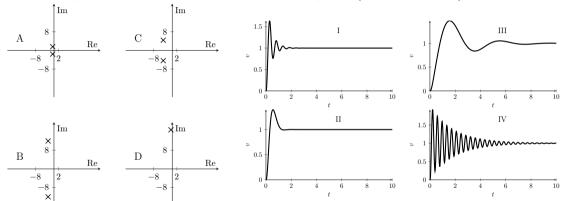
Root locus

Kjartan Halvorsen

2021-07-06

Pole-placement and time-response

Pair the pole-placement with the correct time-response (continuous time)!



Mapping of poles from continuous time to discrete time

Continuous time

Discrete time

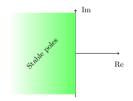
$$Y(s) \triangleq \mathcal{L}\left\{y(t)\right\}$$

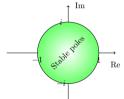
$$Y(z) \triangleq \mathcal{Z}\left\{y(kh)\right\}$$

$$Y(s) = G(s)U(s) = \frac{b}{s+a}U(s)$$

$$Y(z) = H(z)U(z) = \frac{\beta}{z+\alpha}U(z)$$



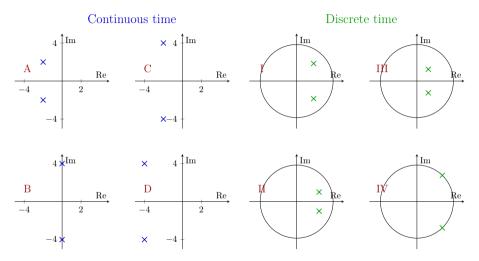




The s-domain of continuous-time systems is related to the z-domain of discrete-time systems through

$$z = e^{sh}$$

Poles in continuous- and discrete time

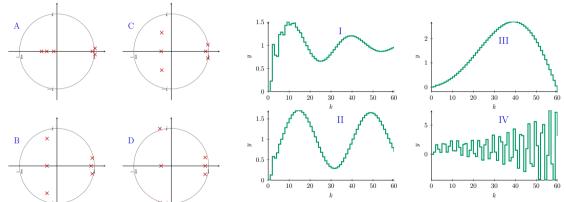


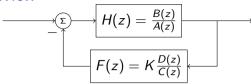
Activity Pair the continuous-time poles with the corresponding discrete-time poles!

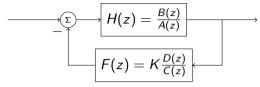
$$h = 0.2$$

Pole-placement and time-response

Pair the pole-placement with the correct time-response (discrete time)!

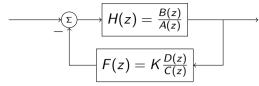






► The loop pulse-transfer function (loop gain) becomes

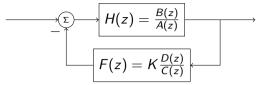
$$L(z) = H(z)F(z) = K\underbrace{\frac{B(z)D(z)}{A(z)C(z)}}_{P(z)} = K\frac{Q(z)}{P(z)}.$$



► The loop pulse-transfer function (loop gain) becomes

$$L(z) = H(z)F(z) = K\underbrace{\frac{B(z)D(z)}{A(z)C(z)}}_{P(z)} = K\frac{Q(z)}{P(z)}.$$

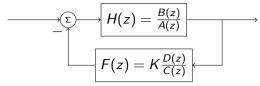
▶ The roots of Q(z) are called the open loop zeros.



► The loop pulse-transfer function (loop gain) becomes

$$L(z) = H(z)F(z) = K\underbrace{\frac{B(z)D(z)}{A(z)C(z)}}_{P(z)} = K\frac{Q(z)}{P(z)}.$$

- ▶ The roots of Q(z) are called the open loop zeros.
- ▶ The roots of P(z) are called the open loop poles.



► The loop pulse-transfer function (loop gain) becomes

$$L(z) = H(z)F(z) = K\underbrace{\frac{B(z)D(z)}{A(z)C(z)}}_{P(z)} = K\frac{Q(z)}{P(z)}.$$

- \triangleright The roots of Q(z) are called the open loop zeros.
- \triangleright The roots of P(z) are called the open loop poles.
- The characteristic equation for the closed-loop system is

$$1 + K \frac{Q(z)}{P(z)} = 0 \quad \Leftrightarrow \quad P(z) + KQ(z) = 0$$



Root locus: Definition

Let

$$\begin{cases} P(z) & = z^n + a_1 z^{n-1} + \dots + a_n = (z - p_1)(z - p_2) \dots (z - p_n) \\ Q(z) & = z^m + b_1 z^{m-1} + \dots + b_m = (s - q_1)(z - q_2) \dots (z - q_m) \end{cases}, \quad n \ge m$$

Root locus: Definition

Let

$$\begin{cases} P(z) &= z^n + a_1 z^{n-1} + \dots + a_n = (z - p_1)(z - p_2) \dots (z - p_n) \\ Q(z) &= z^m + b_1 z^{m-1} + \dots + b_m = (s - q_1)(z - q_2) \dots (z - q_m) \end{cases}, \quad n \ge m$$

The root locus shows how the solution to the characteristic equation

$$P(z) + K \cdot Q(z) = 0, \quad 0 \le K < \infty$$
 (1)

depend on the parameter K. The root locus consists of the set of all points in the complex plane that are solutions to (1) for some non-negative value of K.

Root locus: Characteristics

Start points The n roots of P(z), marked by crosses

End points The m roots of Q(z), marked by circles

Asymptotes Number equal to the pole excess n-m

Real axis Some segments of the real axis belong to the root locus

Root locus: Direction of the asymptotes

The characteristic equation P(z) + KQ(z) = 0 can be written $\frac{P(z)}{Q(z)} = -K$ and for large z it can be approximated as

$$\frac{z^n}{z^m} = -K \quad \Leftrightarrow \quad z^{n-m} = -K.$$

Taking the argument of both sides of the equation gives $(n-m) \arg z = \pi + k2\pi, \ k \in \mathbb{Z}$ So, the directions of the asymptotes are given by the expression

$$\theta_k = \arg z = \frac{(2k+1)\pi}{n-m}, \ k \in \mathbb{Z}$$

Root locus: The asymptotes' intersection with the real axis

$$z_{ip} = \frac{\sum_{i=0}^{n} p_i - \sum_{i=0}^{m} q_i}{n - m},$$

where $\{p_i\}$ are the starting points (open-loop poles) and $\{q_i\}$ are the end points (open-loop zeros).

Root locus exerise: Pair the pulse-trf fcn and root locus

$$G_1(z) = K \frac{(z+2.9)(z+0.2)}{(z-1)^2(z-0.3)}$$

$$G_2(z) = K \frac{(z - 0.5)(z + 0.4)}{(z - 1)(z - 0.3)(z - 0.1)}$$

$$G_3(z) = K \frac{(z - 0.5)(z + 0.8)}{(z - 1)^2(z - 0.3)}$$

$$G_4(z) = K \frac{z - 0.6}{(z - 1)(z - 0.3)}$$

