
WLCSP81	UFQFPN48	LQFP48	LQFP64	LQFP80	TFBGA100	LPQF100	UFBGA121	LPQF128	Pin name (function aft reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
-	-	-	-	-	В3	98	C4	126	PE1	I/O	FT	-	TIM17_CH1, TIM20_CH4, USART1_RX, FMC_NBL1, EVENTOUT	-
-	-	47	63	79	-	99	В3	127	VSS	S	-	-	-	-
A9	48	48	64	80	-	100	А3	128	VDD	S	-	-	-	-

Table 12. STM32G473xB/xC/xE pin definition (continued)

- 1. Function availability depends on the chosen device.
- PC13, PC14 and PC15 are supplied through the power switch. Since the switch only sinks a limited amount of current (3 mA), the use of GPIOs PC13 to PC15 in output mode is limited:

 The speed should not exceed 2 MHz with a maximum load of 30 pF

 - These GPIOs must not be used as current sources (e.g. to drive an LED).
- After a backup domain power-up, PC13, PC14 and PC15 operate as GPIOs. Their function then depends on the content of the RTC registers which are not reset by the system reset. For details on how to manage these GPIOs, refer to the Backup domain and RTC register descriptions in the reference manual RM0440 "STM32G4 Series advanced Arm®-based 32-bit MCUs"
- 4. PG10-NRST pin is FT tolerant if it is configured as PG10 GPIO by option bytes except for the startup time until option bytes are loaded.
- 5. After reset, these pins are configured as JTAG/SW debug alternate functions, and the internal pull-up on PA15, PA13, PB4 pins and the internal pull-down on PA14 pin are activated.
- After reset, a pull-down resistor (Rd = 5.1k Ω from UCPD peripheral) can be activated on PB6, PB4 (UCPD1_CC1, After reset, a pull-down resistor (Rd = 5. 1kΩ from UCPD peripheral) can be activated on PB6, PB4 (UCPD1_CC1). The pull-down on PB6 (UCPD1_CC1) is activated by high level on PA9 (UCPD1_DBCC1). The pull-down on PB4 (UCPD1_CC2) is activated by high level on PA10 (UCPD1_DBC2). This pull-down control (dead battery support on UCPD peripheral) can be disabled by setting bit UCPD1_DBDIS=1 in the PWR_CR3 register. PB4, PB6 have UCPD_CC functionality which implements an internal pull-down resistor (5.1kΩ) which is controlled by the voltage on the UCPD_DBCC pin (PA10, PA9). A high level on the UCPD_DBCC pin activates the pull-down on the UCPD_CC pin. The pull-down effect on the CC lines can be removed by using the bit UCPD1_DBDIS =1 (USB Type-C and power delivery dead battery disable) in the PWR_CR3 register.
- It is recommended to set PB8 in another mode than analog mode after startup to limit consumption if the pin is left unconnected.



4.11 Alternate functions

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Table 13. Alternate function

_																	
		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
	Port	SYS_AF	LPTIM1/ TIM2/5/ 15/16/17	I2C3/ TIM1/2/3/4/5/8/1 5/20/ GPCOMP1	QUADSPI/ I2C3/4/SAI/ USB/ TIM8/15/20/GPC OMP3/ TSC	I2C1/2/3/ 4/ TIM1/8/1 6/17	QUADSPI/SP I1/2/3/4/I2S2/ 3/ I2C4/ UART4/5/TIM 8/ Infrared	QUADSPI/S PI2/3/ I2S2/3/ TIM1/5/8/20 / Infrared	USART1/2/3 /CAN/ GPCOMP5/6 /7	I2C3/4/ UART4/5/L PUART1/G PCOMP1/2/ 3/4/5/6/7	CAN/ TIM1/8/1 5/CAN1/2	QUADSPI/TI M2/3/4/8/17	LPTIM1/ TIM1/8/C AN1/3	SDIO/FMC/ LPUART1/ SAI TIM1	SAI/OPAMP2	UART4/5/ SAI/ TIM2/15/ UCPD	EVENT
	PA0	-	TIM2_CH1	TIM5_CH1	-	-	÷	-	USART2_ CTS	COMP1 _OUT	TIM8_ BKIN	TIM8_ETR	-	-	-	TIM2_ ETR	EVENT OUT
	PA1	RTC_ REFIN	TIM2_CH2	TIM5_CH2	-	-	-	-	USART2_ RTS_DE	-	TIM15_C H1N	-	-	-	-	-	EVENT OUT
	PA2	-	TIM2_CH3	TIM5_CH3	-	-	-	-	USART2_ TX	COMP2 _OUT	TIM15_C H1	QUADSPI_B K1_NCS	-	LPUART1_TX	-	UCPD_ FRSTX	EVENT OUT
	PA3	-	TIM2_CH4	TIM5_CH4	SAI_CK1	-	-	-	USART2_ RX	-	TIM15_C H2	QUADSPI_C LK	-	LPUART1_RX	SAI_MCLK_A	-	EVENT OUT
	PA4	-	-	TIM3_CH2	-	-	SPI1_NSS	SPI3_NSS/I 2S3_WS	USART2_ CK	-	-	-	-	-	SAI_FS_B	-	EVENT OUT
	PA5	-	TIM2_CH1	TIM2_ETR	-	-	SPI1_SCK	-	-	-	-	-	-	-	-	UCPD_ FRSTX	EVENT OUT
	PA6	-	TIM16_CH1	TIM3_CH1	-	TIM8_ BKIN	SPI1_MISO	TIM1_BKIN	-	COMP1 _OUT	-	QUADSPI_B K1_IO3	-	LPUART1_ CTS	-	-	EVENT OUT
tΑ	PA7	-	TIM17_CH1	TIM3_CH2	-	TIM8_ CH1N	SPI1_MOSI	TIM1_CH1 N	-	COMP2_ OUT	-	QUADSPI_B K1_IO2	-	-	-	UCPD_ FRSTX	EVENT OUT
Port A	PA8	мсо	-	-	-	I2C2_ SDA	I2S2_MCK	TIM1_CH1	USART1_ CK	COMP7 _OUT	-	TIM4_ETR	FDCAN3 _RX	SAI_CK2	-	SAI_SCK _A	EVENT OUT
	PA9	-	-	I2C3_SMBA	-	I2C2_ SCL	I2S3_MCK	TIM1_CH2	USART1_ TX	COMP5 _OUT	TIM15_B KIN	TIM2_CH3	FDCAN1 _RX	-	-	SAI_FS_ A	EVENT OUT
	PA10	-	TIM17_BKIN	I2C3_SCL	USB_CRS_ SYNC	I2C2_ SMBA	SPI2_MISO	TIM1_CH3	USART1_ RX	COMP6 _OUT	FDCAN1 _TX	TIM2_CH4	TIM8_ BKIN	SAI_D1	-	SAI_SD_ A	EVENT OUT
	PA11	-	-	-	-	-	SPI2_MOSI/I 2S2_SD	TIM1_CH1 N	USART1_ CTS	COMP1 _OUT	FDCAN1 _RX	TIM4_CH1	TIM1_CH 4	TIM1_BKIN2	-	-	EVENT OUT
	PA12	-,	TIM16_CH1	-	-	-	I2SCKIN	TIM1_CH2 N	USART1_ RTS_DE	COMP2 _OUT	FDCAN1 _TX	TIM4_CH2	TIM1_ ETR	-	-	-	EVENT OUT
	PA13	SWDIO- JTMS	TIM16_CH1N	-	-	-	IR_OUT	-	USART3_ CTS	-	-	TIM4_CH3	-	-	SAI_SD_B	-	EVENT OUT
	PA14	SWCLK- JTCK	LPTIM1_OUT	-	I2C4_SMBA	I2C1_ SDA	TIM8_CH2	TIM1_ BKIN	USART2_ TX	-			FDCAN3 _TX	-	SAI_FS_B	-	EVENT OUT
	PA15	JTDI	TIM2_CH1	TIM8_CH1	-	I2C1_ SCL	SPI1_NSS	SPI3_NSS/I 2S3_WS	USART2_ RX	UART4 _RTS_DE	TIM1_ BKIN	-	FDCAN3 _TX	-	-	TIM2_ET R	EVENT OUT

Pinouts and pin description

Tabl	le 13.	Alt	ernate	fu	nction	(continu	ıed)

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
		SYS_AF	LPTIM1/ TIM2/5/ 15/16/17	I2C3/ TIM1/2/3/4/5/8/1 5/20/ GPCOMP1	QUADSPI/ I2C3/4/SAI/ USB/ TIM8/15/20/GPC OMP3/ TSC	I2C1/2/3/ 4/ TIM1/8/1 6/17	QUADSPI/SP I1/2/3/4/I2S2/ 3/ I2C4/ UART4/5/TIM 8/ Infrared	QUADSPI/S PI2/3/ I2S2/3/ TIM1/5/8/20 / Infrared	USART1/2/3 /CAN/ GPCOMP5/6 /7	I2C3/4/ UART4/5/L PUART1/G PCOMP1/2/ 3/4/5/6/7	CAN/ TIM1/8/1 5/CAN1/2	QUADSPI/TI M2/3/4/8/17	LPTIM1/ TIM1/8/C AN1/3	SDIO/FMC/ LPUART1/ SAI TIM1	SAI/OPAMP2	UART4/5/ SAI/ TIM2/15/ UCPD	EVENT
PortB	PB0	-	-	TIM3_CH3	-	TIM8_ CH2N	-	TIM1_CH2 N	-	-	-	QUADSPI_B K1_IO1	-	-	-	UCPD_ FRSTX	EVENT OUT
	PB1	-	-	TIM3_CH4	-	TIM8_ CH3N	-	TIM1_CH3 N	-	COMP4_O UT	-	QUADSPI_B K1_IO0	-	LPUART1_RTS _DE	-	-	EVENT OUT
	PB2	-	LPTIM1_OUT	TIM5_CH1	TIM20_CH1	I2C3_ SMBA	-	-	-	-	-	QUADSPI_B K2_IO1	-	-	-	-	EVENT OUT
	PB3	JTDO- TRACESWO	TIM2_CH2	TIM4_ETR	USB_CRS_SYN C	TIM8_ CH1N	SPI1_SCK	SPI3_SCK/I 2S3_CK	USART2_TX	-	-	TIM3_ETR	FDCAN3 _RX	-	-	SAI_SCK _B	EVENT OUT
	PB4	JTRST	TIM16_CH1	TIM3_CH1	-	TIM8_ CH2N	SPI1_MISO	SPI3_MISO	USART2_R X	UART5_RT S_DE	-	TIM17_BKIN	FDCAN3 _TX	-	-	SAI_MCL K_B	EVENT OUT
	PB5	-	TIM16_BKIN	TIM3_CH2	TIM8_CH3N	I2C1_ SMBA	SPI1_MOSI	SPI3_MOSI /I2S3_SD	USART2_C K	I2C3_SDA	FDCAN2 _RX	TIM17_CH1	LPTIM1_I N1	SAI_SD_B	-	UART5_ CTS	EVENT OUT
	PB6	-	TIM16_CH1N	TIM4_CH1	-	-	TIM8_CH1	TIM8_ETR	USART1_TX	COMP4_O UT	FDCAN2 _TX	TIM8_BKIN2	LPTIM1_ ETR	-	-	SAI_FS_ B	EVENT OUT
	PB7	=	TIM17_CH1N	TIM4_CH2	I2C4_SDA	I2C1_ SDA	TIM8_BKIN	-	USART1_R X	COMP3_O UT	FDCAN2 _TX	TIM3_CH4	LPTIM1_I N2	FMC_NL	-	UART4_ CTS	EVENT OUT
	PB8	-	TIM16_CH1	TIM4_CH3	SAI_CK1	I2C1_ SCL	-	-	USART3_R X	COMP1_O UT	FDCAN1 _RX	TIM8_CH2	-	TIM1_BKIN	-	SAI_MCL K_A	EVENT OUT
	PB9	-	TIM17_CH1	TIM4_CH4	SAI_D2	I2C1_ SDA	-	IR_OUT	USART3_TX	COMP2_O UT	FDCAN1 _TX	TIM8_CH3	-	TIM1_CH3N	-	SAI_FS_ A	EVENT OUT
	PB10	-	TIM2_CH3	-	-	-	-	-	USART3_TX	LPUART1_ RX	-	QUADSPI_C LK	FDCAN3 _TX	TIM1_BKIN	-	SAI_SCK _A	EVENT OUT
	PB11	-	TIM2_CH4	-	-	-	-	-	USART3_R X	LPUART1_ TX	-	QUADSPI_B K1_NCS	FDCAN3 _RX	-	-	-	EVENT OUT
	PB12		-	TIM5_ETR	-	I2C2_ SMBA	SPI2_NSS/I2 S2_WS	TIM1_BKIN	USART3_C K	LPUART1_ RTS_DE	FDCAN2 _RX		-	-,	-	-	EVENT OUT
	PB13	-	-	-	-	-	SPI2_SCK/I2 S2_CK	TIM1_CH1 N	USART3_CT S	LPUART1_ CTS	FDCAN2 _TX		-	-,	-	-	EVENT OUT
	PB14		TIM15_CH1	-	-	-	SPI2_MISO	TIM1_CH2 N	USART3_RT S_DE	COMP4_O UT	-		-	-,	-	-	EVENT OUT
	PB15	RTC_REFIN	TIM15_CH2	TIM15_CH1N	COMP3_OUT	TIM1_ CH3N	SPI2_MOSI/I 2S2_SD	-	-	-	-	-	-	-	-	-	EVENT OUT





Table 13. Alternate function (continued)

		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
	Port	SYS_AF	LPTIM1/ TIM2/5/ 15/16/17	I2C3/ TIM1/2/3/4/5/8/1 5/20/ GPCOMP1	QUADSPI/ I2C3/4/SAI/ USB/ TIM8/15/20/GPC OMP3/ TSC	I2C1/2/3/ 4/ TIM1/8/1 6/17	QUADSPI/SP I1/2/3/4/12S2/ 3/ I2C4/ UART4/5/TIM 8/ Infrared	QUADSPI/S PI2/3/ I2S2/3/ TIM1/5/8/20 / Infrared	USART1/2/3 /CAN/ GPCOMP5/6 /7	I2C3/4/ UART4/5/L PUART1/G PCOMP1/2/ 3/4/5/6/7	CAN/ TIM1/8/1 5/CAN1/2	QUADSPI/TI M2/3/4/8/17	LPTIM1/ TIM1/8/C AN1/3	SDIO/FMC/ LPUART1/ SAI TIM1	SAI/OPAMP2	UART4/5/ SAI/ TIM2/15/ UCPD	EVENT
	PC0	-	LPTIM1_IN1	TIM1_CH1	-	-	-	-	-	LPUART1_ RX	-	-	-	-	-	-	EVENT OUT
	PC1	-	LPTIM1_OUT	TIM1_CH2	-	-	-	-	-	LPUART1_ TX	-	QUADSPI_ BK2_IO0	-	-	SAI_SD_A	-	EVENT OUT
	PC2	-	LPTIM1_IN2	TIM1_CH3	COMP3_OUT	-	-	TIM20_CH2	-	-	-	QUADSPI_ BK2_IO1	-	-	-	-	EVENT OUT
	PC3	-	LPTIM1_ETR	TIM1_CH4	SAI_D1	-	-	TIM1_BKIN 2	-	-	-	QUADSPI_ BK2_IO2	-	-	SAI_SD_A	-	EVENT OUT
	PC4	-	-	TIM1_ETR	-	I2C2_SC L	-	-	USART1_TX	-	-	QUADSPI_ BK2_IO3	-	-	-	-	EVENT OUT
	PC5	-	-	TIM15_BKIN	SAI_D3	-	-	TIM1_CH4 N	USART1_R X	-	-	-	-	-	-	-	EVENT OUT
	PC6	-	-	TIM3_CH1	-	TIM8_ CH1	-	I2S2_MCK	COMP6_OU T	I2C4_SCL	-	-	-	-	-	-	EVENT OUT
0	PC7	-	-	TIM3_CH2	-	TIM8_ CH2	-	I2S3_MCK	COMP5_OU T	I2C4_SDA	-	-	-	-	-	-	EVENT OUT
Port	PC8	-	-	TIM3_CH3	-	TIM8_ CH3	-	TIM20_CH3	COMP7_OU T	I2C3_SCL	-	-	-	-	-	-	EVENT OUT
	PC9	-	-	TIM3_CH4	-	TIM8_ CH4	I2SCKIN	TIM8_ BKIN2	-	I2C3_SDA	-	-	-	-	-	-	EVENT OUT
	PC10	-	-	-	-	TIM8_ CH1N	UART4_TX	SPI3_SCK/I 2S3_CK	USART3_TX	-	-	-	-	-	-	-	EVENT OUT
	PC11	-	-	-	-	TIM8_ CH2N	UART4_RX	SPI3_MISO	USART3_R X	I2C3_SDA	-	-	-	-	-	-	EVENT OUT
	PC12	-	TIM5_CH2	-	-	TIM8_C H3N	UART5_TX	SPI3_MOSI /I2S3_SD	USART3_C K	-	-	-	-	-	-	UCPD_ FRSTX	EVENT OUT
	PC13	-	-	TIM1_BKIN	-	TIM1_ CH1N	-	TIM8_CH4 N	-	-	-	-	-	-	-	-	EVENT OUT
	PC14	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT
	PC15	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT

Pinouts and pin description

Table 13. Alternate function (continued)

		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
	Port	SYS_AF	LPTIM1/ TIM2/5/ 15/16/17	I2C3/ TIM1/2/3/4/5/8/1 5/20/ GPCOMP1	QUADSPI/ I2C3/4/SAI/ USB/ TIM8/15/20/GPC OMP3/ TSC	I2C1/2/3/ 4/ TIM1/8/1 6/17	QUADSPI/SP I1/2/3/4/I2S2/ 3/ I2C4/ UART4/5/TIM 8/ Infrared	QUADSPI/S PI2/3/ I2S2/3/ TIM1/5/8/20 / Infrared	USART1/2/3 /CAN/ GPCOMP5/6 /7	I2C3/4/ UART4/5/L PUART1/G PCOMP1/2/ 3/4/5/6/7	CAN/ TIM1/8/1 5/CAN1/2	QUADSPI/TI M2/3/4/8/17	LPTIM1/ TIM1/8/C AN1/3	SDIO/FMC/ LPUART1/ SAI TIM1	SAI/OPAMP2	UART4/5/ SAI/ TIM2/15/ UCPD	EVENT
	PD0	-	-	-	-	-	-	TIM8_CH4 N	-	-	FDCAN1 _RX	-	-	FMC_D2	-	-	EVENT OUT
	PD1	-	-	-	-	TIM8_ CH4	-	TIM8_BKIN	-	-	FDCAN1 _TX	-	-	FMC_D3	-	-	EVENT OUT
	PD2	-	-	TIM3_ETR	-	TIM8_ BKIN	UART5_RX	-	-	-	-	-	-	-	-	-	EVENT OUT
	PD3	-	-	TIM2_CH1/TIM2 _ETR	-	-	-	-	USART2_CT S	-	-	QUADSPI_B K2_NCS	-	FMC_CLK	-	-	EVENT OUT
	PD4	-	-	TIM2_CH2	-	-	-	-	USART2_RT S_DE	-	FDCAN1 _RX	QUADSPI_B K2_IO0	-	FMC_NOE	-	-	EVENT OUT
	PD5	-	-	-	-	-	-	-	USART2_TX	-	FDCAN1 _TX	QUADSPI_B K2_IO1	-	FMC_NWE	-	-	EVENT OUT
	PD6	-	-	TIM2_CH4	SAI_D1	-	-	-	USART2_ RX	-	FDCAN2 _RX	QUADSPI_B K2_IO2	-	FMC_NWAIT	SAI_SD_A	-	EVENT OUT
τĐ	PD7	-	-	TIM2_CH3	-	-	-	-	USART2_ CK	-	-	QUADSPI_B K2_IO3	-	FMC_NCE/FM C_NE1	-	-	EVENT OUT
Port D	PD8	-	-	-	-	-	-	-	USART3_TX	-	-	-	-	FMC_D13	-	-	EVENT OUT
	PD9	-	-	-	-	-	-	-	USART3_ RX	-	FDCAN2 _RX	-	-	FMC_D14	-	-	EVENT OUT
	PD10	-	-	-	-	-	-	-	USART3_ CK	-	FDCAN2 _TX	-	-	FMC_D15	-	-	EVENT OUT
	PD11	-	TIM5_ETR	-	-	I2C4_ SMBA	-	-	USART3_ CTS	-	-	-	-	FMC_A16	-	-	EVENT OUT
	PD12	-	-	TIM4_CH1	-	-	-	-	USART3_ RTS_DE	-	-	-	-	FMC_A17	-	-	EVENT OUT
	PD13	-	-	TIM4_CH2	-	-	-	-	-	-	-	-	-	FMC_A18	-	-	EVENT OUT
	PD14	-	-	TIM4_CH3	-	-	-	-	-	-	-	-	ï	FMC_D0	-	-	EVENT OUT
	PD15	-	-	TIM4_CH4	-	-	-	SPI2_NSS	-	-	-	-	-	FMC_D1	-	-	EVENT OUT



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		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
	Port	SYS_AF	LPTIM1/ TIM2/5/ 15/16/17	I2C3/ TIM1/2/3/4/5/8/1 5/20/ GPCOMP1	QUADSPI/ I2C3/4/SAI/ USB/ TIM8/15/20/GPC OMP3/ TSC	I2C1/2/3/ 4/ TIM1/8/1 6/17	QUADSPI/SP I1/2/3/4/I2S2/ 3/ I2C4/ UART4/5/TIM 8/ Infrared	QUADSPI/S PI2/3/ I2S2/3/ TIM1/5/8/20 / Infrared	USART1/2/3 /CAN/ GPCOMP5/6 /7	I2C3/4/ UART4/5/L PUART1/G PCOMP1/2/ 3/4/5/6/7	CAN/ TIM1/8/1 5/CAN1/2	QUADSPI/TI M2/3/4/8/17	LPTIM1/ TIM1/8/C AN1/3	SDIO/FMC/ LPUART1/ SAI TIM1	SAI/OPAMP2	UART4/5/ SAI/ TIM2/15/ UCPD	EVENT
	PE0	-	-	TIM4_ETR	TIM20_CH4N	TIM16_ CH1	-	TIM20_ETR	USART1_ TX	-	FDCAN1 _RX	-	-	FMC_NBL0	-	-	EVENT OUT
	PE1	-	-	-	-	TIM17_ CH1	-	TIM20_CH4	USART1_ RX	-	FDCAN1 _TX	-	-	FMC_NBL1	-	-	EVENT OUT
	PE2	TRACECK	-	TIM3_CH1	SAI_CK1	-	SPI4_SCK	TIM20_CH1	-	-	-	-	-	FMC_A23	SAI_MCLK_A	-	EVENT OUT
	PE3	TRACED0	-	TIM3_CH2	-	-	SPI4_NSS	TIM20_CH2	-	-	-	-	-	FMC_A19	SAI_SD_B	-	EVENT OUT
	PE4	TRACED1	-	TIM3_CH3	SAI_D2	-	SPI4_NSS	TIM20_CH1 N	-	-	-	-	-	FMC_A20	SAI_FS_A	-	EVENT OUT
	PE5	TRACED2	-	TIM3_CH4	SAI_CK2	-	SPI4_MISO	TIM20_CH2 N	-	-	-	-	-	FMC_A21	SAI_SCK_A	-	EVENT OUT
	PE6	TRACED3	-	-	SAI_D1	-	SPI4_MOSI	TIM20_CH3 N	-	-	-	-	-	FMC_A22	SAI_SD_A	-	EVENT OUT
n T	PE7	-	-	TIM1_ETR	-	-	-	-	-	-	-	-	-	FMC_D4	SAI_SD_B	-	EVENT OUT
Port	PE8	-	TIM5_CH3	TIM1_CH1N	-	-	-	-	-	-	-	-	-	FMC_D5	SAI_SCK_B	-	EVENT OUT
	PE9	-	TIM5_CH4	TIM1_CH1	-	-	-	-	-	-	-	-	-	FMC_D6	SAI_FS_B	-	EVENT OUT
	PE10	-	-	TIM1_CH2N	-	-	-	-	-	-	-	QUADSPI_ CLK	-	FMC_D7	SAI_MCLK_B	-	EVENT OUT
	PE11	-	-	TIM1_CH2	-	-	SPI4_NSS	-	-	-	-	QUADSPI_ BK1_NCS	-	FMC_D8	-	-	EVENT OUT
	PE12	-	-	TIM1_CH3N	-	-	SPI4_SCK	-	-	-	-	QUADSPI_ BK1_IO0	-	FMC_D9	-	-	EVENT OUT
	PE13	-	-	TIM1_CH3	-	-	SPI4_MISO	-	-	-	-	QUADSPI_ BK1_IO1	-	FMC_D10	-	-	EVENT OUT
	PE14	-	-	TIM1_CH4	-	-	SPI4_MOSI	TIM1_ BKIN2	-	-	-	QUADSPI_ BK1_IO2	-	FMC_D11	-	-	EVENT OUT
	PE15	-	-	TIM1_BKIN	-	-	-	TIM1_ CH4N	USART3_ RX	-	-	QUADSPI_ BK1_IO3	-	FMC_D12	-	-	EVENT OUT

Pinouts and pin description

Table 13. Alternate function (continued)

		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
	Port	SYS_AF	LPTIM1/ TIM2/5/ 15/16/17	I2C3/ TIM1/2/3/4/5/8/1 5/20/ GPCOMP1	QUADSPI/ I2C3/4/SAI/ USB/ TIM8/15/20/GPC OMP3/ TSC	I2C1/2/3/ 4/ TIM1/8/1 6/17	QUADSPI/SP I1/2/3/4/I2S2/ 3/ I2C4/ UART4/5/TIM 8/ Infrared	QUADSPI/S PI2/3/ I2S2/3/ TIM1/5/8/20 / Infrared	USART1/2/3 /CAN/ GPCOMP5/6 /7	I2C3/4/ UART4/5/L PUART1/G PCOMP1/2/ 3/4/5/6/7	CAN/ TIM1/8/1 5/CAN1/2	QUADSPI/TI M2/3/4/8/17	LPTIM1/ TIM1/8/C AN1/3	SDIO/FMC/ LPUART1/ SAI TIM1	SAI/OPAMP2	UART4/5/ SAI/ TIM2/15/ UCPD	EVENT
	PF0	-	-	-	-	I2C2_ SDA	SPI2_NSS/I2 S2_WS	TIM1_CH3 N	-	-	-	-	-	-	-	-	EVENT OUT
	PF1	-	-	-	-	-	SPI2_SCK/I2 S2_CK	-	-	-	-	-	-	-	-	-	EVENT OUT
	PF2	-	-	TIM20_CH3	-	I2C2_ SMBA	-	-	-	-	-	-	-	FMC_A2	-	-	EVENT OUT
	PF3	-	-	TIM20_CH4	-	I2C3_ SCL	-	-	-	-	-	-	-	FMC_A3	-	-	EVENT OUT
	PF4	-	-	COMP1_OUT	TIM20_CH1N	I2C3_ SDA	-	-	-	-	-	-	-	FMC_A4	-	-	EVENT OUT
	PF5	-	-	TIM20_CH2N	-	-	-	-	-	-	-	-	-	FMC_A5	-	-	EVENT OUT
	PF6	-	TIM5_ETR	TIM4_CH4	SAI_SD_B	I2C2_ SCL	-	TIM5_CH1	USART3_ RTS	-	-	QUADSPI_ BK1_IO3	-	-	-	-	EVENT OUT
Щ	PF7	-	-	TIM20_BKIN	-	-	-	TIM5_CH2	-	-	-	QUADSPI_ BK1_IO2	-	FMC_A1	SAI_MCLK_B	-	EVENT OUT
To d	PF8	-	-	TIM20_BKIN2	-	-	-	TIM5_CH3	-	-	-	QUADSPI_ BK1_IO0	-	FMC_A24	SAI_SCK_B	-	EVENT OUT
	PF9	-	-	TIM20_BKIN	TIM15_CH1	-	SPI2_SCK	TIM5_CH4	-	-	-	QUADSPI_ BK1_IO1	-	FMC_A25	SAI_FS_B	-	EVENT OUT
	PF10	-	-	TIM20_BKIN2	TIM15_CH2	-	SPI2_SCK	-	-	-	-	QUADSPI_ CLK	-	FMC_A0	SAI_D3	-	EVENT OUT
	PF11	-	-	TIM20_ETR	-	-	-	-	-	-	-	-	-	FMC_NE4	-	-	EVENT OUT
	PF12	-	-	TIM20_CH1	-	-	-	-	-	-	-	-	-	FMC_A6	-	-	EVENT OUT
	PF13	-	-	TIM20_CH2	-	I2C4_ SMBA	-	-	-	-	-	-	-	FMC_A7	-	-	EVENT OUT
	PF14	-	-	TIM20_CH3	-	I2C4_ SCL	-	-	-	-	-	-	-	FMC_A8	-	-	EVENT OUT
	PF15	-	-	TIM20_CH4	-	I2C4_ SDA	-	-	-	-	-	-	-	FMC_A9	-	-	EVENT OUT



Table 13. Alternate function (continued)

							e is. Ait			(,						
		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
	Port	SYS_AF	LPTIM1/ TIM2/5/ 15/16/17	I2C3/ TIM1/2/3/4/5/8/1 5/20/ GPCOMP1	QUADSPI/ I2C3/4/SAI/ USB/ TIM8/15/20/GPC OMP3/ TSC	I2C1/2/3/ 4/ TIM1/8/1 6/17	QUADSPI/SP I1/2/3/4/I2S2/ 3/ I2C4/ UART4/5/TIM 8/ Infrared	QUADSPI/S PI2/3/ I2S2/3/ TIM1/5/8/20 / Infrared	USART1/2/3 /CAN/ GPCOMP5/6 /7	I2C3/4/ UART4/5/L PUART1/G PCOMP1/2/ 3/4/5/6/7	CAN/ TIM1/8/1 5/CAN1/2	QUADSPI/TI M2/3/4/8/17	LPTIM1/ TIM1/8/C AN1/3	SDIO/FMC/ LPUART1/ SAI TIM1	SAI/OPAMP2	UART4/5/ SAI/ TIM2/15/ UCPD	EVENT
	PG0	-	-	TIM20_CH1N	-	-	-	-	-	-	-	-	-	FMC_A10	-	-	EVENT OUT
	PG1	-	-	TIM20_CH2N	-	-	-	-	-	-	-	-	-	FMC_A11	-	-	EVENT OUT
	PG2	-	1	TIM20_CH3N	-	-	SPI1_SCK	1	-	-	-	-	ı	FMC_A12	-	1	EVENT OUT
	PG3	-	1	TIM20_BKIN	-	I2C4_ SCL	SPI1_MISO	TIM20_CH4 N	-	-	-	-	ı	FMC_A13	-	1	EVENT OUT
	PG4	-	-	TIM20_BKIN2	-	I2C4_ SDA	SPI1_MOSI	-	-	-	-	-	-	FMC_A14	-	-	EVENT OUT
Port G	PG5	-	-	TIM20_ETR	-	-	SPI1_NSS	-	-	LPUART1_ CTS	-	-	-	FMC_A15	-	-	EVENT OUT
	PG6	-	-	TIM20_BKIN	-	I2C3_ SMBA	-	-	-	LPUART1_ RTS_DE	-	-	-	FMC_INT	-	-	EVENT OUT
	PG7	-	-	-	SAI_CK1	I2C3_ SCL	-	-	-	LPUART1_ TX	-	-	-	FMC_INT	SAI_MCLK_A	-	EVENT OUT
	PG8	-	-	-	-	I2C3_ SDA	-	-	-	LPUART1_ RX	-	-	-	FMC_NE3	-	-	EVENT OUT
	PG9	-	-	-	-	-	-	SPI3_SCK	USART1_TX	-	-	-	-	FMC_NCE/FM C_NE2	-	TIM15_C H1N	EVENT OUT
	PG10	МСО	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-

5 Electrical characteristics

5.1 Parameter conditions

Unless otherwise specified, all voltages are referenced to V_{SS}.

5.1.1 Minimum and maximum values

Unless otherwise specified, the minimum and maximum values are guaranteed in the worst conditions of ambient temperature, supply voltage and frequencies by tests in production on 100% of the devices with an ambient temperature at $T_A = 25\,^{\circ}C$ and $T_A = T_A max$ (given by the selected temperature range).

Data based on characterization results, design simulation and/or technology characteristics are indicated in the table footnotes and are not tested in production. Based on characterization, the minimum and maximum values refer to sample tests and represent the mean value plus or minus three times the standard deviation (mean $\pm 3\sigma$).

5.1.2 Typical values

Unless otherwise specified, typical data are based on $T_A = 25$ °C, $V_{DD} = V_{DDA} = 3$ V. They are given only as design guidelines and are not tested.

Typical ADC accuracy values are determined by characterization of a batch of samples from a standard diffusion lot over the full temperature range, where 95% of the devices have an error less than or equal to the value indicated (mean $\pm 2\sigma$).

5.1.3 Typical curves

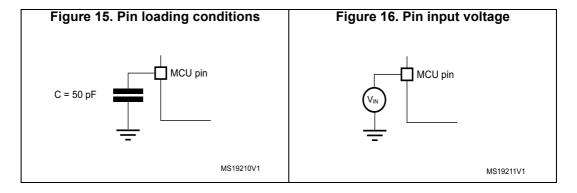
Unless otherwise specified, all typical curves are given only as design guidelines and are not tested.

5.1.4 Loading capacitor

The loading conditions used for pin parameter measurement are shown in *Figure 15*.

5.1.5 Pin input voltage

The input voltage measurement on a pin of the device is described in *Figure 16*.



5.1.6 Power supply scheme

VBAT Backup circuitry (LSE, RTC, 1.55 - 3.6 V Backup registers) V_{CORE} n x VDD Regulator V_{DDIO} OUT Level shifter Kernel logic Ю (CPU, Digital n x 100 nF GPIOs logic & Memories) +1 x 4.7 µF n x VSS VDDA Reset block V_{DDA} Temp. sensor PLL, HSI16, HSI48 VREF+ V_{REF} ADCs/ VRFF+ Standby circuitry DACs/ 10 nF (Wakeup logic, OPAMPs/ COMPs/ VREF-IWDG) VREFBUF VSSA MS60206V1

Figure 17. Power supply scheme

Caution:

Each power supply pair (V_{DD}/V_{SS} , V_{DDA}/V_{SSA} etc.) must be decoupled with filtering ceramic capacitors as shown above. These capacitors must be placed as close as possible to, or below, the appropriate pins on the underside of the PCB to ensure the good functionality of the device.

5.1.7 Current consumption measurement

Figure 18. Current consumption measurement

The I_{DD_ALL} parameters given in *Table 21* to *Table 25* represent the total MCU consumption including the current supplying V_{DD} , V_{DDA} and V_{BAT} .

5.2 Absolute maximum ratings

Stresses above the absolute maximum ratings listed in *Table 14: Voltage characteristics*, *Table 15: Current characteristics* and *Table 16: Thermal characteristics* may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability. Exposure to maximum rating conditions for extended periods may affect device reliability. Device mission profile (application conditions) is compliant with JEDEC JESD47 qualification standard, extended mission profiles are available on demand.

Symbol	Ratings	Min	Max	Unit
V _{DD} - V _{SS}	External main supply voltage (including V_{DD} , V_{DDA} , V_{BAT} and V_{REF+})	-0.3	4.0	
	Input voltage on FT_xxx pins except FT_c pins	V _{SS} -0.3	min (V _{DD} , V _{DDA}) + 4.0 ⁽³⁾⁽⁴⁾	V
V _{IN} ⁽²⁾	Input voltage on FT_c pins	V _{SS} -0.3	5.5	
	Input voltage on TT_xx pins	V _{SS} -0.3	4.0	
	Input voltage on any other pins	V _{SS} -0.3	4.0	
ΔV _{DDx}	Variations between different V _{DDX} power pins of the same domain	-	50	mV
V _{SSx} -V _{SS}	Variations between all the different ground pins ⁽⁵⁾	-	50	
V _{REF+} -V _{DDA}	Allowed voltage difference for V _{REF+} > V _{DDA}	-	0.4	V

Table 14. Voltage characteristics⁽¹⁾

All main power (V_{DD}, V_{DDA}, V_{BAT}) and ground (V_{SS}, V_{SSA}) pins must always be connected to the external power supply, in the permitted range.

- V_{IN} maximum must always be respected. Refer to Table 15: Current characteristics for the maximum allowed injected current values.
- 3. This formula has to be applied only on the power supplies related to the IO structure described in the pin definition table.
- 4. To sustain a voltage higher than 4 V the internal pull-up/pull-down resistors must be disabled.
- 5. Include VREF- pin.

Table 15. Current characteristics

Symbol	Ratings	Max	Unit
ΣIV_{DD}	Total current into sum of all V _{DD} power lines (source) ⁽¹⁾	150	
ΣIV _{SS}	Total current out of sum of all V _{SS} ground lines (sink) ⁽¹⁾	150	
IV _{DD(PIN)}	Maximum current into each V _{DD} power pin (source) ⁽¹⁾	100	
IV _{SS(PIN)}	Maximum current out of each V _{SS} ground pin (sink) ⁽¹⁾	100	
	Output current sunk by any I/O and control pin except FT_f	20	
I _{IO(PIN)}	Output current sunk by any FT_f pin	20	mA
	Output current sourced by any I/O and control pin	20	
71	Total output current sunk by sum of all I/Os and control pins ⁽²⁾	100	
ΣΙ _{ΙΟ(PIN)}	Total output current sourced by sum of all I/Os and control pins ⁽²⁾	100	
I _{INJ(PIN)} ⁽³⁾	Injected current on FT_xxx, TT_xx, NRST pins	-5/0 ⁽⁴⁾	
Σ I _{INJ(PIN)}	Total injected current (sum of all I/Os and control pins) ⁽⁵⁾	±25	

- All main power (V_{DD}, V_{DDA}, V_{BAT}) and ground (V_{SS}, V_{SSA}) pins must always be connected to the external power supplies, in the permitted range.
- This current consumption must be correctly distributed over all I/Os and control pins. The total output current must not be sunk/sourced between two consecutive power supply pins referring to high pin count LQFP packages.
- 3. Positive injection (when $V_{IN} > V_{DD}$) is not possible on these I/Os and does not occur for input voltages lower than the specified maximum value.
- A negative injection is induced by VIN < VSS. IINJ(PIN) must never be exceeded. Refer also to Table 14: Voltage characteristics for the minimum allowed input voltage values.
- When several inputs are submitted to a current injection, the maximum ∑|I_{INJ(PIN)}| is the absolute sum of the negative injected currents (instantaneous values).

Table 16. Thermal characteristics

Symbol	Ratings	Value	Unit
T _{STG}	Storage temperature range	-65 to +150	°C
T _J	Maximum junction temperature	150	°C



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5.3 Operating conditions

5.3.1 General operating conditions

Table 17. General operating conditions

Symbol	Parameter	Conditions	Min	Max	Unit
f _{HCLK}	Internal AHB clock frequency	-	0	170	
f _{PCLK1}	Internal APB1 clock frequency	-	0	170	MHz
f _{PCLK2}	Internal APB2 clock frequency	-	0	170	
V_{DD}	Standard operating voltage	-	1.71 ⁽¹⁾	3.6	V
		ADC or COMP used	1.62	3.6	
		DAC 1 MSPS or DAC 15 MSPS	1.71	3.0	
V_{DDA}	Analog supply voltage	OPAMP used	2.0	3.6	V
- DDA	, manag cappi, remage	VREFBUF used	2.4		
		ADC, DAC, OPAMP, COMP, VREFBUF not used	0	3.6	
V_{BAT}	Backup operating voltage	-	1.55	3.6	V
		TT_xx	-0.3	V _{DD} +0.3	
		FT_c	-0.3	5	
V _{IN}	I/O input voltage	All I/O except TT_xx and FT_c	-0.3	MIN(MIN(V _{DD} , V _{DDA})+3.6 V, 5.5 V) ⁽²⁾⁽³⁾	V
P _D	Power dissipation	See Section 6.11: Thermal character appropriate thermal resistance and Power dissipation is then calculate temperature (T _A) and maximum juselected thermal resistance.	d package. ed accordir	ng ambient	mW
	Ambient temperature for the	Maximum power dissipation	-40	85	
-	suffix 6 version	Low-power dissipation ⁽⁴⁾	-40	105	
T _A	Ambient temperature for the	Maximum power dissipation	-40	125	°C
	suffix 3 version	Low-power dissipation ⁽⁴⁾	-40	130	
T	lunction tomporature resea	Suffix 6 version	-40	105	- °C
T_J	Junction temperature range	Suffix 3 version	-40	130	

^{1.} When RESET is released functionality is guaranteed down to V_{BOR0} Min.

This formula has to be applied only on the power supplies related to the IO structure described by the pin definition table.
 Maximum I/O input voltage is the smallest value between MIN(V_{DD}, V_{DDA})+3.6 V and 5.5V.

^{3.} For operation with voltage higher than Min (V_{DD}, V_{DDA}) +0.3 V, the internal Pull-up and Pull-Down resistors must be disabled.

In low-power dissipation state, T_A can be extended to this range as long as T_J does not exceed T_{Jmax} (see Section 6.11: Thermal characteristics).

5.3.2 Operating conditions at power-up / power-down

The parameters given in *Table 18* are derived from tests performed under the ambient temperature condition summarized in *Table 17*.

Table 18. Operating conditions at power-up / power-down

Symbol	Parameter	Conditions	Min	Max	Unit
+	V _{DD} rise time rate		0	8	µs/V
t _{VDD}	V _{DD} fall time rate	-	10	8	μ5/ ν
4	V _{DDA} rise time rate		0	8	µs/V
[₹] VDDA	V _{DDA} fall time rate	-	10	8	μ5/ ν

5.3.3 Embedded reset and power control block characteristics

The parameters given in *Table 19* are derived from tests performed under the ambient temperature conditions summarized in *Table 17: General operating conditions*.

Table 19. Embedded reset and power control block characteristics

Symbol	Parameter	Conditions ⁽¹⁾	Min	Тур	Max	Unit
t _{RSTTEMPO} (2)	Reset temporization after BOR0 is detected	V _{DD} rising	-	250	400	μs
V _{BOR0} ⁽²⁾	Brown-out reset threshold 0	Rising edge	1.62	1.66	1.7	V
VBOR0`	Brown-out reset timeshold o	Falling edge	1.6	1.64	1.69	V
V	Brown-out reset threshold 1	Rising edge	2.06	2.1	2.14	V
V _{BOR1}	Brown-out reset threshold i	Falling edge	1.96	2	2.04	V
V	Brown-out reset threshold 2	Rising edge	2.26	2.31	2.35	V
V _{BOR2}	Brown-out reset tilleshold 2	Falling edge	2.16	2.20	2.24	V
V	Brown-out reset threshold 3	Rising edge	2.56	2.61	2.66	V
V _{BOR3}	Brown-out reset tilleshold 3	Falling edge	2.47	2.52	2.57	V
V	Brown-out reset threshold 4	Rising edge	2.85	2.90	2.95	V
V _{BOR4}	Brown-out reset tilleshold 4	Falling edge	2.76	2.81	2.86	V
V	Programmable voltage	Rising edge	2.1	2.15	2.19	V
V _{PVD0}	detector threshold 0	Falling edge	2	2.05	2.1	V
V	PVD threshold 1	Rising edge	2.26	2.31	2.36	V
V _{PVD1}	FVD tillesiloid i	Falling edge	2.15	2.20	2.25	V
V	DVD throshold 2	Rising edge	2.41	2.46	2.51	V
V _{PVD2}	PVD threshold 2	Falling edge	2.31	2.36	2.41	V
V	PVD threshold 3	Rising edge	2.56	2.61	2.66	V
V _{PVD3}	T VD tilleshold 3	Falling edge	2.47	2.52	2.57	V



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Table 19. Embedded reset and power control block characteristics (continued)

Symbol	Parameter	Conditions ⁽¹⁾	Min	Тур	Max	Unit
V	PVD threshold 4	Rising edge	2.69	2.74	2.79	V
V_{PVD4}	FVD tillesiloid 4	Falling edge	2.59	2.64	2.69	V
V	PVD threshold 5	Rising edge	2.85	2.91	2.96	V
V _{PVD5}	F VD tillesiloid 5	Falling edge	2.75	2.81	2.86	V
V	PVD threshold 6	Rising edge	2.92	2.98	3.04	V
V _{PVD6}	1 VD tillesiloid 0	Falling edge	2.84	2.90	2.96	V
t _{PVD_STUP}	PVD startup time	-	-	-	8	us_stup
V _{hyst_BORH0}	Hysteresis voltage of BORH0	Hysteresis in continuous mode	-	20	-	mV
7.2		Hysteresis in other mode	-	30	-	
V _{hyst_BOR_PVD}	Hysteresis voltage of BORH (except BORH0) and PVD	-	-	100	-	mV
I _{DD} (BOR_PVD) ⁽²⁾	BOR ⁽³⁾ (except BOR0) and PVD consumption from V _{DD}	-	-	1.1	1.6	μΑ
V	V _{DDA} peripheral voltage	Rising edge	1.61	1.65	1.69	V
V _{PVM1}	monitoring (COMP/ADC)	Falling edge	1.6	1.64	1.68	V
V	V _{DDA} peripheral voltage	Rising edge	1.78	1.82	1.86	V
V_{PVM2}	monitoring (OPAMP/DAC)	Falling edge	1.77	1.81	1.85	V
V _{hyst_PVM1}	PVM1 hysteresis	-	-	10	-	mV
V _{hyst_PVM2}	PVM2 hysteresis	-	-	10	-	mV
I _{DD} (PVM1/PVM2)	PVM1 and PVM2 consumption from V _{DD}	-	-	2	-	μΑ

Continuous mode means Run/Sleep modes, or temperature sensor enable in Low-power run/Low-power sleep modes.

^{2.} Guaranteed by design - Not tested in production.

^{3.} BOR0 is enabled in all modes (except shutdown) and its consumption is therefore included in the supply current characteristics tables.

5.3.4 Embedded voltage reference

The parameters given in *Table 20* are derived from tests performed under the ambient temperature and supply voltage conditions summarized in *Table 17: General operating conditions*.

Table 20. Embedded internal voltage reference

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V _{REFINT}	Internal reference voltage	-40 °C < T _A < +130 °C	1.182	1.212	1.232	V
t _{S_vrefint} (1)	ADC sampling time when reading the internal reference voltage	-	4 ⁽²⁾	-	-	μs
t _{start_vrefint}	Start time of reference voltage buffer when ADC is enable	-	-	8	12 ⁽²⁾	μs
I _{DD} (V _{REFINTBUF})	V _{REFINT} buffer consumption from V _{DD} when converted by ADC	-	-	12.5	20 ⁽²⁾	μΑ
ΔV_{REFINT}	Internal reference voltage spread over the temperature range	V _{DD} = 3 V	-	5	7.5 ⁽²⁾	mV
T _{Coeff}	Average temperature coefficient	-40°C < T _A < +130°C	-	30	50 ⁽²⁾	ppm/°C
A _{Coeff}	Long term stability	1000 hours, T = 25°C	-	300	1000 ⁽²⁾	ppm
V _{DDCoeff}	Average voltage coefficient	3.0 V < V _{DD} < 3.6 V	-	250	1200 ⁽²⁾	ppm/V
V _{REFINT_DIV1}	1/4 reference voltage		24	25	26	0/
V _{REFINT_DIV2}	1/2 reference voltage	-	49	50	51	% V _{REFINT}
V _{REFINT_DIV3}	3/4 reference voltage		74	75	76	IXLI IIVI

^{1.} The shortest sampling time is determined in the application by multiple iterations.

^{2.} Guaranteed by design - Not tested in production.

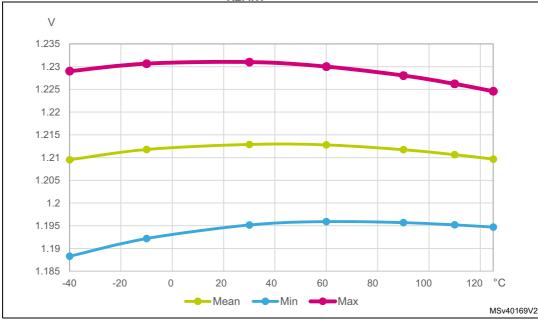


Figure 19. V_{REFINT} versus temperature

5.3.5 Supply current characteristics

The current consumption is a function of several parameters and factors such as the operating voltage, ambient temperature, I/O pin loading, device software configuration, operating frequencies, I/O pin switching rate, program location in memory and executed binary code

The current consumption is measured as described in *Figure 18: Current consumption measurement*.

Typical and maximum current consumption

The MCU is placed under the following conditions:

- All I/O pins are in analog input mode
- All peripherals are disabled except when explicitly mentioned
- The flash memory access time is adjusted with the minimum wait states number, depending on the f_{HCLK} frequency (refer to the table "number of wait states according to CPU clock (HCLK) frequency" available in the reference manual RM0440 "STM32G4 Series advanced Arm[®]-based 32-bit MCUs").
- When the peripherals are enabled f_{PCLK} = f_{HCLK}
- The voltage scaling Range 1 is adjusted to f_{HCLK} frequency as follows:
 - Voltage Range 1 Boost mode for 150 MHz < f_{HCLK} ≤ 170 MHz
 - Voltage Range 1 Normal mode for 26 MHz < f_{HCLK} ≤ 150 MHz

The parameters given in *Table 21* to *Table 25* are derived from tests performed under ambient temperature and supply voltage conditions summarized in *Table 17: General operating conditions*.



Table 21. Current consumption in Run and Low-power run modes, code with data processing running from flash in single Bank, ART enable (Cache ON Prefetch OFF)

		Condition	n				Тур					Max			
Symbol	Parameter	-	Voltage scaling	f _{HCLK}	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	Unit
				26 MHz	3.65	3.85	4.45	5.1	6.45	4.40	6.60	11.0	16.0	22.0	
				16 MHz	2.30	2.55	3.1	3.8	5.15	3.00	5.00	9.00	14.9	21.0	
				8 MHz	1.25	1.50	2.05	2.8	4.1	2.00	3.6	7.70	13.0	19.0	
			Range 2	4 MHz	0.75	0.955	1.5	2.3	3.6	1.40	3.00	7.00	12.0	19.0	
				2 MHz	0.47	0.69	1.25	2	3.35	0.990	2.60	6.70	12.0	19.0	
				1 MHz	0.34	0.55	1.1	1.9	3.2	0.830	2.50	6.50	12.0	18.0	
				100 KHz	0.22	0.43	0.98	1.75	3.1	0.690	2.30	6.30	11.0	18.0	
IDD (Run)	Supply current	f _{HCLK} = f _{HSE} up to 48 MHz included, bypass mode PLL ON above 48	Range 1 Boost mode	170 MHz	29.50	29.5	31	32	34.5	31.0	35.0	42.0	48.0	56.0	mA
(*,	in Run mode	MHz all		150 MHz	24.50	26	27	28	30	26.0	28.0	34.0	44.0	47.0	
		peripherals disable		120 MHz	19.50	20	20.5	21.5	23.5	21.0	23.0	32.0	38.0	43.0	
				80 MHz	13.00	13.5	14	15.5	17	15.0	17.0	25.0	30.0	37.0	
				72 MHz	12.00	12	13	14	15.5	13.0	16.0	23.0	29.0	36.0	
			Range 1	64 MHz	10.50	11	11.5	12.5	14.5	12.0	14.0	21.0	27.0	34.0	
				48 MHz	7.90	8.2	9	9.7	11.5	9.10	13.0	19.0	25.0	32.0	
				32 MHz	5.40	5.65	6.4	7.2	8.85	6.50	9.60	15.0	21.0	29.0	
				24 MHz	4.10	4.35	5.1	5.95	7.6	5.20	8.00	14.0	20.0	28.0	
				16 MHz	2.80	3.1	3.8	4.7	6.3	4.30	6.40	12.0	18.0	26.0	

Electrical characteristics

Table 21. Current consumption in Run and Low-power run modes, code with data processing running from flash in single Bank, ART enable (Cache ON Prefetch OFF) (continued)

		Condition	1				Тур					Max			
Symbol	Parameter	-	Voltage scaling	f _{HCLK}	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	Unit
				2 MHz	455	725	1350	2250	3800	1200	3200	8100	14000	22000	
		SYSCLK source is	HSE	1 MHz	280	545	1200	2100	3600	1100	3000	7900	14000	22000	
		in bypass mode all peripherals disa	ble	250 KHz	160	435	1100	2000	3500	840	2800	7700	14000	22000	
IDD (LPRun)	Supply current			62.5 KHz	130	405	1050	1950	3500	810	2700	7600	14000	22000	
IDD (LFKull)	in Low-power run mode			2 MHz	920	1200	1850	2750	4250	1900	3800	8700	15000	22000	μA
		SYSCLK source is	HSI16	1 MHz	780	1100	1700	2650	4150	1700	3700	8600	14000	22000	
		all peripherals disa	ble	250 KHz	725	980	1600	2500	4050	1600	3600	8400	14000	22000	
				62.5 KHz	720	955	1600	2500	4000	1500	3500	8400	14000	22000	



Table 22. Current consumption in Run and Low-power run modes, code with data processing running from flash in dual bank, ART enable (Cache ON Prefetch OFF)

		Condi	tions				Тур					Max ⁽¹⁾			
Symbol	Parameter	-	Voltage scaling	fHCLK	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	Unit
				26 MHz	3.70	3.9	4.45	5.15	6.45	4.40	6.60	11.0	16.0	22.0	
				16 MHz	2.35	2.55	3.1	3.85	5.15	3.00	5.00	9.00	14.0	21.0	
				8 MHz	1.25	1.5	2.05	2.8	4.15	2.00	3.60	7.70	13.0	19.0	
			Range 2	4 MHz	0.75	0.97	1.5	2.3	3.6	1.40	3.00	7.00	12.0	19.0	
	f _H			2 MHz	0.47	0.7	1.25	2.05	3.35	0.990	2.60	6.70	12.0	19.0	
				1 MHz	0.34	0.56	1.1	1.9	3.2	0.830	2.50	6.50	12.0	18.0	
		f _{HCLK} = f _{HSE} up to 48MHz		100 KHz	0.22	0.44	0.975	1.8	3.1	0.690	2.30	6.30	11.0	18.0	
IDD	Supply	included, bypass mode	Range 1 Boost mode	170 MHz	29.50	30	31	32	34.5	31.0	35.0	42.0	48.0	56.0	
(Run)	current in Run mode	PLL ON above 48		150 MHz	24.50	24.5	25.5	26.5	28.5	26.0	28.0	34.0	44.0	47.0	mA
		MHz all		120 MHz	19.50	20	20.5	22	23.5	21.0	23.0	32.0	38.0	43.0	
		peripherals disable		80 MHz	13.00	13.5	14.5	15.5	17	15.0	17.0	25.0	30.0	37.0	
				72 MHz	12.00	12.5	13	14	15.5	13.0	16.0	23.0	29.0	36.0	
			Range 1	64 MHz	10.50	11	11.5	13	14.5	12.0	14.0	21.0	27.0	34.0	
				48 MHz	7.95	8.3	9	10	11.5	9.10	13.0	19.0	25.0	32.0	
				32 MHz	5.40	5.7	6.45	7.25	8.9	6.50	9.60	15.0	21.0	29.0	
				24 MHz	4.10	4.4	5.1	6	7.65	5.20	8.00	14.0	20.0	28.0	
				16 MHz	2.85	3.15	3.8	4.75	6.35	4.30	6.40	12.0	18.0	26.0	

Table 22. Current consumption in Run and Low-power run modes, code with data processing running from flash in dual bank, ART enable (Cache ON Prefetch OFF) (continued)

		Condi	tions				Тур					Max ⁽¹⁾			
Symbol	Parameter	-	Voltage scaling	fHCLK	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	Unit
				2 MHz	450	725	1350	2250	3800	1200	3200	8100	14000	22000	
		SYSCLK sourd in bypass mod		1 MHz	270	575	1200	2150	3650	1100	3000	7900	14000	22000	
	Supply	all peripherals		250 KHz	185	460	1050	2000	3550	840	2800	7700	14000	22000	
IDD	current in			62.5 KHz	130	430	1050	2000	3500	810	2700	7600	14000	22000	
(LPRun)	Low-power run mode			2 MHz	970	1200	1850	2750	4300	1900	3800	8700	15000	22000	μA
	Turi mode	SYSCLK source	ce is HSI16	1 MHz	800	1100	1700	2650	4150	1700	3700	8600	14000	22000	
		all peripherals	disable	250 KHz	680	990	1600	2550	4050	1600	3600	8400	14000	22000	
				62.5 KHz	695	965	1600	2500	4050	1500	3500	8400	14000	22000	

^{1.} Evaluated by characterization - Not tested in production, unless otherwise specified.





Table 23. Current consumption in Run and Low-power run modes, code with data processing running from SRAM1

		Condi	itions				Тур					Max ⁽¹⁾			
Symbol	Parameter	-	Voltage scaling	fHCLK	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	Unit
				26 MHz	3.35	3.55	4.1	4.95	6.45	4.00	6.20	11.0	15.0	22.0	
				16 MHz	2.15	2.35	2.9	3.7	5.25	3.10	4.70	8.70	14.0	20.0	
				8 MHz	1.15	1.35	1.9	2.7	4.2	1.90	3.50	7.50	13.0	19.0	
			Range 2	4 MHz	0.69	0.855	1.4	2.2	3.7	1.30	2.90	6.90	12.0	19.0	
				2 MHz	0.43	0.595	1.15	1.95	3.45	0.960	2.60	6.60	12.0	18.0	
				1 MHz	0.30	0.47	1	1.8	3.3	0.810	2.40	6.40	12.0	18.0	
		f _{HCLK} = f _{HSE}		100 KHz	0.19	0.355	0.89	1.7	3.2	0.680	2.30	6.30	11.0	18.0	
	Supply	up to 48MHz included, bypass mode	Range 1 Boost mode	170 MHz	26.00	26.5	27.5	28.5	30.5	28.0	32.0	39.0	45.0	53.0 ⁽²⁾	
IDD(Run)	current in Run mode	PLL ON above 48 MHz all		150 MHz	21.50	22	22.5	23.5	25.5	23.0	25.0	31.0	41.0	46.0 ⁽²⁾	mA
		peripherals		120 MHz	17.50	17.5	18.5	19.5	21.5	19.0	21.0	30.0	36.0	41.0	
		disable		80 MHz	11.50	12	12.5	13.5	15.5	13.0	15.0	23.0	29.0	35.0	
				72 MHz	10.50	11	11.5	12.5	14.5	12.0	14.0	21.0	27.0	34.0	
			Range 1	64 MHz	9.45	9.7	10.5	11.5	13.5	11.0	13.0	20.0	26.0	33.0	
				48 MHz	7.25	7.5	8.2	9.25	11	8.10	12.0	17.0	23.0	31.0	
				32 MHz	4.90	5.15	5.85	6.9	8.7	6.00	8.90	15.0	21.0	29.0	
				24 MHz	3.75	4	4.7	5.7	7.5	4.80	7.50	13.0	19.0	27.0	
				16 MHz	2.60	2.85	3.5	4.5	6.3	4.00	6.10	12.0	18.0	26.0	

Table 23. Current consumption in Run and Low-power run modes, code with data processing running from SRAM1 (continued)

		Condi	tions				Тур					Max ⁽¹⁾			
Symbol	Parameter	-	Voltage scaling	fHCLK	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	Unit
				2 MHz	365	570	1200	2150	3850	1200	3100	7900	14000	22000	
		SYSCLK sour		1 MHz	240	425	1050	2000	3650	960	2900	7700	14000	22000	
	Cumply	in bypass mod all peripherals		250 KHz	135	315	945	1850	3550	840	2800	7600	13000	22000	
IDD	Supply current in			62.5 KHz	105	285	915	1850	3550	780	2700	7600	13000	22000	uA
(LPRun)	Low-power run mode			2 MHz	835	1050	1650	2600	4300	1800	3700	8600	14000	22000	μΑ
	Turrinoue	SYSCLK sour	ce is HSI16	1 MHz	775	940	1550	2500	4150	1700	3600	8500	14000	22000	
		all peripherals	disable	250 KHz	640	860	1450	2400	4100	1500	3500	8400	14000	22000	
				62.5 KHz	640	830	1450	2350	4050	1600	3500	8400	14000	22000	

^{1.} Evaluated by characterization - Not tested in production, unless otherwise specified.



^{2.} Guaranteed by test in production.



Table 24. Typical current consumption in Run and Low-power run modes, with different codes running from flash memory, ART enable (Cache ON Prefetch OFF)

Symbol	Parameter	Condi	tions Voltage scaling	Code	Typ Single Bank Mode 25°C	Typ Dual Bank Mode 25°C	Unit	Typ Single Bank Mode 25°C	Typ Dual Bank Mode 25°C	Unit
				Reduced code ⁽¹⁾	3.65	3.7		140	142	
			Range2	Coremark	3.65	3.7	-	140	142	
			f _{HCLK} =26MHz	Dhrystone2.1	3.65	3.7	mA	140	142	μΑ/MHz
				Fibonacci	4.55	4.2		175	162	
				While(1)	2.90	3		112	115	
		f _{HCLK} =f _{HSE} up to 48 MHZ		Reduced code ⁽¹⁾	24.5	24.5		163	163	
IDD	Supply	included, bypass	Range 1	Coremark	24	24		160	160	
(Run)	current in Run mode	mode PLL ON above 48 MHz all	f _{HCLK} = 150 MHz	Dhrystone2.1	24.5	24.5	mA	163	163	µA/MHz
		peripherals disable		Fibonacci	22.5	28		150	187	
		disable		While(1)	19.5	20		130	133	
				Reduced code ⁽¹⁾	29.5	29.5		174	174	
			Range 1	Coremark	29	29		171	171	
			Boost mode f _{HCLK} = 170 MHz	Dhrystone2.1	29.5	29.5	mA	174	174	µA/MHz
			HOLK	Fibonacci	38	35		224	206	
				While(1)	23.5	24		138	141	

Table 24. Typical current consumption in Run and Low-power run modes, with different codes running from flash memory, ART enable (Cache ON Prefetch OFF) (continued)

Symbol	Parameter	Condi	tions	Code	Typ Single Bank Mode	Typ Dual Bank Mode	Unit	Typ Single Bank Mode	Typ Dual Bank Mode	Unit
		-	Voltage scaling		25°C	25°C		25°C	25°C	
	DD current in			Reduced code ⁽¹⁾	920	970		460	485	
loo		SYSCLK source is	HSI16	Coremark	905	985		453	493	
(LPRun)		f _{HCLK} = 2 MHz all peripherals disa	ble	Dhrystone2.1	915	915	μA	458	458	μΑ/MHz
				Fibonacci	1,050	950		525	475	
				While(1)	930	875		465	438	

^{1.} Reduced code used for characterization results provided in *Table 21*, *Table 23*.



Table 25. Typical current consumption in Run and Low-power run modes, with different codes running from SRAM1

		Conditions			Тур		Тур	
Symbol	Parameter	-	Voltage scaling	Code	25°C	Unit	25°C	Unit
				Reduced code ⁽¹⁾	3.25		125	
			Range2	Coremark	3.35		129	
			f _{HCLK} =26 M	Dhrystone2.1	3.30	mA	127	μA/MHz
			Hz	Fibonacci	3.30		127	
				While(1)	3.40		131	
				Reduced code ⁽¹⁾	21.50		143	
		$f_{HCLK} = f_{HSE}$ up to 48 MHZ	Range 1	Coremark	22.50		150	
IDD (Run)	Supply current in Run mode	included, bypass mode PLL ON above 48 MHz all	f _{HCLK} = 150	Dhrystone2.1	21.50	mA	143	μA/MHz
		peripherals disable	MHz	Fibonacci	22.50		150	
				While(1)	20.00		133	
				Reduced code ⁽¹⁾	26.00		153	
			Range 1	Coremark	27.00		159	
			Boost mode f _{HCLK} =	Dhrystone2.1	26.00	mA	153	μA/MHz
			170 MHz	Fibonacci	27.50		162	
				While(1)	24.50		144	
				Reduced code ⁽¹⁾	955		478	
IDD				Coremark	890		445	
IDD (LPRun)	Supply current in Low-power run	f _{HCLK} = f _{HSE} = 2 MHz all peripherals disable		Dhrystone2.1	915	μΑ	458	μA/MHz
(12)				Fibonacci	880		440	
				While(1)	905		453	

^{1.} Reduced code used for characterization results provided in *Table 21*, *Table 23*.

Table 26. Typical current consumption in Run and Low-power run modes, with different codes running from SRAM2

		Conditions			Тур		Тур	
Symbol	Parameter	-	Voltage scaling	f _{HCLK}	Single bank mode	Unit	Single bank mode	Unit
				Reduced code ⁽¹⁾	2.65		102	
			Range2	Coremark	2.80		108	
			f _{HCLK} =26 M	Dhrystone2.1	2.65	mA	102	μΑ/MHz
			Hz	Fibonacci	2.60		100	
				While(1)	2.45		94	
				Reduced code ⁽¹⁾	17.50		117	
	Cupply ourront in	f _{HCLK} = f _{HSE} up to 48 MHZ	Range 1	Coremark	18.00		120	
IDD (Run)	Supply current in Run mode	included, bypass mode PLL ON above 48 MHz all	f _{HCLK} = 150	Dhrystone2.1	17.50	mA	117	μA/MHz
		peripherals disable	MHz	Fibonacci	17.00		113	
				While(1)	16		107	
				Reduced code ⁽¹⁾	21.00		124	
			Range 1 Boost mode	Coremark	22.00		129	
			f _{HCLK} =	Dhrystone2.1	21.00	mA	124	μA/MHz
			170 MHz	Fibonacci	20.50		121	
				While(1)	19.50		115	
				Reduced code ⁽¹⁾	890		445	
IDD	Cupply ourrort :-	SYSCLK source is HSI16		Coremark	830		415	
(LPRun)	Supply current in Low-power run	F _{HCLK} = 2MHz		Dhrystone2.1	825	μΑ	413	μΑ/MHz
	·	all peripherals disable		Fibonacci	830		415	
				While(1)	815		408	

^{1.} Reduced code used for characterization results provided in *Table 21*, *Table 23*.





Table 27. Typical current consumption in Run and Low-power run modes, with different codes running from CCMSRAM

		Conditions			Тур		Тур	
Symbol	Parameter	-	Voltage scaling	f _{HCLK}	Single bank mode	Unit	Single bank mode	Unit
				Reduced code ⁽¹⁾	2.75		106	
			Range2	Coremark	2.85		110	
			f _{HCLK} =26 M	Dhrystone2.1	2.75	mA	106	μA/MHz
			Hz	Fibonacci	2.95		113	
				While(1)	2.60		100	
				Reduced code ⁽¹⁾	18.00		120	
		$f_{HCLK} = f_{HSE}$ up to 48 MHZ	Range 1	Coremark	18.50		123	
IDD (Run)	Supply current in Run mode	included, bypass mode PLL ON above 48 MHz all	f _{HCLK} = 150	Dhrystone2.1	18.00	mA	120	μΑ/MHz
		peripherals disable	MHz	Fibonacci	19.00		127	
		PLL ON above 48 MHz all		While(1)	17.00		113	
				Reduced code ⁽¹⁾	22.00		129	
			Range 1	Coremark	22.50		132	
			Boost mode f _{HCLK} =	Dhrystone2.1	22.00	mA	129	μΑ/MHz
			170 MHz	Fibonacci	23.50		138	
				While(1)	20.50		121	
			•	Reduced code ⁽¹⁾	900		450	
	Supply current in Low-power run	SYSCLK source is HSI16		Coremark	850		425	
IDD (LPRun)		F _{HCLK} = 2MHz		Dhrystone2.1	870	μΑ	435	μΑ/MHz
(=. 1 (011)		all peripherals disable		Fibonacci	850		425	
				While(1)	810		405	

1. Reduced code used for characterization results provided in *Table 21*, *Table 23*.

Table 28. Current consumption in Sleep and Low-power sleep mode flash memory ON

		Condition					Тур					Max			
Symbol	Parameter	-	Voltage scaling	f _{HCLK}	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	Unit
				26 MHz	0.98	1.1	1.75	2.4	3.75	1.90	3.50	7.60	13.0	19.0	
				16 MHz	0.67	0.835	1.45	2.15	3.5	1.50	3.00	7.10	12.0	19.0	
				8 MHz	0.44	0.605	1.25	2	3.35	1.10	2.70	6.70	12.0	19.0	
			Range 2	4 MHz	0.33	0.5	1.1	1.9	3.25	0.860	2.50	6.50	12.0	18.0	
				2 MHz	0.27	0.445	1.05	1.85	3.2	0.760	2.40	6.40	11.0	18.0	
				1 MHz	0.24	0.415	1.05	1.8	3.15	0.720	2.30	6.40	11.0	18.0	
	f _{HCLK} = f _{HSE}		100 KHz	0.21	0.385	0.995	1.8	3.1	0.670	2.30	6.30	11.0	18.0		
IDD (Sleep)	Supply current	f _{HCLK} = f _{HSE} up to 48 MHz included, bypass	Range 1 Boost mode	170 MHz	6.60	6.95	7.8	8.9	10.5	8.00	12.0	18.0	24.0	33.0	mA
івв (оксер)	in sleep mode	up to 48 MHz included, bypass		150 MHz	5.50	5.8	6.55	7.55	9.25	6.40	9.50	15.0	21.0	29.0	
		peripherals disable		120 MHz	4.50	4.75	5.5	6.55	8.2	5.40	8.20	14.0	20.0	28.0	
				80 MHz	3.15	3.45	4.2	5.15	6.8	4.50	6.60	12.0	18.0	26.0	
				72 MHz	2.85	3.15	3.9	4.9	6.55	4.20	6.30	12.0	18.0	26.0	
			Range 1	64 MHz	2.60	2.9	3.65	4.6	6.3	3.50	6.00	12.0	18.0	26.0	
				48 MHz	1.90	2.2	3	3.65	5.3	3.20	5.30	11.0	17.0	25.0	
				32 MHz	1.40	1.65	2.4	3.2	4.85	2.70	4.80	11.0	17.0	25.0	
				24 MHz	1.10	1.35	2.1	3	4.65	2.30	4.50	9.80	16.0	25.0	
				16 MHz	0.83	1.1	1.85	2.75	4.35	1.90	4.10	9.40	16.0	24.0	

Table 28. Current consumption in Sleep and Low-power sleep mode flash memory ON (continued)

		Condition					Тур					Max			
Symbol	Parameter	-	Voltage scaling	f _{HCLK}	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	Unit
				2 MHz	205	430	1150	2050	3600	1600	2900	7800	14000	22000	
		SYSCLK source is I in bypass mode	HSE	1 MHz	165	400	1100	2000	3550	1100	2900	7700	14000	22000	μA
		all peripherals disab	le	250 KHz	145	370	1100	2000	3550	820	2800	7700	13000	22000	μΑ
IDD	Supply current in Low-power			62.5 KHz	140	365	1050	2000	3550	810	2800	7700	13000	22000	
(LPSleep)	sleep mode			2 MHz	700	925	1650	2550	4100	1600	3600	8400	14000	22000	
	sleep mode SYS	SYSCLK source is I	HSI16	1 MHz	710	925	1600	2550	4100	1600	3600	8400	14000	22000	uА
		all peripherals disab	le	250 KHz	670	910	1600	2500	4050	1600	3600	8400	14000	22000	μΑ
				62.5 KHz	685	910	1600	2500	4050	1600	3600	8400	14000	22000	

Table 29. Current consumption in low-power sleep modes, flash memory in power-down

		Condi	tion		-	<u>-</u>	Тур		<u> </u>			Max			
Symbol	Parameter	-	Voltage scaling	f _{HCLK}	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	Unit
				2 MHz	210	385	1150	2050	3550	910	2900	7800	14000	22000	
			YSCLK source is HSE bypass mode	1 MHz	150	360	1100	2000	3550	860	2900	7700	14000	22000	
		7.	i periprierais disable	250 KHz	120	330	1050	2000	3500	820	2700	7600	13000	21000	
IDD	Supply current		oypass mode peripherals disable	62.5 KHz	110	330	1050	1950	3500	810	2700	7600	13000	21000	
(LPSleep)	in low-power sleep mode			2 MHz	675	900	1600	2500	4050	1600	3600	8500	14000	22000	μA
		SYSCLK source	1	1 MHz	695	890	1600	2500	4050	1600	3600	8400	14000	22000	
		all peripherals	SCLK Source is HSI IO	250 KHz	640	885	1600	2500	4050	1600	3600	8500	14000	22000	
				62.5 KHz	690	880	1600	2500	4050	1400	3000	7000	12000	19000	

Table 30. Current consumption in Stop 1 mode

		Conditions				Тур					Max ⁽¹⁾			
Symbol	Parameter	-	V DD	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	Unit
	Supply current		1.8 V	80	250	830	1550	2850	630	2100	5900	11000	18000	
IDD	in Stop 1	RTC disabled	2.4 V	80	250	835	1600	2850	640	2100	5900	11000	18000	
(Stop 1)	mode, RTC disabled	Tro diddblod	3.0 V	80.5	255	840	1600	2900	640	2200	6000	11000	18000	
	aloabioa		3.6 V	81.5	255	845	1600	2900	640	2200	6000	11000	18000	
			1.8 V	80.5	255	830	1550	2850	640	2100	5900	11000	18000	
		RTC clocked by LSI	2.4 V	81	255	835	1600	2850	640	2200	5900	11000	18000	
		t RTC clocked by LSE	3.0 V	81.5	255	835	1600	2850	640	2200	6000	11000	18000	
			3.6 V	82	255	845	1600	2900	650	2200	6000	11000	18000	μΑ
IDD	Supply current		1.8 V	80	255	830	1550	2850	-	-	-	-	ı	μ.
(Stop 1	in Stop 1		2.4 V	80.5	255	830	1600	2850	-	-	-	-	ı	
with RTC)	mode, RTC enabled	bypassed at 32768 Hz	3.0 V	81.5	255	835	1600	2900	-	-	-	-	-	
10)	Chablea		3.6 V	83	260	845	1600	2900	-	-	-	-	-	
			1.8 V	83.5	220	655	1300	1	-	-	-	-	-	
		RTC clocked by LSE guartz in low drive mode	2.4 V	84	220	660	1300	-	-	-	-	-	-	
		at 32768 Hz	3.0 V	84.5	220	660	1300	-	-	-	-	-	-	
			3.6 V	87	220	660	1300	-	-	-	-	-	-	
IDD	Supply current	Wakeup clock is HSI6, voltage Range 1	3.0 V	1.73	-	-	-	-	-	-	-	-	-	
(wakeu p from Stop 1	during wakeup from Stop 1 mode	Wakeup clock is HSI6 = 4 MHz, (HPRE = 4), voltage Range 2	3.0 V	1.29	-	-	-	-	-	-	-	-	-	mA

^{1.} Evaluated by characterization - Not tested in production, unless otherwise specified.





Table 31. Current consumption in Stop 0 mode

Symbol	Doromotor	Condit	ions			Тур					Max ⁽¹⁾			l leit
Symbol	Parameter -	-	V DD	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	Unit
			1.8 V	190	380	980	1750	3100	790	2400	6500	11000	19000	
IDD(Stop 0)	Supply current in Stop 0 mode,		2.4 V	190	380	985	1750	3100	790	2400	6400	11000	19000	
(Stop 0)	RTC disabled	-	3 V	190	380	985	1750	3100	800	2400	6500	12000	19000	μΑ
			3.6 V	190	380	985	1750	3100	800	2500	6500	12000	12000	

^{1.} Evaluated by characterization - Not tested in production, unless otherwise specified.

Table 32. Current consumption in Standby mode

Symbol	Parameter	Conditio	ns			Тур					Max	(1)		Unit
Symbol	Parameter	-	VDD	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	Unit
			1.8 V	100	275	1350	3450	8450	200	1100	4100	9700	27000	
		No independent	2.4 V	110	325	1600	4100	10000	220	1200	4800	12000	31000	
	Complete surrent in Ctandles	watchdog	3 V	130	385	1900	4850	12000	240	1400	5500	13000	35000	
IDD	Supply current in Standby mode (backup registers		3.6 V	180	530	2400	6050	14500	360	1700	6300	15000	40000	^
(Standby)	retained), RTC disabled		1.8 V	300	-	-	-	-	-	-	-	-	-	- nA
	KTC disabled	With	2.4 V	365	-	-	-	-	-	-	-	-	-	
		independent watchdog	3 V	435	-	-	-	-	-	-	-	-	-	
			3.6 V	545	-	-	-	-	-	-	-	-	-	

Electrical characteristics

Table 32. Current consumption in Standby mode (continued)

Combal	Down water	Conditio	ns			Тур					Max ⁽	1)		11:4
Symbol	Parameter	-	V DD	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	Unit
		DTO als also d	1.8 V	540	725	1800	3850	8850	660	1500	4600	11000	27000	
		RTC clocked by LSI, no	2.4 V	700	920	2150	4650	10500	860	1900	5300	12000	31000	
		independent watchdog	3 V	885	1150	2650	5550	12500	1100	2200	6300	14000	36000	
			3.6 V	1100	1450	3350	7000	15500	1400	2700	7400	16000	41000	nA
		RTC clocked	1.8 V	580	-	-	-	-	-	-	-	-	-	
		by LSI, with	2.4 V	760	-	-	-	-	-	-	-	-	-	
	Supply current in Standby	independent watchdog	3 V	960	-	-	-	-	-	-	-	-	-	
IDD (Standby with	mode (backup registers	g	3.6 V	1200	-	-	-	-	-	-	-	-	-	
RTC)	retained), RTC enabled	RTC clocked	1.8 V	410	580	1600	3650	8600	-	-	-	-	-	
	Tri o chablod	by LSE	2.4 V	545	750	1950	4450	10500	-	-	-	-	-	
		bypassed at 32768 Hz	3 V	830	1150	2750	5800	13000	-	-	-	-	-	
			3.6 V	2200	3050	5550	9550	18000	-	-	-	-	-	nA
		RTC clocked	1.8 V	370	570	1350	3150	7100	-	-	-	-	-	
		by LSE guartz ⁽²⁾ in	2.4 V	495	715	1650	3800	8350	-	-	-	-	-	
		low drive	3 V	655	915	2100	4550	9850	-	-	-	-	-	
		mode	3.6 V	875	1350	2800	5750	12000	-	-	-	-	-	
			1.8 V	300	825	2950	6300	12550	-	-	-	-	-	
IDD (2)	Supply current to be added in Standby mode when SRAM2	_	2.4 V	305	875	2900	6400	12500	-	-	-	-	-	nA
(SRAM2) ⁽³⁾	is retained		3 V	305	865	2950	6150	12500	-	-	-	-	-	""
			3.6 V	310	870	3000	6450	13000	-	-	-	-	-	





Table 32. Current consumption in Standby mode (continued)

Symbol	Parameter	Conditio	ns			Тур					Max ⁽	1)		Unit
Symbol	Farameter	-	V DD	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	Ullit
	Supply current during wakeup from Standby mode	Wakeup clock is HSI16 = 16 MHz ⁽⁴⁾	3 V	2.46	-	-	-	-	-	-	-	-	-	mA

- 1. Evaluated by characterization Not tested in production, unless otherwise specified.
- 2. Based on characterization done with a 32.768 kHz crystal (MC306-G-06Q-32.768, manufacturer JFVNY) with two 6.8 pF loading capacitors.
- 3. The supply current in Standby with SRAM2 mode is: IDD_ALL(Standby) + IDD_ALL(SRAM2). The supply current in Standby with RTC with SRAM2 mode is: IIDD_ALL(Standby + RTC) + IDD_ALL(SRAM2).
- 4. Wakeup with code execution from flash memory. Average value given for a typical wakeup time as specified in Table 36: Low-power mode wakeup timings.

Table 33. Current consumption in Shutdown mode

Cumbal	Davamatav	Conditi	ons			Тур					Max ⁽¹⁾			Unit
Symbol	Parameter	-	VDD	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	Unit
	Supply current		1.8 V	19	140	885	2500	6600	78.0	490	3100	8100	24000	
IDD	in Shutdown mode (backup		2.4 V	28	180	1050	2950	7800	94.0	570	3600	9300	27000	nA
(Shutdown)	registers retained) RTC	-	3 V	43	230	1300	3600	9300	130	680	4100	11000	31000	IIA
	disabled		3.6 V	87	360	1750	4700	12000	190	870	4900	13000	35000	

Table 33. Current consumption in Shutdown mode (continued)

Symbol	Parameter	Condition	ons			Тур					Max ⁽¹⁾			Unit
Symbol	Farameter	-	V DD	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	Oiiit
		RTC	1.8 V	330	445	1150	2700	6800	-	ı	-	-	ı	
		clocked by LSE	2.4 V	460	605	1450	3350	8150	-	i	-	-	i	
	Supply current	bypassed at 32768	3 V	745	1000	2200	4550	10500	-	-	-	-	-	
IDD	in Shutdown mode (backup	Hz	3.6 V	2100	2850	4900	8150	15500	-	-	-	-	-	nA
(Shutdown with RTC)	registers retained) RTC	RTC	1.8 V	285	450	1050	2500	-	-	-	-	-	-	IIA
	enabled	clocked by LSE	2.4 V	410	585	1300	3050	-	-	-	-	-	-	
		quartz ⁽²⁾ in low drive	3 V	565	770	1750	3750	-	-	-	-	-	-	
		mode	3.6 V	780	1200	2400	4850	-	-	-	-	-	-	
IDD(wakeup from Shutdown)	Supply current during wakeup from Shutdown mode	Wakeup clock is HSI16 = 16 MHz ⁽³⁾	3 V	1.6	1	-	-	-	-	-	-	-	ı	mA

^{1.} Evaluated by characterization - Not tested in production, unless otherwise specified.



^{2.} Based on characterization done with a 32.768 kHz crystal (MC306-G-06Q-32.768, manufacturer JFVNY) with two 6.8 pF loading capacitors.

^{3.} Wakeup with code execution from flash memory. Average value given for a typical wakeup time as specified in Table 36: Low-power mode wakeup timings.



Table 34. Current consumption in VRAT mode

Symbol	Parameter	Condition	ons			Тур					Max ⁽¹⁾			Unit
Зушьог	Parameter	-	V_{BAT}	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	Unit
			1.8 V	4	17	92	245	600	-	-	-	-	-	
		RTC	2.4 V	5	20	105	280	690	-	-	-	-	-	
		disabled	3 V	6	24	125	330	805	-	-	-	-	-	
			3.6 V	16	54	260	675	1650	-	-	-	-	-	
		RTC	1.8 V	310	315	350	470	-	-	-	-	-	-	
IDD(V _{BAT})	Backup domain	enabled and clocked by	2.4 V	435	440	500	665	-	-	-	-	-	-	nA
IDD(VBAT)	supply current	LSE bypassed at	3 V	720	815	1050	1350	-	-	-	-	-	-	IIA
		32768 Hz	3.6 V	2150	2600	3400	4050	-	-	-	-	-	-	
			1.8 V	270	345	455	715	835	-	-	-	-	-	
		RTC enabled and clocked by	2.4 V	385	455	650	910	910	-	-	-	-	-	
		LSE	3 V	525	600	910	1150	1000	-	-	-	-	-	
		quartz ⁽²⁾	3.6 V	710	995	1250	1700	1900	-	-	_	-	-	

^{1.} Evaluated by characterization results - Not tested in production, unless otherwise specified.

^{2.} Based on characterization done with a 32.768 kHz crystal (MC306-G-06Q-32.768, manufacturer JFVNY) with two 6.8 pF loading capacitors.

IO system current consumption

The current consumption of the I/O system has two components: static and dynamic.

I/O static current consumption

All the I/Os used as inputs with pull-up generate current consumption when the pin is externally held low. The value of this current consumption can be simply computed by using the pull-up/pull-down resistors values given in *Table 54: I/O static characteristics*.

For the output pins, any external pull-down or external load must also be considered to estimate the current consumption.

Additional I/O current consumption is due to I/Os configured as inputs if an intermediate voltage level is externally applied. This current consumption is caused by the input Schmitt trigger circuits used to discriminate the input value. Unless this specific configuration is required by the application, this supply current consumption can be avoided by configuring these I/Os in analog mode. This is notably the case of ADC, OPAMP, COMP input pins which should be configured as analog inputs.

Caution:

Any floating input pin can also settle to an intermediate voltage level or switch inadvertently, as a result of external electromagnetic noise. To avoid current consumption related to floating pins, they must either be configured in analog mode, or forced internally to a definite digital value. This is done either by using pull-up/down resistors or by configuring the pins in output mode.

I/O dynamic current consumption

In addition to the internal peripheral current consumption measured previously (see *Table 36: Low-power mode wakeup timings*), the I/Os used by an application also contribute to the current consumption. When an I/O pin switches, it uses the current from the I/O supply voltage to supply the I/O pin circuitry and to charge/discharge the capacitive load (internal or external) connected to the pin:

$$I_{SW} = V_{DDIOx} \times f_{SW} \times C$$

where

 I_{SW} is the current sunk by a switching I/O to charge/discharge the capacitive load V_{DD} is the I/O supply voltage

f_{SW} is the I/O switching frequency

C is the total capacitance seen by the I/O pin: $C = C_{INT} + C_{EXT} + C_{S}$

C_S is the PCB board capacitance including the pad pin.

The test pin is configured in push-pull output mode and is toggled by software at a fixed frequency.

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On-chip peripheral current consumption

The current consumption of the on-chip peripherals is given in *Table 35*. The MCU is placed under the following conditions:

- All I/O pins are in Analog mode
- The given value is calculated by measuring the difference of the current consumptions:
 - when the peripheral is clocked on
 - when the peripheral is clocked off
- Ambient operating temperature and supply voltage conditions summarized in Table 14: Voltage characteristics
- The power consumption of the digital part of the on-chip peripherals is given in *Table 35*. The power consumption of the analog part of the peripherals (where applicable) is indicated in each related section of the datasheet.

Table 35. Peripheral current consumption

Bus	Peripheral	Range 1 Boost mode	Range 1 Normal mode	Range 2	Low-power run and sleep	Unit
-	Bus Matrix	6.12	5.69	4.70	6.11	μΑ/MHz
	AHB1 to APB1 bridge	0.26	0.25	0.22	0.03	
	AHB1 to APB2 bridge	0.39	0.37	0.32	0.03	
	FSMC	10.21	9.52	7.87	10.28	
	QUADSPI	3.51	3.27	2.69	3.51	
AHB1	CORDIC	1.28	1.19	0.98	0.78	μΑ/MHz
	CRC	0.74	0.68	0.57	0.63	
	DMA 1	2.83	2.64	2.17	2.75	
	DMA 2	3.11	2.90	2.39	2.43	
	DMAMUX	6.71	6.26	5.17	6.68	
	SRAM1	0.58	0.54	0.44	0.54	
	FLASH	6.46	6.01	4.95	6.15	
	FMAC	4.59	4.29	3.57	3.83	



Table 35. Peripheral current consumption (continued)

Bus	Peripheral	Range 1 Boost mode	Range 1 Normal mode	Range 2	Low-power run and sleep	Unit
АНВ2	ADC1/ADC2	6.24	5.80	4.77	5.88	μΑ/MHz
	ADC3/ADC4/ADC5	8.21	7.64	6.29	8.14	
	DAC1	4.70	4.38	3.63	4.40	
	DAC2	2.51	2.34	1.93	2.14	
	DAC3	4.62	4.31	3.57	4.15	
	DAC4	4.31	4.01	3.32	3.90	
	GPIOA	0.09	0.08	0.07	0.14	
	GPIOB	0.10	0.09	0.07	0.03	
	GPIOC	0.10	0.09	0.08	0.03	
	GPIOD	0.06	0.06	0.03	0.05	
	GPIOE	0.23	0.22	0.18	0.10	
	GPIOF	0.07	0.07	0.05	0.02	
	GPIOG	0.25	0.24	0.20	0.24	
	SRAM2	0.39	0.37	0.29	0.28	
	CCM SRAM	0.29	0.27	0.23	0.22	
	RNG	2.09	1.95	NA	NA	

Table 35. Peripheral current consumption (continued)

Bus	Peripheral	Range 1 Boost mode	Range 1 Normal mode	Range 2	Low-power run and sleep	Unit
	CRS	0.74	0.68	0.57	0.51	
	FDCAN1/FDCAN2/FDCAN3	22.20	20.68	17.10	21.15	
	I2C1	1.29	1.20	0.99	1.28	
	I2C2	1.29	1.20	0.99	1.28	
	I2C3	1.25	1.17	0.96	1.56	
	I2C4	1.25	1.16	0.96	1.97	
	LPTIM1	1.11	1.03	0.85	1.42	
	LPUART1	1.91	1.78	1.47	2.03	
	PWR	0.71	0.65	0.53	0.53	
	RTC	2.64	2.46	2.07	3.26	
	SPI2/I2S2	4.05	3.77	3.11	4.16	
	SPI3/I2S3	4.08	3.81	3.13	4.49	
APB1	TIM2	7.97	7.42	6.16	8.29	μΑ/MHz
	TIM3	6.37	5.93	4.92	6.81	
	TIM4	6.43	5.98	4.97	6.50	
	TIM5	8.28	7.71	6.38	8.11	
	TIM6	1.22	1.13	0.94	1.45	
	TIM7	1.28	1.18	0.98	1.56	
	UART4	2.51	2.33	1.92	3.14	
	UART5	2.79	2.60	2.14	3.34	
	USART2	2.75	2.56	2.12	3.11	
	USART3	2.71	2.52	2.08	2.47	
	USB	0.46	0.43	NA	NA	
	UCPD	2.46	2.28	1.89	NA	
	WWDG	0.42	0.39	0.31	0.42	

Table 35. Peripheral current consumption (continued)

Bus	Peripheral	Range 1 Boost mode	Range 1 Normal mode	Range 2	Low-power run and sleep	Unit
	SAI1	2.67	2.48	2.05	2.64	
	SPI1	1.99	1.86	1.54	2.02	
	SPI4	1.99	1.86	1.54	2.02	
	TIM1	10.85	10.13	8.40	9.93	
	TIM8	10.67	9.96	8.25	9.82	
APB2	TIM15	4.81	4.48	3.71	4.57	μΑ/MHz
	TIM16	3.71	3.45	2.88	3.45	
	TIM17	3.66	3.41	2.83	3.81	
	TIM20	10.71	9.99	8.29	10.00	
	USART1	2.49	2.31	1.91	2.49	
	SYSCFG/COMP/OPAMP/VREFBUF	1.63	1.52	1.25	0.91	

Table 35. Peripheral current consumption (continued)

Bus		Peripheral	Range 1 Boost mode	Range 1 Normal mode	Range 2	Low-power run and sleep	Unit
	ADC1/ ADC2	independent clock domain	0.72	0.67	0.53	0.63	
	ADC3/ ADC4/ ADC5	independent clock domain	0.67	0.62	0.50	0.22	
	FDCAN1/ FDCAN2/ FDCAN3	independent clock domain	11.62	10.84	8.95	10.24	
	I2C1	independent clock domain	4.03	3.76	3.12	4.15	
	I2C2	independent clock domain	3.78	3.52	2.93	3.23	
	I2C3	independent clock domain	2.72	2.55	2.11	2.65	
	I2C4	independent clock domain	3.95	3.67	3.04	2.81	
	I2S2	independent clock domain	1.49	1.40	1.15	1.63	μΑ/MHz
Independent clock domain	I2S3	independent clock domain	1.52	1.43	1.16	2.15	
	LPTIM1	independent clock domain	4.00	3.71	3.08	3.57	
	LPUART1	independent clock domain	4.43	4.13	3.45	4.02	
	QUADSPI	independent clock domain	0.54	0.51	0.44	0.75	
	RNG	independent clock domain	0.83	0.87	NA	NA	
	USB	independent clock domain	1.10	1.17	NA	NA	
	SAI1	independent clock domain	3.36	3.14	2.58	3.25	
	UART4	independent clock domain	6.60	6.17	5.14	6.02	
	UART5	independent clock domain	6.60	6.16	5.12	6.12	
	USART1	independent clock domain	7.62	7.12	5.89	6.90	
	USART2	independent clock domain	7.37	6.86	5.70	6.72	
	USART3	independent clock domain	7.98	7.44	6.17	8.21	
All	-		369.00	316.04	266.18	325.00	μΑ/MHz



5.3.6 Wakeup time from low-power modes and voltage scaling transition times

The wakeup times given in *Table 36* are the latency between the event and the execution of the first user instruction.

The device goes in low-power mode after the WFE (Wait For Event) instruction.

Table 36. Low-power mode wakeup timings⁽¹⁾

Symbol	Parameter		Conditions	Тур	Max	Unit
t _{WUSLEEP}	Wakeup time from Sleep mode to Run mode		-	11	12	Nb of
t _{WULPSLEEP}	Wakeup time from Low- power sleep mode to Low- power run mode		-	10	11	CPU cycles
	Wake up time from Stop 0	Range 1	Wakeup clock HSI16 = 16 MHz	5.8	6	
	mode to Run mode in flash memory	Range 2	Wakeup clock HSI16 = 16 MHz	18.4	19.1	
t _{wustop0}	Wake up time from Stop 0	Range 1	Wakeup clock HSI16 = 16 MHz	2.8	3	
	mode to Run mode in SRAM1	Range 2	Wakeup clock HSI16 = 16 MHz	2.9	3	
	Wake up time from Stop 1	Range 1	Wakeup clock HSI16 = 16 MHz	9.5	9.8	
	mode to Run in flash memory	Range 2	Wakeup clock HSI16 = 16 MHz	21.9	22.7	
	Wake up time from Stop 1	Range 1	Wakeup clock HSI16 = 16 MHz	6.6	6.9	ſ
	mode to Run mode in SRAM1	Range 2	Wakeup clock HSI16 = 16 MHz	6.4	6.6	
twustop1	Wake up time from Stop 1 mode to Low-power run mode in flash memory	Regulator in low-power Wakeup clock	Wakeup clock HSI16 = 16 MHz,	26.1	27.1 ⁽²⁾	μs
	Wake up time from Stop 1 mode to Low-power run mode in SRAM1	mode (LPR=1 in PWR_CR1)	with HPRE = 8	14.4	15 ⁽²⁾	
t _{WUSTBY}	Wakeup time from Standby mode to Run mode	Range 1	Wakeup clock HSI16 = 16 MHz	29.7	33.8	
t _{WUSTBY} SRAM2	Wakeup time from Standby with SRAM2 to Run mode	Range 1	Wakeup clock HSI16 = 16 MHz	29.7	33.5	
t _{WUSHDN}	Wakeup time from Shutdown mode to Run mode	Range 1	Wakeup clock HSI16 = 16 MHz	267.9	274.6 ⁽²⁾	
twulprun	Wakeup time from Low- power run mode to Run mode ⁽³⁾	Wakeup clock with HPRE = 8	HSI16 = 16 MHz	5	7	

^{1.} Evaluated by characterization - Not tested in production.

^{2.} Characterization results for temperature range from 0°C to 125°C.

^{3.} Time until REGLPF flag is cleared in PWR_SR2.

Table 37	Regulator	modes	transition	times ⁽¹⁾
Iable Ji.	Negulator	IIIOGES	uansinon	unics

Symbol	Parameter	Conditions	Тур	Max	Unit
	Regulator transition time from Range 2 to Range 1 or Range 1 to Range 2 ⁽²⁾	Wakeup clock HSI16 = 16 MHz with HPRE = 8	20	40	μs

- 1. Evaluated by characterization Not tested in production.
- 2. Time until VOSF flag is cleared in PWR_SR2.

Table 38. Wakeup time using USART/LPUART⁽¹⁾

Symbol	Parameter	Conditions	Тур	Max	Unit
Wakeup time needed to calculate the maximum USART/LPUART baudrate	Stop 0 mode	ı	1.7		
t _{WUUSART} t _{WULPUART}	allowing to wakeup up from stop mode when USART/LPUART clock source is HSI16	Stop 1 mode	-	8.5	μs

^{1.} Guaranteed by design - Not tested in production.

5.3.7 External clock source characteristics

High-speed external user clock generated from an external source

In bypass mode the HSE oscillator is switched off and the input pin is a standard GPIO.

The external clock signal has to respect the I/O characteristics in *Section 5.3.14*. However, the recommended clock input waveform is shown in *Figure 20: High-speed external clock source AC timing diagram*.

Table 39. High-speed external user clock characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Тур	Max	Unit	
f _{HSE_ext}	User external clock	Voltage scaling Range 1	-	8	48	- MHz	
	source frequency	Voltage scaling Range 2	-	8	26	IVII IZ	
V _{HSEH}	OSC_IN input pin high level voltage	-	0.7 V _{DD}	-	V_{DD}	V	
V _{HSEL}	OSC_IN input pin low level voltage	-	V_{SS}	-	0.3 V _{DD}		
t _{w(HSEH)}	OSC_IN high or low time	Voltage scaling Range 1	7	-	-	ns	
	OSC_IN HIGH OF IOW WITE	Voltage scaling Range 2	18	-	-	115	

^{1.} Guaranteed by design - Not tested in production.



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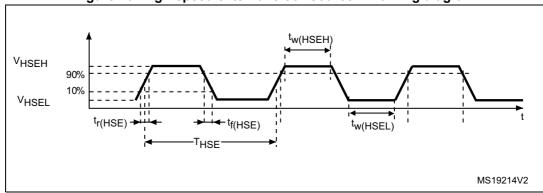


Figure 20. High-speed external clock source AC timing diagram

Low-speed external user clock generated from an external source

In bypass mode the LSE oscillator is switched off and the input pin is a standard GPIO.

The external clock signal has to respect the I/O characteristics in Section 5.3.14. However, the recommended clock input waveform is shown in Figure 21.

Symbol Parameter Conditions Min Max Unit Тур User external clock source 1000 f_{LSE_ext} 32.768 kHz frequency OSC32_IN input pin high $0.7~V_{DD}$ V_{LSEH} V_{DD} level voltage V OSC32_IN input pin low level V_{LSEL} V_{SS} $0.3 V_{DD}$ voltage tw(LSEH) OSC32_IN high or low time 250 ns t_{w(LSEL)}

Table 40. Low-speed external user clock characteristics⁽¹⁾

^{1.} Guaranteed by design - Not tested in production.

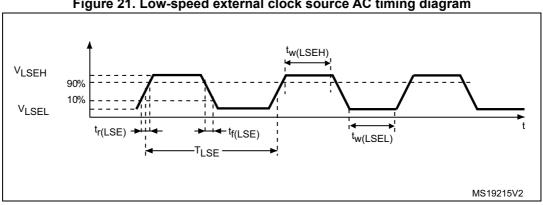


Figure 21. Low-speed external clock source AC timing diagram

High-speed external clock generated from a crystal/ceramic resonator

The high-speed external (HSE) clock can be supplied with a 4 to 48 MHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on design simulation results obtained with typical external components specified in *Table 41*. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Symbol	Parameter	Conditions ⁽²⁾	Min	Тур	Max	Unit
f _{OSC_IN}	Oscillator frequency	-	4	8	48	MHz
R _F	Feedback resistor	-	-	200	-	kΩ
		During startup ⁽³⁾	-	-	5.5	
		V_{DD} = 3 V, Rm = 30 Ω , CL = 10 pF@8 MHz	-	0.44	-	
		V_{DD} = 3 V, Rm = 45 Ω , CL = 10 pF@8 MHz	-	0.45	-	
I _{DD(HSE)}	HSE current consumption	$V_{DD} = 3 \text{ V},$ $Rm = 30 \Omega,$ CL = 5 pF@48 MHz	-	0.68	-	mA
		$V_{DD} = 3 V,$ $Rm = 30 \Omega,$ CL = 10 pF@48 MHz	-	0.94	-	
		V _{DD} = 3 V, Rm = 30 Ω, CL = 20 pF@48 MHz	-	1.77	-	
G _m	Maximum critical crystal transconductance	Startup	-	-	1.5	mA/V
t _{SU(HSE)} ⁽⁴⁾	Startup time	V _{DD} is stabilized	-	2	-	ms

Table 41. HSE oscillator characteristics⁽¹⁾

For C_{L1} and C_{L2} , it is recommended to use high-quality external ceramic capacitors in the 5 pF to 20 pF range (typ.), designed for high-frequency applications, and selected to match the requirements of the crystal or resonator (see *Figure 22*). C_{L1} and C_{L2} are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of C_{L1} and C_{L2} . PCB and MCU pin capacitance must be included (10 pF can be used as a rough estimate of the combined pin and board capacitance) when sizing C_{L1} and C_{L2} .



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^{1.} Guaranteed by design - Not tested in production.

^{2.} Resonator characteristics given by the crystal/ceramic resonator manufacturer.

^{3.} This consumption level occurs during the first 2/3 of the $t_{\text{SU(HSE)}}$ startup time

^{4.} t_{SU(HSE)} is the startup time measured from the moment it is enabled (by software) to a stabilized 8 MHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer

Note: For information on selecting the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website www.st.com.

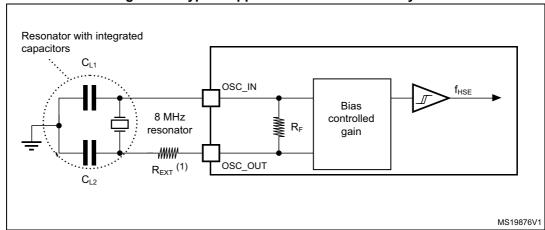


Figure 22. Typical application with an 8 MHz crystal

1. R_{EXT} value depends on the crystal characteristics.

Low-speed external clock generated from a crystal resonator

The low-speed external (LSE) clock can be supplied with a 32.768 kHz crystal resonator oscillator. All the information given in this paragraph are based on design simulation results obtained with typical external components specified in *Table 42*. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).



Symbol	Parameter	Conditions ⁽²⁾	Min	Тур	Max	Unit
		LSEDRV[1:0] = 00 Low drive capability	-	250	-	
	LSE current consumption	LSEDRV[1:0] = 01 Medium low drive capability	-	315	-	nA
I _{DD(LSE)}	Loc current consumption	LSEDRV[1:0] = 10 Medium high drive capability	-	500	-	IIA
		LSEDRV[1:0] = 11 High drive capability	-	630	-	
		LSEDRV[1:0] = 00 Low drive capability	-	-	0.5	
Gm	Maximum critical crystal	LSEDRV[1:0] = 01 Medium low drive capability	-	-	0.75	μΑ/V
Gm _{critmax}	gm	LSEDRV[1:0] = 10 Medium high drive capability	-	1	1.7	μΑνν
		LSEDRV[1:0] = 11 High drive capability	-	-	2.7	
t _{SU(LSE)} ⁽³⁾	Startup time	V _{DD} is stabilized	-	2	ı	S

Table 42. LSE oscillator characteristics $(f_{LSE} = 32.768 \text{ kHz})^{(1)}$

- 1. Guaranteed by design Not tested in production.
- 2. Refer to the note and caution paragraphs below the table, and to the application note AN2867 "Oscillator design guide for ST microcontrollers".
- t_{SU(LSE)} is the startup time measured from the moment it is enabled (by software) to a stabilized 32.768 kHz oscillation is reached. This value is measured for a standard crystal and it can vary significantly with the crystal manufacturer

Note: For information on selecting the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website www.st.com.

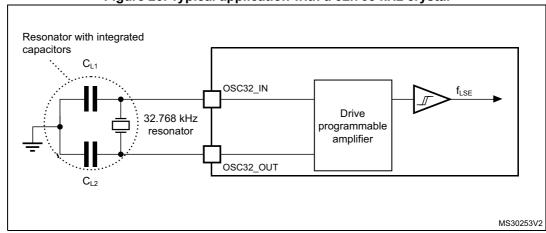


Figure 23. Typical application with a 32.768 kHz crystal

Note:

An external resistor is not required between OSC32_IN and OSC32_OUT and it is forbidden to add one.



5.3.8 Internal clock source characteristics

The parameters given in *Table 43* are derived from tests performed under ambient temperature and supply voltage conditions summarized in *Table 17: General operating conditions*. The provided curves are characterization results, not tested in production.

High-speed internal (HSI16) RC oscillator

Table 43. HSI16 oscillator characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f _{HSI16}	HSI16 Frequency	V _{DD} =3.0 V, T _A =30 °C	15.88	-	16.08	MHz
TRIM	HSI16 user trimming step	Trimming code is not a multiple of 64	0.2	0.3	0.4	%
	Thomas distributing step	Trimming code is a multiple of 64	-4	-6	-8	70
DuCy(HSI16) ⁽²⁾	Duty Cycle	-	45	-	55	%
. (110140)	HSI16 oscillator frequency	T _A = 0 to 85 °C	-1	-	1	%
$\Delta_{Temp}(HSI16)$	drift over temperature	T _A = -40 to 125 °C	-2	-	1.5	%
Δ _{VDD} (HSI16)	HSI16 oscillator frequency drift over V _{DD}	V _{DD} =1.62 V to 3.6 V	-0.1	-	0.05	%
t _{su} (HSI16) ⁽²⁾	HSI16 oscillator start-up time	-	ı	0.8	1.2	μs
t _{stab} (HSI16) ⁽²⁾	HSI16 oscillator stabilization time	-	-	3	5	μs
I _{DD} (HSI16) ⁽²⁾	HSI16 oscillator power consumption	-	-	155	190	μΑ

^{1.} Evaluated by characterization - Not tested in production.

^{2.} Guaranteed by design - Not tested in production.

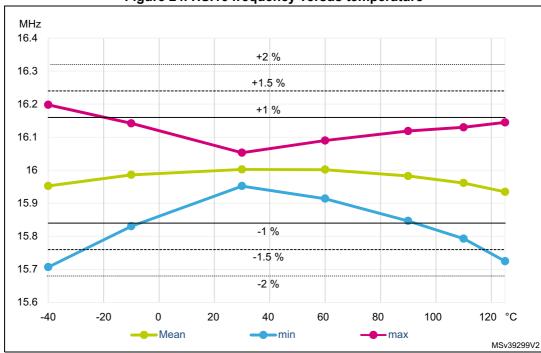


Figure 24. HSI16 frequency versus temperature

High-speed internal 48 MHz (HSI48) RC oscillator

Table 44. HSI48 oscillator characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f _{HSI48}	HSI48 Frequency	V _{DD} =3.0V, T _A =30°C	-	48	-	MHz
TRIM	HSI48 user trimming step	-	-	0.11 ⁽²⁾	0.18 ⁽²⁾	%
USER TRIM COVERAGE	HSI48 user trimming coverage	±32 steps	±3 ⁽³⁾	±3.5 ⁽³⁾	-	%
DuCy(HSI48)	Duty Cycle	-	45 ⁽²⁾	-	55 ⁽²⁾	%
ACC	Accuracy of the HSI48 oscillator over temperature	V _{DD} = 3.0 V to 3.6 V, T _A = -15 to 85 °C	-	-	±3 ⁽³⁾	%
ACC _{HSI48_REL}	(factory calibrated)	V_{DD} = 1.65 V to 3.6 V, T_A = -40 to 125 °C	-	-	±4.5 ⁽³⁾	/0
D (HSIV8)	HSI48 oscillator frequency	V _{DD} = 3 V to 3.6 V	-	0.025 ⁽³⁾	0.05 ⁽³⁾	%
D _{VDD} (HSI48)	drift with V _{DD}	V _{DD} = 1.65 V to 3.6 V	-	0.05 ⁽³⁾	0.1 ⁽³⁾	/0
t _{su} (HSI48)	HSI48 oscillator start-up time	-	-	2.5 ⁽²⁾	6 ⁽²⁾	μs
I _{DD} (HSI48)	HSI48 oscillator power consumption	-	-	340 ⁽²⁾	380 ⁽²⁾	μA

Symbol Parameter		Conditions	Min	Тур	Max	Unit
N _T jitter	Next transition jitter Accumulated jitter on 28 cycles ⁽⁴⁾	-	-	+/-0.15 ⁽²⁾	-	ns
P _T jitter	Paired transition jitter Accumulated jitter on 56 cycles ⁽⁴⁾	-	-	+/-0.25 ⁽²⁾	-	ns

Table 44. HSI48 oscillator characteristics⁽¹⁾ (continued)

- 1. V_{DD} = 3 V, T_A = -40 to 125°C unless otherwise specified.
- 2. Guaranteed by design Not tested in production.
- 3. Evaluate by characterization Not tested in production.
- 4. Jitter measurement are performed without clock source activated in parallel.

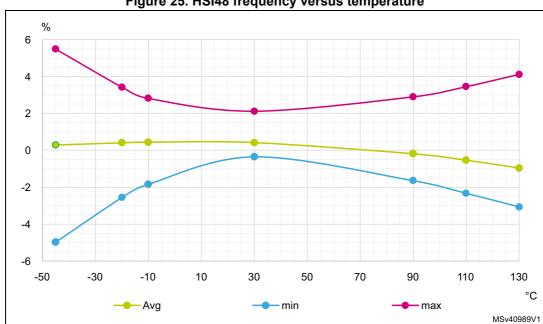


Figure 25. HSI48 frequency versus temperature

Low-speed internal (LSI) RC oscillator

Conditions Symbol Unit **Parameter** Min Typ Max $V_{DD} = 3.0 V,$ 32.96 31.04 $T_A = 30 \, ^{\circ}C$ LSI Frequency kHz $V_{DD} = 1.62 \text{ to } 3.6 \text{ V},$ 29.5 34 $T_A = -40 \text{ to } 125 \,^{\circ}\text{C}$ LSI oscillator start-up $t_{SU}(LSI)^{(2)} \\$ 130 80 μs time

Table 45. LSI oscillator characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
t _{STAB} (LSI) ⁽²⁾	LSI oscillator stabilization time	5% of final frequency	-	125	180	μs
I _{DD} (LSI) ⁽²⁾	LSI oscillator power consumption	-	-	110	180	nA

Table 45. LSI oscillator characteristics⁽¹⁾ (continued)

5.3.9 PLL characteristics

The parameters given in *Table 46* are derived from tests performed under temperature and V_{DD} supply voltage conditions summarized in *Table 17: General operating conditions*.

Parameter Conditions Symbol Min Typ Max Unit PLL input clock⁽²⁾ 16 MHz 2.66 f_{PLL_IN} PLL input clock duty cycle % 55 45 Voltage scaling Range 1 2.0645 170 Boost mode PLL multiplier output clock P f_{PLL_P_OUT} 2.0645 Voltage scaling Range 1 150 Voltage scaling Range 2 2.0645 26 Voltage scaling Range 1 8 170 Boost mode PLL multiplier output clock Q f_{PLL_Q_OUT} Voltage scaling Range 1 8 150 MHz Voltage scaling Range 2 26 8 Voltage scaling Range 1 8 170 Boost mode PLL multiplier output clock R f_{PLL_R_OUT} Voltage scaling Range 1 8 150 Voltage scaling Range 2 26 Voltage scaling Range 1 96 344 PLL VCO output f_{VCO_OUT} Voltage scaling Range 2 96 128 PLL lock time 15 40 _ μs t_{LOCK} RMS cycle-to-cycle jitter 28.6 Jitter System clock 150 MHz **±ps** RMS period jitter 21.4 -VCO freq = 96 MHz 200 260 PLL power consumption on $I_{DD}(PLL)$ VCO freq = 192 MHz 300 380 μΑ $V_{DD}^{(1)}$ VCO freq = 344 MHz 520 650

Table 46. PLL characteristics⁽¹⁾

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^{1.} Evaluated by characterization - Not tested in production.

^{2.} Guaranteed by design - Not tested in production.

^{1.} Guaranteed by design - Not tested in production.

^{2.} Take care of using the appropriate division factor M to obtain the specified PLL input clock values.

5.3.10 Flash memory characteristics

Table 47. Flash memory characteristics⁽¹⁾

Symbol	Parameter	Conditions	Тур	Max	Unit	
t _{prog}	64-bit programming time	-	81.7	83.35	μs	
4	One row (32 double word) programming time	Normal programming	2.61	2.7		
t _{prog_row}		Fast programming	1.91	1.95		
	One page (2 Kbytes)	Normal programming	20.91	21.34	ms	
^t prog_page	programming time	Fast programming	15.29	15.6		
t _{ERASE}	Page (2 Kbytes) erase time	-	22.02	24.47		
	One bank (256 Kbyte) programming time	Normal programming	2.68	2.73		
^t prog_bank		Fast programming	1.96	2	S	
t _{ME}	Mass erase time (one or two banks)	-	22.13	24.6	ms	
	Average consumption	Write mode	3.5	-		
	from V _{DD}	Erase mode	3.5	-		
I _{DD}	Maximum ourrant (noak)	Write mode	7 (for 6 μs)	-	- mA	
	Maximum current (peak)	Erase mode	7 (for 67 μs)	-		

^{1.} Guaranteed by design - Not tested in production.

Table 48. Flash memory endurance and data retention

Symbol	Parameter	Conditions	Min ⁽¹⁾	Unit	
N _{END}	Endurance	T _A = -40 to +105 °C	10	kcycles	
		1 kcycle ⁽²⁾ at T _A = 85 °C	30		
	Data retention	1 kcycle ⁽²⁾ at T _A = 105 °C	15		
4		1 kcycle ⁽²⁾ at T _A = 125 °C	7	Vooro	
t _{RET}		10 kcycles ⁽²⁾ at T _A = 55 °C	30	Years	
		10 kcycles ⁽²⁾ at T _A = 85 °C	10 kcycles ⁽²⁾ at T _A = 85 °C	15	
		10 kcycles ⁽²⁾ at T _A = 105 °C	10		

^{1.} Evaluated by characterization - Not tested in production.

^{2.} Cycling performed over the whole temperature range.

5.3.11 EMC characteristics

Susceptibility tests are performed on a sample basis during device characterization.

Functional EMS (electromagnetic susceptibility)

While a simple application is executed on the device (toggling 2 LEDs through I/O ports). the device is stressed by two electromagnetic events until a failure occurs. The failure is indicated by the LEDs:

- Electrostatic discharge (ESD) (positive and negative) is applied to all device pins until a functional disturbance occurs. This test is compliant with the IEC 61000-4-2 standard.
- FTB: A Burst of Fast Transient voltage (positive and negative) is applied to V_{DD} and V_{SS} through a 100 pF capacitor, until a functional disturbance occurs. This test is compliant with the IEC 61000-4-4 standard.

A device reset allows normal operations to be resumed.

The test results are given in *Table 49*. They are based on the EMS levels and classes defined in application note AN1709.

Level/ **Symbol Conditions Parameter** Class $V_{DD} = 3.3 \text{ V}, T_A = +25 ^{\circ}\text{C},$ Voltage limits to be applied on any I/O pin $f_{HCLK} = 170 \text{ MHz}.$ 3B V_{FESD} to induce a functional disturbance conforming to IEC 61000-4-2 Fast transient voltage burst limits to be $V_{DD} = 3.3 \text{ V}, T_A = +25 ^{\circ}\text{C},$ $f_{HCLK} = 170 \text{ MHz},$ applied through 100 pF on V_{DD} and V_{SS} $\mathsf{V}_{\mathsf{EFTB}}$ 5A pins to induce a functional disturbance conforming to IEC 61000-4-4

Table 49. EMS characteristics

Designing hardened software to avoid noise problems

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software. It should be noted that good EMC performance is highly dependent on the user application and the software in particular.

Therefore it is recommended that the user applies EMC software optimization and pregualification tests in relation with the EMC level requested for his application.

Software recommendations

The software flowchart must include the management of runaway conditions such as:

- Corrupted program counter
- Unexpected reset
- Critical Data corruption (control registers...)

Prequalification trials

Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the NRST pin or the Oscillator pins for 1 second.

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To complete these trials, ESD stress can be applied directly on the device, over the range of specification values. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring (see application note AN1015).

Electromagnetic Interference (EMI)

The electromagnetic field emitted by the device are monitored while a simple application is executed (toggling 2 LEDs through the I/O ports). This emission test is compliant with IEC 61967-2 standard which specifies the test board and the pin loading.

Max vs. [f_{HSE}/f_{HCLK}] Monitored **Symbol Conditions** Unit **Parameter** frequency band 8 MHz / 170 MHz 0.1 MHz to 30 MHz 4 30 MHz to 130 MHz 0 $V_{DD} = 3.6 \text{ V}, T_A = 25 ^{\circ}\text{C},$ dB_µV Peak level LQFP128 package 130 MHz to 1 GHz 16 S_{FMI} compliant with IEC 61967-2 1 GHz to 2 GHz 11 EMI Level 3.5

Table 50. EMI characteristics

5.3.12 Electrical sensitivity characteristics

Based on three different tests (ESD, LU) using specific measurement methods, the device is stressed in order to determine its performance in terms of electrical sensitivity.

Electrostatic discharge (ESD)

Electrostatic discharges (a positive then a negative pulse separated by 1 second) are applied to the pins of each sample according to each pin combination. The sample size depends on the number of supply pins in the device (3 parts × (n+1) supply pins). This test conforms to the ANSI/JEDEC standard.

Symbol	Ratings	Conditions		Class	Maximum value ⁽¹⁾	Unit
V _{ESD(HBM)}	Electrostatic discharge voltage (human body model)	T_A = +25 °C, conforming to ANSI/ESDA/JEDEC JS-001		2	2000	V
V	Electrostatic discharge	T _A = +25 °C, conforming to	LQFP100 and LQFP128	C1	250	V
V _{ESD(CDM)}	voltage (charge device model)	ANSI/ESDA/JEDEC JS- 002 Other packages		C2a	500	V

Table 51. ESD absolute maximum ratings

1. Evaluated by characterization - Not tested in production.



Static latch-up

Two complementary static tests are required on three parts to assess the latch-up performance:

- A supply over-voltage is applied to each power supply pin.
- A current injection is applied to each input, output and configurable I/O pin.

These tests are compliant with EIA/JESD 78E IC latch-up standard.

Table 52. Electrical sensitivities

Symbol	Parameter	Conditions	Class
LU	Static latch-up class	TA = +125 °C conforming to JESD78E	Class II level A

5.3.13 I/O current injection characteristics

As a general rule, current injection to the I/O pins, due to external voltage below V_{SS} or above V_{DD} (for standard, 3.3 V-capable I/O pins) should be avoided during normal product operation. However, in order to give an indication of the robustness of the microcontroller in cases when abnormal injection accidentally happens, susceptibility tests are performed on a sample basis during device characterization.

Functional susceptibility to I/O current injection

While a simple application is executed on the device, the device is stressed by injecting current into the I/O pins programmed in floating input mode. While current is injected into the I/O pin, one at a time, the device is checked for functional failures.

The failure is indicated by an out of range parameter: ADC error above a certain limit (higher than 5 LSB TUE), out of conventional limits of induced leakage current on adjacent pins (out of the -5 μ A/+0 μ A range) or other functional failure (for example reset occurrence or oscillator frequency deviation).

The characterization results are given in *Table 53*.

Negative induced leakage current is caused by negative injection and positive induced leakage current is caused by positive injection.

Table 53. I/O current injection susceptibility

Symbol	Description		Functional susceptibility Description		
Symbol			Negative injection	Positive injection	Unit
		All except TT_a, PF10, PB8-BOOT0, PC10	-5	NA	
I _{INJ} ⁽¹⁾	Injected current on pin	PF10, PB8-BOOT0, PC10	-0	NA	mA
	TT_a pins		-5	0	

^{1.} Evaluated by characterization - Not tested in production.



5.3.14 I/O port characteristics

General input/output characteristics

Unless otherwise specified, the parameters given in *Table 54* are derived from tests performed under the conditions summarized in *Table 17: General operating conditions*. All I/Os are designed as CMOS- and TTL-compliant.

Table 54. I/O static characteristics

Symbol	Parameter		Conditions	Min	Тур	Max	Unit			
		All except	1621/2/ 2261/			0.3xV _{DD}				
V _{II} ⁽¹⁾⁽²⁾	I/O input low level	FT_c	1.62 V <v<sub>DD<3.6 V</v<sub>	-	-	0.39xV _{DD} -0.06 ⁽³⁾	V			
VIL. //	voltage	ET o	1.62 V <v<sub>DD<3.6 V</v<sub>			0.3xV _{DD}	V			
		FT_c	1.02 V \ V _{DD} \ 3.0 V	_	_	0.25xV _{DD}				
	I/O input high level	All except	1.62 V <v<sub>DD<3.6 V</v<sub>	0.7xV _{DD}	-	-				
V _{IH} ⁽¹⁾⁽²⁾ high le		FT_c	1.02 0 1000 10.0 0	0.49xV _{DD} +0.26 ⁽³⁾	-	-	V			
	voltage	FT_c	1.62 V <v<sub>DD<3.6 V</v<sub>	0.7xV _{DD}	-	-				
V _{HYS} ⁽³⁾	Input hysteresis	TT_xx, FT_xxx, NRST	1.62 V <v<sub>DD<3.6 V</v<sub>	-	200	-	mV			
		FT_xx	$0 < V_{IN} \le V_{DD}$	-	-	±100				
			except	$V_{DD} \le V_{IN} \le V_{DD} + 1 V$	-	-	650 ⁽⁴⁾			
		FT_c	V _{DD} +1 V < V _{IN} ≤ 5.5 V	-	-	200 ⁽⁴⁾				
		ГТ	$0 \le V_{IN} \le V_{DDMAX}$	-	-	2000				
	Input leakage current ⁽³⁾	FT_c	V _{DD} ≤ V _{IN} <0.5 V	-	-	3000				
		leakage	leakage	leakage		$0 \le V_{IN} \le V_{DD}$	-	-	±150	nA
I _{leak}					FT_u, PC3	$V_{DD} \le V_{IN} \le V_{DD} + 1 V$	-	-	±2500	IIA
								$V_{DD} \le V_{IN} \le 5.5 \text{ V}$	-	-
		ET d	$0 \le V_{IN} \le V_{DD}$	-	-	±4500				
		FT_d	$V_{DD} + 1V \le V_{IN} \le 5.5 \text{ V}$	-	-	±9000				
		TT_xx	$0 \le V_{IN} \le V_{DD}$	-	-	±150				
		11_^^	$V_{DD} \le V_{IN} \le 3.6 \text{ V}$	-	-	2000				
R _{PU}	Weak pull- up equivalent resistor ⁽⁵⁾		$V_{IN} = V_{SS}$	25	40	55	1.0			
R _{PD}	Weak pull- down equivalent resistor ⁽⁵⁾		$V_{IN} = V_{DD}$	25	40	55	kΩ			
C _{IO}	I/O pin capacitance	I/O pin capacitance	-	-	5	-	pF			

^{1.} Refer to Figure 26: I/O input characteristics

- 2. Data based on characterization results, not tested in production
- 3. Guaranteed by design Not tested in production.
- 4. This value represents the pad leakage of the I/O itself. The total product pad leakage is provided by this formula: $I_{Total_Ileak_max} = 10 \ \mu A + [number of I/Os where VIN is applied on the pad]_x \ I_{lkg}(Max)$.
- Pull-up and pull-down resistors are designed with a true resistance in series with a switchable PMOS/NMOS. This PMOS/NMOS contribution to the series resistance is minimal (~10% order).

Note:

For more information about GPIO properties, refer to the application note AN4899 "STM32 GPIO configuration for hardware settings and low-power consumption" available from the ST website www.st.com.

All I/Os are CMOS- and TTL-compliant (no software configuration required). Their characteristics cover more than the strict CMOS-technology or TTL parameters. The coverage of these requirements is shown in *Figure 26* for standard I/Os, and 5 V tolerant I/Os (except FT_c).

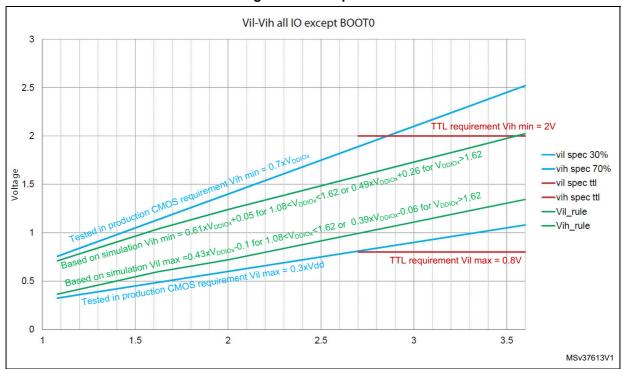


Figure 26. I/O input characteristics

Output driving current

The GPIOs (general purpose input/outputs) can sink or source up to ± 8 mA, and sink or source up to ± 20 mA (with a relaxed V_{OI}/V_{OH}).

5

In the user application, the number of I/O pins which can drive current must be limited to respect the absolute maximum rating specified in Section 5.2:

- The sum of the currents sourced by all the I/Os on V_{DD}, plus the maximum consumption of the MCU sourced on V_{DD}, cannot exceed the absolute maximum rating ΣI_{VDD} (see *Table 14: Voltage characteristics*).
- The sum of the currents sunk by all the I/Os on V_{SS}, plus the maximum consumption of the MCU sunk on V_{SS}, cannot exceed the absolute maximum rating ΣI_{VSS} (see Table 14: Voltage characteristics).

Output voltage levels

Unless otherwise specified, the parameters given in the table below are derived from tests performed under the ambient temperature and supply voltage conditions summarized in *Table 17: General operating conditions*. All I/Os are CMOS- and TTL-compliant (FT OR TT unless otherwise specified).

Table 33. Output Voltage characteristics									
Symbol	Parameter	Conditions	Min	Max	Unit				
V _{OL} ⁽³⁾	Output low level voltage for an I/O pin	CMOS port	-	0.4					
V _{OH} ⁽³⁾	Output high level voltage for an I/O pin	$ I_{IO} $ = 2 mA for FT_c I/Os = 8 mA for other $I/Os V_{DD}$ $\geq 2.7 V$	V _{DD} -0.4	-					
V _{OL} ⁽³⁾	Output low level voltage for an I/O pin	TTL port	-	0.4					
V _{OH} ⁽³⁾	Output high level voltage for an I/O pin	$ I_{IO} = 2 \text{ mA for FT_c}$ I/Os = 8 mA for other I/Os $V_{DD} \ge 2.7 \text{ V}$	2.4	-					
V _{OL} ⁽³⁾	Output low level voltage for an I/O pin	All I/Os except FT_c	-	1.3	V				
V _{OH} ⁽³⁾	Output high level voltage for an I/O pin	I _{IO} = 20 mA V _{DD} ≥ 2.7 V	V _{DD} -1.3	-	\ \ \				
V _{OL} ⁽³⁾	Output low level voltage for an I/O pin	I _{IO} = 1 mA for FT_c	-	0.4					
V _{OH} ⁽³⁾	Output high level voltage for an I/O pin	I/Os = 4 mA for other I/Os V _{DD} ≥ 1.62 V	V _{DD} -0.45	-					
V _{OLFM+}	Output low level voltage for an FT I/O	$ I_{IO} = 20 \text{ mA}$ $V_{DD} \ge 2.7 \text{ V}$	-	0.4					
(3) W	pin in FM+ mode (FT I/O with "f" option) $ I_{IO} = 10 \text{ mA} $ $V_{DD} \ge 1.62 \text{ V} $		-	0.4					

Table 55. Output voltage characteristics⁽¹⁾⁽²⁾

Input/output AC characteristics

The definition and values of input/output AC characteristics are given in *Figure 27* and *Table 56*, respectively.

Unless otherwise specified, the parameters given are derived from tests performed under the ambient temperature and supply voltage conditions summarized in *Table 17: General operating conditions*.



The I_{IO} current sourced or sunk by the device must always respect the absolute maximum rating specified in Table 14:
 Voltage characteristics, and the sum of the currents sourced or sunk by all the I/Os (I/O ports and control pins) must always respect the absolute maximum ratings ΣI_{IO}.

^{2.} TTL and CMOS outputs are compatible with JEDEC standards JESD36 and JESD52.

^{3.} Guaranteed by design - Not tested in production.

Table 56. I/O (except FT_c) AC characteristics⁽¹⁾ (2)

Speed	Symbol	Parameter	Conditions	Min	Max	Unit	
			C=50 pF, 2.7 V≤V _{DD} ≤3.6 V	-	5		
		Maximum	C=50 pF, 1.62 V≤V _{DD} ≤2.7 V	-	1	NAL 1-	
	Fmax	frequency	C=10 pF, 2.7 V≤V _{DD} ≤3.6 V	-	10	MHz	
00			C=10 pF, 1.62 V≤V _{DD} ≤2.7 V	-	1.5		
00			C=50 pF, 2.7 V≤V _{DD} ≤3.6 V	-	25		
	T=/Tf	Output rise and	C=50 pF, 1.62 V≤V _{DD} ≤2.7 V	-	52		
	Tr/Tf	fall time	C=10 pF, 2.7 V≤V _{DD} ≤3.6 V	-	17	ns	
			C=10 pF, 1.62 V≤V _{DD} ≤2.7 V	-	37		
			C=50 pF, 2.7 V≤V _{DD} ≤3.6 V	-	25		
	-	Maximum	C=50 pF, 1.62 V≤V _{DD} ≤2.7 V	-	10		
	Fmax	frequency	C=10 pF, 2.7 V≤V _{DD} ≤3.6 V	-	50	MHz	
04	0.4		C=10 pF, 1.62 V≤V _{DD} ≤2.7 V	-	15		
01				C=50 pF, 2.7 V≤V _{DD} ≤3.6 V	-	9	
	T/Tf	Tr/Tf Output rise and fall time	C=50 pF, 1.62 V≤V _{DD} ≤2.7 V	-	16	ne	
	1 17/11		C=10 pF, 2.7 V≤V _{DD} ≤3.6 V	-	4.5	ns	
			C=10 pF, 1.62 V≤V _{DD} ≤2.7 V	-	9		
			C=50 pF, 2.7 V≤V _{DD} ≤3.6 V	-	50		
		Maximum	C=50 pF, 1.62 V≤V _{DD} ≤2.7 V	-	25	NAL 1—	
	Fmax	frequency	C=10 pF, 2.7 V≤V _{DD} ≤3.6 V	-	100 ⁽³⁾	MHz	
40			C=10 pF, 1.62 V≤V _{DD} ≤2.7 V	-	37.5		
10			C=50 pF, 2.7 V≤V _{DD} ≤3.6 V	-	5.8		
	T/Tf	Output rise and	C=50 pF, 1.62 V≤V _{DD} ≤2.7 V	-	11		
	Tr/Tf	fall time	C=10 pF, 2.7 V≤V _{DD} ≤3.6 V	-	2.5	ns	
			C=10 pF, 1.62 V≤V _{DD} ≤2.7 V	-	5		
			C=30 pF, 2.7 V≤V _{DD} ≤3.6 V	-	120 ⁽³⁾		
		Maximum	C=30 pF, 1.62 V≤V _{DD} ≤2.7 V	-	50	NAL 1-	
Fmax	Fillax	frequency	C=10 pF, 2.7 V≤V _{DD} ≤3.6 V	-	180 ⁽³⁾	MHz	
			C=10 pF, 1.62 V≤V _{DD} ≤2.7 V	-	75		
11			C=30 pF, 2.7 V≤V _{DD} ≤3.6 V	-	3.3		
	T,,/T4	Output rise and	C=30 pF, 1.62 V≤V _{DD} ≤2.7 V	-	6		
	Tr/Tf	fall time ⁽⁴⁾	C=10 pF, 2.7 V≤V _{DD} ≤3.6 V	-	1.7	ns	
				C=10 pF, 1.62 V≤V _{DD} ≤2.7 V	-	3.3	



Table 56. I/O (except FT	c) AC characte	eristics ⁽¹⁾ (2)	continued)
Table 30. I/O (Except I I	or AC Charact	まいろいしろ゛´`´い	COILLIIGE

Speed	Symbol	Parameter	Conditions	Min	Max	Unit
	Fmax ⁽⁵⁾	Maximum frequency		-	1	MHz
FM+	Tr/TF ⁽⁴⁾	Output high to low level fall time	C=50 pF, 1.6 V≤V _{DD} ≤3.6 V	-	5	ns

- The I/O speed is configured using the OSPEEDRy[1:0] bits. The Fm+ mode is configured in the SYSCFG_CFGR1 register. Refer to the reference manual RM0440 "STM32G4 Series advanced Arm®-based 32-bit MCUs" for a description of GPIO Port configuration register.
- 2. Guaranteed by design Not tested in production.
- 3. This value represented the I/O capability but maximum system frequency is 170 MHz.
- 4. The fall time is defined between 70% and 30% of the output waveform accordingly to I2C specification.
- 5. The maximum frequency is defined with the following conditions:

 - (Tr+ Tf) ≤ 2/3 T. 45%<Duty cycle<55%

Table 57. I/O FT_c AC characteristics⁽¹⁾ (2)

Speed	Symbol	Parameter	Conditions	Min	Max	Unit	
	Fmax	Maximum	C=50 pF, 2.7 V≤V _{DD} ≤3.6 V	50 pF, 2.7 V≤V _{DD} ≤3.6 V -		MHz	
0 Tr/Tf	FIIIAX	frequency	C=50 pF, 1.6 V≤V _{DD} ≤2.7 V	-	1	IVITZ	
	Output H/L to	C=50 pF, 2.7 V≤V _{DD} ≤3.6 V	-	170			
	Ir/If	L/H level fall time	C=50 pF, 1.6 V≤V _{DD} ≤2.7 V	-	330	ns	
	Fmax	Maximum	C=50 pF, 2.7 V≤V _{DD} ≤3.6 V	-	10	MHz	
	Fillax	frequency	C=50 pF, 1.6 V≤V _{DD} ≤2.7 V	-	5	IVIHZ	
1		Output H/L to	C=50 pF, 2.7 V≤V _{DD} ≤3.6 V	-	35		
	Tr/Tf	L/H level fall time	C=50 pF, 1.6 V≤V _{DD} ≤2.7 V - 65		65	ns	

- The I/O speed is configured using the OSPEEDRy[1:0] bits. The Fm+ mode is configured in the SYSCFG_CFGR1 register. Refer to the reference manual RM0440 "STM32G4 Series advanced Arm based 32-bit MCUs" for a description of GPIO Port configuration register.
- 2. Guaranteed by design Not tested in production.

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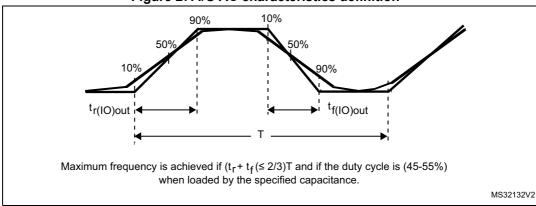


Figure 27. I/O AC characteristics definition⁽¹⁾

1. Refer to Table 56: I/O (except FT_c) AC characteristics.

5.3.15 NRST pin characteristics

The NRST pin input driver uses the CMOS technology. It is connected to a permanent pull-up resistor, R_{PU} .

Unless otherwise specified, the parameters given in the table below are derived from tests performed under the ambient temperature and supply voltage conditions summarized in *Table 17: General operating conditions*.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V _{IL(NRST)}	NRST input low level voltage	-	-	-	$0.3_{x}V_{DD}$	V
V _{IH(NRST)}	Voltago		0.7 _x V _{DD}	-	-	V
V _{hys(NRST)}	NRST Schmitt trigger voltage hysteresis	-	-	200	-	mV
R _{PU}	Weak pull-up equivalent resistor ⁽²⁾	V _{IN} = V _{SS}	25	40	55	kΩ
V _{F(NRST)}	NRST input filtered pulse	-	-	-	70	ns
V _{NF(NRST)}	NRST input not filtered pulse	1.71 V ≤ V _{DD} ≤ 3.6 V	350	-	-	ns

Table 58. NRST pin characteristics⁽¹⁾

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^{1.} Guaranteed by design - Not tested in production.

The pull-up is designed with a true resistance in series with a switchable PMOS. This PMOS contribution to the series resistance is minimal (~10% order).

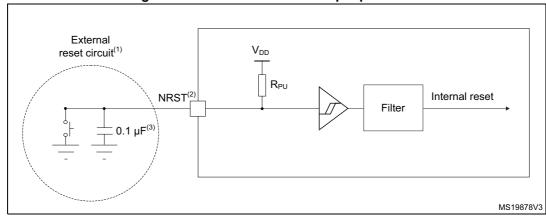


Figure 28. Recommended NRST pin protection

- 1. The reset network protects the device against parasitic resets.
- 2. The user must ensure that the level on the NRST pin can go below the V_{IL(NRST)} max level specified in *Table 58: NRST pin characteristics*. Otherwise the reset is not taken into account by the device.
- 3. The external capacitor on NRST must be placed as close as possible to the device.

5.3.16 Extended interrupt and event controller input (EXTI) characteristics

The pulse on the interrupt input must have a minimal length in order to guarantee that it is detected by the event controller.

Table 59. EXTI input characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
PLEC	Pulse length to event controller	-	20	-	-	ns

^{1.} Guaranteed by design - Not tested in production.

5.3.17 Analog switches booster

Table 60. Analog switches booster characteristics⁽¹⁾

Symbol	Parameter	Min	Тур	Max	Unit
V_{DD}	Supply voltage	1.62	-	3.6	V
t _{SU(BOOST)}	Booster startup time	-	-	240	μs
I _{DD(BOOST)}	Booster consumption for 1.62 V ≤ V _{DD} ≤ 2.0 V	-	-	250	
	Booster consumption for $2.0 \text{ V} \le \text{V}_{\text{DD}} \le 2.7 \text{ V}$	-	-	500	μΑ
	Booster consumption for $2.7 \text{ V} \le \text{V}_{DD} \le 3.6 \text{ V}$	-	-	900	

1. Guaranteed by design - Not tested in production.

5.3.18 Analog-to-digital converter characteristics

Unless otherwise specified, the parameters given in *Table 61* are preliminary values derived from tests performed under ambient temperature, f_{PCLK} frequency and V_{DDA} supply voltage conditions summarized in *Table 17: General operating conditions*.

Note: It is recommended to perform a calibration after each power-up.

Table 61. ADC characteristics⁽¹⁾ (2)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V_{DDA}	Analog supply voltage	-	1.62	-	3.6	V
.,	Positive	V _{DDA} ≥ 2 V	2	-	V_{DDA}	V
V_{REF+}	reference voltage	V _{DDA} < 2 V		V_{DDA}		V
V _{REF-}	Negative reference voltage	-		V _{SSA}		V
V _{CMIN}	Input common mode	Differential	(V _{REF+} +V _{REF-})/2 - 0.18	(V _{REF+} + V _{REF-})/2	(V _{REF+} + V _{REF-})/2 + 0.18	V
		Range 1, single ADC operation	0.14	-	60	
		Range 2	-	-	26	
	ADC clock frequency	Range 1, all ADCs operation, single ended mode $V_{DDA} \ge 2.7 \text{ V}$	0.14	- 52		
f _{ADC}		Range 1, all ADCs operation, single ended mode V _{DDA} ≥ 1.62 V	0.14	-	42	MHz
		Range 1, all ADCs operation, differential mode $V_{DDA} \ge 1.62 \text{ V}$	0.14	- 56		
f _s	Sampling rate, continuous mode	For given resolution and sampling time cycles (t _s)	0.001	f _{ADC} / (s [cycles] + r	sampling time resolution [bits] + 0.5)	Msps
T _{TRIG}	External trigger	Considering trigger conversion latency time (t _{LATR} or t _{LATRINJ})	-	- 1ms		-
	period	Resolution = 12 bits, fADC=60 MHz	tconv + [t _{LATR} or t _{LATRINJ}]			
V _{AIN} (3)	Conversion voltage range	-	0	-	V _{REF+}	V



Table 61. ADC characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit	
R _{AIN} ⁽⁴⁾	External input impedance	-	-	-	50	kΩ	
C _{ADC}	Internal sample and hold capacitor	-	-	5	-	pF	
t _{STAB}	Power-up time	-		1		conversion cycle	
4	Calibration time	f _{ADC} = 60 MHz		1.93		μs	
t _{CAL}	Calibration time	-		116		1/f _{ADC}	
	Trigger	CKMODE = 00	1.5	2	2.5		
	conversion latency Regular	CKMODE = 01	-	-	2.0		
t _{LATR}	and injected	CKMODE = 10	-	-	2.25	1/f _{ADC}	
	channels without conversion abort	CKMODE = 11	-	-	2.125		
	Trigger	CKMODE = 00	2.5	3	3.5		
	conversion latency Injected channels aborting a regular conversion	CKMODE = 01	-	-	3.0		
^t LATRINJ		CKMODE = 10	-	-	3.25	1/f _{ADC}	
		CKMODE = 11	-	-	3.125		
	a	f _{ADC} = 60 MHz	0.0416	-	10.675	μs	
t _s	Sampling time	-	2.5	-	640.5	1/f _{ADC}	
t _{ADCVREG_STUP}	ADC voltage regulator start-up time	-	-	-	20	μs	
t _{CONV}	Total conversion time (including	f _{ADC} = 60 MHz Resolution = 12 bits	0.25	-	10.883	μs	
	sampling time)	-	t _s [cycles] + resc	olution [bits]	+0.5 = 15 to 653	1/f _{ADC}	
	ADC	fs = 4 Msps	-	590	730		
I _{DDA} (ADC)	consumption from the V _{DDA}	fs = 1 Msps	-	160	220	μΑ	
	supply	fs = 10 ksps	-	16	50		
	ADC	fs = 4 Msps	-	110	140		
I _{DDV_S} (ADC)	consumption from the V _{REF+}	fs = 1 Msps	-	30	40	μΑ	
, DDV_8(, 12-6)	single ended mode	fs = 10 ksps	-	0.6	2	L.,	
	ADC	fs = 4 Msps	-	220	270		
I _{DDV_D} (ADC)	consumption	fs = 1 Msps	-	60	70	μΑ	
	differential mode	fs = 10 ksps	-	1.3	3		

^{1.} Guaranteed by design - Not tested in production.

- 2. The I/O analog switch voltage booster is enabled when V_{DDA} < 2.4 V (BOOSTEN = 1 in the SYSCFG_CFGR1 when V_{DDA} < 2.4V). It is disabled when $V_{DDA} \ge 2.4$ V.
- 3. V_{REF+} can be internally connected to V_{DDA}, depending on the package. Refer to *Section 4: Pinouts and pin description* for further details.
- 4. The maximum value of RAIN can be found in *Table 62: Maximum ADC RAIN*.



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The maximum value of R_{AIN} can be found in *Table 62: Maximum ADC RAIN*.

Table 62. Maximum ADC R_{AIN}⁽¹⁾⁽²⁾

Decelution	Sampling cycle	Sampling time	R _{AIN} m	nax (Ω)
Resolution	@60 MHz	[ns]	Fast channels ⁽³⁾	Slow channels ⁽⁴⁾
	2.5	41.67	100	N/A
	6.5	108.33	330	100
	12.5	208.33	680	470
12 bits	24.5	408.33	1500	1200
12 DIIS	47.5	791.67	2200	1800
	92.5	1541.67	4700	3900
	247.5	4125	12000	10000
	640.5	10675	39000	33000
	2.5	41.67	120	N/A
	6.5	108.33	390	180
	12.5	208.33	820	560
10 hita	24.5	408.33	1500	1200
10 bits	47.5	791.67	2200	1800
	92.5	1541.67	5600	4700
	247.5	4125	12000	10000
	640.5	10675	47000	39000
	2.5	41.67	180	N/A
	6.5	108.33	470	270
	12.5	208.33	1000	680
O bito	24.5	408.33	1800	1500
8 bits	47.5	791.67	2700	2200
	92.5	1541.67	6800	5600
	247.5	4125	15000	12000
	640.5	10675	50000	50000
	2.5	41.67	220	N/A
	6.5	108.33	560	330
	12.5	208.33	1200	1000
6 bits	24.5	408.33	2700	2200
บ มแร	47.5	791.67	3900	3300
	92.5	1541.67	8200	6800
	247.5	4125	18000	15000
	640.5	10675	50000	50000

- 1. Guaranteed by design Not tested in production.
- 2. The I/O analog switch voltage booster is enabled when V_{DDA} < 2.4 V (BOOSTEN = 1 in the SYSCFG_CFGR1 when V_{DDA} < 2.4V). It is disabled when $V_{DDA} \ge 2.4$ V.
- 3. Fast channels are: ADCx_IN1 to ADCx_IN5.
- 4. Slow channels are: all ADC inputs except the fast channels.



Table 63. ADC accuracy - limited test conditions $1^{(1)(2)(3)}$

Symbol	Parameter	Co	nditions ⁽⁴⁾		Min	Тур	Max	Unit
			Single	Fast channel (max speed)	-	5.9	6.9	
ET	Total		ended	Slow channel (max speed)	-	5.5	6.9	
<u> </u>	unadjusted error		Differential	Fast channel (max speed)	-	4.6	5.6	-
			Differential	Slow channel (max speed)	-	4	5.6	
			Single	Fast channel (max speed)	-	2.5	4	
EO	Offset error		ended	Slow channel (max speed)	-	1.9	4	
EO	Oliset elloi		Differential	Fast channel (max speed)	-	1.8	2.8	
			Dillerential	Slow channel (max speed)	-	1.1	2.8	
			Single	Fast channel (max speed)	-	4.6	6.6	
EG	Coin orror		ended	Slow channel (max speed)	-	4.5	6.6	LSB
EG Gain error		Differential	Fast channel (max speed)	-	3.6	4.6	LOD	
			Dillerential	Slow channel (max speed)	-	3.3	4.6	
			Single	Fast channel (max speed)	-	1.1	1.9	
Differential	0; 1 ADQ (; ADQ) 1	ended	Slow channel (max speed)	-	1.3	1.9		
ED	linearity error	frequency ≤ 60 MHz, V _{DDA} = VREF+ = 3 V, TA = 25 °C Continuous mode, sampling rate: Fast channels@4Msps	Differential	Fast channel (max speed)	-	1.3	1.6	- - - -
			Dillerential	Slow channel (max speed)	-	1.4	1.6	
			Single ended	Fast channel (max speed)	-	2.3	3.4	
EL	Integral			Slow channel (max speed)	-	2.4	3.4	
EL	linearity error		Differential	Fast channel (max speed)	-	2.1	3.2	
				Slow channel (max speed)	-	2.2	3.2	
			Single	Fast channel (max speed)	10.4	10.6	-	
ENOB	Effective number of		ended	Slow channel (max speed)	10.4	10.6	-	bits
ENOB	bits		Differential	Fast channel (max speed)	10.8	10.9	-	DIIS
			Dillerential	Slow channel (max speed)	10.8	10.9	-	
	Ciamal ta		Single	Fast channel (max speed)	64.4	65.6	-	
SINAD	Signal-to- noise and		ended	Slow channel (max speed)	64.4	65.6	-	
SINAD	distortion ratio		Differential	Fast channel (max speed)	66.8	67.5	-	
	Tallo		Differential	Slow channel (max speed)	66.8	67.5	-	1
			Single	Fast channel (max speed)	65	66.9	1	dB
	Signal to		ended	Slow channel (max speed)	65	66.9	1	1
SNR	Signal-to- noise ratio			Fast channel (max speed)	67	69	1	
			Differential	Slow channel (max speed)	67	69	-	

Table 63. ADC accuracy - limited test conditions $1^{(1)(2)(3)}$ (continued)

Symbol	Parameter	Co	Min	Тур	Max	Unit		
		I fraccionali / CO MIII-	Single	Fast channel (max speed)	-	-73	-72	
			ended	Slow channel (max speed)	-	-73	-72	
THD				Fast channel (max speed)	-	-73	-72	dB
			Differential	Slow channel (max speed)	-	-73	-72	

- 1. Evaluated by characterization Not tested in production.
- 2. ADC DC accuracy values are measured after internal calibration.
- 3. ADC accuracy vs. negative Injection Current: Injecting negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative current.
- 4. The I/O analog switch voltage booster is enabled when V_{DDA} < 2.4 V (BOOSTEN = 1 in the SYSCFG_CFGR1 when V_{DDA} < 2.4 V). It is disabled when V_{DDA} \geq 2.4 V. No oversampling.



Table 64. ADC accuracy - limited test conditions $2^{(1)(2)(3)}$

Sym- bol	Parameter		Conditions ⁽⁴	d test conditions 2000	Min	Тур	Max	Unit
			Single	Fast channel (max speed)	-	5.9	8.4	
	Total		ended	Slow channel (max speed)	-	5.5	8	
ET	unadjusted error		Differential	Fast channel (max speed)	-	4.6	6.6	
			Differential	Slow channel (max speed)	-	4	6	
			Single	Fast channel (max speed)	-	2.5	6	
F0	Offset error		ended	Slow channel (max speed)	-	1.9	6.9	
EO	Oliset error		Differential	Fast channel (max speed)	-	1.8	3.3	
			Differential	Slow channel (max speed)	-	1.1	3.3	
			Single	Fast channel (max speed)	-	4.6	8.1	
F0	0-1		ended	Slow channel (max speed)	-	4.5	8.1	
EG	Gain error		Differential	Fast channel (max speed)	-	3.6	4.6	LSB
			Differential	Slow channel (max speed)	-	3.3	4.6	
	Differential ED linearity Single ADC operation	Single	Fast channel (max speed)	-	1.1	1.8		
- FD		Single ADC operation ADC clock frequency	ended	Slow channel (max speed)	-	1.3	1.8	-
Eυ	error		Differential	Fast channel (max speed)	-	1.3	1.6	
		≤ 60 MHz, 2 V ≤ V _{DDA}		Slow channel (max speed)	-	1.4	1.6	
		Continuous mode, sampling rate:	Single	Fast channel (max speed)	-	2.3	4.4	
	Integral	Fast channels@4Msps	ended	Slow channel (max speed)	-	2.4	4.4	- - -
EL	linearity error	Slow channels@2Msps	D:" " !	Fast channel (max speed)	-	2.1	4.1	
			Differential	Slow channel (max speed)	-	2.2	3.7	
			Single	Fast channel (max speed)	10	10.6	-	
ENOD	Effective		ended	Slow channel (max speed)	10	10.6	-	h:to
ENOB	number of bits		Differential	Fast channel (max speed)	10.7	10.9	-	bits
			Differential	Slow channel (max speed)	10.7	10.9	-	
	0:11-		Single	Fast channel (max speed)	62	65.6	-	
CINIAD	Signal-to- noise and		ended	Slow channel (max speed)	62	65.6	-	
SINAD	distortion		Differential	Fast channel (max speed)	65	67.5	-	
	ratio		Differential	Slow channel (max speed)	65	67.5	-	
			Single	Fast channel (max speed)	64	66.9	-	dB
CNID	Signal-to-		ended	Slow channel (max speed)	64	66.9	-	-
SNR	noise ratio		Differential	Fast channel (max speed)	66.5	69	-	
			Differential	Slow channel (max speed)	66.5	69	-	

Table 64. ADC accuracy - limited test conditions $2^{(1)(2)(3)}$ (continued)

Sym- bol	Parameter		Min	Тур	Max	Unit		
		Single ADC operation ADC clock frequency ≤ 60 MHz, 2 V ≤ V _{DDA} Continuous mode, sampling	Single	Fast channel (max speed)	-	-73	-65	
	Total		ended	Slow channel (max speed)	-	-73	-67	
THD	harmonic		Differential	Fast channel (max speed)	-	-73	-70	dB
	distortion	rate: Fast channels@4Msps Slow channels@2Msps		Slow channel (max speed)	-	-73	-71	

- 1. Evaluated by characterization Not tested in production.
- 2. ADC DC accuracy values are measured after internal calibration.
- 3. ADC accuracy vs. negative Injection Current: Injecting negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative current.
- 4. The I/O analog switch voltage booster is enabled when V_{DDA} < 2.4 V (BOOSTEN = 1 in the SYSCFG_CFGR1 when V_{DDA} < 2.4 V). It is disabled when $V_{DDA} \ge 2.4$ V. No oversampling.



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Table 65. ADC accuracy - limited test conditions $3^{(1)(2)(3)}$

Sym- bol	Parameter		Conditions	s ⁽⁴⁾	Min	Тур	Max	Unit
			Single	Fast channel (max speed)	-	5.9	7.9	
	Total		ended	Slow channel (max speed)	-	5.5	7.5	
ET	unadjusted error		D:#fanantial	Fast channel (max speed)	-	4.6	7.6	
			Differential	Slow channel (max speed)	-	4	5.5	
			Single	Fast channel (max speed)	-	2.5	5.5	
EO	Offset error		ended	Slow channel (max speed)	-	1.9	5.5	
EO	Oliset error		Differential	Fast channel (max speed)	-	1.8	3.5	
			Differential	Slow channel (max speed)	-	1.1	3	
			Single	Fast channel (max speed)	-	4.6	7.1	
F0	Cain arrar		ended	Slow channel (max speed)	-	4.5	7	LOD
EG	Gain error		Differential	Fast channel (max speed)	-	3.6	4.1	LSB
			Dillerential	Slow channel (max speed)	-	3.3	4.8	
Differential ED linearity error		Single ended	Fast channel (max speed)	-	1.1	1.9		
	Single ADC operation		Slow channel (max speed)	-	1.3	1.9		
	1	ADC clock frequency ≤ 60 MHz,	Differential	Fast channel (max speed)	-	1.3	1.6	-
		$1.62 \text{ V} \le \text{V}_{DDA} = \text{V}_{REF+}$ $\le 3.6 \text{ V},$	Dillerential	Slow channel (max speed)	-	1.4	1.6	
		Continuous mode, sampling rate:	Single ended	Fast channel (max speed)	-	2.3	4.4	
EL	Integral linearity			Slow channel (max speed)	-	2.4	4.4	
	error	Fast channels@4Msps Slow channels@2Msps	Differential	Fast channel (max speed)	-	2.1	3.7	
			Differential	Slow channel (max speed)	-	2.2	3.7	
			Single	Fast channel (max speed)	10	10.6	-	
ENOB	Effective number of		ended	Slow channel (max speed)	10	10.6	-	bits
ENOB	bits		Differential	Fast channel (max speed)	10.6	10.9	-	טונס
			Dillerential	Slow channel (max speed)	10.6	10.9	-	
	Cianal to		Single	Fast channel (max speed)	62	65.6	-	
SINAD	Signal-to- noise and		ended	Slow channel (max speed)	62	65.6	-	
SINAD	distortion		Differential	Fast channel (max speed)	65	67.5	-	dB
	ratio		Dillerential	Slow channel (max speed)	65	67.5	-	
			Single	Fast channel (max speed)	63	66.9	-	
SNR	Signal-to-		ended	Slow channel (max speed)	63	66.9	-	
SINK	noise ratio		Differential	Fast channel (max speed)	66	69	-	
			חווכוכווומו	Slow channel (max speed)	66	69	-	

Table 65. ADC accuracy - limited test conditions $3^{(1)(2)(3)}$ (continued)

Sym- bol	Parameter	Conditions ⁽⁴⁾				Тур	Max	Unit	
THD	Total harmonic distortion	Single ADC operation ADC clock frequency ≤ 60 MHz, 1.62 V ≤ V _{DDA} = V _{REF+}	Single ended	Fast channel (max speed)	-	-73	-67		
				Slow channel (max speed)	-	-73	-67		
			Differential	Fast channel (max speed)	-	-73	-71		
		≤ 3.6 V, Continuous mode, sampling rate: Fast channels@4Msps Slow channels@2Msps		Slow channel (max speed)	-	-73	-71	dB	

- 1. Evaluated by characterization Not tested in production.
- 2. ADC DC accuracy values are measured after internal calibration.
- 3. ADC accuracy vs. negative Injection Current: Injecting negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative current.
- 4. The I/O analog switch voltage booster is enabled when V_{DDA} < 2.4 V (BOOSTEN = 1 in the SYSCFG_CFGR1 when V_{DDA} < 2.4 V). It is disabled when $V_{DDA} \ge 2.4$ V. No oversampling.



Table 66. ADC accuracy (Multiple ADCs operation) - limited test conditions 1⁽¹⁾⁽²⁾⁽³⁾

Symbol	Parameter	Conditions ⁽⁴⁾		Min	Тур	Max	Unit
ET	Total unadjusted		Single ended	-	4.5	-	- LSB
	error		Differential	-	4.1	-	
EO	Offset error		Single ended	-	1.3	-	
	Oliset elloi		Differential	-	0.4	-	
EG	Gain error		Single ended	-	3.9	-	
	Gairrerror	Multiple ADC operation	Differential	-	3.4	-	
ED	Differential	ADC clock frequency:	Single ended	-	1.5	-	
	linearity error	single ended ≤ 52 MHz, differential ≤ 56 MHz,	Differential	-	1.2	-	
EL	Integral linearity	$V_{DDA} = V_{REF} = 3.3 V$	Single ended	-	1.7	-	
EL	error	25°C, Continuous mode, sampling time: Fast channels: 2.5 cycles Slow channels: 6.5 cycles LQFP100 package	Differential	-	2.1	-	
ENOB	Effective		Single ended	-	10.7	-	- bits
ENOB	number of bits		Differential	-	10.9	-	
	Signal-to-noise		Single ended	-	66.3	-	. dB
SINAD	and distortion ratio		Differential	-	67.2	-	
SNR	Signal-to-noise		Single ended	-	67.3	-	
SINK	ratio		Differential	-	68.6	-	
THD	Total harmonic		Single ended	-	-73.5	-	- dB
	distortion		Differential	-	-73	-	

^{1.} Data based on characterization result, not tested in production.

^{2.} ADC DC accuracy values are measured after internal calibration.

^{3.} ADC accuracy vs. negative Injection Current: Injecting negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative current.

^{4.} The I/O analog switch voltage booster is enabled when V_{DDA} < 2.4 V (BOOSTEN = 1 in the SYSCFG_CFGR1 when V_{DDA} < 2.4 V). It is disabled when $V_{DDA} \ge 2.4$ V. No oversampling.

Table 67. ADC accuracy (Multiple ADCs operation) - limited test conditions 2⁽¹⁾⁽²⁾⁽³⁾

Symbol	Parameter	Conditions ⁽⁴⁾		Min	Тур	Max	Unit	
ET	Total unadjusted		Single ended	-	7.1	-		
E1	error		Differential	-	4.6	-		
EO	Offset error		Single ended	-	4.2	-		
EO	Offset error		Differential	-	2.8	-		
EG	Gain error		Single ended	-	6.8	-	LSB	
EG	Gain enoi	Multiple ADC operation ADC clock frequency: single ended ≤ 52 MHz, differential ≤ 56 MHz,	Differential	-	4.3	-	LOD	
ED	Differential		Single ended	-	1.5	-		
ED	linearity error		Differential	-	1.7	-		
EL	Integral linearity	V _{DDA} ≥ 2.7 V, V _{REF} ≥ 1.62 V,	Single ended	-	3.1	-		
EL	error	-40 to 125°C, Continuous mode,	Differential	-	2.4	-		
ENOB	Effective	sampling time:	Single ended	-	10.2	-	bits	
ENOB	number of bits	Fast channels: 2.5 cycles Slow channels: 6.5 cycles	Differential	-	10.6	-	Dits	
	Signal-to-noise	LQFP100 package	Single ended	-	62.9	-		
SINAD	and distortion ratio		Differential	-	65.3	-	dB	
SNR	Signal-to-noise		Single ended	-	63.6	-		
SINK	ratio		Differential	-	66.3	-		
THD	Total harmonic		Single ended	-	-70.9	-	dB	
טחו	distortion	Differential	-	-71.8	-	ub		

^{1.} Data based on characterization result, not tested in production.

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^{2.} ADC DC accuracy values are measured after internal calibration.

^{3.} ADC accuracy vs. negative Injection Current: Injecting negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative current.

^{4.} The I/O analog switch voltage booster is enabled when V_{DDA} < 2.4 V (BOOSTEN = 1 in the SYSCFG_CFGR1 when V_{DDA} < 2.4 V). It is disabled when $V_{DDA} \ge 2.4$ V. No oversampling.

Table 68. ADC accuracy (Multiple ADCs operation) - limited test conditions 3⁽¹⁾⁽²⁾⁽³⁾

Symbol	Parameter	Conditions ⁽⁴	1)	Min	Тур	Max	Unit
ET	Total unadjusted		Single ended	-	7.4	-	
	error		Differential	-	4.6	-	
EO	Offset error		Single ended	-	4	-	
EO	Offset effor		Differential	-	2.8	-	
EG	Gain error		Single ended	-	7.2	-	LSB
EG	Gain enoi	Multiple ADC operation	Differential	-	4.3	-	
ED	Differential	single ended ≤ 42 MHz, differential ≤ 56 MHz, V _{DDA} = V _{REF} ≥ 1.62 V,	Single ended	-	1.8	-	
ED	linearity error		Differential	-	1.7	-	
EL	Integral linearity		Single ended	-	3.1	-	
EL	error	-40 to 125°C, Continuous mode,	Differential	-	2.4	-	
ENOB	Effective	sampling time:	Single ended	-	10.1	-	bits
ENOB	number of bits	Fast channels: 2.5 cycles Slow channels: 6.5 cycles	Differential	-	10.6	-	DILS
	Signal-to-noise	LQFP100 package	Single ended	-	62.6	-	
SINAD	and distortion ratio		Differential	-	65.3	-	dB
CND	SNR Signal-to-noise ratio		Single ended	-	63.2	-	45
SINK			Differential	-	66.3	-	
THD	Total harmonic		Single ended	-	-70.6	-	dB
טחו	distortion	·	Differential	-	-71.8	-	uБ

^{1.} Data based on characterization result, not tested in production.

^{2.} ADC DC accuracy values are measured after internal calibration.

^{3.} ADC accuracy vs. negative Injection Current: Injecting negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative current.

^{4.} The I/O analog switch voltage booster is enabled when V_{DDA} < 2.4 V (BOOSTEN = 1 in the SYSCFG_CFGR1 when V_{DDA} < 2.4 V). It is disabled when $V_{DDA} \ge 2.4$ V. No oversampling.

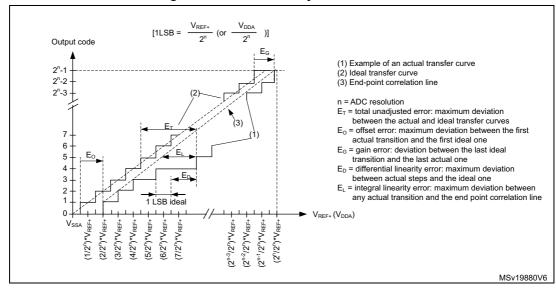
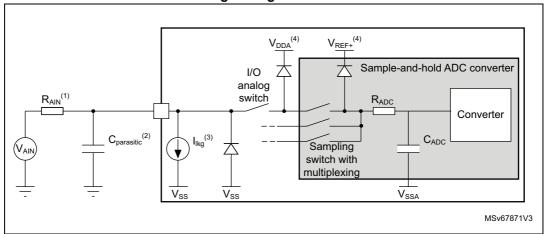


Figure 29. ADC accuracy characteristics

Figure 30. Typical connection diagram when using the ADC with FT/TT pins featuring analog switch function



- 1. Refer to Table 61: ADC characteristics for the values of R_{AIN} and C_{ADC} .
- C_{parasitic} represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the
 pad capacitance (refer to *Table 54: I/O static characteristics* for the value of the pad capacitance). A high
 C_{parasitic} value downgrades conversion accuracy. To remedy this, f_{ADC} should be reduced.
- 3. Refer to Table 54: I/O static characteristics for the values of I_{lkq}.
- 4. Refer to Figure 17: Power supply scheme.

General PCB design guidelines

Power supply decoupling must be performed as shown in *Figure 17: Power supply scheme*. The decoupling capacitor on V_{DDA} must be ceramic (good quality) and it must be placed as close as possible to the chip.

4

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5.3.19 Digital-to-Analog converter characteristics

Table 69. DAC 1MSPS characteristics⁽¹⁾

Symbol	Parameter	Co	onditions	Min	Тур	Max	Unit
V_{DDA}	Analog supply voltage for DAC ON	DAC output buffer OFF, DAC_OUT pin not connected (internal connection only)		1.71	-	3.6	
		Other modes		1.80	-		
V _{REF+}	Positive reference voltage	DAC output bu pin not connec connection only		1.71	-	V _{DDA}	V
		Other modes		1.80	-		
V _{REF-}	Negative reference voltage		-		V _{SSA}		
R_L	Resistive load	DAC output	connected to V _{SSA}	5	-	-	kΩ
IXL	Tresistive load	buffer ON	connected to V _{DDA}	25	-	-	KSZ
R _O	Output Impedance	DAC output bu	9.6	11.7	13.8	kΩ	
В	Output impedance sample and hold mode, output	V _{DD} = 2.7 V		-	-	2	kO.
R _{BON}	buffer ON	V _{DD} = 2.0 V		-	-	3.5	kΩ
_	Output impedance sample	V _{DD} = 2.7 V		-	-	16.5	
R _{BOFF}	and hold mode, output buffer OFF	V _{DD} = 2.0 V		-	-	18.0	kΩ
C_L	One anitive land	DAC output buffer ON		-	-	50	pF
C _{SH}	- Capacitive load	Sample and ho	old mode	-	0.1	1	μF
V _{DAC_OUT}	Voltage on DAC_OUT	DAC output bu	ffer ON	0.2	-	V _{REF+} - 0.2	V
	output	DAC output bu	ffer OFF	0	-	V _{REF+}	
			±0.5 LSB	-	1.7	3	
	Settling time (full scale: for	Normal mode DAC output	±1 LSB	-	1.6	2.9	
	a 12-bit code transition between the lowest and the	buffer ON	±2 LSB	-	1.55	2.85	
t _{SETTLING}	highest input codes when	CL ≤ 50 pF, RL ≥ 5 kΩ	±4 LSB	-	1.48	2.8	μs
	DAC_OUT reaches final value)		±8 LSB	-	1.4	2.75	
	,	Normal mode I OFF, ±1LSB, C	DAC output buffer CL = 10 pF	-	2	2.5	
. (2)	Wakeup time from off state (setting the ENx bit in the	Normal mode DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ		-	4.2	7.5	
t _{WAKEUP} ⁽²⁾	DAC Control register) until final value ±1 LSB	Normal mode DAC output buffer OFF, CL ≤ 10 pF		-	2 5	μs	
PSRR	V _{DDA} supply rejection ratio	Normal mode [CL ≤ 50 pF, RL	DAC output buffer ON . = 5 kΩ, DC	-	-80	-28	dB

Table 69. DAC 1MSPS characteristics⁽¹⁾ (continued)

Symbol	Parameter	Co	onditions	Min	Тур	Max	Unit
T _{W_to_W}	Minimal time between two consecutive writes into the DAC_DORx register to guarantee a correct DAC_OUT for a small variation of the input code (1 LSB) DAC_MCR:MODEx[2:0] = 000 or 001 DAC_MCR:MODEx[2:0] = 010 or 011	CL ≤ 50 pF, RL CL ≤ 10 pF	· · · · · · · · · · · · · · · · · · ·			-	μѕ
		DAC_OUT	DAC output buffer ON, C _{SH} = 100 nF	-	0.7	3.5	me
	Sampling time in sample and hold mode (code transition between the	pin connected	DAC output buffer OFF, C _{SH} = 100 nF	-	10.5	18	ms
t _{SAMP} lowest input code ar highest input code w	lowest input code and the highest input code when DACOUT reaches final	DAC_OUT pin not connected (internal connection only)	DAC output buffer OFF	-	2	3.5	μs
I _{leak}	Output leakage current	Sample and ho DAC_OUT pin		-	-	_(3)	nA
Cl _{int}	Internal sample and hold capacitor		-	5.2	7	8.8	pF
t _{TRIM}	Middle code offset trim time	DAC output bu	ffer ON	50	-	-	μs
V	Middle code offset for 1 trim	V _{REF+} = 3.6 V		-	1500	-	μV
V _{offset}	code step	V _{REF+} = 1.8 V		-	750	-	μν
		DAC output	No load, middle code (0x800)	-	315	500	
		buffer ON	No load, worst code (0xF1C)	-	450	670	
I _{DDA} (DAC)	DAC consumption from V _{DDA}	DAC output No load, middle buffer OFF code (0x800)		-	-	0.2	μA
		Sample and hold mode, C _{SH} = 100 nF		-	315 x Ton/(Ton +Toff) (4)	670 x Ton/(Ton +Toff) (4)	



Symbol	Parameter	Co	onditions	Min	Тур	Max	Unit
	DAC consumption from V _{REF+}	DAC output		-	185	240	
		buffer ON	No load, worst code (0xF1C)	-	340	400	
		DAC output buffer OFF No load, middle code (0x800)		-	155	205	
I _{DDV} (DAC)		Sample and hold mode, buffer ON, C _{SH} = 100 nF, worst case		-	185 _x Ton/(Ton +Toff) (4)	400 x Ton/(Ton +Toff) (4)	μΑ
		Sample and hold mode, buffer OFF, C _{SH} = 100 nF, worst case			155 x Ton/(Ton +Toff) (4)	205 x Ton/(Ton +Toff) (4)	

Table 69. DAC 1MSPS characteristics⁽¹⁾ (continued)

- 1. Guaranteed by design Not tested in production.
- 2. In buffered mode, the output can overshoot above the final value for low input code (starting from min value).
- 3. Refer to Table 54: I/O static characteristics.
- Ton is the Refresh phase duration. Toff is the Hold phase duration. Refer to the reference manual RM0440 "STM32G4 Series advanced Arm®-based 32-bit MCUs" for more details.

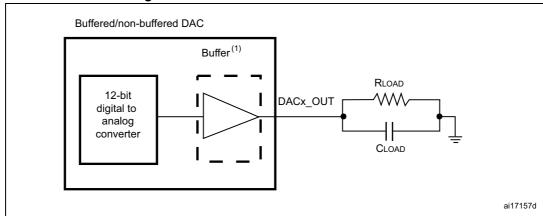


Figure 31. 12-bit buffered / non-buffered DAC

The DAC integrates an output buffer to reduce the output impedance and to drive external loads directly
without the use of an external operational amplifier. The buffer can be bypassed by configuring the BOFFx
bit in the DAC_CR register.

Table 70. DAC 1MSPS accuracy⁽¹⁾

Symbol	Parameter	Conditio	ns	Min	Тур	Max	Unit
DAII	Differential non	DAC output buffer ON		-	-	±2	
DNL	linearity (2)	DAC output buffer OFF		-	-	±2	
-	monotonicity	10 bits		(Guarantee	d	
INL	Integral non	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ		-	-	±4	
INL	linearity ⁽³⁾	DAC output buffer OFF CL ≤ 50 pF, no RL		-	-	±4	
		DAC output buffer ON	V _{REF+} = 3.6 V	-	-	±12	1.00
Offset	Offset error at code 0x800 ⁽³⁾	CL ≤ 50 pF, RL ≥ 5 kΩ	V _{REF+} = 1.8 V	-	-	±25	LSB
		DAC output buffer OFF CL ≤ 50 pF, no RL		-	-	±8	
Offset1	Offset error at code 0x001 ⁽⁴⁾	DAC output buffer OFF CL ≤ 50 pF, no RL	_	-	-	±5	
Offset Error at code 0x800	DAC output buffer ON	V _{REF+} = 3.6 V	-	-	±5		
OlisetGal	after calibration	CL ≤ 50 pF, RL ≥ 5 kΩ	V _{REF+} = 1.8 V	-	-	±7	
Coin	Gain error ⁽⁵⁾	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ		-	-	±0.5	%
Gain	Gain enois?	DAC output buffer OFF CL ≤ 50 pF, no RL		-	-	±0.5	70
TUE	Total unadjusted	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ		-	-	±30	LSB
TOE	error	DAC output buffer OFF CL ≤ 50 pF, no RL		-	-	±12	LOD
TUECal	Total unadjusted error after calibration	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ		-	-	±23	LSB
SNR	Signal-to-noise	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ 1 kHz, BW 500 kHz		-	71.2	-	٩Đ
SINK	ratio	DAC output buffer OFF CL ≤ 50 pF, no RL, 1 kHz BW 500 kHz		-	71.6	-	- dB
THD	Total harmonic	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ, 1	kHz	-	-78	-	dB
וחט	distortion	DAC output buffer OFF CL ≤ 50 pF, no RL, 1 kHz		-	-79	-	uΔ



Table 70. DAC 1MSPS	accuracy ⁽¹⁾ ((continued)
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Symbol	Parameter	Conditions	Min	Тур	Max	Unit	
SINAD	Signal-to-noise and distortion ratio	DAC output buffer ON CL \leq 50 pF, RL \geq 5 k Ω , 1 kHz	-	70.4	-	dB	
		DAC output buffer OFF CL ≤ 50 pF, no RL, 1 kHz	-	71	-		
I FNOR I	Effective number of bits	DAC output buffer ON CL \leq 50 pF, RL \geq 5 k Ω , 1 kHz	-	11.4	-	hita	
		DAC output buffer OFF CL ≤ 50 pF, no RL, 1 kHz	-	11.5	-	bits	

- 1. Guaranteed by design Not tested in production.
- 2. Difference between two consecutive codes 1 LSB.
- 3. Difference between measured value at Code i and the value at Code i on a line drawn between Code 0 and last Code 4095.
- 4. Difference between the value measured at Code (0x001) and the ideal value.
- 5. Difference between ideal slope of the transfer function and measured slope computed from code 0x000 and 0xFFF when buffer is OFF, and from code giving 0.2 V and $(V_{REF+} 0.2)$ V when buffer is ON.

Table 71. DAC 15MSPS characteristics⁽¹⁾

Symbol	Parameter	Conditions	;	Min	Тур	Max	Unit	
V_{DDA}	Analog supply voltage for DAC ON	-		1.71	-	3.6		
V _{REF+}	Positive reference voltage	-		1.71	-	V_{DDA}	V	
V _{REF-}	Negative reference voltage	-			V _{SSA}			
V _{DAC_OUT}	Voltage on DAC_OUT output	-		0	-	V _{REF+}	V	
		V _{DDA} >2,7V With One comparator on DAC output	10%-90%	-	16	22		
			5%-95%	-	21	29		
			1%-99%	-	33	46		
	Settling time (full scale: for		32lsb	-	40	53		
4	a 12-bit code transition between the lowest and the		1lsb	-	64	87	20	
t _{SETTLING}	highest input codes when DAC OUT reaches final		10%-90%	-	24	32	ns	
	value)	V _{DDA} >2,7V	5%-95%	-	32	43		
		With One comparator and OPAMP on DAC	1%-99%	-	49	67		
		output	32lsb	-	57	75		
			1lsb	-	93	125		

Table 71. DAC 15MSPS characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions	S	Min	Тур	Max	Unit
			10%-90%	-	16	88	
		V _{DDA} <2,7V	5%-95%	-	21	116	
		With One comparator	1%-99%	-	33	181	
	Settling time (full scale: for a 12-bit code transition	on DAC output	32lsb	-	40	196	
+	between the lowest and the		1lsb	-	64	332	ns
t _{SETTLING}	highest input codes when DAC_OUT reaches final		10%-90%	-	24	128	115
	value)	V _{DDA} <2,7V	5%-95%	-	32	170	
		With One comparator and OPAMP on DAC	1%-99%	-	49	265	
		output	32lsb	-	57	284	
		1Is	1lsb	-	93	483	
t _{WAKEUP} ⁽²⁾	Wakeup time from off state (setting the ENx bit in the DAC Control register) until final value ±1 LSB	Normal mode CL ≤ 10 p	-	1.4	3.5	μs	
DODD	V aumply rejection ratio	V _{DD} > 2.7 V		65	85	-	٩D
PSRR	V _{DDA} supply rejection ratio	V _{DD} <2.7 V		40	85	-	- dB
^t SAMP	Sampling time in sample and hold mode (code transition between the lowest input code and the highest input code when DACOUT reaches final value ±1LSB)	-		-	0.7	-	μs
Cl _{int}	Internal sample and hold capacitor	-		-	4	5	pF
dV/dt (hold phase)	Voltage decay rate in Sample and hold mode, during hold phase	CSH = 4 pF T = 55°C		-	50	-	mV/ms
I _{DDA} (DAC)	DAC consumption from V_{DDA}	No load, middle code (0x800)		-	-	0.2	μA
I _{DDV} (DAC)	DAC consumption from V _{REF+}	No load, middle code (0	0x800) ⁽³⁾	-	720	955	μΛ

^{1.} Guaranteed by design - Not tested in production.

^{2.} In buffered mode, the output can overshoot above the final value for low input code (starting from min value).

^{3.} Worst case consumption is at code 0x800.

Table 72. DAC 15MSPS accuracy⁽¹⁾

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
DNL	Differential non linearity (2)	-	-2	-	2	
INL	Integral non linearity ⁽³⁾	CL ≤ 50 pF, no RL	-5	-	5	
TUE	Total unadjusted error	CL ≤ 50 pF, no RL	-5	-	5	LSB
DCS	Dynamic code spike	Spike amplitude on DAC voltage when DAC output value is decreasing	-	0	4	

- 1. Guaranteed by design Not tested in production.
- 2. Difference between two consecutive codes 1 LSB.
- 3. Difference between measured value at code i and the value at code i on a line drawn between code 0 and last code 4095. Offset error is included.

5.3.20 Voltage reference buffer characteristics

Table 73. VREFBUF characteristics⁽¹⁾

Symbol	Parameter	Conditio	ns	Min	Тур	Max	Unit
			VRS = 00	2.4	-	3.6	
		Normal mode	VRS = 01	2.8	-	3.6	
V	Analog supply		VRS = 10	3.135	-	3.6	
V_{DDA}	voltage		VRS= 00	1.65	-	2.4	
		Degraded mode ⁽²⁾	VRS = 01	1.65	-	2.8	
			VRS= 10	1.65	-	3.135	1 ,
			VRS= 00	2.044	2.048	2.052	V
		Normal mode ⁽³⁾	VRS= 01	2.496	2.5	2.504	
V _{REFBUF_}	Voltage reference output		VRS = 10	2.896	2.9	2.904	
OUT			VRS= 00	V _{DDA} -250 mV	-	V_{DDA}	
		Degraded mode ⁽²⁾	VRS = 01	V _{DDA} -250 mV	-	V_{DDA}	
			VRS = 10	V _{DDA} -250 mV	-	V_{DDA}	
V _{REFOUT}	Voltage reference output spread over the temperature range	V _{DDA} = 3V		-	-	See Figure 32, Figure 33, Figure 34	mV
TRIM	Trim step resolution	-		-	±0.05	±0.1	%
CL	Load capacitor	-		0.5	1	1.5	μF
esr	Equivalent Serial Resistor of Cload	-		-	-	2	Ω
I _{load}	Static load current	-		-	-	6.5	mA
I _{line_reg} (4)	Line regulation	-		-	1000	2000	ppm/V
I _{load_reg}	Load regulation	500 μA ≤ I _{load} ≤4 mA	Normal mode	-	50	500	ppm/m A
	Temperature	-40 °C < TJ < +125	°C	-	-	Tcoeff_vr	
T _{Coeff}	coefficient	0 °C < TJ < +50 °C		-	-	efint + 50 ⁽⁵⁾	ppm/ °C
DCDD	Power supply	DC 100 kHz		40	55	-	٩D
PSRR	rejection			25	40	-	dB
		$CL = 0.5 \mu F^{(6)}$		-	300	350	
t _{START}	Start-up time	$CL = 1.1 \ \mu F^{(6)}$		-	500	650	μs
		$CL = 1.5 \mu F^{(6)}$		-	650	800	



Symbol	Parameter	Conditions	Min	Тур	Max	Unit
Inrush	Control of maximum DC current drive on VREFBUF_ OUT during start- up phase (7)	-	-	8	-	mA
		I _{load} = 0 μA	-	16	25	
I _{DDA} (VREF	VREFBUF consumption from	I _{load} = 500 μA	-	18	30	μA
BUF)	V _{DDA}	I _{load} = 4 mA	-	35	50	μΑ
		I _{load} = 6.5 mA	-	45	80	

Table 73. VREFBUF characteristics⁽¹⁾ (continued)

- 1. Guaranteed by design, unless otherwise specified.
- In degraded mode, the voltage reference buffer can not maintain accurately the output voltage which follows (V_{DDA} drop voltage).
- 3. Guaranteed by characterization results.
- 4. Line regulation is given for overall supply variation, in normal mode.
- 5. Tcoeff_vrefint refer to Tcoeff parameter in the embedded voltage reference section.
- 6. The capacitive load must include a 100 nF low ESR capacitor in order to cut-off the high frequency noise.
- To correctly control the VREFBUF inrush current during start-up phase and scaling change, the V_{DDA} voltage should be in the range [2.4 V to 3.6 V], [2.8 V to 3.6 V] and [3.135 V to 3.6 V] respectively for VRS=0,1 and 2.

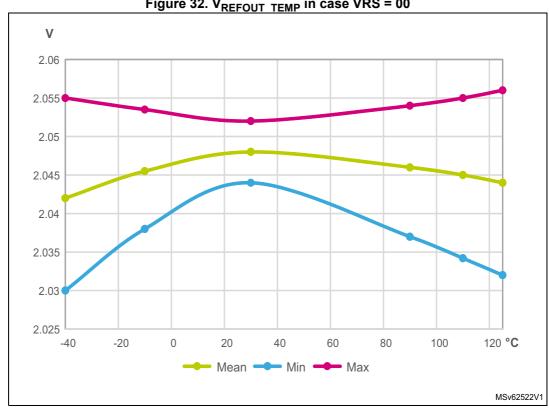


Figure 32. V_{REFOUT TEMP} in case VRS = 00

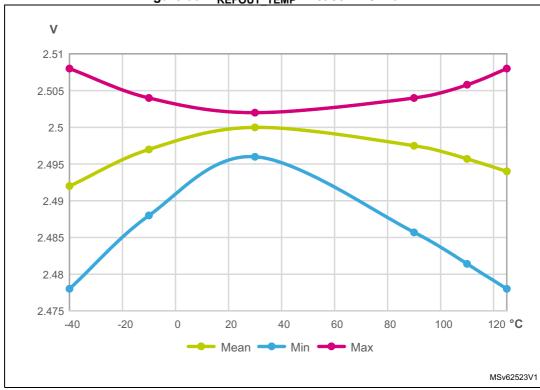
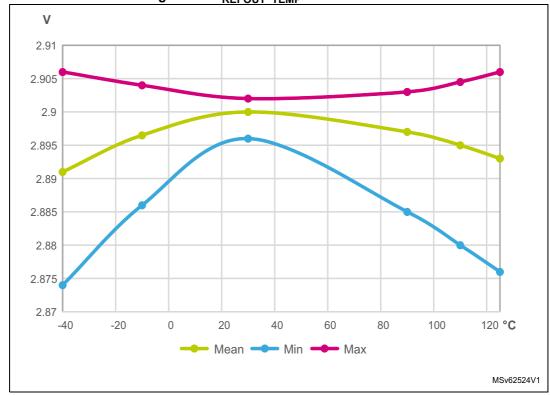


Figure 33. $V_{REFOUT\ TEMP}$ in case VRS = 01





5.3.21 Comparator characteristics

Table 74. COMP characteristics⁽¹⁾

Symbol	Parameter	Conc	litions	Min	Тур	Max	Unit
V_{DDA}	Analog supply voltage		-	1.62	-	3.6	
V _{IN}	Comparator input voltage range		-	0	-	V_{DDA}	V
V _{BG} ⁽²⁾	Scaler input voltage	-		\	/REFIN	Γ	
V _{SC} ⁽³⁾	Scaler offset voltage		-	-	±5	±10	mV
I (SCALED)	Scaler static consumption from	BRG_EN=0 (br	idge disable)	-	200	300	nA
I _{DDA} (SCALER)	V_{DDA}	BRG_EN=1 (bri	idge enable)	-	8.0	1	μΑ
t _{START_SCALER}	Scaler startup time		-	-	100	200	μs
t _{START}	Comparator startup time to reach propagation delay specification	-		-	-	5	μs
t _D ⁽⁴⁾	Propagation delay (From COMP input pin to COMP output pin) for 200 mV step with 100 mV overdrive		V _{DDA} < 2.7 V	-	-	35	ns
		50pF load on output	V _{DDA} ≥2.7 V	-	16.7	31	ns
V _{offset} ⁽³⁾	Comparator offset error	Full V _{DDA} voltage temperature rar		-9	-6/+2	3	mV
		HYST[2:0] = 0		-	0	-	
		HYST[2:0] =1		4	9	16	
		HYST[2:0] = 2		7	18	32	
V	Comparator hysteresis	HYST[2:0] = 3		11	27	47	mV
V_{hys}	Comparator hysteresis	HYST[2:0] = 4		15	36	63	
		HYST[2:0] = 5		19	45	79	
		HYST[2:0] = 6		23	54	95	
		HYST[2:0] = 7		26	63	110	
	Comparator consumption from	Static		-	450	720	
I _{DDA} (COMP)	V _{DDA}	With 50 kHz ±100 mV overdrive square signal		-	450	-	μΑ

^{1.} Guaranteed by design, unless otherwise specified.

^{2.} Refer to Table 20: Embedded internal voltage reference.

^{3.} Guaranteed by characterization results.

^{4.} Typical value (3V) is an average for all comparators propagation delay.

5.3.22 Operational amplifiers characteristics

Table 75. OPAMP characteristics⁽¹⁾ (2)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V_{DDA}	Analog supply voltage	-	2	3.3	3.6	V
CMIR	Common mode input range	-	0	-	V_{DDA}	V
VI (3)	Input offset voltage	25 °C, No Load on output.	-	-	±1.5	m\/
VI _{OFFSET} ⁽³⁾	Input offset voltage	All voltage/temperature.	-	-	±3	mV
ΔVI _{OFFSET}	Input offset voltage drift	-	-	±10	-	μV/°C
TRIMOFFSE TP	Offset trim step at low common input voltage (0.1 x V _{DDA})	-	-	1.1	1.2	mV
TRIMOFFSE TN	Offset trim step at high common input voltage (0.9 x V _{DDA})	-	-	1.3	1.65	liiv
I _{LOAD}	Drive current	-	-	-	500	
I _{LOAD_PGA}	Drive current in PGA mode	-	-	-	270	μA
C _{LOAD}	Capacitive load	-	-	-	50	pF
CMRR	Common mode rejection ratio	-	-	60	-	dB
PSRR	Power supply rejection ratio	$C_{LOAD} \le 50 \text{ pf},$ $R_{LOAD} \ge 4 \text{ k}\Omega \text{ DC Vcom=V}_{DDA}/2$	-	80	-	dB
GBW	Gain Bandwidth Product	100mV ≤ Output dynamic range ≤ V _{DDA} - 100mV	7	13	-	MHz
op(3)	Slew rate	Normal mode	2.5	6.5	-	.,,
SR ⁽³⁾	(from 10 and 90% of output voltage)	High-speed mode	18	45	-	V/µs
40		100mV ≤ Output dynamic range ≤ V _{DDA} - 100mV	65	95	-	ID.
AO	Open loop gain	200mV ≤ Output dynamic range ≤ V _{DDA} ₋ 200mV	75	95	-	dB
V _{OHSAT} ⁽³⁾	High saturation voltage	I_{load} = max or R_{load} = min Input at V_{DDA} . Follower mode	V _{DDA} - 100	-	-	m\/
V _{OLSAT} ⁽³⁾	Low saturation voltage	I _{load} = max or R _{load} = min Input at 0. Follower mode		-	100	- mV
ϕ_{m}	Phase margin	Follower mode, Vcom=V _{DDA} /2	-	65	-	0
GM	Gain margin	Follower mode, Vcom=V _{DDA} /2	-	10	_	dB



Table 75. OPAMP characteristics⁽¹⁾ (continued)

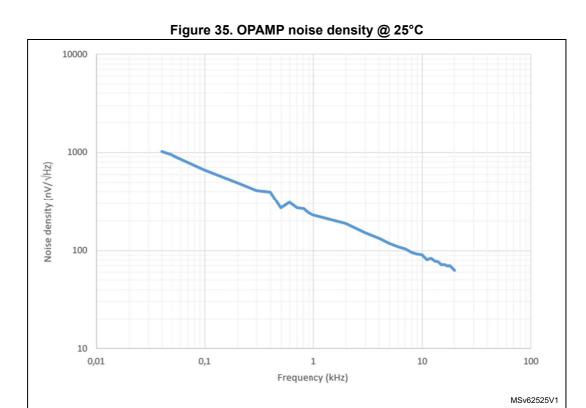
Symbol	Parameter	Conditions	3	Min	Тур	Max	Unit
	Wake up time from	Normal mode	$C_{LOAD} \le 50 \text{ pf},$ $R_{LOAD} \ge 4 \text{ k}\Omega$ follower configuration	-	3	6	
t _{WAKEUP} ⁽³⁾	Wake up time from OFF state.	High-speed mode	$C_{LOAD} \le 50 \text{ pf},$ $R_{LOAD} \ge$ $20 \text{ k}\Omega$ follower configuration	-	3	6	μs
I _{bias}	OPAMP input bias current	See I _{leak} parameter in To	able 54: I/O statio	c charac	cteristics f	or given	pin.
	Non inverting gain value ⁽⁴⁾	PGA Gain = 2 0.1 ≤ Out	V _{DDA} < 2.2	-2	-	2	
		dynamic range ≤ V _{DDA} - 0.1	V _{DDA} ≥ 2.2	-1	-	1	
		PGA Gain=4, 100mV ≤ Output dynamic range ≤ V _{DDA} - 100mV		-1	-	1	%
		PGA Gain=8 100mV ≤ Output dynamic range ≤ V _{DDA} - 100mV		-1	-	1	
		PGA Gain=16, 100mV ≤ Output dynamic range ≤ V _{DDA} - 100mV		-1	-	1	
		PGA Gain=32 200mV ≤ Output ≤ V _{DDA} - 200mV		-2	-	2	
DCA main		PGA Gain=64 200mV ≤ Output dynamic range ≤ V _{DDA} - 200mV		-2	-	2	
PGA gain		PGA Gain = -1	V _{DDA} < 2.2	-2	-	2	
		100mV ≤ Output dynamic range ≤ V _{DDA} - 100mV	V _{DDA} ≥ 2.2	-1	1	1	
		PGA Gain=-3, 100mV ≤ Output dynamic range ≤ V _{DDA} - 100mV		-1	ı	1	
	Inverting gain value	PGA Gain=-7 100mV ≤ Ou range ≤ V _{DDA} - 100mV	tput dynamic	-1	-	1	%
		PGA Gain=-15, 100mV ≤ Output dynamic range ≤ V _{DDA} - 100mV		-1	-	1	
		PGA Gain=-31 200mV ≤ O 200mV	output ≤ V _{DDA} -	-2	-	2	
		PGA Gain=-63 200mV ≤ O range ≤ V _{DDA} - 200mV	utput dynamic	-5	ı	2	

Table 75. OPAMP characteristics⁽¹⁾ (continued)

Symbol	Parameter	Condition	s	Min	Тур	Max	Unit
		PGA Gain = 2		-	10/10	-	
	DO/D4 into we al	PGA Gain = 4			30/10	-	
	R2/R1 internal resistance values in	PGA Gain = 8		-	70/10	-	
	non-inverting PGA mode ⁽⁵⁾	PGA Gain = 16		-	150/10	-	
	mode(e)	PGA Gain = 32		-	310/10	-	
Б		PGA Gain = 64		-	630/10	-	kΩ/k
R _{network}		PGA Gain = -1		-	10/10	-	Ω
		PGA Gain = -3		-	30/10	-	
	R2/R1 internal	PGA Gain = -7		-	70/10	-	
	resistance values in inverting PGA mode ⁽⁵⁾	PGA Gain = -15		-	150/10	-	
	-	PGA Gain = -31		-	310/10	-	
		PGA Gain = -63		-	630/10	-	
Delta R	Resistance variation (R1 or R2)	-		-15	-	+15	%
		Gain = 2		-	GBW/2	-	
		Gain = 4		-	GBW/4	-	
	PGA bandwidth for different non inverting gain	Gain = 8		-	GBW/8	-	MHz
		Gain = 16		-	GBW/16	-	IVII IZ
		Gain = 32		-	GBW/32	-	
DOA DW		Gain = 64		-	GBW/64	-	
PGA BW		Gain = -1		-	GBW/2	-	
		Gain = -3		-	GBW/4	-	
	PGA bandwidth for	Gain = -7		-	GBW/8	-	N 41 1-
	different inverting gain	Gain = -15		-	GBW/16	-	MHz
		Gain = -31		-	GBW/32	-	
		Gain = -63		-	GBW/64	-	
a.N.I	Valtaga paiga dagaitu	at 1 kHz, Output loaded w	rith 4 kΩ	-	250	-	nV/√
eN	Voltage noise density	at 10 kHz, Output loaded	with 4 kΩ	-	90	-	Hz
I (ODAMD)	OPAMP consumption	Normal mode	No load,	-	1.3	2.2	A
I _{DDA} (OPAMP)	from V _{DDA}	High-speed mode	follower mode	-	1.4	2.6	mA
	ADC sampling time	V _{DDA} < 2V	•	300	-	-	
UT OPA	when reading the OPAMP output. OPAINTOEN=1	V _{DDA} ≥ 2V		200	-	-	ns
I _{DDA} (OPAMPI	OPAMP consumption	Normal mode	no load,	-	0.45	0.7	
NT)	from V _{DDA} . OPAINTOEN=1	High-speed mode	follower mode	-	0.5	0.8	mA



- 1. Guaranteed by design, unless otherwise specified.
- 2. Data guaranteed on normal and high speed mode unless otherwise specified.
- 3. Guaranteed by characterization results.
- 4. Valid also for inverting gain configuration with external bias.
- 5. R2 is the internal resistance between OPAMP output and OPAMP inverting input. R1 is the internal resistance between OPAMP inverting input and ground. The PGA gain =1+R2/R1



5.3.23 Temperature sensor characteristics

Table 76. TS characteristics

Symbol	Parameter	Min	Тур	Max	Unit
T _L ⁽¹⁾	V _{TS} linearity with temperature	-	±1	±2	°C
Avg_Slope ⁽¹⁾	Average slope		2.5	2.7	mV/°C
V ₃₀	Voltage at 30°C (±5 °C) ⁽²⁾		0.76	0.785	V
t _{START-RUN} ⁽¹⁾	Start-up time in Run mode (start-up of buffer)	-	8	15	μs
t _{START_CONT} (3)	Start-up time when entering in continuous mode	-	70	120	μs
t _{S_temp} ⁽¹⁾	ADC sampling time when reading the temperature	5	-	-	μs
I _{DD} (TS) ⁽¹⁾	Temperature sensor consumption from VDD, when selected by ADC	-	4.7	7	μΑ

^{1.} Guaranteed by design - Not tested in production.

5.3.24 V_{BAT} monitoring characteristics

Table 77. V_{BAT} monitoring characteristics⁽¹⁾

Symbol	Parameter		Тур	Max	Unit
R	Resistor bridge for V _{BAT}	-	3x39	-	kΩ
Q	Ratio on V _{BAT} measurement	-	3	-	-
Er ⁽²⁾	Error on Q	-10	-	10	%
t _{S_vbat} ⁽²⁾	ADC sampling time when reading the $V_{\mbox{\footnotesize{BAT}}}$	12	-	-	μs

^{1.} $1.55 \text{ V} < \text{V}_{BAT} < 3.6 \text{ V}.$

Table 78. V_{BAT} charging characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
R _{BC}	Battery	VBRS = 0	-	5	-	
	charging resistor	VBRS = 1	-	1.5	-	kΩ

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^{2.} Measured at V_{DDA} = 3.0 V ±10 mV. The V_{30} ADC conversion result is stored in the TS_CAL1 byte. Refer to *Table 5: Temperature sensor calibration values*.

^{3.} Continuous mode means RUN mode or Temperature Sensor ON.

^{2.} Guaranteed by design - Not tested in production.

5.3.25 Timer characteristics

The parameters given in the following tables are guaranteed by design.

Refer to *Section 5.3.14: I/O port characteristics* for details on the input/output alternate function characteristics (output compare, input capture, external clock, PWM output).

Table 79. TIMx⁽¹⁾ characteristics⁽²⁾

Symbol	Parameter	Conditions	Min	Max	Unit
+	Timer resolution time	-	1	-	t _{TIMxCLK}
t _{res(TIM)}	Timer resolution time	f _{TIMxCLK} = 170 MHz	5.88	-	ns
	Timer external clock	-	0	f _{TIMxCLK} /2	MHz
f _{EXT}	frequency on CH1 to CH4	f _{TIMxCLK} = 170 MHz	0	85	MHz
Res _{TIM}	Timer resolution	TIMx (except TIM2 and TIM5)	-	16	bit
		TIM2 and TIM5	-	32	
+	16-bit counter clock	-	1	65536	t _{TIMxCLK}
^t COUNTER	period	f _{TIMxCLK} = 170 MHz	0.00588	385.5	μs
	Maximum possible	-	-	65536 × 65536	t _{TIMxCLK}
t _{MAX_COUNT}	count with 32-bit counter	f _{TIMxCLK} = 170 MHz	-	25.26	s
f	Encoder frequency on	-	0	f _{TIMxCLK} /4	MHz
f _{ENC}	TI1 and TI2 input pins	f _{TIMxCLK} = 170MHz	0	42.5	MHz
t _{W(INDEX)}	Index pulsewidth on ETR input	-	2	-	Tck
t _{W(TI1, TI2)}	Min pulsewidth on TI1 and TI2 inputs in all encoder modes except directional clock x1	-	2	-	Tck
	Min pulsewidth on TI1 and TI2 inputs in directional clock x1	-	3	-	Tck

^{1.} TIMx is used as a general term in which x stands for 1,2,3,4,5,6,7,8,15,16, 17 or 20.

^{2.} Guaranteed by design - Not tested in production.

Prescaler divider	PR[2:0] bits	Min timeout RL[11:0]= 0x000	Max timeout RL[11:0]= 0xFFF	Unit
/4	0	0.125	512	
/8	1	0.250	1024	
/16	2	0.500	2048	
/32	3	1.0	4096	ms
/64	4	2.0	8192	
/128	5	4.0	16384	
/256	6 or 7	8.0	32768	

Table 80. IWDG min/max timeout period at 32 kHz (LSI)⁽¹⁾⁽²⁾

The exact timings still depend on the phasing of the APB interface clock versus the LSI clock so that there is always a full RC period of uncertainty.

	Table 51: WWD S Hill/Hax timeout value at 175 Hill 2 (1 5ER)						
Unit	Max timeout value	Min timeout value	Prescaler WDGTB				
	1.542	0.0241	0	1			
mo	3.084	0.0482	1	2			
ms	6.168	0.0964	2	4			
	12.336	0.1928	3	8			

Table 81. WWDG min/max timeout value at 170 MHz (PCLK)(1)

5.3.26 Communication interfaces characteristics

I²C interface characteristics

The I2C interface meets the timings requirements of the I²C-bus specification and user manual rev. 03 for:

- Standard-mode (Sm): with a bit rate up to 100 kbit/s
- Fast-mode (Fm): with a bit rate up to 400 kbit/s
- Fast-mode Plus (Fm+): with a bit rate up to 1 Mbit/s.

The I2C timings requirements are guaranteed by design when the I2C peripheral is properly configured (refer to reference manual RM0440 "STM32G4 Series advanced Arm®-based 32-bit MCUs") and when the I2CCLK frequency is greater than the minimum shown in the table below.



^{1.} Guaranteed by design - Not tested in production.

^{1.} Guaranteed by design - Not tested in production.

Symbol	Parameter	Condition		Min	Unit
f(I2CCLK)		Standard mode		2	
		Fast-mode	Analog Filtre ON DNF=0	8	
	I2CCLK frequency	Analog Filtre OFF	Analog Filtre OFF DNF=1	9	MHz
		Fast-mode	Analog Filtre ON DNF=0	17	
		Plus Analog Filtre OFF DNF=1		16	

Table 82. Minimum I2CCLK frequency in all I2C modes

The SDA and SCL I/O requirements are met with the following restrictions:

- The SDA and SCL I/O pins are not "true" open-drain. When configured as open-drain, the PMOS connected between the I/O pin and V_{DDIOx} is disabled, but is still present.
- The 20mA output drive requirement in Fast-mode Plus is supported partially. This limits the maximum load Cload supported in Fm+, which is given by these formulas:
 - $t_r(SDA/SCL)=0.8473 \times R_p \times C_{load}$
 - R_D(min)= (V_{DD} V_{OL}(max)) / I_{OL}(max)

Where Rp is the I2C lines pull-up. Refer to Section 5.3.14: I/O port characteristics for the I2C I/Os characteristics.

All I2C SDA and SCL I/Os embed an analog filter. Refer to *Table 83* below for the analog filter characteristics:

Table 83. I2C analog filter characteristics⁽¹⁾

Symbol	Parameter	Min	Max	Unit
t _{AF}	Maximum pulse width of spikes that are suppressed by the analog filter	50 ⁽²⁾	90(3)	ns

- 1. Guaranteed by design Not tested in production.
- 2. Spikes with widths below $t_{AF(min)}$ are filtered.
- 3. Spikes with widths above $t_{\text{AF}(\text{max})}$ are not filtered

SPI characteristics

Unless otherwise specified, the parameters given in *Table 84* for SPI are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and supply voltage conditions summarized in *Table 17: General operating conditions*.

- Output speed is set to OSPEEDRy[1:0] = 11
- Capacitive load C = 30 pF
- Measurement points are done at CMOS levels: 0.5 x V_{DD}

Refer to Section 5.3.14: I/O port characteristics for more details on the input/output alternate function characteristics (NSS, SCK, MOSI, MISO for SPI).



Table 84. SPI characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Тур	Max ⁽²⁾	Unit
		Master mode 2.7 V < V _{DD} < 3.6 V Voltage Range V1			75	
		Master mode 1.71 V < V _{DD} < 3.6 V Voltage Range V1			50	
		Master transmitter mode 1.71 V < V _{DD} < 3.6 V Voltage Range V1			50	
f _{SCK} 1/t _{c(SCK)}	SPI clock frequency	Slave receiver mode 1.71 V < V _{DD} < 3.6 V Voltage Range V1	-	-	50	MHz
		Slave mode transmitter/full duplex 2.7 V < V _{DD} < 3.6 V Voltage Range V1			41	
		Slave mode transmitter/full duplex 1.71 V < V _{DD} < 3.6 V Voltage Range V1			27	
		1.71 V < V _{DD} < 3.6 V Voltage Range V2			13	
t _{su(NSS)}	NSS setup time	Slave mode	4*T _{pclk}	-	-	-
t _{h(NSS)}	NSS hold time	Slave mode	2*T _{pclk}	-	-	-
$\begin{matrix} t_{w(\text{SCKH})} \\ t_{w(\text{SCKL})} \end{matrix}$	SCK high and low time	Master mode, SPI prescaler = 2	T _{pclk} -1	T _{pclk}	T _{pclk} +1	ns
t _{su(MI)}	Data input setup time	Master mode	4	ı	-	ns
t _{su(SI)}	Data input setup time	Slave mode	3	-	-	113
t _{h(MI)}	Data input hold time	Master mode	4	-	-	ns
t _{h(SI)}	Data input noid time	Slave mode	1	-	-	113
t _{a(SO)}	Data output access time	Slave mode	9	-	34	ns
t _{dis(SO)}	Data output disable time	Slave mode	9	-	16	ns



Symbol	Parameter	Conditions	Min	Тур	Max ⁽²⁾	Unit
		Slave mode 2.7 V < V _{DD} < 3.6 V Voltage Range V1	-	9	12	
t _{v(SO)}	(SO) Data output valid time	Slave mode 1.71 V < V _{DD} < 3.6 V Voltage Range V1	-	9	18	
		Slave mode 1.71 V < V _{DD} < 3.6 V Voltage Range V2	-	13	22	ns
t _{v(MO)}		Master mode	-	3.5	4.5	
t		Slave mode 1.71 V < V _{DD} < 3.6 V	6	-	-	
t _{h(SO)}	Data output hold time	Slave mode Range V2	9	-	-	
t _{h(MO)}		Master mode	2	-	-	

Table 84. SPI characteristics⁽¹⁾ (continued)

The maximum frequency in Slave transmitter mode is determined by the sum of tv(SO) and tsu(MI) which has to fit into SCK low or high-phase preceding the SCK sampling edge. This value can be achieved when the SPI communicates with a master having tsu(MI) = 0 while Duty(SCK) = 50%.

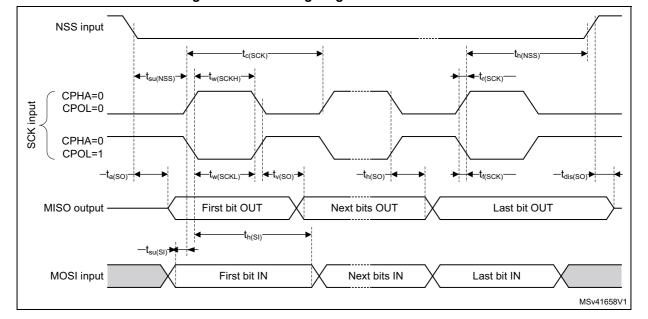


Figure 36. SPI timing diagram - slave mode and CPHA = 0

^{1.} Guaranteed by characterization results.

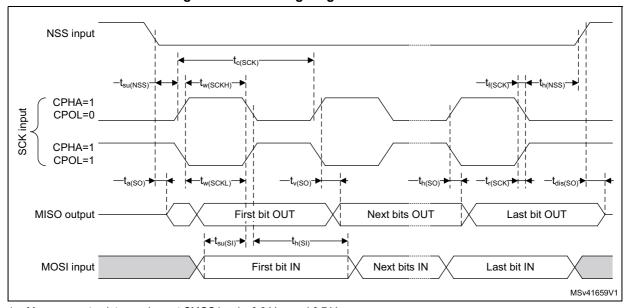


Figure 37. SPI timing diagram - slave mode and CPHA = 1

1. Measurement points are done at CMOS levels: 0.3 $\rm V_{DD}$ and 0.7 $\rm V_{DD.}$

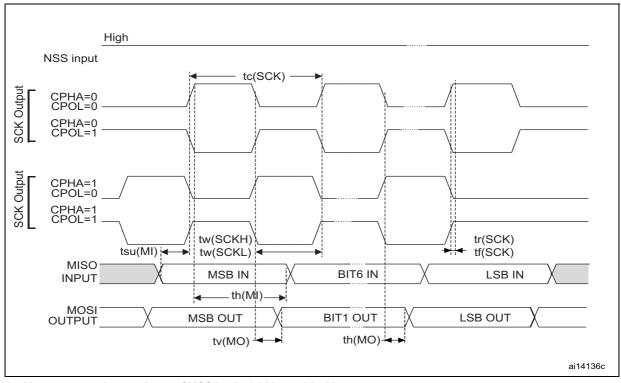


Figure 38. SPI timing diagram - master mode

1. Measurement points are done at CMOS levels: 0.3 $\rm V_{DD}$ and 0.7 $\rm V_{DD.}$

I2S characteristics

Unless otherwise specified, the parameters given in *Table 85* for I2S are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and V_{DD} supply voltage conditions summarized in *Table 17: General operating conditions*, with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 10
- Capacitive load C=30pF
- Measurement points are done at CMOS levels: 0.5 V_{DD}

Refer to Section 5.3.14: I/O port characteristics for more details on the input/output alternate function characteristics (CK,SD,WS).

Table 85. I2S characteristics⁽¹⁾

Symbol	Parameter	Condit	ions	Min	Max	Unit
f _{MCLK}	I2S Main clock output	-		256x8 K	256 *Fs ⁽²⁾	MHz
f.	I2S clock frequency	Master data		-	64xFs	MHz
f _{CK}	123 Clock frequency	Slave data		-	64xFs	IVII IZ
D _{CK}	I2S clock frequency duty cycle	Slave receiver		30	70	%
t _{v(WS)}	WS valid time	Master mode		-	6	
•	WS hold time	Master mode		3	-	
t _{h(WS)}		Slave mode	Slave mode		-	
t _{su(WS)}	WS setup time	Slave mode		4	-	
t _{su(SD_MR)}	Data input setup	Master receiver		3	-	
t _{su(SD_SR)}	time	Slave receiver		4	-	
t _{h(SD_MR)}	Data input hold time	Master receiver		4	-	ns
t _{h(SD_SR)}	Data input noid time	Slave receiver		2	-	
+		Slave transmitter (after	$2.7 \text{ V} \le \text{V}_{DD} \le 3.6 \text{ V}$	-	15	
t _{v(SD_ST)}	Data output valid time	enable edge)	1.65 V ≤ V _{DD} ≤ 3.6 V	-	22	
t _{v(SD_MT)}		Master transmitter (after enable edge)		-	3	
t _{h(SD_ST)}	Data output hold	Slave transmitter (after enable edge)		7	-	
t _{h(SD_MT)}	4ina a	Master transmitter (after el	nable edge)	1	-	

^{1.} Guaranteed by characterization results, not tested in production.

Note:

Refer to the reference manual RM0440 "STM32G4 Series advanced Arm®-based 32-bit MCUs" I2S section for more details about the sampling frequency (Fs), f_{MCK} , f_{CK} , D_{CK} values reflect only the digital peripheral behavior, source clock precision might slightly change the values D_{CK} depends mainly on ODD bit value. Digital contribution leads to a min of (I2SDIV/(2*I2SDIV+ODD) and a max (I2SDIV+ODD)/(2*I2SDIV+ODD) and Fs max supported for each mode/condition.



^{2. 256}xFs maximum is 49.152 MHz.

SAI characteristics

Unless otherwise specified, the parameters given in *Table 86* for SAI are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and V_{DD} supply voltage conditions summarized in *Table 17: General operating conditions*, with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 10
- Capacitive load C = 30 pF
- Measurement points are done at CMOS levels: 0.5 x V_{DD}

Refer to Section 5.3.14: I/O port characteristics for more details on the input/output alternate function characteristics (CK,SD,FS).



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Table 86. SAI characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Max	Unit	
f _{MCLK}	SAI Main clock output	-	-	50	MHz	
		Master transmitter 2.7 V ≤ V _{DD} ≤ 3.6 V Voltage Range 1	-	33		
		Master transmitter 1.71 V ≤ V _{DD} ≤ 3.6 V Voltage Range 1	-	22		
		Master receiver Voltage Range 1	-	22		
f _{CK}	SAI clock frequency ⁽²⁾	Slave transmitter 2.7 V \leq V _{DD} \leq 3.6 V Voltage Range 1	-	45	MHz	
		Slave transmitter 1.71 V \leq V _{DD} \leq 3.6 V Voltage Range 1	-	29		
		Slave receiver Voltage Range 1	-	50		
		Slave transmitter Voltage Range 2	-	13		
+	FS valid time	Master mode 2.7 V ≤ V _{DD} ≤ 3.6 V	-	15	ns	
t _{v(FS)}	rs valid time	Master mode 1.71 V ≤ V _{DD} ≤ 3.6 V	-	22	115	
t _{h(FS)}	FS hold time	Master mode	10	-	ns	
t _{su(FS)}	FS setup time	Slave mode	2	-	ns	
t _{h(FS)}	FS hold time	Slave mode	1	-	ns	
t _{su(SD_A_MR)}	Data input setup time	Master receiver	2.5	-	ne	
t _{su(SD_B_SR)}	Data input setup time	Slave receiver	1	-	ns	
t _{h(SD_A_MR)}	Data input hold time	Master receiver	5	-	ne	
t _{h(SD_B_SR)}	Data input noid time	Slave receiver	1	-	ns	
		Slave transmitter (after enable edge) 2.7 $V \le V_{DD} \le 3.6 V$	-	11		
$t_{v(SD_B_ST)}$	Data output valid time	Slave transmitter (after enable edge) 1.71 V \leq V _{DD} \leq 3.6 V	-	17	ns	
		Slave transmitter (after enable edge) voltage range V2	-	20		
t _{h(SD_B_ST)}	Data output hold time	Slave transmitter (after enable edge)	10	-	ns	

(
Symbol	Parameter	Conditions	Min	Max	Unit	
t	5	Master transmitter (after enable edge) 2.7 $V \le V_{DD} \le 3.6 V$	-	14	ns	
t _v (SD_A_MT)	^L v(SD_A_MT)	Data output valid time	Master transmitter (after enable edge) 1.71 V \leq V _{DD} \leq 3.6 V	-	21	113
t _{h(SD_A_MT)}	Data output hold time	Master transmitter (after enable edge)	10	-	ns	

Table 86. SAI characteristics⁽¹⁾ (continued)

- 1. Guaranteed by characterization results.
- 2. APB clock frequency must be at least twice SAI clock frequency.

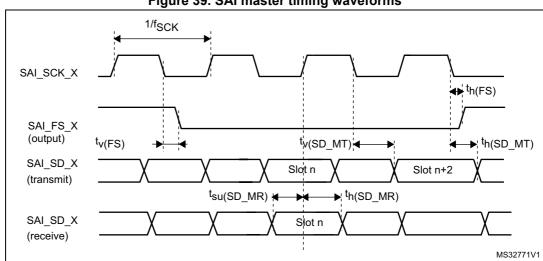
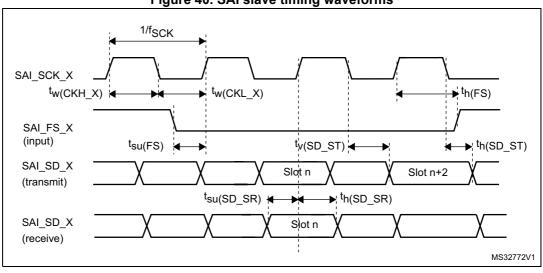


Figure 39. SAI master timing waveforms





CAN (controller area network) interface

Refer to Section 5.3.14: I/O port characteristics for more details on the input/output alternate function characteristics (FDCANx_TX and FDCANx_RX).

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USB characteristics

The device USB interface is fully compliant with the USB specification version 2.0 and is USB-IF certified (for Full-speed device operation).

Table 87. USB electrical characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V _{DD}	JSB transceiver operating voltage		3.0 ⁽²⁾	-	3.6	V
t _{Crystal_less}	USB crystal less operation ter	USB crystal less operation temperature		-	85	°C
R _{PUI}	Embedded USB_DP pull-up v	alue during idle	900	1250	1500	Ω
R _{PUR}	Embedded USB_PD pull-up v	Embedded USB_PD pull-up value during reception		2300	3200	12
Z _{sDRV} ⁽³⁾	Output driver impedance ⁽⁴⁾	Driving high and low	28	36	44	Ω

^{1.} TA = -40 to 125 °C unless otherwise specified.

USART interface characteristics

Unless otherwise specified, the parameters given in *Table 88* for USART are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and V_{DD} supply voltage conditions summarized in *Table 88*, with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 10
- Capacitive load C=30 pF
- Measurement points are done at CMOS levels: 0.5 V_{DD}

Refer to Section 5.3.14: I/O port characteristics for more details on the input/output alternate function characteristics (NSS, CK, TX, RX for USART).

Table 88. USART electrical characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f.	USART clock frequency	Master mode	-	-	21	MHz
f _{CK}	OSANT Clock frequency	Slave mode	-	-	22	IVII IZ
t _{su} (NSS)	NSS setup time	Slave mode	t _{ker} + 2	-	-	ne
t _h (NSS)	NSS hold time	Slave mode	2	-	-	ns
t _w (CKH) t _w (CKL)	CK high and low time	Master mode	1/f _{ck} /2-1	1/f _{ck} /2	1/f _{ck} /2+1	ns
t _{su} (RX)	Data input setup time	Master mode	t _{ker} + 2	-	-	
i _{Su} (IVX)	Data input setup time	Slave mode	2	-	-	ns
t. (DY)	Data input hold time	Master mode	1	-	-	115
t _h (RX)	Data input hold time	Slave mode	0.5	-	-	



^{2.} The device USB functionality is ensured down to 2.7 V but not the full USB electrical characteristics, which are degraded in the 2.7-to-3.0 V voltage range.

^{3.} Guarantee by design.

^{4.} No external termination series resistors are required on USB_PD (D+) and USB_DM (D-); the matching impedance is already included in the embedded driver.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
t (TV)	Data output valid time	Master mode	-	0.5	1.5	
ι _ν (1 Λ)	t _v (TX) Data output valid time	Slave mode	-	10	22	200
+ (DV)	t _h (RX) Data output hold time	Master mode	0	-	-	ns
ι _h (κλ)		Slave mode	7	-	-	

Table 88. USART electrical characteristics⁽¹⁾ (continued)

5.3.27 FSMC characteristics

Unless otherwise specified, the parameters given in *Table 89* to *Table 102* for the FMC interface are derived from tests performed under the ambient temperature, f_{HCLK} frequency and V_{DD} supply voltage conditions summarized in *Table 17*, with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 11
- Capacitive load C = 30 pF
- Measurement points are done at CMOS levels: 0.5 x V_{DD}

Refer to Section 5.3.14: I/O port characteristics for more details on the input/output characteristics.

Asynchronous waveforms and timings

Figure 41 through Figure 44 represent asynchronous waveforms and Table 89 through Table 96 provide the corresponding timings. The results shown in these tables are obtained with the following FMC configuration:

- AddressSetupTime = 0x1
- AddressHoldTime = 0x1
- DataHoldTime = 0x1
- ByteLaneSetup = 0x1
- DataSetupTime = 0x1 (except for asynchronous NWAIT mode, DataSetupTime = 0x5)
- BusTurnAroundDuration = 0x0

In all timing tables, the THCLK is the HCLK clock period.



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^{1.} Based on characterization, not tested in production.

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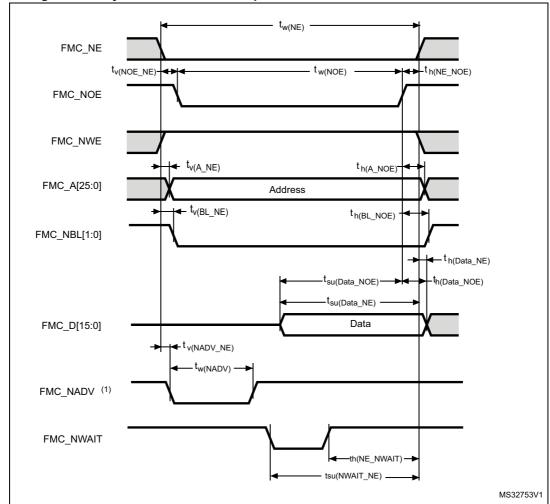


Figure 41. Asynchronous non-multiplexed SRAM/PSRAM/NOR read waveforms

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Table 89. Asynchronous non-multiplexed SRAM/PSRAM/NOR read timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
t _{w(NE)}	FMC_NE low time	3 T _{HCLK} - 0.5	3T _{HCLK} + 1	
t _{v(NOE_NE)}	FMC_NEx low to FMC_NOE low	0	1	
t _{w(NOE)}	FMC_NOE low time	2 T _{HCLK} - 0.5	2 T _{HCLK} + 1	
t _{h(NE_NOE)}	FMC_NOE high to FMC_NE high hold time	T _{HCLK}	-	
t _{v(A_NE)}	FMC_NEx low to FMC_A valid	-	2	
t _{h(A_NOE)}	Address hold time after FMC_NOE high	2 T _{HCLK} – 1	ı	ns
t _{su(Data_NE)}	Data to FMC_NEx high setup time	T _{HCLK} + 20	-	113
t _{su(Data_NOE)}	Data to FMC_NOEx high setup time	20	-	
t _{h(Data_NOE)}	Data hold time after FMC_NOE high	0	-	
t _{h(Data_NE)}	Data hold time after FMC_NEx high	0	-	
t _{v(NADV_NE)}	FMC_NEx low to FMC_NADV low	-	1.5	
t _{w(NADV)}	FMC_NADV low time	-	T _{HCLK} + 8	

^{1.} CL = 30 pF.

Table 90. Asynchronous non-multiplexed SRAM/PSRAM/NOR read-NWAIT timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
t _{w(NE)}	FMC_NE low time	-	8 T _{HCLK} + 1	
t _{w(NOE)}	FMC_NWE low time	7 T _{HCLK} - 1	7 T _{HCLK} + 0.5	
t _{w(NWAIT)}	FMC_NWAIT low time	T _{HCLK}	-	ns
t _{su(NWAIT_NE)}	FMC_NWAIT valid before FMC_NEx high	5 T _{HCLK} + 17	-	
t _{h(NE_NWAIT)}	FMC_NEx hold time after FMC_NWAIT invalid	4 T _{HCLK} + 17	-	

^{1.} CL = 30 pF.

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^{2.} Guaranteed by characterization results.

^{2.} Guaranteed by characterization results.

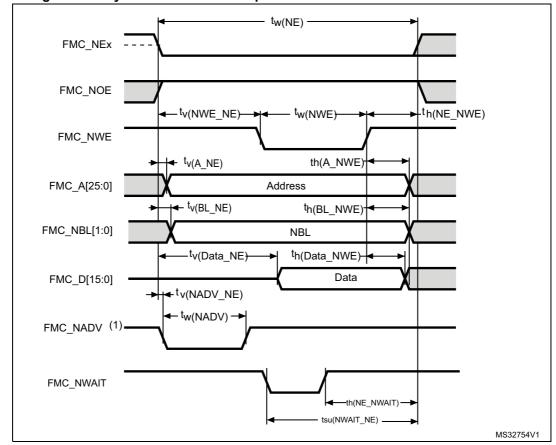


Figure 42. Asynchronous non-multiplexed SRAM/PSRAM/NOR write waveforms

Table 91. Asynchronous non-multiplexed SRAM/PSRAM/NOR write timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
t _{w(NE)}	FMC_NE low time	3 T _{HCLK} - 0.5	3 T _{HCLK} + 1	
t _{v(NWE_NE)}	FMC_NEx low to FMC_NWE low	T _{HCLK} - 0.5	T _{HCLK} + 1	
t _{w(NWE)}	FMC_NWE low time	T _{HCLK} -2	T _{HCLK} + 1	
t _{h(NE_NWE)}	FMC_NWE high to FMC_NE high hold time	T _{HCLK} - 0.5	-	
t _{v(A_NE)}	FMC_NEx low to FMC_A valid	-	0	
t _{h(A_NWE)}	Address hold time after FMC_NWE high	T _{HCLK} - 1	-	ns
t _{v(BL_NE)}	FMC_NEx low to FMC_BL valid	-	0	115
t _{h(BL_NWE)}	FMC_BL hold time after FMC_NWE high	T _{HCLK} + 0.5	-	
t _{v(Data_NE)}	Data to FMC_NEx low to Data valid	-	T _{HCLK} + 2	
t _{h(Data_NWE)}	Data hold time after FMC_NWE high	T _{HCLK} + 6	-	
t _{v(NADV_NE)}	FMC_NEx low to FMC_NADV low	-	1.5	
t _{w(NADV)}	FMC_NADV low time	-	T _{HCLK} + 0.5	

^{1.} CL = 30 pF.

^{2.} Guaranteed by characterization results.

Table 92. Asynchronous non-multiplexed SRAM/PSRAM/NOR write-NWAIT $timings^{(1)(2)}$

Symbol	Parameter	Min	Max	Unit		
t _{w(NE)}	FMC_NE low time	9 T _{HCLK} - 1	9 T _{HCLK} + 1			
t _{w(NWE)}	FMC_NWE low time	6 T _{HCLK} - 1	6 T _{HCLK} + 1	ne		
t _{su(NWAIT_NE)}	FMC_NWAIT valid before FMC_NEx high	7 T _{HCLK} + 17	-	ns		
t _{h(NE_NWAIT)}	FMC_NEx hold time after FMC_NWAIT invalid	7 T _{HCLK} + 17	-			

- 1. CL = 30 pF.
- 2. Guaranteed by characterization results.

Figure 43. Asynchronous multiplexed PSRAM/NOR read waveforms

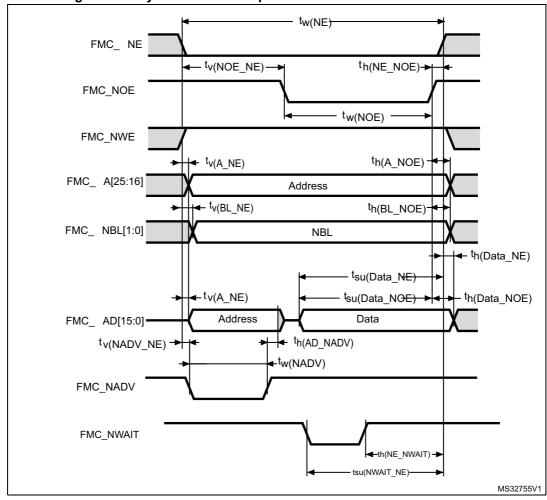


Table 93. Asynchronous multiplexed PSRAM/NOR read timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
t _{w(NE)}	FMC_NE low time	3 T _{HCLK} - 0.5	3 T _{HCLK} + 1	
t _{v(NOE_NE)}	FMC_NEx low to FMC_NOE low	0	1	
t _{w(NOE)}	FMC_NOE low time	2 T _{HCLK} - 0.5	2 T _{HCLK} + 0.5	
t _{h(NE_NOE)}	FMC_NOE high to FMC_NE high hold time	T _{HCLK}	-	
t _{v(A_NE)}	FMC_NEx low to FMC_A valid	-	2	
t _{v(NADV_NE)}	FMC_NEx low to FMC_NADV low	0.5	1.5	
t _{w(NADV)}	FMC_NADV low time	T _{HCLK}	T _{HCLK} + 1.5	
t _{h(AD_NADV)}	FMC_AD(address) valid hold time after FMC_NADV high	T _{HCLK} - 0.3	-	ns
t _{h(A_NOE)}	Address hold time after FMC_NOE high	Address hold until next read operation	-	
t _{su(Data_NE)}	Data to FMC_NEx high setup time	T _{HCLK} + 20	1	
t _{su(Data_NOE)}	Data to FMC_NOE high setup time	20	-	
t _{h(Data_NE)}	Data hold time after FMC_NEx high	0	-	
t _{h(Data_NOE)}	Data hold time after FMC_NOE high	0	-	

^{1.} CL = 30 pF.

Table 94. Asynchronous multiplexed PSRAM/NOR read-NWAIT timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
t _{w(NE)}	FMC_NE low time	8 T _{HCLK} - 1	8 T _{HCLK} + 1	
t _{w(NOE)}	FMC_NWE low time	7 T _{HCLK} - 1	7 T _{HCLK} + 0.5	ns
t _{su(NWAIT_NE)}	FMC_NWAIT valid before FMC_NEx high	5 T _{HCLK} + 17	-	113
t _{h(NE_NWAIT)}	FMC_NEx hold time after FMC_NWAIT invalid	4 T _{HCLK} + 17	-	

^{1.} CL = 30 pF.

2. Guaranteed by characterization results.

^{2.} Guaranteed by characterization results.

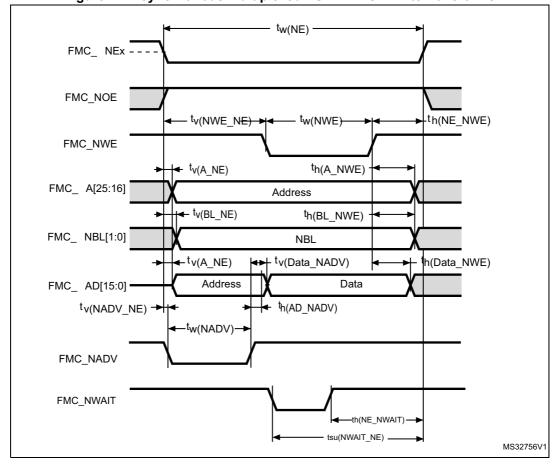


Figure 44. Asynchronous multiplexed PSRAM/NOR write waveforms

Table 95. Asynchronous multiplexed PSRAM/NOR write timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
t _{w(NE)}	FMC_NE low time	3 T _{HCLK} - 0.5	3 T _{HCLK} + 1	
t _{v(NWE_NE)}	FMC_NEx low to FMC_NWE low	T _{HCLK} - 0.5	T _{HCLK} + 1	
t _{w(NWE)}	FMC_NWE low time	T _{HCLK} - 2	T _{HCLK} + 1	
t _{h(NE_NWE)}	FMC_NWE high to FMC_NE high hold time	T _{HCLK} - 0.5	-	
t _{v(A_NE)}	FMC_NEx low to FMC_A valid	-	0	
t _{v(NADV_NE)}	FMC_NEx low to FMC_NADV low	0	1.5	
t _{w(NADV)}	FMC_NADV low time	T _{HCLK} + 0.5	T _{HCLK} + 1.5	
t _{h(AD_NADV)}	FMC_AD(address) valid hold time after FMC_NADV high	T _{HCLK} - 3	-	ns
t _{h(A_NWE)}	Address hold time after FMC_NWE high	Address hold until next write operation	-	
t _{h(BL_NWE)}	FMC_BL hold time after FMC_NWE high	T _{HCLK} - 0.5	-	
t _{v(BL_NE)}	FMC_NEx low to FMC_BL valid	-	0	
t _{v(Data_NADV)}	FMC_NADV high to Data valid	-	T _{HCLK} + 2	
t _{h(Data_NWE)}	Data hold time after FMC_NWE high	T _{HCLK} + 6	-	

^{1.} CL = 30 pF.

Table 96. Asynchronous multiplexed PSRAM/NOR write-NWAIT timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
t _{w(NE)}	FMC_NE low time	9 T _{HCLK} - 1	9 T _{HCLK} + 1	
t _{w(NWE)}	FMC_NWE low time	6 T _{HCLK} - 1	6 T _{HCLK} + 0.5	ns
t _{su(NWAIT_NE)}	FMC_NWAIT valid before FMC_NEx high	7 T _{HCLK} + 17	-	
t _{h(NE_NWAIT)}	FMC_NEx hold time after FMC_NWAIT invalid	5 T _{HCLK} + 17	-	

^{1.} CL = 30 pF.

^{2.} Guaranteed by characterization results.

^{2.} Guaranteed by characterization results.

Synchronous waveforms and timings

Figure 45 through Figure 48 represent synchronous waveforms and Table 97 through Table 100 provide the corresponding timings. The results shown in these tables are obtained with the following FMC configuration:

- BurstAccessMode = FMC_BurstAccessMode_Enable
- MemoryType = FMC_MemoryType_CRAM
- WriteBurst = FMC_WriteBurst_Enable
- CLKDivision = 1
- DataLatency = 1 for NOR Flash; DataLatency = 0 for PSRAM
 In all timing tables, the T_{HCl K} is the HCLK clock period.
- For 2.7 V \leq V_{DD} \leq 3.6 V, maximum FMC_CLK = 60 MHz for CLKDIV = 0x1 and 54 MHz for CLKDIV = 0x0 at CL = 30 pF (on FMC_CLK).
- For 1.71 V \leq V_{DD} \leq 2.7 V, maximum FMC_CLK = 60 MHz for CLKDIV = 0x1 and 32 MHz for CLKDIV = 0x0 at CL= 20 pF (on FMC_CLK).

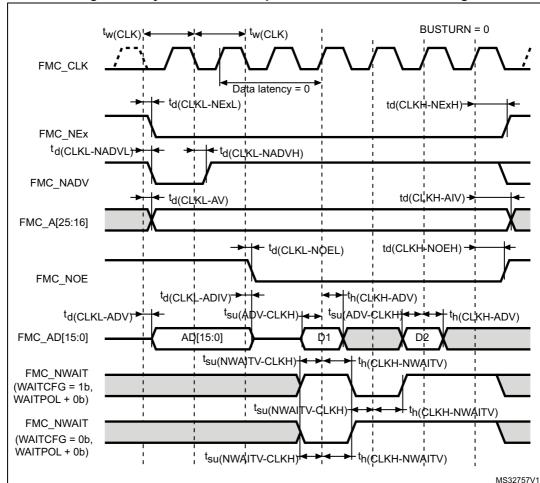


Figure 45. Synchronous multiplexed NOR/PSRAM read timings

Table 97. Synchronous multiplexed NOR/PSRAM read timings⁽¹⁾⁽²⁾⁽³⁾

	· · · · · · · · · · · · · · · · · · ·			
Symbol	Parameter	Min	Max	Unit
t _{w(CLK)}	FMC_CLK period	R*T _{HCLK} - 0.5	-	
t _{d(CLKL-NExL)}	FMC_CLK low to FMC_NEx low (x=02)	-	1.5	
t _{d(CLKH_NExH)}	FMC_CLK high to FMC_NEx high (x= 02)	R*T _{HCLK} /2 + 1	-	
t _{d(CLKL-NADVL)}	FMC_CLK low to FMC_NADV low	-	2.5	
t _{d(CLKL-NADVH)}	FMC_CLK low to FMC_NADV high	3.5	-	
t _{d(CLKL-AV)}	FMC_CLK low to FMC_Ax valid (x=1625)	-	4	
t _{d(CLKH-AIV)}	FMC_CLK high to FMC_Ax invalid (x=1625)	R*T _{HCLK} /2 + 1	-	
t _{d(CLKL-NOEL)}	FMC_CLK low to FMC_NOE low	-	2	ns
t _{d(CLKH-NOEH)}	FMC_CLK high to FMC_NOE high	R*T _{HCLK} /2 + 1	-	
t _{d(CLKL-ADV)}	FMC_CLK low to FMC_AD[15:0] valid	-	3	
t _{d(CLKL-ADIV)}	FMC_CLK low to FMC_AD[15:0] invalid	0	-	
t _{su(ADV-CLKH)}	FMC_A/D[15:0] valid data before FMC_CLK high	2	-	
t _{h(CLKH-ADV)}	FMC_A/D[15:0] valid data after FMC_CLK high	4	-	
t _{su(NWAIT-CLKH)}	FMC_NWAIT valid before FMC_CLK high	1.5	-	
t _{h(CLKH-NWAIT)}	FMC_NWAIT valid after FMC_CLK high	4	-	

^{1.} CL = 30 pF.

^{2.} Guaranteed by characterization results.

^{3.} Clock ratio R = (HCLK period /FMC_CLK period).

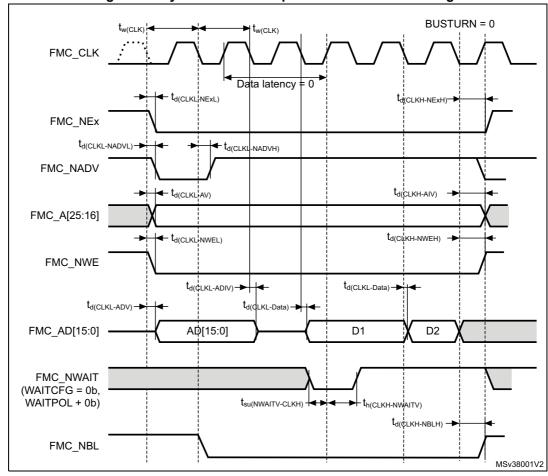


Figure 46. Synchronous multiplexed PSRAM write timings

Table 98. Synchronous multiplexed PSRAM write timings⁽¹⁾⁽²⁾⁽³⁾

Symbol	Parameter	Min	Max	Unit
t _{w(CLK)}	FMC_CLK period	R*T _{HCLK} - 0.5	-	
t _{d(CLKL-NExL)}	FMC_CLK low to FMC_NEx low (x=02)	-	1.5	
t _{d(CLKH-NExH)}	FMC_CLK high to FMC_NEx high (x= 02)	R*T _{HCLK} /2 + 1	-	
t _{d(CLKL-NADVL)}	FMC_CLK low to FMC_NADV low	-	2.5	
t _{d(CLKL-NADVH)}	FMC_CLK low to FMC_NADV high	3.5	-	
t _{d(CLKL-AV)}	FMC_CLK low to FMC_Ax valid (x=1625)	-	4	
t _{d(CLKH-AIV)}	FMC_CLK high to FMC_Ax invalid (x=1625)	R*T _{HCLK} /2 + 1	-	
t _{d(CLKL-NWEL)}	FMC_CLK low to FMC_NWE low	-	2	200
t _{d(CLKH-NWEH)}	FMC_CLK high to FMC_NWE high	R*T _{HCLK} /2 + 1	-	ns
t _{d(CLKL-ADV)}	FMC_CLK low to FMC_AD[15:0] valid	-	3	
t _{d(CLKL-ADIV)}	FMC_CLK low to FMC_AD[15:0] invalid	0	-	
t _{d(CLKL-DATA)}	FMC_A/D[15:0] valid data after FMC_CLK low	-	3	
t _{d(CLKL-NBLL)}	FMC_CLK low to FMC_NBL low	1	-	
t _{d(CLKH-NBLH)}	FMC_CLK high to FMC_NBL high	R*T _{HCLK} /2 + 1.5	-	
t _{su(NWAIT-CLKH)}	FMC_NWAIT valid before FMC_CLK high	1.5	-	
t _{h(CLKH-NWAIT)}	FMC_NWAIT valid after FMC_CLK high	4	-	

^{1.} CL = 30 pF.

^{2.} Guaranteed by characterization results.

^{3.} Clock ratio R = (HCLK period /FMC_CLK period).

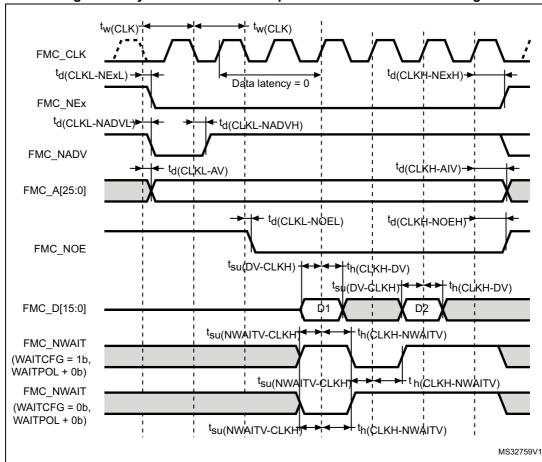


Figure 47. Synchronous non-multiplexed NOR/PSRAM read timings

Table 99. Synchronous non-multiplexed NOR/PSRAM read timings $^{(1)(2)(3)}$

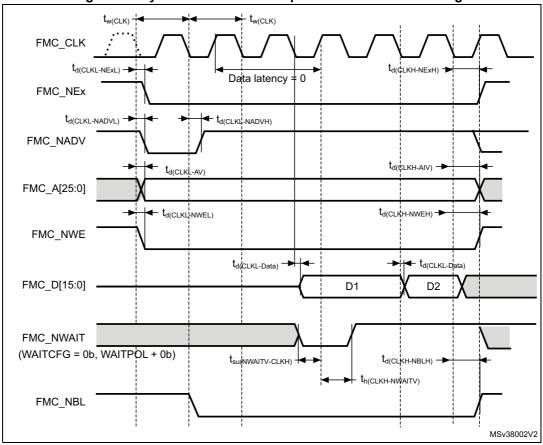
Symbol	Symbol Parameter		Max	Unit
t _{w(CLK)}	FMC_CLK period	R*T _{HCLK} - 0.5	-	
t _{d(CLKL-NExL)}	FMC_CLK low to FMC_NEx low (x=02)	-	1.5	
t _{d(CLKH-NExH)}	FMC_CLK high to FMC_NEx high (x= 02)	R*T _{HCLK} /2 + 1	-	
t _{d(CLKL-NADVL)}	FMC_CLK low to FMC_NADV low	-	2.5	
t _{d(CLKL-NADVH)}	FMC_CLK low to FMC_NADV high	3.5	-	
t _{d(CLKL-AV)}	t _{d(CLKL-AV)} FMC_CLK low to FMC_Ax valid (x=1625)		4	ns
t _{d(CLKH-AIV)}	FMC_CLK high to FMC_Ax invalid (x=1625)	R*T _{HCLK} /2+- 1	-	
t _{d(CLKL-NOEL)}	FMC_CLK low to FMC_NOE low	-	2	
t _{d(CLKH-NOEH)}	FMC_CLK high to FMC_NOE high	R*T _{HCLK} /2 + 1	-	
t _{su(DV-CLKH)}	FMC_D[15:0] valid data before FMC_CLK high	2	-	
t _{h(CLKH-DV)}	FMC_D[15:0] valid data after FMC_CLK high	4	-	
t _{su(NWAIT-CLKH)}	FMC_NWAIT valid before FMC_CLK high	1.5	-	ne
t _{h(CLKH-NWAIT)}	FMC_NWAIT valid after FMC_CLK high	4	ı	ns



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- 1. CL = 30 pF.
- 2. Guaranteed by characterization results.
- 3. Clock ratio R = (HCLK period /FMC_CLK period).





Symbol	Parameter	Min	Max	Unit
t _{w(CLK)}	FMC_CLK period	R*T _{HCLK} - 0.5	-	
t _{d(CLKL-NExL)}	FMC_CLK low to FMC_NEx low (x=02)	-	1.5	
t _{d(CLKH-NExH)}	FMC_CLK high to FMC_NEx high (x= 02)	R*T _{HCLK} /2 + 1	-	
t _{d(CLKL-NADVL)}	FMC_CLK low to FMC_NADV low	-	2.5	
t _{d(CLKL-NADVH)}	FMC_CLK low to FMC_NADV high	3.5	-	
t _{d(CLKL-AV)}	FMC_CLK low to FMC_Ax valid (x=1625)	-	4	
t _{d(CLKH-AIV)}	FMC_CLK high to FMC_Ax invalid (x=1625)	R*T _{HCLK} /2 + 1	-	ns
t _{d(CLKL-NWEL)}	FMC_CLK low to FMC_NWE low	-	2	115
t _{d(CLKH-NWEH)}	FMC_CLK high to FMC_NWE high	R*T _{HCLK} /2 + 1	-	
t _{d(CLKL-Data)}	FMC_D[15:0] valid data after FMC_CLK low	-	3	
t _{d(CLKL-NBLL)}	FMC_CLK low to FMC_NBL low	1	-	
t _{d(CLKH-NBLH)}	FMC_CLK high to FMC_NBL high	R*T _{HCLK} /2 + 1.5	-	
t _{su(NWAIT-CLKH)}	FMC_NWAIT valid before FMC_CLK high	1.5	-	
t _{h(CLKH-NWAIT)}	FMC_NWAIT valid after FMC_CLK high	4	ı	

Table 100. Synchronous non-multiplexed PSRAM write timings⁽¹⁾⁽²⁾⁽³⁾

NAND controller waveforms and timings

Figure 49 through Figure 52 represent synchronous waveforms, and Table 101 and Table 102 provide the corresponding timings. The results shown in these tables are obtained with the following FMC configuration:

- COM.FMC_SetupTime = 0x01
- COM.FMC WaitSetupTime = 0x03
- COM.FMC_HoldSetupTime = 0x02
- COM.FMC_HiZSetupTime = 0x01
- ATT.FMC SetupTime = 0x01
- ATT.FMC_WaitSetupTime = 0x03
- ATT.FMC_HoldSetupTime = 0x02
- ATT.FMC HiZSetupTime = 0x01
- Bank = FMC_Bank_NAND
- MemoryDataWidth = FMC_MemoryDataWidth_16b
- ECC = FMC ECC Enable
- ECCPageSize = FMC_ECCPageSize_512Bytes
- TCLRSetupTime = 0
- TARSetupTime = 0

In all timing tables, the T_{HCLK} is the HCLK clock period.



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^{1.} CL = 30 pF.

^{2.} Guaranteed by characterization results.

^{3.} Clock ratio R = (HCLK period /FMC_CLK period).

FMC_NCEx

ALE (FMC_A17)
CLE (FMC_A16)

FMC_NWE

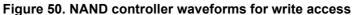
Th(NOE-ALE)

FMC_NOE (NRE)

Th(NOE-ALE)

MSv38003V1

Figure 49. NAND controller waveforms for read access



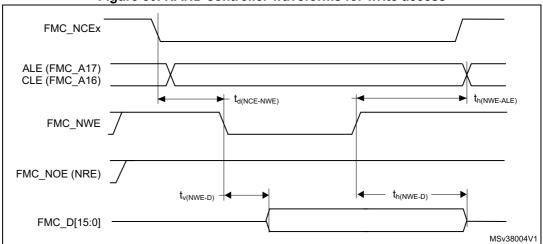
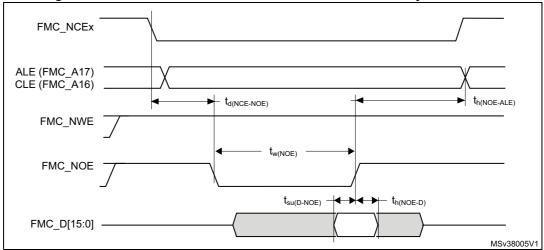


Figure 51. NAND controller waveforms for common memory read access



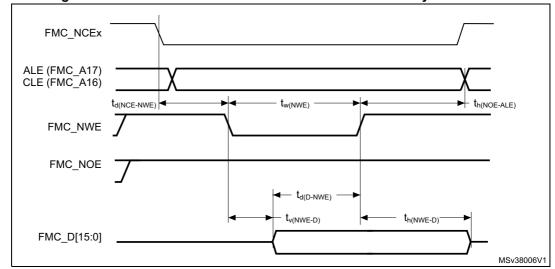


Figure 52. NAND controller waveforms for common memory write access

Table 101. Switching characteristics for NAND flash memory read cycles (1)(2)

Symbol	Parameter	Min	Max	Unit
T _{w(N0E)}	FMC_NOE low width	4 T _{HCLK} - 1	4 T _{HCLK}	
T _{su(D-NOE)}	FMC_D[15-0] valid data before FMC_NOE high	19	-	
T _{h(NOE-D)}	FMC_D[15-0] valid data after FMC_NOE high	0	-	ns
T _{d(NCE-NOE)}	FMC_NCE valid before FMC_NOE low	-	3 T _{HCLK}	
T _{h(NOE-ALE)}	FMC_NOE high to FMC_ALE invalid	3 T _{HCLK}	-	

^{1.} CL = 30 pF.

Table 102. Switching characteristics for NAND flash memory write cycles⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
T _{w(NWE)}	FMC_NWE low width	4 T _{HCLK} -1	4 T _{HCLK}	
T _{v(NWE-D)}	FMC_NWE low to FMC_D[15-0] valid	0	-	
T _{h(NWE-D)}	FMC_NWE high to FMC_D[15-0] invalid	high to FMC_D[15-0] invalid 3 T _{HCLK} - 1		ns
T _{d(D-NWE)}	T _{d(D-NWE)} FMC_D[15-0] valid before FMC_NWE high		ı	113
T _{d(NCE_NWE)}	FMC_NCE valid before FMC_NWE low	-	3 T _{HCLK}	
T _{h(NWE-ALE)}	FMC_NWE high to FMC_ALE invalid	3 T _{HCLK}	-	

^{1.} CL = 30 pF.

2. Guaranteed by characterization results.

^{2.} Guaranteed by characterization results.

5.3.28 QUADSPI characteristics

Unless otherwise specified, the parameters given in *Table 103* and *Table 104* for Quad SPI are derived from tests performed under the ambient temperature, f_{AHB} frequency and V_{DD} supply voltage conditions summarized in *Table 17: General operating conditions*, with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 11
- Capacitive load C = 15 or 20 pF
- Measurement points are done at CMOS levels: 0.5 x V_{DD}

Refer to Section 5.3.14: I/O port characteristics for more details on the input/output alternate function characteristics.

Table 103. Quad SPI characteristics in SDR mode⁽¹⁾

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
E(OCK)	Quad SPI clock	1.71 < V _{DD} < 3.6 V, C _{LOAD} = 15 pF Voltage Range 1	-	-	50	MHz
frequency		1.71 < V _{DD} < 3.6 V, C _{LOAD} = 20 pF Voltage Range 2	-	-	110	IVII IZ
t _{w(CKH)}	Quad SPI clock high	PRESCALER [7:0]	t _(CK) /2-0.5	-	t _(CK) /2+1	
t _{w(CKL)}	and low time Even division	n =0,1, 3, 5	t _(CK) /2-1	-	t _(CK) /2+0.5	
t _{w(CKH)}	Quad SPI clock high	PRESCALER [7:0]	(n/2)*t _(CK) /(n+1) - 0.5	-	(n/2)*t _(CK) /(n+1) + 1	
t _{w(CKL)}	and low time Odd division	n =2,4, 6, 8	(n/2+1)*t _(CK) /(n+1) - 1	-	(n/2+1)*t(_{CK)} /(n+1) +0.5	20
t _{s(IN)}	Data input setup time	1.71 < V _{DD} < 3.6 V	1	-	-	ns
t _{h(IN)}	Data input hold time	1.71 < V _{DD} < 3.6 V	5	-	-	
t _{v(OUT)}	Data output valid time	1.71 < V _{DD} < 3.6 V	-	1	1.5	
t _{h(OUT)}	Data output hold time	1.71 < V _{DD} < 3.6 V	0.5	-	-	

^{1.} Guaranteed by characterization results.

Table 104. QUADSPI characteristics in DDR mode⁽¹⁾

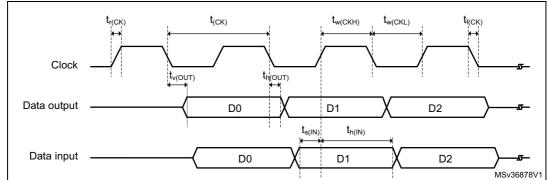
Symbol	Parameter	Conditions	Min	Тур	Мах	Unit
E(OCK)	F(QCK) Quad SPI clock frequency	1.71 < V _{DD} < 3.6 V, C _{LOAD} = 15 pF Voltage Range 1	-	ı	50	MHz
T (QON)		$1.71 < V_{DD} < 3.6 V$, $C_{LOAD} = 20 pF$ Voltage Range 2	-	1	70	IVII IZ

Table 104. QUADSPI characteristics in DDR mode⁽¹⁾ (continued)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
t _{w(CKH)}	Quad SPI clock high	PRESCALER [7:0]	t _(CK) /2	-	t _(CK) /2+1	
t _{w(CKL)}	and low time Even division	n =0,1, 3, 5	t _(CK) /2-1	-	t _(CK) /2	
t _{w(CKH)}	Quad SPI clock high and	PRESCALER [7:0]	(n/2)*t _(CK) /(n+1)	-	$(n/2)*t_{(CK)}/(n+1) + 1$	
t _{w(CKL)}	low time Odd division	n =2,4, 6, 8	(n/2+1)*t _(CK) /(n+1) - 1	-	$(n/2+1)*t(_{CK)}/(n+1)$	
t _{sr(IN)}	Data input setup time on rising edge	1.71 < V _{DD} < 3.6 V	1	-	-	
t _{sf(IN)}	Data input setup time on falling edge	1.71 < V _{DD} < 3.6 V	1	-	-	
t _{hr(IN)}	Data input hold time on rising edge	1.71 < V _{DD} < 3.6 V	6	-	-	
t _{hf(IN)}	Data input hold time on falling edge	1.71 < V _{DD} < 3.6 V	5	-	-	
	Data output valid time on	1.71 < V _{DD} < 3.6 V DHHC = 0		7.5	8	ns
t _{vr(OUT)}	rising edge	1.71 < V _{DD} < 3.6 V DHHC = 1	-	Thclk/2 +1	Thclk/2+1.5	
	Data output valid time	1.71 < V _{DD} < 3.6 V DHHC = 0		7	10	
t _{vf(OUT)}	Data output valid time	1.71 < V _{DD} < 3.6 V DHHC = 1	-	Thclk/2 +1	Thclk/2+2	
4	Data output hold time on	1.71 < V _{DD} < 3.6 V DHHC = 0	2	-	-	
t _{hr(OUT)}	rising edge	1.71 < V _{DD} < 3.6 V DHHC = 1	Thclk/2+ 0.5	-	-	
	Data output hold time on	1.71 < V _{DD} < 3.6 V DHHC = 0	3	-	-	
t _{hf(OUT)}	falling edge	1.71 < V _{DD} < 3.6 V DHHC = 1	Thclk/2+0.5	-	-	

1. Guaranteed by characterization results.

Figure 53. Quad SPI timing diagram - SDR mode



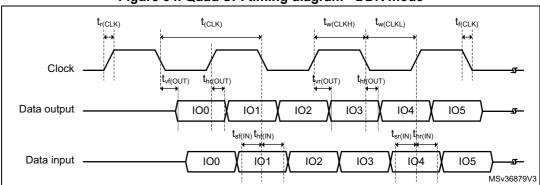


Figure 54. Quad SPI timing diagram - DDR mode

5.3.29 UCPD characteristics

UCPD1 controller complies with USB Type-C Rev.1.2 and USB Power Delivery Rev. 3.0 specifications.

Table 105. UCPD characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V _{DD} L	UCPD operating supply voltage	Sink mode only	3.0	3.3	3.6	V
		Sink and source mode	3.135	3.3	3.465	٧

6 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK packages, depending on their level of environmental compliance. ECOPACK specifications, grade definitions and product status are available at: www.st.com. ECOPACK is an ST trademark.

6.1 Device marking

Refer to technical note "Reference device marking schematics for STM32 microcontrollers and microprocessors" (TN1433) available on www.st.com, for the location of pin 1 / ball A1 as well as the location and orientation of the marking areas versus pin 1 / ball A1.

Parts marked as "ES", "E" or accompanied by an engineering sample notification letter, are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.



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6.2 WLCSP81 package information (B068)

This WLCSP is a 81-ball, 4.02 x 4.27 mm, 0.4 mm pitch wafer level chip scale package.

DETAIL A A1 BALL LOCATION D 0000000 Ġ e2 Ε 00000000 00000000 △ eee (4X) BOTTOM VIEW TOP VIEW SIDE VIEW | A2 øь FRONT VIEW Ż SEATING PLANE DETAIL A ROTATED 90° B068_WLCSP81_DIE469_ME_V1

Figure 55. WLCSP81 - Outline

- 1. Drawing is not to scale.
- 2. Dimension is measured at the maximum bump diameter parallel to primary datum Z.
- 3. Primary datum Z and seating plane are defined by the spherical crowns of the bump.
- 4. Bump position designation per JESD 95-1, SPP-010.

Table 106. WLCSP81 - Mechanical data

Symbol	millimeters		inches ⁽¹⁾			
	Min	Тур	Max	Min	Тур	Max
A ⁽²⁾	-	-	0.59	-	-	0.023
A1	-	0.18	-	-	0.007	-
A2	-	0.38	-	-	0.015	-
A3	-	0.025	-	-	0.001	-
b	0.22	0.25	0.28	0.009	0.010	0.011
D	4.00	4.02	4.04	0.157	0.158	0.159

14515 1001 112001 01 11100114111041 4444 (0011111404)							
0	millimeters			inches ⁽¹⁾			
Symbol	Min	Тур	Max	Min	Тур	Max	
E	4.25	4.27	4.29	0.167	0.168	0.169	
е	-	0.40	-	-	0.016	-	
e1	-	3.20	-	-	0.126	-	
e2	-	3.20	-	-	0.126	-	
F ⁽³⁾	-	0.410	-	-	0.016	-	
G ⁽³⁾	-	0.535	-	-	0.021	-	
aaa	-	-	0.10	-	-	0.004	
bbb	-	-	0.10	-	-	0.004	
ccc	-	-	0.10		-	0.004	
ddd	-	-	0.05	-	-	0.002	
eee	-	-	0.05	-	-	0.002	

Table 106. WLCSP81 - Mechanical data (continued)

- 1. Values in inches are converted from mm and rounded to 3 decimal digits.
- 2. The maximum total package height is calculated by the RSS method (Root Sum Square) using nominal and tolerances values of A1 and A2.
- 3. Calculated dimensions are rounded to the 3rd decimal place

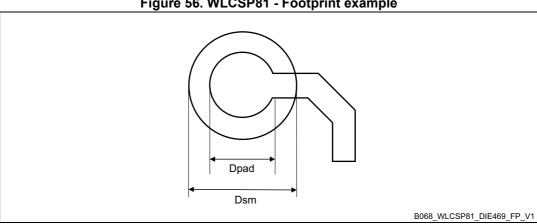


Figure 56. WLCSP81 - Footprint example

Table 107. WLCSP81 - Example of PCB design rules

Dimension	Values
Pitch	0.4 mm
Dpad	0,225 mm
Dsm	0.290 mm typ. (depends on soldermask registration tolerance)
Stencil opening	0.250 mm
Stencil thickness	0.100 mm



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6.3 UFQFPN48 package information (A0B9)

This UFQFPN is a 48-lead, 7x7 mm, 0.5 mm pitch, ultra thin fine pitch quad flat package.

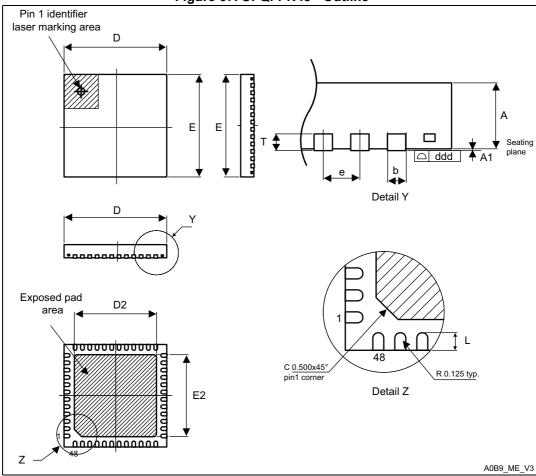


Figure 57. UFQFPN48 - Outline

- 1. Drawing is not to scale.
- 2. All leads/pads should also be soldered to the PCB to improve the lead/pad solder joint life.
- There is an exposed die pad on the underside of the UFQFPN48 package. It is recommended to connect and solder this back-side pad to PCB ground.

inches⁽¹⁾ millimeters **Symbol** Min Max Min Тур Max Typ 0.500 0.550 0.600 0.0197 0.0217 0.0236 Α Α1 0.000 0.020 0.050 0.0000 0.0008 0.0020 D 6.900 7.000 7.100 0.2717 0.2756 0.2795 6.900 7.000 7.100 0.2717 0.2795 Ε 0.2756 D2 5.500 5.600 5.700 0.2165 0.2205 0.2244 E2 5.500 5.600 5.700 0.2165 0.2205 0.2244 L 0.300 0.400 0.500 0.0118 0.0157 0.0197 Т 0.152 0.0060 b 0.200 0.250 0.300 0.0079 0.0098 0.0118 0.500 0.0197 е 0.080 0.0031 ddd

Table 108. UFQFPN48 - Mechanical data

^{1.} Values in inches are converted from mm and rounded to 4 decimal digits.

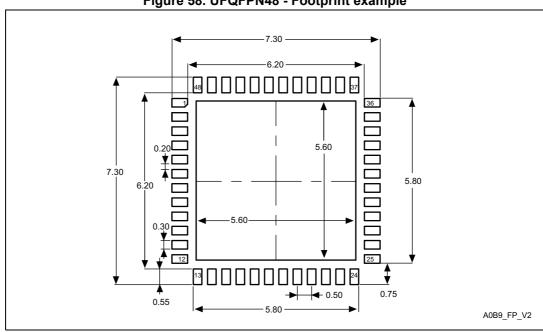


Figure 58. UFQFPN48 - Footprint example

1. Dimensions are expressed in millimeters.

6.4 LQFP48 package information (5B)

This LQFP is a 48-pin, 7 x 7 mm low-profile quad flat package.

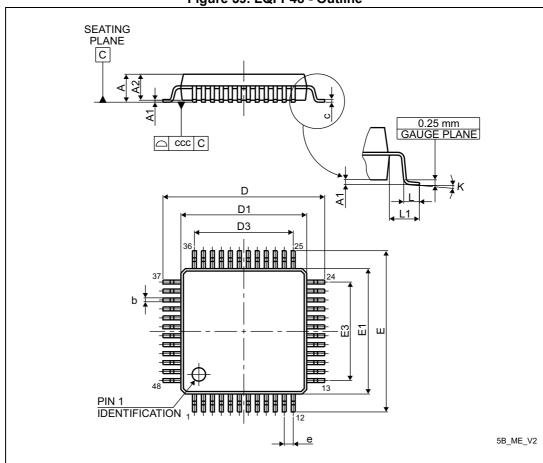


Figure 59. LQFP48 - Outline

1. Drawing is not to scale.

Table 109. LQFP48 - Mechanical data

Cumbal	millimeters			inches ⁽¹⁾		
Symbol	Min	Тур	Max	Min	Тур	Max
Α	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
С	0.090	-	0.200	0.0035	-	0.0079
D	8.800	9.000	9.200	0.3465	0.3543	0.3622
D1	6.800	7.000	7.200	0.2677	0.2756	0.2835
D3	-	5.500	-	-	0.2165	-
E	8.800	9.000	9.200	0.3465	0.3543	0.3622
E1	6.800	7.000	7.200	0.2677	0.2756	0.2835
E3	-	5.500	-	-	0.2165	-
е	-	0.500	-	-	0.0197	-
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
k	0°	3.5°	7°	0°	3.5°	7°
ccc	-	-	0.080	-	-	0.0031

^{1.} Values in inches are converted from mm and rounded to 4 decimal digits.

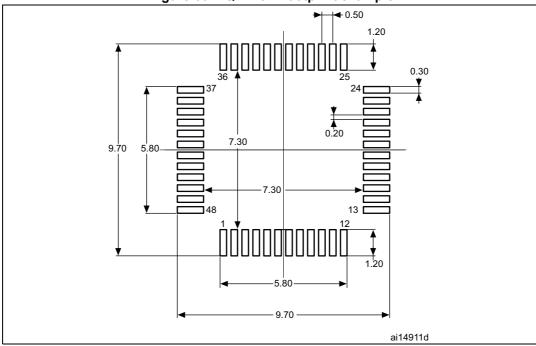


Figure 60. LQFP48 - Footprint example

1. Dimensions are expressed in millimeters.

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6.5 LQFP64 package information (5W)

This LQFP is a 64-pin, 10 x 10 mm low-profile quad flat package.

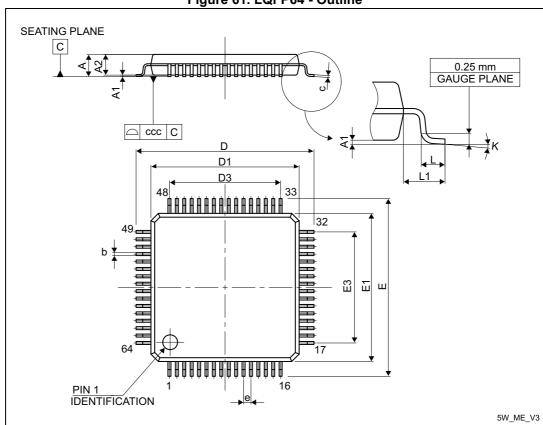


Figure 61. LQFP64 - Outline

1. Drawing is not to scale.

Table 110. LQFP64 - Mechanical data

Cumbal	millimeters			inches ⁽¹⁾		
Symbol	Min	Тур	Max	Min	Тур	Max
Α	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
С	0.090	-	0.200	0.0035	-	0.0079
D	-	12.000	-	-	0.4724	-
D1	-	10.000	-	-	0.3937	-
D3	-	7.500	-	-	0.2953	-
E	-	12.000	-	-	0.4724	-
E1	-	10.000	-	-	0.3937	-

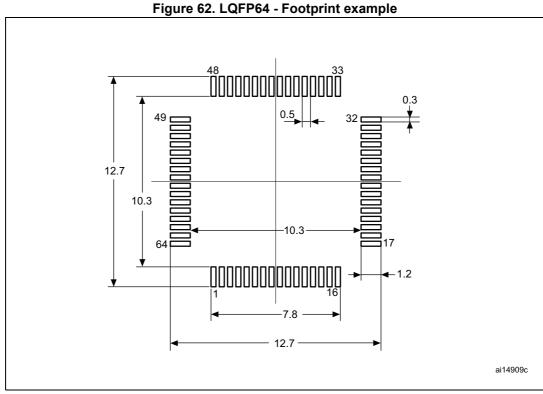


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Symbol	millimeters			inches ⁽¹⁾		
	Min	Тур	Max	Min	Тур	Max
E3	-	7.500	-	-	0.2953	-
е	-	0.500	-	-	0.0197	-
K	0°	3.5°	7°	0°	3.5°	7°
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
ccc	-	-	0.080	-	-	0.0031

Table 110. LQFP64 - Mechanical data (continued)

^{1.} Values in inches are converted from mm and rounded to 4 decimal digits.



1. Dimensions are expressed in millimeters.

6.6 LQFP80 package information (9X)

This LQFP is a 80 pin, 12 x 12 mm low-profile quad flat package.

Note: See list of notes in the notes section.

Figure 63. LQFP80 - Outline⁽¹⁵⁾

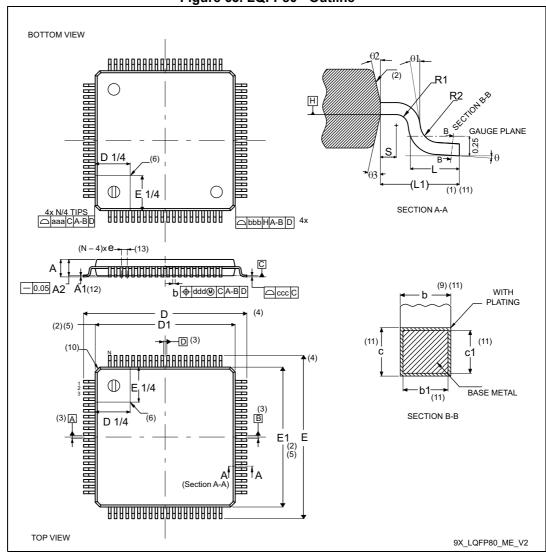


Table 111. LQFP80 - Mechanical data

Dise	mm			inches ⁽¹⁴⁾		
Dim.	Min	Тур	Max	Min	Max	
Α	-	-	1.60	-	-	0.0630
A1 ⁽¹²⁾	0.05	-	0.15	0.0020	-	0.0059
A2	1.35	1.40	1.45	0.0531	0.0551	0.0571
b ⁽⁹⁾⁽¹¹⁾	0.17	0.22	0.27	0.0067	0.0087	0.0106
b1 ⁽¹¹⁾	0.17	0.20	0.23	0.0067	0.0078	0.0090
c ⁽¹¹⁾	0.09	-	0.20	0.0038	-	0.0067
c1 ⁽¹¹⁾	0.09	-	0.16	0.0038	-	0.0063
D		14.00 BSC			0.5512 BSC	
D1		12.00 BSC			0.4724 BSC	
E		14.00 BSC		0.5512 BSC		
E1		12.00 BSC		0.4724 BSC		
е		0.50 BSC		0.0197 BSC		
L	0.45	0.60	0.75	0.0177	0.0236	0.0295
L1		1.00 REF		0.0394 REF		
N ⁽¹³⁾			;	30		
θ	0°	3.5°	7°	0°	3.5°	7°
θ1	0°	-	-	0°	-	-
θ2	10°	12°	14°	10°	12°	14°
θ3	10°	12°	14°	10°	12°	14°
R1	0.08	-	-	0.0031	-	-
R2	0.08	-	0.20	0.0031	-	0.0079
S	0.20	-	-	0.0079	-	-
aaa ⁽¹⁾	0.20			0.0079		
bbb ⁽¹⁾	0.20			0.0079		
ccc ⁽¹⁾		0.08		0.0031		
ddd ⁽¹⁾		0.08			0.0031	

Notes:

- Dimensioning and tolerancing schemes conform to ASME Y14.5M-1994.
- 2. The Top package body size may be smaller than the bottom package size by as much as 0.15 mm.
- 3. Datums A-B and D to be determined at datum plane H.
- To be determined at seating datum plane C.
- Dimensions D1 and E1 do not include mold flash or protrusions. Allowable mold flash or protrusions is "0.25 mm" per side. D1 and E1 are Maximum plastic body size dimensions including mold mismatch.
- Details of pin 1 identifier are optional but must be located within the zone indicated.
- All Dimensions are in millimeters.
- 8. No intrusion allowed inwards the leads.
- Dimension "b" does not include dambar protrusion. Allowable dambar protrusion shall not cause the lead width to exceed the maximum "b" dimension by more than 0.08 mm. Dambar cannot be located on the lower radius or the foot. Minimum space between protrusion and an adjacent lead is 0.07 mm for 0.4 mm and 0.5 mm pitch packages.
- 10. Exact shape of each corner is optional.
- 11. These dimensions apply to the flat section of the lead between 0.10 mm and 0.25 mm from the lead tip.
- 12. A1 is defined as the distance from the seating plane to the lowest point on the package
- 13. "N" is the number of terminal positions for the specified body size.
- 14. Values in inches are converted from mm and rounded to 4 decimal digits.
- 15. Drawing is not to scale.

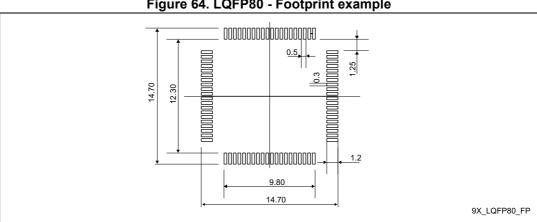


Figure 64. LQFP80 - Footprint example

1. Dimensions are expressed in millimeters.

6.7 TFBGA100 package information (A08Q)

This TFBGA is a 100-ball, 8 x 8 mm, 0.8 mm pitch fine pitch ball grid array package.

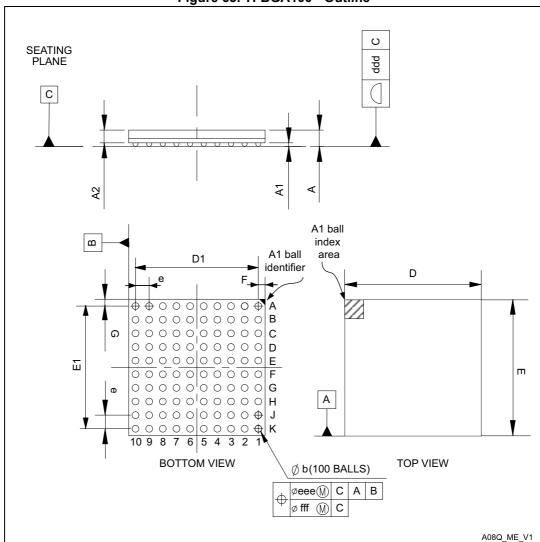


Figure 65. TFBGA100 - Outline

Table 112. TFBGA100 - Mechanical data

Comple of	millimeters			inches ⁽¹⁾		
Symbol	Min	Тур	Max	Min	Тур	Max
Α	-	-	1.100	-	-	0.0433
A1	0.150	-	-	0.0059	-	-
A2	-	0.760	-	-	0.0299	-
b	0.350	0.400	0.450	0.0138	0.0157	0.0177
D	7.850	8.000	8.150	0.3091	0.3150	0.3209
D1	-	7.200	-	-	0.2835	-
E	7.850	8.000	8.150	0.3091	0.3150	0.3209
E1	-	7.200	-	-	0.2835	-
е	-	0.800	-	-	0.0315	-
F	-	0.400	-	-	0.0157	-
G	-	0.400	-	-	0.0157	-
ddd	-	-	0.100	-	-	0.0039
eee	-	-	0.150	-	-	0.0059
fff	-	-	0.080	-	-	0.0031

^{1.} Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 66. TFBGA100 - Footprint example

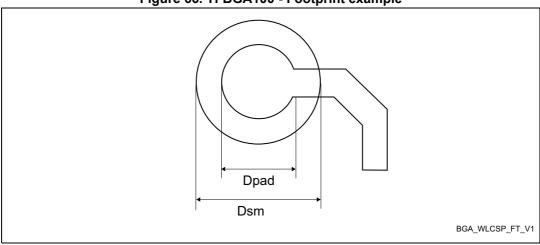


Table 113. TFBGA100 - Example of PCB design rules

Dimension	Recommended values		
Pitch	0.8		
Dpad	0.400 mm		
Dsm	0.470 mm typ. (depends on the soldermask registration tolerance)		

Table 113. TFBGA100 - Example of PCB design rules (continued)

Dimension	Recommended values		
Stencil opening	0.400 mm		
Stencil thickness	Between 0.100 mm and 0.125 mm		
Pad trace width	0.120 mm		

6.8 LQFP100 package information (1L)

This LQFP is 100 lead, 14 x 14 mm low-profile quad flat package.

Note: See list of notes in the notes section.

Figure 67. LQFP100 - Outline⁽¹⁵⁾

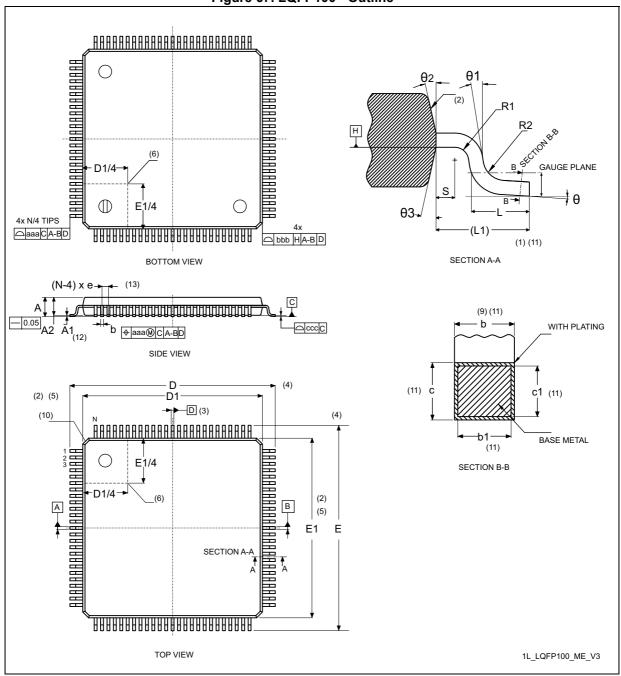


Table 114. LQFP100 - Mechanical data

O make at	millimeters			inches ⁽¹⁴⁾		
Symbol	Min	Тур	Max	Min	Тур	Max
Α	-	1.50	1.60	-	0.0590	0.0630
A1 ⁽¹²⁾	0.05	-	0.15	0.0019	-	0.0059
A2	1.35	1.40	1.45	0.0531	0.0551	0.0570
b ⁽⁹⁾⁽¹¹⁾	0.17	0.22	0.27	0.0067	0.0087	0.0106
b1 ⁽¹¹⁾	0.17	0.20	0.23	0.0067	0.0079	0.0090
c ⁽¹¹⁾	0.09	-	0.20	0.0035	-	0.0079
c1 ⁽¹¹⁾	0.09	-	0.16	0.0035	-	0.0063
D ⁽⁴⁾		16.00 BSC			0.6299 BSC	
D1 ⁽²⁾⁽⁵⁾		14.00 BSC			0.5512 BSC	
E ⁽⁴⁾		16.00 BSC		0.6299 BSC		
E1 ⁽²⁾⁽⁵⁾		14.00 BSC		0.5512 BSC		
е		0.50 BSC		0.0197 BSC		
L	0.45	0.60	0.75	0.177	0.0236	0.0295
L1 ⁽¹⁾⁽¹¹⁾		1.00		-	0.0394	-
N ⁽¹³⁾			1	00		
θ	0°	3.5°	7°	0°	3.5°	7°
θ1	0°	-	-	0°	-	-
θ2	10°	12°	14°	10°	12°	14°
θ3	10°	12°	14°	10°	12°	14°
R1	0.08	-	-	0.0031	-	-
R2	0.08	-	0.20	0.0031	-	0.0079
S	0.20	-	-	0.0079	-	-
aaa ⁽¹⁾	0.20				0.0079	
bbb ⁽¹⁾	0.20			0.0079		
ccc ⁽¹⁾		0.08		0.0031		
ddd ⁽¹⁾		0.08			0.0031	

Notes:

- Dimensioning and tolerancing schemes conform to ASME Y14.5M-1994.
- 2. The Top package body size may be smaller than the bottom package size by as much as 0.15 mm.
- 3. Datums A-B and D to be determined at datum plane H.
- To be determined at seating datum plane C.
- Dimensions D1 and E1 do not include mold flash or protrusions. Allowable mold flash or protrusions is "0.25 mm" per side. D1 and E1 are Maximum plastic body size dimensions including mold mismatch.
- Details of pin 1 identifier are optional but must be located within the zone indicated.
- All Dimensions are in millimeters.
- 8. No intrusion allowed inwards the leads.
- Dimension "b" does not include dambar protrusion. Allowable dambar protrusion shall not cause the lead width to exceed the maximum "b" dimension by more than 0.08 mm. Dambar cannot be located on the lower radius or the foot. Minimum space between protrusion and an adjacent lead is 0.07 mm for 0.4 mm and 0.5 mm pitch packages.
- 10. Exact shape of each corner is optional.
- 11. These dimensions apply to the flat section of the lead between 0.10 mm and 0.25 mm from the lead tip.
- 12. A1 is defined as the distance from the seating plane to the lowest point on the package
- 13. "N" is the number of terminal positions for the specified body size.
- 14. Values in inches are converted from mm and rounded to 4 decimal digits.
- 15. Drawing is not to scale.

16.7 1L LQFP100 FP V1

Figure 68. LQFP100 - Footprint example

1. Dimensions are expressed in millimeters.

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6.9 LQFP128 package information (TC)

This LQFP128 is a 128-pin, 14 x 14 mm low-profile quad flat package.

Note: See list of notes in the notes section.

Figure 69. LQFP128 - Outline⁽¹⁵⁾

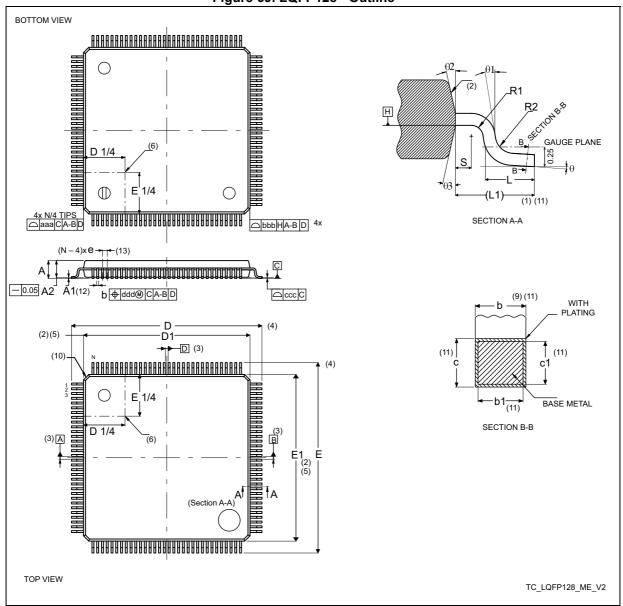


Table 115. LQFP128 - Mechanical data

Oh al	Millimeters			Inches ⁽¹⁴⁾		
Symbol	Min.	Тур.	Max.	Min.	Тур.	Max.
Α	-	-	1.60	-	-	0.0630
A1 ⁽¹²⁾	0.05	-	0.15	0.0020		0.0059
A2	1.35	1.40	1.45	0.0531	0.0551	0.0571
b ⁽⁹⁾⁽¹¹⁾	0.13	0.18	0.23	0.0051	0.0071	0.0091
b1 ⁽¹¹⁾	0.13	0.16	0.19	0.0051	0.0063	0.0075
c ⁽¹¹⁾	0.09	-	0.20	0.0035	-	0.0079
c1 ⁽¹¹⁾	0.09	-	0.16	0.0035	-	0.0063
D ⁽⁴⁾		16.00 BSC			0.6299 BSC	
D1 ⁽⁵⁾		14.00 BSC			0.5512 BSC	
E ⁽⁴⁾		16.00 BSC		0.6299 BSC		
E1 ⁽²⁾⁽⁵⁾		14.00 BSC		0.5512 BSC		
е		0.40 BSC		0.0157 BSC		
L	0.45	0.60	0.75	0.0177	0.0236	0.0295
L1		1.00 REF		0.0394 REF		
N ⁽¹³⁾			12	28		
θ	0°	3.5°	7°	0°	3.5°	7°
θ1	0°	-	-	0°	-	-
θ2	10°	12°	14°	10°	12°	14°
θ3	10°	12°	14°	10°	12°	14°
R1	0.08	-	-	0.0031	-	-
R2	0.08	-	0.20	0.0031	-	0.0079
S	0.20	-	-	0.0079	-	-
aaa ⁽¹⁾	0.20			0.0079		
bbb ⁽¹⁾	0.20			0.0079		
ccc ⁽¹⁾		0.08		0.0031		
ddd ⁽¹⁾		0.07			0.0027	

Notes:

- 1. Dimensioning and tolerancing schemes conform to ASME Y14.5M-1994.
- 2. The Top package body size may be smaller than the bottom package size by as much as 0.15 mm.
- 3. Datums A-B and D to be determined at datum plane H.
- To be determined at seating datum plane C.
- 5. Dimensions D1 and E1 do not include mold flash or protrusions. Allowable mold flash or protrusions is "0.25 mm" per side. D1 and E1 are Maximum plastic body size dimensions including mold mismatch.
- 6. Details of pin 1 identifier are optional but must be located within the zone indicated.
- 7. All Dimensions are in millimeters.
- 8. No intrusion allowed inwards the leads.
- 9. Dimension "b" does not include dambar protrusion. Allowable dambar protrusion shall not cause the lead width to exceed the maximum "b" dimension by more than 0.08 mm. Dambar cannot be located on the lower radius or the foot. Minimum space between protrusion and an adjacent lead is 0.07 mm for 0.4 mm and 0.5 mm pitch packages.
- 10. Exact shape of each corner is optional.
- 11. These dimensions apply to the flat section of the lead between 0.10 mm and 0.25 mm from the lead tip.
- 12. A1 is defined as the distance from the seating plane to the lowest point on the package body.
- 13. "N" is the number of terminal positions for the specified body size.
- 14. Values in inches are converted from mm and rounded to 4 decimal digits.
- 15. Drawing is not to scale.

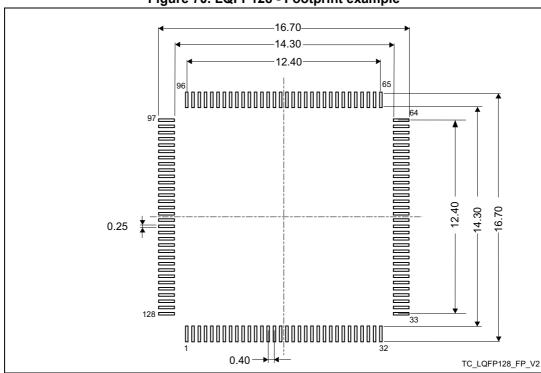


Figure 70. LQFP128 - Footprint example

1. Dimensions are expressed in millimeters.

6.10 UFBGA121 package information (B0CU)

This UFBGA is a 121 balls, 6 x 6 mm, 0.5 mm pitch, fine pitch, square ball grid array package.

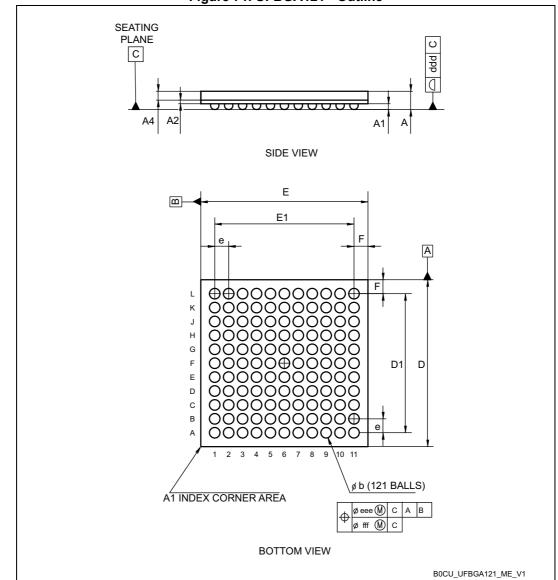


Figure 71. UFBGA121 - Outline

- 1. Drawing is not to scale.
- The terminal A1 corner must be identified on the top surface by using a corner chamfer, ink or metalized
 markings, or other feature of package body or integral heat slug.
 A distinguishing feature is allowable on the bottom surface of the package to identify the terminal A1
 corner. Exact shape of each corner is optional.

Table 116. UFBGA121 - Mechanical data

Symbol	millimeters			inches ⁽¹⁾		
	Min	Тур	Max	Min	Тур	Max
A ⁽²⁾	-	-	0.60	-	-	0.0236
A1	-	-	0.11	-	-	0.0043
A2	-	0.13	-	-	0.0051	-
A4	-	0.32	-	-	0.0126	-
b ⁽³⁾	0.24	0.29	0.34	0.0094	0.0114	0.0134
D	5.85	6.00	6.15	0.2303	0.2362	0.2421
D1	-	5.00	-	-	0.1969	-
Е	5.85	6.00	6.15	0.2303	0.2362	0.2421
E1	-	5.00	-	-	0.1969	-
е	-	0.50	-	-	0.0197	-
F	-	0.50	-	-	0.0197	-
ddd	-	-	0.08	-	-	0.0031
eee ⁽⁴⁾	-	-	0.15	-	-	0.0059
fff ⁽⁵⁾	-	-	0.05	-	-	0.0020

- 1. Values in inches are converted from mm and rounded to 4 decimal digits.
- 2. UFBGA stands for Ultra-Thin Profile Fine Pitch Ball Grid Array.
 Ultra Thin profile: 0.50 < A ≤ 0.65mm / Fine pitch: e < 1.00mm pitch.
 The total profile height (Dim A) is measured from the seating plane to the top of the component
 The maximum total package height is calculated by the following methodology:
 A Max = A1 Typ + A2 Typ + A4 Typ + √ (A1²+A2²+A4² tolerance values)
- 3. The typical balls diameters before mounting is 0.20 mm
- The tolerance of position that controls the location of the pattern of balls with respect to datum A and B. For each ball there is a cylindrical tolerance zone eee perpendicular to datum C and located on true position with respect to datum A and B as defined by e. The axis perpendicular to datum C of each ball must lie within this tolerance zone.
- The tolerance of position that controls the location of the balls within the matrix with respect to each other. For each ball there is a cylindrical tolerance zone fff perpendicular to datum C and located on true position as defined by e. The axis perpendicular to datumC of each ball must lie within this tolerance zone. Each tolerance zone fff in the array is contained entirely in the respective zone eee above. The axis of each ball must lie within the contained entirely in the respective zone eee above. ball must lie simultaneously in both tolerance zones.



Dpad
Dsm

BGA_WLCSP_FT_V1

Figure 72. UFBGA121 - Footprint example

Table 117. UFBGA121 - Example of PCB design rules

Dimension	Values
Pitch	0.5 mm
Dpad	0.225 mm
Dsm	0.290 mm typ. (depends on soldermask registration tolerance)
Stencil opening	0.250 mm
Stencil thickness	0.100 mm

6.11 Thermal characteristics

The maximum chip-junction temperature, T_J max, in degrees Celsius, may be calculated using the following equation:

 $T_J \max = T_A \max + (P_D \max x \Theta_{JA})$

Where:

- T_A max is the maximum ambient temperature in °C,
- Θ_{JA} is the package junction-to-ambient thermal resistance, in °C/W,
- P_D max is the sum of P_{INT} max and $P_{I/O}$ max (P_D max = P_{INT} max + $P_{I/O}$ max),
- P_{INT} max is the product of I_{DD} and V_{DD}, expressed in Watts. This is the maximum chip internal power.

P_{I/O} max represents the maximum power dissipation on output pins where:

$$P_{I/O}$$
 max = Σ ($V_{OL} \times I_{OL}$) + Σ (($V_{DDIOx} - V_{OH}$) × I_{OH}),

taking into account the actual V_{OL} / I_{OL} and V_{OH} / I_{OH} of the I/Os at low and high level in the application.

Table 118. Package thermal characteristics

Symbol	Parameter	Value	Unit
Θ_{JA}	Thermal resistance junction-ambient LQFP128 - 14 × 14 mm	43.0	
	Thermal resistance junction-ambient LQFP100 - 14 × 14 mm	46.2	
	Thermal resistance junction-ambient LQFP80 - 12 × 12 mm	46.8	
	Thermal resistance junction-ambient LQFP64 - 10 × 10 mm	47.9	
	Thermal resistance junction-ambient LQFP48 - 7 × 7 mm	55.2	°C/W
	Thermal resistance junction-ambient TFBGA100 - 8 × 8 mm	30.8	
	Thermal resistance junction-board UFBGA121 - 6 × 6 mm	TBD	
	Thermal resistance junction-ambient UFQFPN48 - 7 × 7 mm	26.8	
	Thermal resistance junction-ambient WLCSP81 - 4.02 X 4.27 mm	45	



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Table 118. Package thermal characteristics (continued)

Symbol	Parameter	Value	Unit
	Thermal resistance junction-case LQFP128 - 14 × 14 mm	7.0	
	Thermal resistance junction-case LQFP100 - 14 × 14 mm	8.3	
	Thermal resistance junction-case LQFP80 - 12 × 12 mm	8.2	
	Thermal resistance junction-case LQFP64 - 10 × 10 mm	8.0	
Θ _{JC}	Thermal resistance junction-case LQFP48 - 7 × 7 mm	9.6	°C/W
	Thermal resistance junction-case TFBGA100 - 8 × 8 mm	13	
	Thermal resistance junction-board UFBGA121 - 6 × 6 mm	TBD	
	Thermal resistance junction-case UFQFPN48 - 7 × 7 mm	2 ⁽¹⁾ 7.5	
	Thermal resistance junction-case WLCSP81 - 4.02 X 4.27 mm	1.46	
	Thermal resistance junction-board LQFP128 - 14 × 14 mm	19.9	
$\Theta_{ extsf{JB}}$	Thermal resistance junction-board LQFP100 - 14 × 14 mm	22.9	
	Thermal resistance junction-board LQFP80 - 12 × 12 mm	22.3	
	Thermal resistance junction-board LQFP64 - 10 × 10 mm	21.8	
	Thermal resistance junction-board LQFP48 - 7 × 7 mm	24.3	°C/W
	Thermal resistance junction-board TFBGA100 - 8 × 8 mm	13.42	
	Thermal resistance junction-board UFBGA121 - 6 × 6 mm	TBD	
	Thermal resistance junction-board UFQFPN48 - 7 × 7 mm	11	
	Thermal resistance junction-board WLCSP81 - 4.02 X 4.27 mm	27.45	

^{1.} Thermal resistance junction-case where the case is the bottom thermal pad on the UFQFPN package.

6.11.1 Reference document

JESD51-2 Integrated Circuits Thermal Test Method Environment Conditions - Natural Convection (Still Air). Available from www.jedec.org



6.11.2 Selecting the product temperature range

When ordering the microcontroller, the temperature range is specified in the ordering information scheme shown in *Section 7: Ordering information*.

Each temperature range suffix corresponds to a specific guaranteed ambient temperature at maximum dissipation and, to a specific maximum junction temperature.

As applications do not commonly use the STM32G473xE at maximum dissipation, it is useful to calculate the exact power consumption and junction temperature to determine which temperature range is best suited to the application.

The following examples show how to calculate the temperature range needed for a given application.

Example 1: High-performance application

Assuming the following application conditions:

Maximum ambient temperature T_{Amax} = 82 °C (measured according to JESD51-2), I_{DDmax} = 50 mA, V_{DD} = 3.5 V, maximum 20 I/Os used at the same time in output at low level with I_{OL} = 8 mA, V_{OL} = 0.4 V and maximum 8 I/Os used at the same time in output at low level with I_{OL} = 20 mA, V_{OL} = 1.3 V

 $P_{INTmax} = 50 \text{ mA} \times 3.5 \text{ V} = 175 \text{ mW}$

 $P_{IOmax} = 20 \times 8 \text{ mA} \times 0.4 \text{ V} + 8 \times 20 \text{ mA} \times 1.3 \text{ V} = 272 \text{ mW}$

This gives: P_{INTmax} = 175 mW and P_{IOmax} = 272 mW:

 $P_{Dmax} = 175 + 272 = 447 \text{ mW}$

Using the values obtained in T_{Jmax} is calculated as follows:

For LQFP100, 42 °C/W

 $T_{\text{lmax}} = 82 \,^{\circ}\text{C} + (42 \,^{\circ}\text{C/W} \times 447 \,^{\circ}\text{mW}) = 82 \,^{\circ}\text{C} + 18.774 \,^{\circ}\text{C} = 100.774 \,^{\circ}\text{C}$

This is within the range of the suffix 6 version parts ($-40 < T_J < 105$ °C) see Section 7: Ordering information.

In this case, parts must be ordered at least with the temperature range suffix 6 (see Section 7: Ordering information).

Note:

With this given P_{Dmax} we can find the TAmax allowed for a given device temperature range (order code suffix 6 or 7).

Suffix 6:
$$T_{Amax} = T_{Jmax} - (42 \text{ °C/W} \times 447 \text{ mW}) = 105 - 18.774 = 86.226 \text{ °C}$$

Suffix 3:
$$T_{Amax} = T_{Jmax} - (42 \degree C/W \times 447 \ mW) = 130-18.774 = 111.226 \degree C$$

Example 2: High-temperature application

Using the same rules, it is possible to address applications that run at high ambient temperatures with a low dissipation, as long as junction temperature T_J remains within the specified range.



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Assuming the following application conditions:

Maximum ambient temperature T_{Amax} = 100 °C (measured according to JESD51-2), I_{DDmax} = 20 mA, V_{DD} = 3.5 V, maximum 20 I/Os used at the same time in output at low level with I_{OL} = 8 mA, V_{OL} = 0.4 V

 $P_{INTmax} = 20 \text{ mA} \times 3.5 \text{ V} = 70 \text{ mW}$

 $P_{IOmax = 20} \times 8 \text{ mA} \times 0.4 \text{ V} = 64 \text{ mW}$

This gives: $P_{INTmax} = 70 \text{ mW}$ and $P_{IOmax} = 64 \text{ mW}$:

 $P_{Dmax} = 70 + 64 = 134 \text{ mW}$

Thus: P_{Dmax} = 134 mW

Using the values obtained in T_{Jmax} is calculated as follows:

For LQFP100, 42 °C/W

 T_{Jmax} = 100 °C + (42 °C/W × 134 mW) = 100 °C + 5.628 °C = 105.628 °C

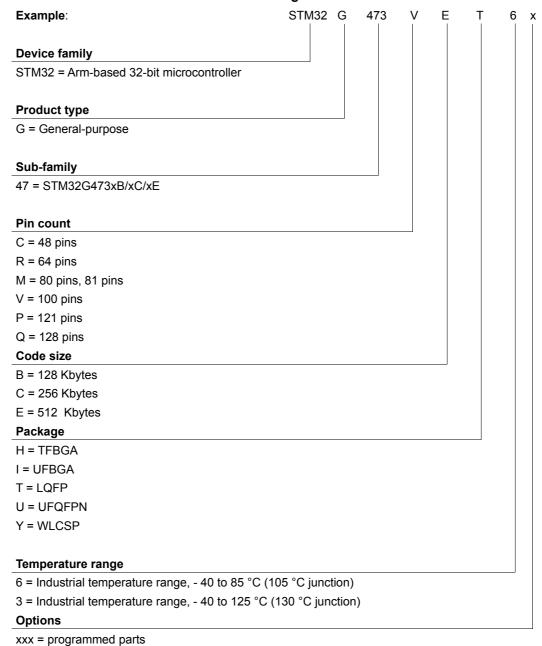
This is above the range of the suffix 6 version parts ($-40 < T_J < 105$ °C).

In this case, parts must be ordered at least with the temperature range suffix 3 (see Section 7: Ordering information) unless we reduce the power dissipation in order to be able to use suffix 6 parts.



7 Ordering information

Table 119. Ordering information



For a list of available options (memory, package, and so on) or for further information on any aspect of this device, contact the nearest ST sales office.

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TR = tape and reel

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9 Revision history

Table 120. Document revision history

Table 120. Document revision history				
Date	Revision	Changes		
08-May-2019	1	initial release		
17-Oct-2019	2	 Updated: Section 2: Description, Section 3.5: Embedded SRAM, Table 2: STM32G473xB/xC/xE features and peripheral counts, Table 17: General operating conditions, Table 35: Peripheral current consumption, Table 61: ADC characteristics, Table 62: Maximum ADC RAIN, Table 84: SPI characteristics, Table 118: Package thermal characteristics, Table 119: Ordering information Added: Table 66: ADC accuracy (Multiple ADCs operation) - limited test conditions 1, Table 68: ADC accuracy (Multiple ADCs operation) - limited test conditions 3, Table 68: ADC accuracy (Multiple ADCs operation) - limited test conditions 3 		
30-Oct-2020	3	Updated: - Table 1: Device summary - Table 12: STM32G473xB/xC/xE pin definition - Section 3.18: Analog-to-digital converter (ADC) - Table 17: General operating conditions - Table 21: Current consumption in Run and Low-power run modes, code with data processing running from flash in single Bank, ART enable (Cache ON Prefetch OFF) - Table 22: Current consumption in Run and Low-power run modes, code with data processing running from flash in dual bank, ART enable (Cache ON Prefetch OFF) - Table 23: Current consumption in Run and Low-power run modes, code with data processing running from SRAM1 - Table 28: Current consumption in Sleep and Low-power run modes, code with data processing running from SRAM1 - Table 28: Current consumption in Sleep and Low-power sleep mode flash memory ON - Table 29: Current consumption in low-power sleep modes, flash memory in power-down - Table 30: Current consumption in Stop 1 mode - Table 31: Current consumption in Stop 0 mode - Table 32: Current consumption in Standby mode - Table 51: ESD absolute maximum ratings - Table 71: DAC 15MSPS characteristics - Table 75: OPAMP characteristics - Table 75: OPAMP characteristics - Table 75: DPAMP characteristics - Table 85: I2S characteristics - Table 16: UFBGA121 - Mechanical data - Table 116: UFBGA121 - Recommended PCB design rules - Table 119: Ordering information - Figure 71: UFBGA121 - Outline - Figure 72: UFBGA121 - Footprint example		



Table 120. Document revision history (continued)

Date	Revision	Changes	
30-Oct-2020	3 (Cont.)	Added: - Figure 102: LQFP128 - Recommended footprint - Table 14: STM32G473xB/xC/xE UFBGA121 pinout, - Table 116: UFBGA121 - Mechanical data - Table 117: UFBGA121 - Recommended PCB design rules - Table 71: UFBGA121 - Outline Table 72: UFBGA121 - Footprint example	
16-Nov-2021	4	Updated: - Features - Section 2: Description - Section 3.4: Embedded flash memory - Table 2: STM32G473xB/xC/xE features and peripheral counts - Table 5: Temperature sensor calibration values - Table 12: STM32G473xB/xC/xE pin definition - Figure 14: STM32G473xB/xC/xE UFBGA121 pinout - Table 30: Current consumption in Stop 1 mode - Table 31: Current consumption in Stop 0 mode - Table 32: Current consumption in Standby mode - Section 5.3.14: I/O port characteristics - Table 63: ADC accuracy - limited test conditions 1 - Table 64: ADC accuracy - limited test conditions 2 - Table 65: ADC accuracy - limited test conditions 3 - Figure 29: ADC accuracy characteristics - Figure 30: Typical connection diagram when using the ADC with FT/TT pins featuring analog switch function - Figure 102: LQFP128 - Recommended footprint Added: - Note 1 in Table 77: V _{BAT} monitoring characteristics	
20-Apr-2023	5	Updated: - Section 3.11.1: Power supply schemes - Table 2: STM32G473xB/xC/xE features and peripheral counts - Table 13: Alternate function - Table 19: Embedded reset and power control block characteristics - Section 6: Package information Added: - Figure 3: Power-up/down sequence - Section 6.1: Device marking - Section 8: Important security notice	

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