Computerized control - Final Exam - modified from Fall 2015

Kjartan Halvorsen

2015-11-24

The dynamic model of a ship with input u being the rudder angle and the output y being the heading (see figure 1) can be described as a continuous-time second order system with a pole in the origin

$$G(s) = \frac{K}{s(s+a)}.$$

For fully loaded, large tankers this dynamics is often unstable, meaning that $a < 0^{-1}$.

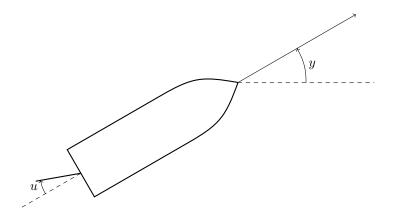


Figure 1: Heading of a ship controlled by rudder input.

Consider for this exam the normalized continuous-time model of the tanker

$$G(s) = \frac{1}{s(s-1)}.$$

with the discrete-time model obtained by zero-order hold

$$H(z) = \frac{(-1 + e^h - h)z + 1 - (1 - h)e^h}{(z - 1)(z - e^h)}.$$

Specifically, use sampling time h = 0.2, which gives the (approximate) model

$$H(z) = \frac{0.02z + 0.02}{(z - 1)(z - 1.2)} = \frac{0.02z + 0.02}{z^2 - 2.2z + 1.2}.$$
 (1)

All answers should be well motivated!

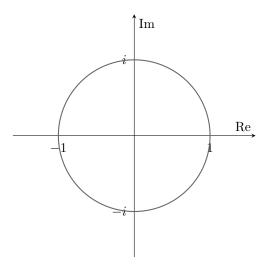


Figure 2: Problem 1: Plot the poles and zeros of the discrete-time system.

Problem 1

- 1. In figure 2 draw the poles (crosses) and zero (circle) for the discrete-time pulse-transfer function in (1).
- 2. Assume that the tanker with model (1) is stabilized using error-feedback and a PD-controller. The Bode-diagram of the resulting **closed-loop** system is given in figure 3. What is the bandwidth of the closed-loop system? At what frequency is the resonance peak?

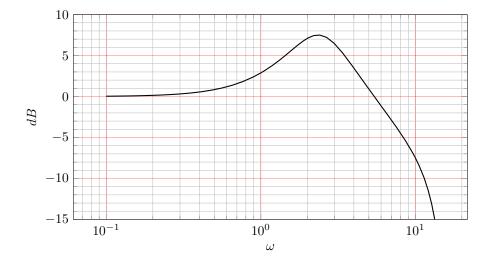


Figure 3: Problem 1: Bode diagram of closed-loop system with PD-control

Problem 2

Figure 4 shows a system controlled with an RST controller. Note that the system includes an anti-aliasing filter modelled as a pure time-delay of two sampling periods. What is the closed-loop pulse-transfer function from the disturbance d to the output y? You do not need to multiply the polynomials. It is sufficient to state your answer in terms of A(z), B(z), B

¹Fossen, Thor I. Handbook of marine craft hydrodynamics and motion control. John Wiley & Sons, 2011.

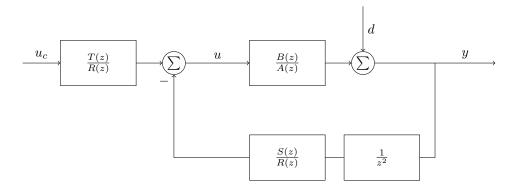


Figure 4: Problem 2: Two-degree-of-freedom controller with anti-aliasing filter.

Problem 3

When designing an RST-controller for the system in Problem 2, R(z) and S(z) are determined from a diophantine equation, based on the required placement of the closed-loop poles. Assume the following desired closed-loop denominator:

$$A_{cl} = \underbrace{(z - p_1)(z - p_2)z^2}_{A_c} \underbrace{(z - p_3)^3}_{A_c} \tag{2}$$

- 1. Write the diophantine equation in terms of $A_c(z)$, $A_o(z)$, A(z), B(z), R(z), S(z) and Z^2 .
- 2. Let the controller polynomials R(z) and S(z) have the same order. Determine this order, so that all the controller parameters can be determined from the diophantine equation. Note that you only need to determine the **order** of the controller. You do not need to write the equation for the controller parameters.

Problem 4

The controllable canonical state-space representation of (1) is given by

$$x(k+1) = \begin{bmatrix} 2.2 & -1.2 \\ 1 & 0 \end{bmatrix} (k) + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u(k)$$
$$y(k) = \begin{bmatrix} 0.02 & 0.02 \end{bmatrix} x(k),$$
 (3)

with

$$x(k) = \begin{bmatrix} x_1(k) \\ x_2(k) \end{bmatrix}.$$

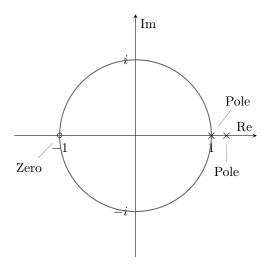
Introduce the state-feedback law $u(k) = -l_1x_1(k) - l_2x_2(k)$ and determine l_1 and l_2 so that the closed-loop system has the characteristic polynomial

$$(z - 0.9 + 0.1i)(z - 0.9 - 0.1i) = z^2 - 1.8z + 0.82.$$

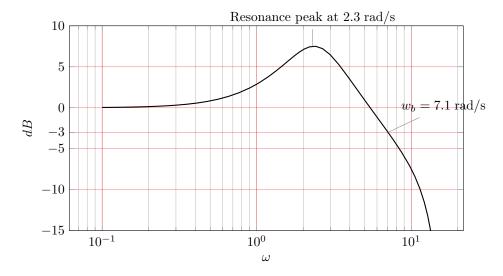
Solution

Problem 1

1. Poles and zeros



2. Bandwidth and resonance



Problem 2

To compute the pulse-transfer function from d to y, assume $u_c = 0$. We get

$$Y = D + \frac{B}{A}U = D - \frac{B}{A}\frac{S}{R}\frac{1}{z^2}Y$$

$$Y + \frac{BS}{ARz^2}Y = D$$

$$Y = \frac{1}{1 + \frac{BS}{ARz^2}}D$$

$$= \frac{A(z)R(z)z^2}{A(z)R(z)z^2 + B(z)S(z)}D$$

Problem 3

1. The diophantine equation becomes

$$A(z)R(z)z^2 + B(z)S(z) = A_c(z)A_o(z)$$

2. The right hand side of the diophantine equation has order 7, hence the left hand side must have the same order. Since $A(z)z^2$ has order 4, then R(z) must have order 3. We choose R(z) and S(z)

to have the same order (which is a smart choice because then the pulse-transfer function of the controller does not introduce a time-delay), we get

$$\frac{S(z)}{R(z)} = \frac{s_0 z^3 + s_1 z^2 + s_2 z + s_3}{z^3 + r_1 z^2 + r_2 z + r_3}$$

which has 7 parameters. The diophantine equation gives 7 equations to determine uniquely the 7 control parameters. Note that the terms $A(z)R(z)z^2$ and B(z)S(z) do **not** have to have the same order.

Problem 4

With the control law we get the closed-loop system

$$x(k+1) = \begin{bmatrix} 2.2 & -1.2 \\ 1 & 0 \end{bmatrix} x(k) - \begin{bmatrix} l_1 & l_2 \\ 0 & 0 \end{bmatrix} x(k)$$
$$= \begin{bmatrix} 2.2 - l_1 & -1.2 - l_2 \\ 1 & 0 \end{bmatrix} x(k)$$

which is also on controllable canonical form. Thus we can immediately write the characteristic polynomial of the closed-loop system as

$$z^2 + (-2.2 + l_1)z + (1.2 + l_2).$$

It is also straight-forward to write the characteristic polynomial using the formula

$$\det \left(zI - \begin{bmatrix} 2.2 - l_1 & -1.2 - l_2 \\ 1 & 0 \end{bmatrix} \right) = \det \begin{bmatrix} z - 2.2 + l_1 & 1.2 + l_2 \\ -1 & z \end{bmatrix}$$
$$= (z - 2.2 + l_1)z + (1.2 + l_2) = z^2 + (-2.2 + l_1)z + (1.2 + l_2).$$

Comparing coefficients with the desired characteristic polynomial

$$z^2 - 1.8 + 0.82$$

gives the solution

$$l_1 = -1.8 + 2.2 = 0.4$$

 $l_2 = 0.82 - 1.2 = -0.38$

With

$$m_0 = \frac{A_c(1)}{B(1)} = \frac{1 - 1.8 + 0.82}{0.02 + 0.02} = 0.5$$

and

$$u = -Lx + m_0 u_c,$$

the Bode-diagram of the closed-loop system becomes

