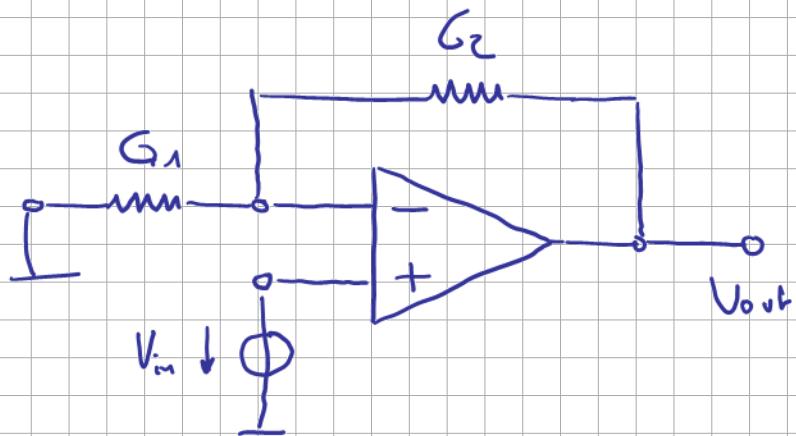
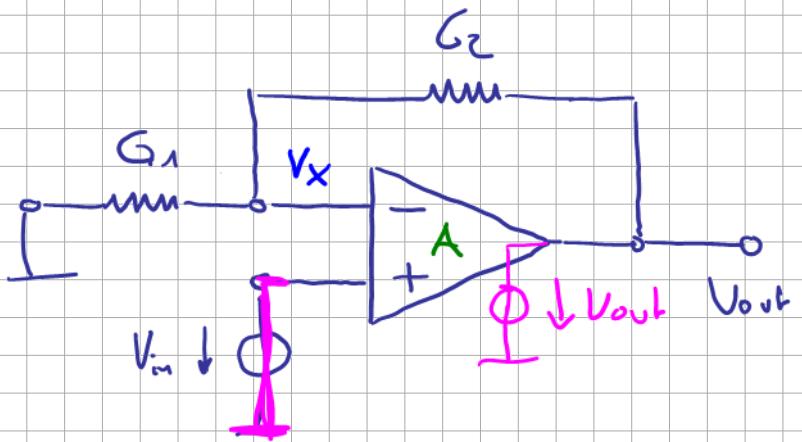
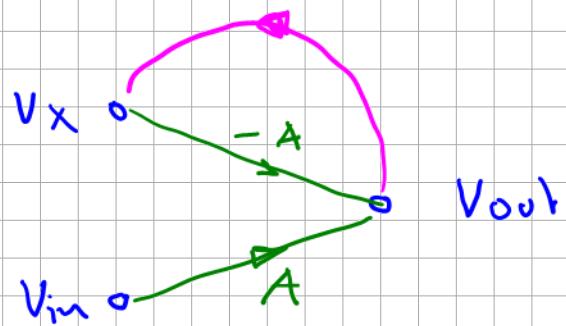
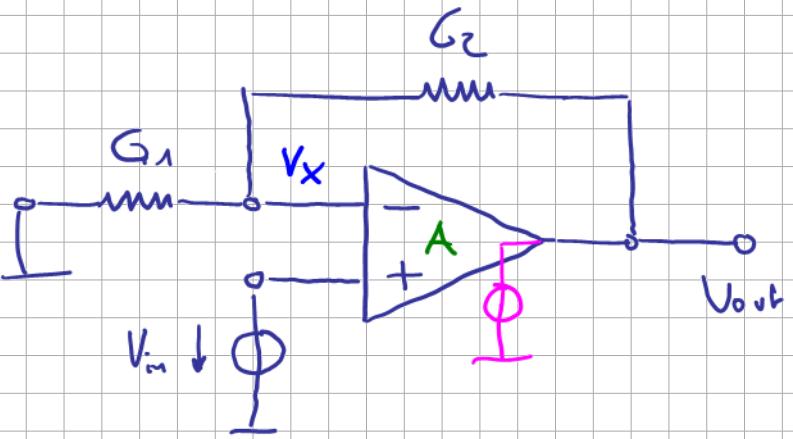


# Simple amplifier with an OpAmp: complete analysis

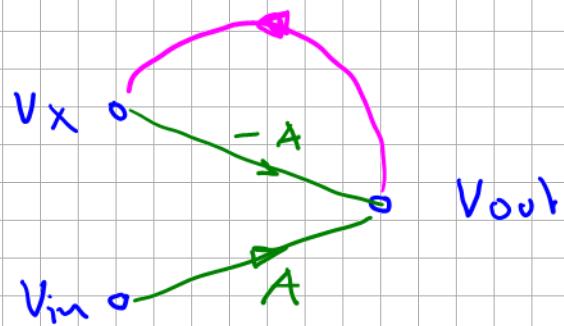
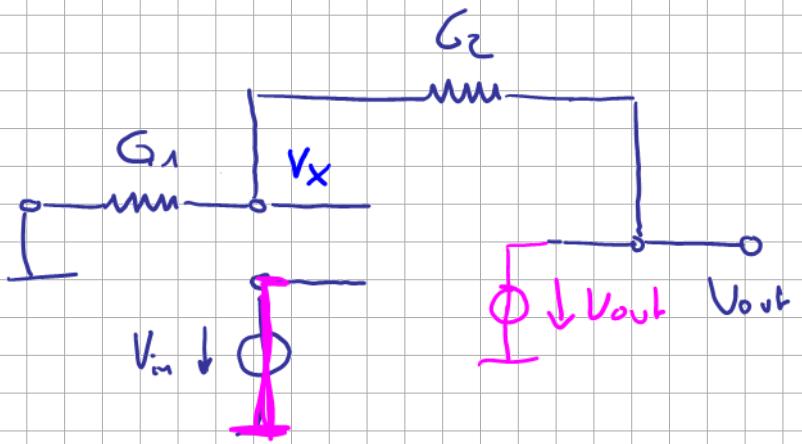
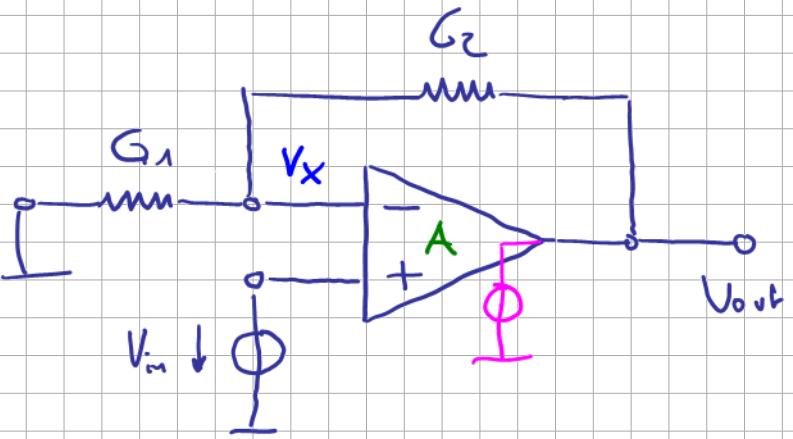


I will assume that you  
(know Mason's gain rule  
( $\rightarrow$  teaching videos))

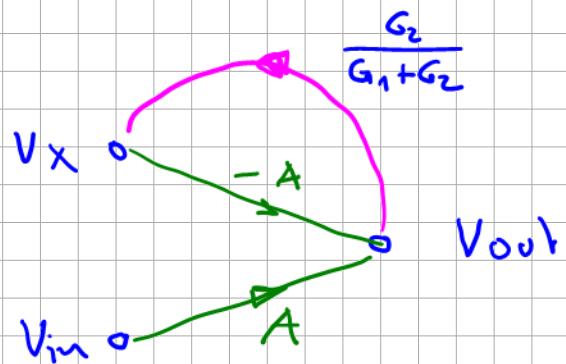
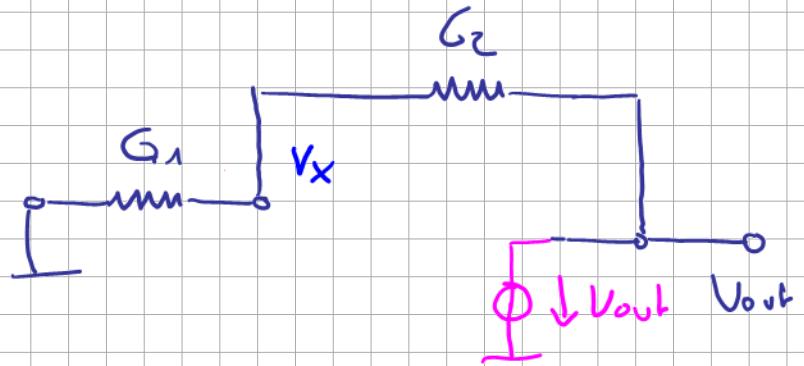
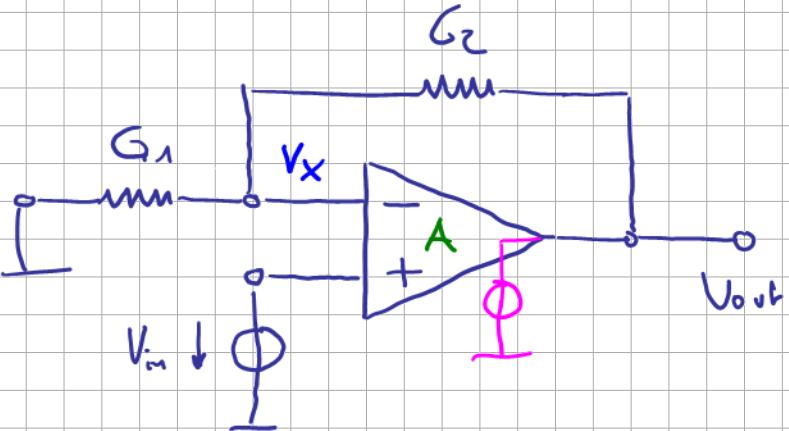
- Analysis with conventional SFG
- Loop gain, open-loop gain, closed-loop gain
- Analysis with driving-point SFG
- Positive vs. negative input
- Bandwidth, gain-bandwidth product
- Stability and phase margin
- Output resistance
- Noise analysis (resistor and amplifier noise)
- slew rate constraints
- SFG for an amplifier with the AD812 CFB OpAmp



$$V_{in} = 0$$



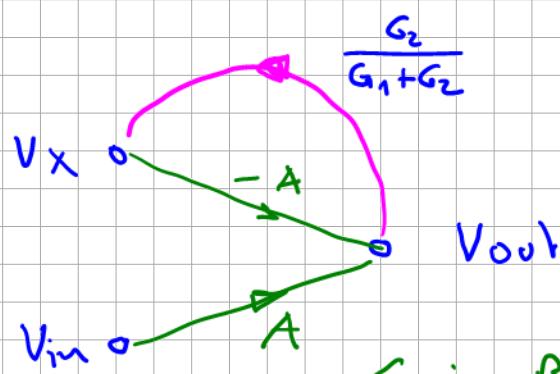
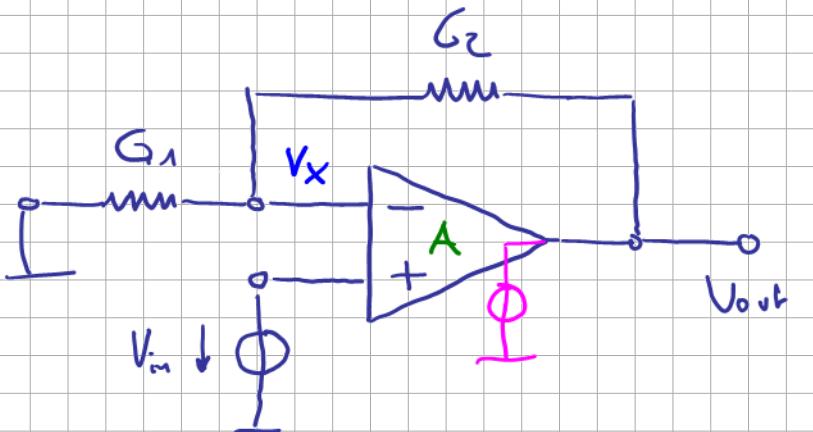
$$V_{in} = 0$$



$$\frac{V_x}{V_{out}} = \frac{G_2}{G_1 + G_2}$$



$$\frac{U_c}{V_{in}} = \frac{Y_{C1}}{Y_{C1} + Y_{C2}} = \frac{j\omega C_1}{j\omega C_1 + j\omega C_2}$$



$$\Rightarrow A \approx \frac{\omega_1}{s}$$

$$PM = \text{Phase} + 180^\circ$$

Gain-BW-Product  $\hat{=} 100 \text{ MHz} \cdot 2\pi$   
 Phase =  $-90^\circ$

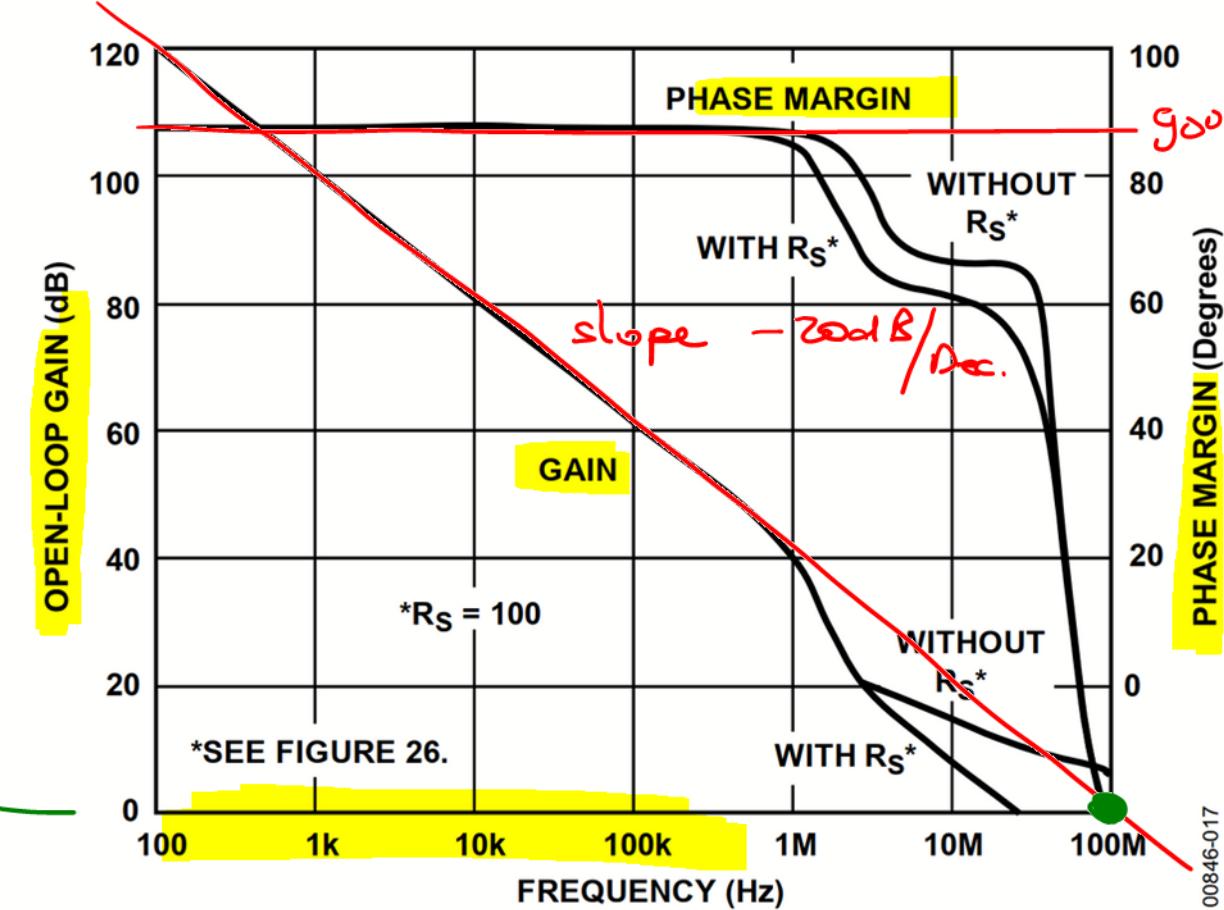
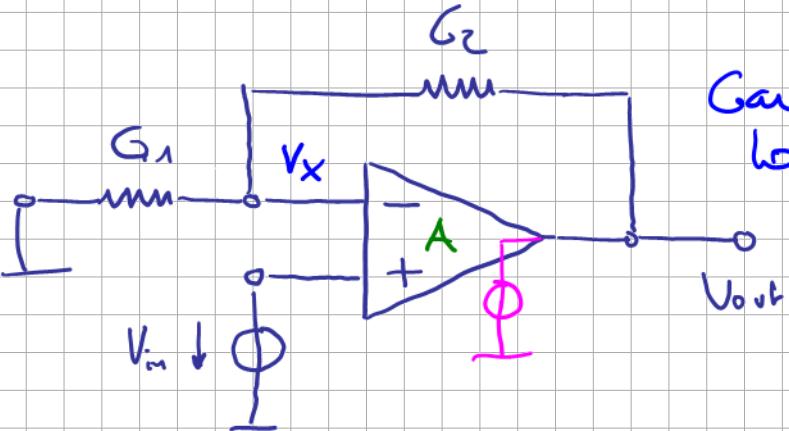


Figure 16. Open-Loop Gain and Phase Margin vs. Frequency



*Gain of the loop*

$$A \approx \frac{\omega_1}{s}$$

*Open-loop gain (Opt<sub>loop</sub>)*

$$L_1 = -A \cdot \frac{G_2}{G_1 + G_2}$$

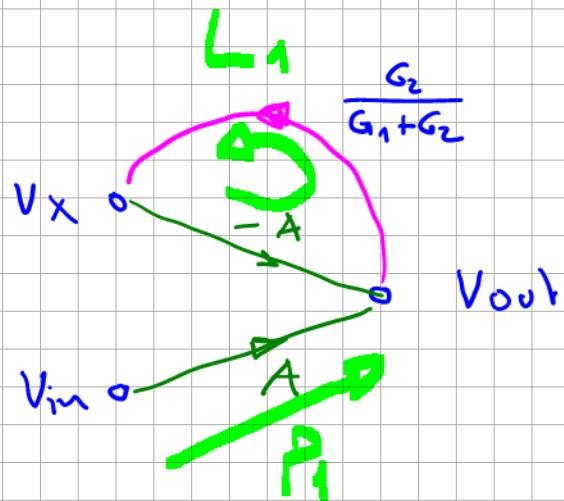
*Loop gain*

$$\Delta = 1 - L_1 = 1 + A \frac{G_2}{G_1 + G_2}$$

$$(G_1 + G_2) \Delta = G_1 + G_2 + A G_2$$

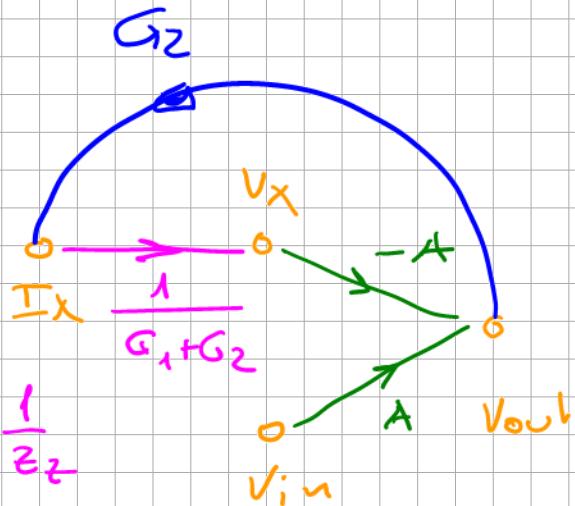
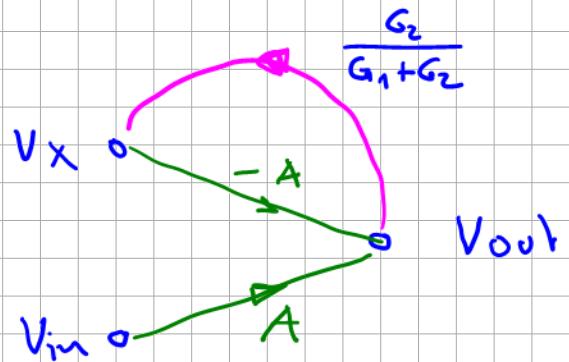
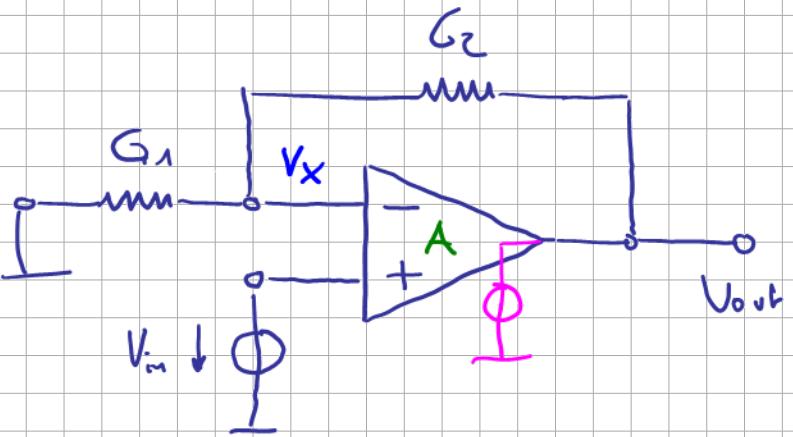
$$P_1 = A \quad \Delta_1 = 1$$

$$(G_1 + G_2) P_1 \Delta_1 = A (G_1 + G_2)$$

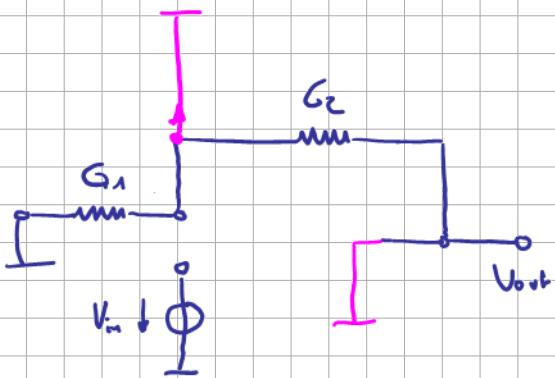


$$\frac{V_{out}}{V_{in}} = \frac{A (G_1 + G_2)}{G_1 + G_2 + A G_2}$$

*Closed-loop gain  
(result Mason)*

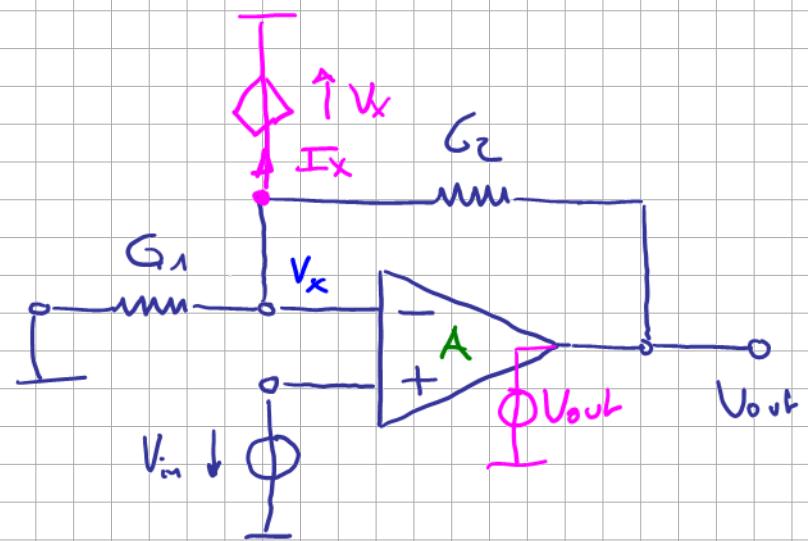
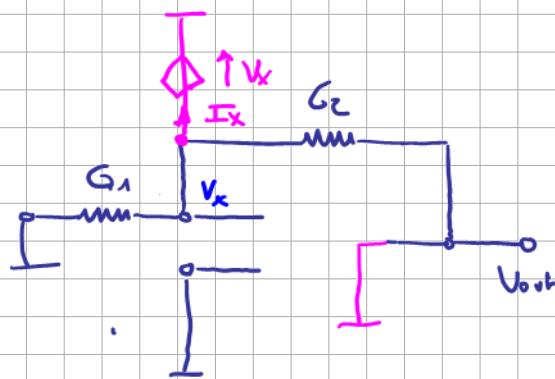
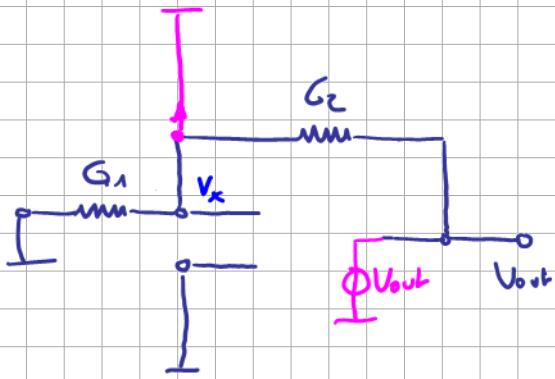


$$Y_x = \frac{1}{Z_2} = G_1 + G_2$$



$$V_{in} ?$$

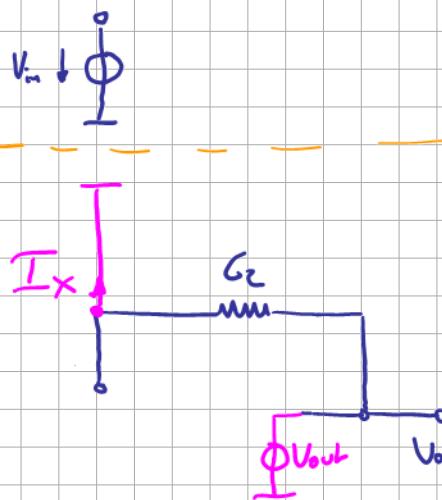
$$V_{out} = V_x = 0$$



$\uparrow I_x$

$V_{in} ?$

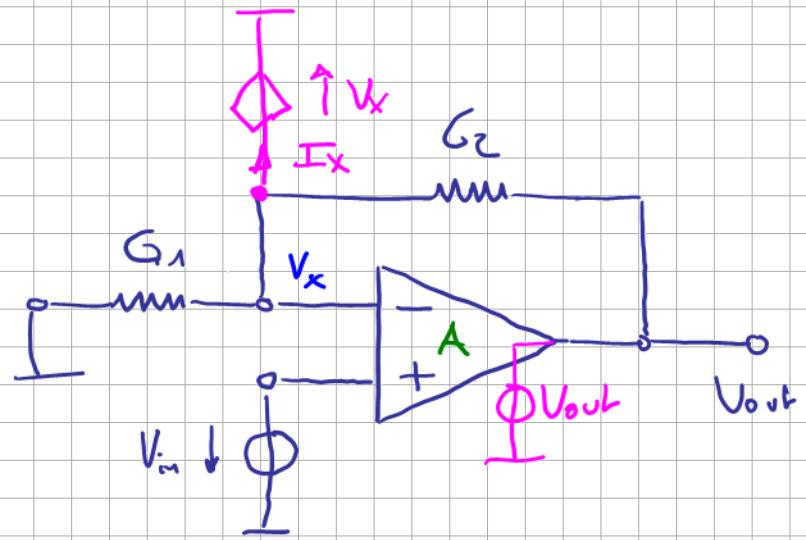
$$V_{out} = V_x = 0$$



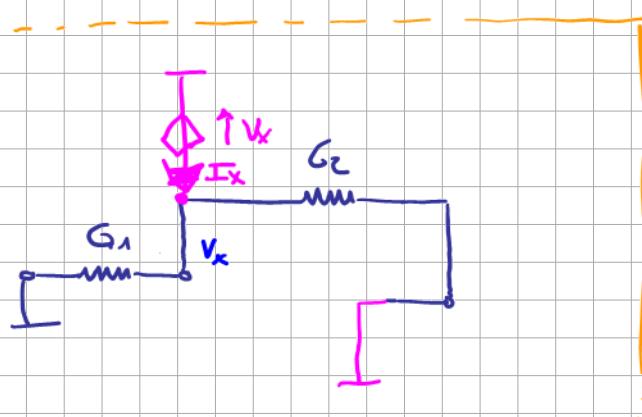
$I_x$  in Graph is for

$$V_x = 0$$

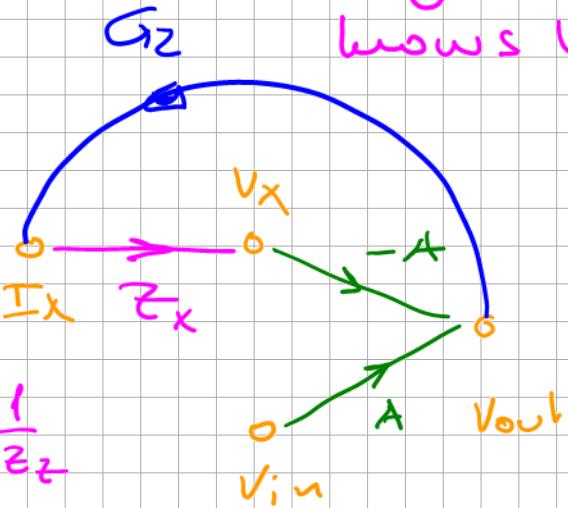
short-circuit current

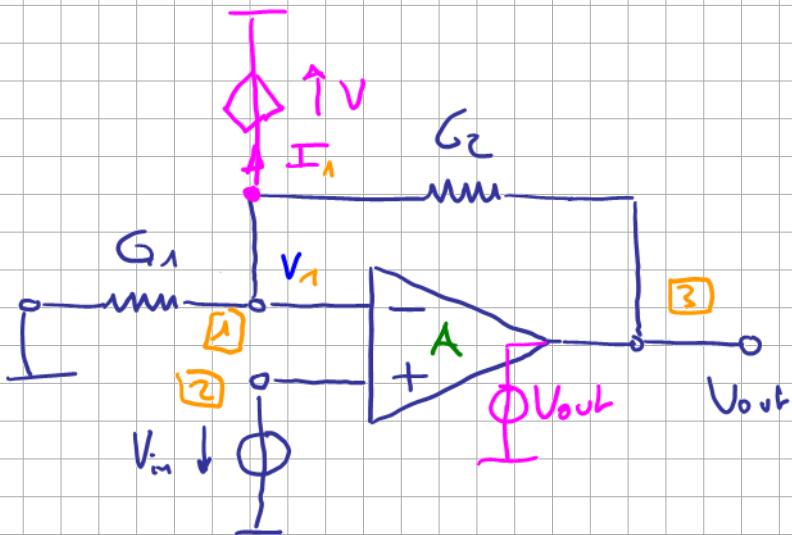


Magic Source: it already  
knows  $V_x$ ,  $\rightarrow \underline{\underline{I_x = 0}}$



$$Y_x = \frac{1}{Z_x} = G_1 + G_2$$





Gain :   
 1000      60 dB  
 100      40 dB  
 10      20 dB

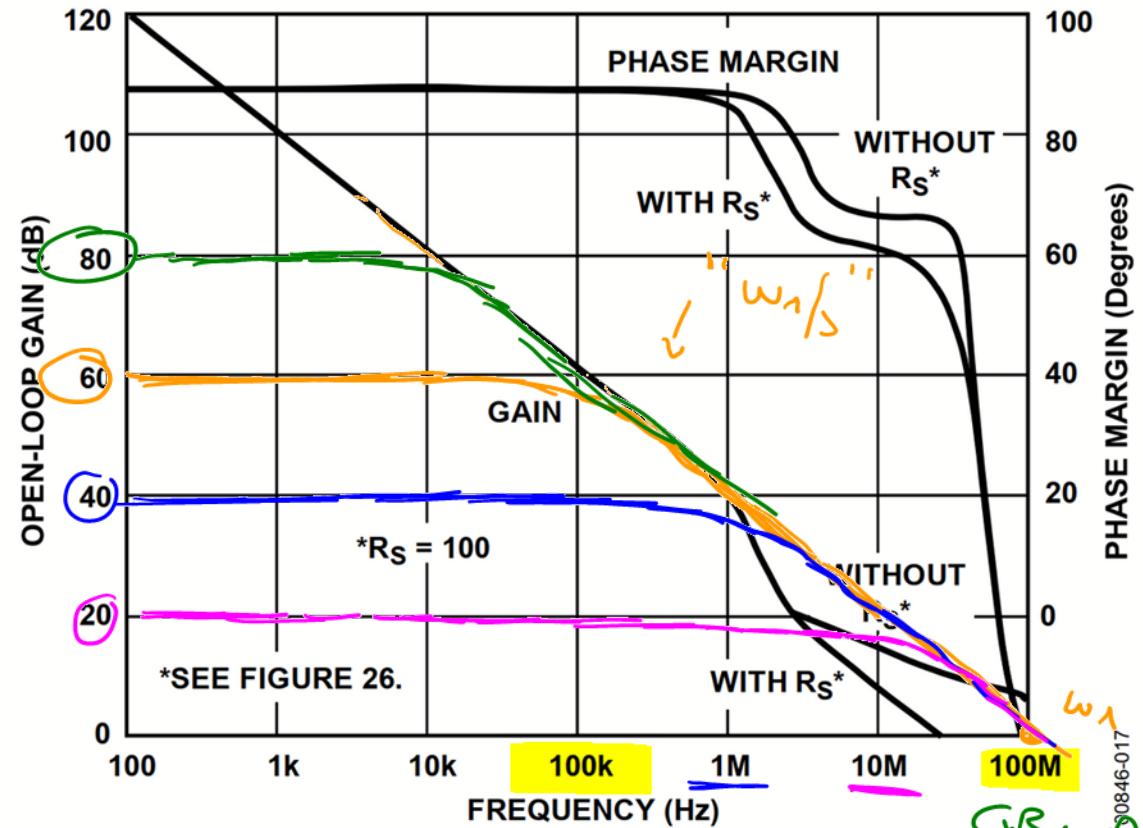


Figure 16. Open-Loop Gain and Phase Margin vs. Frequency

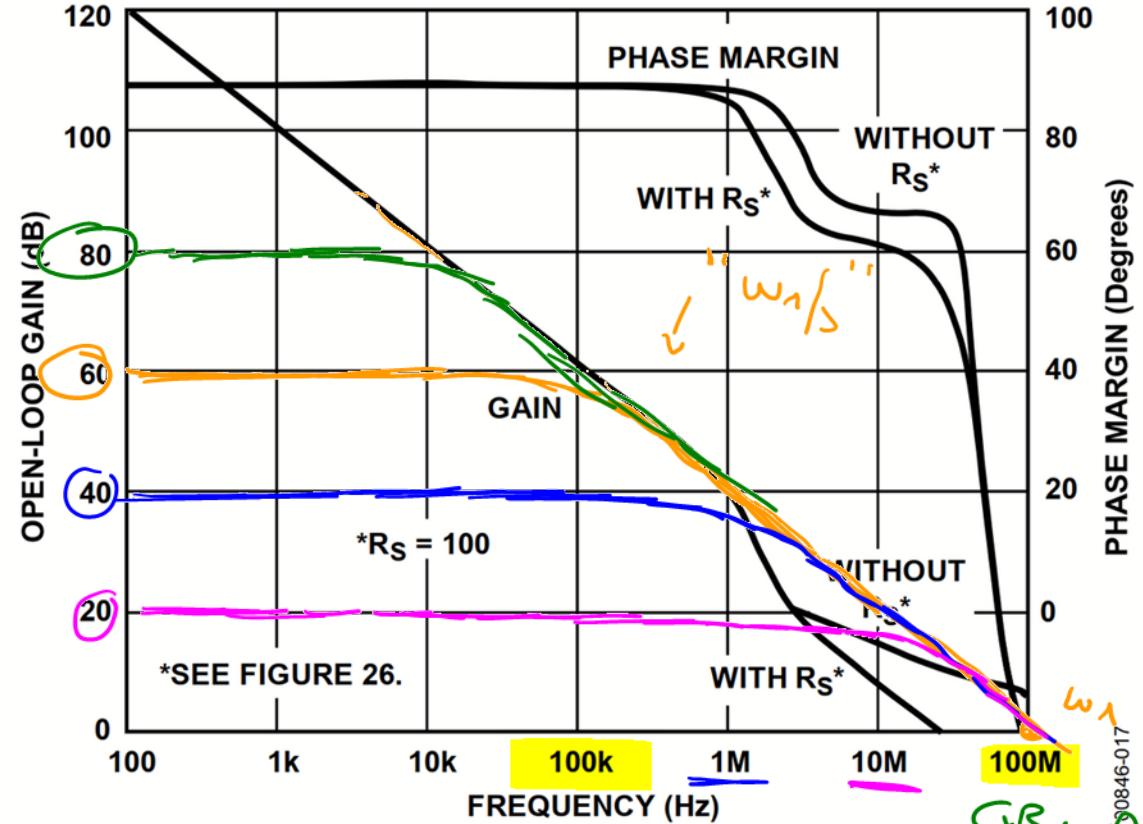
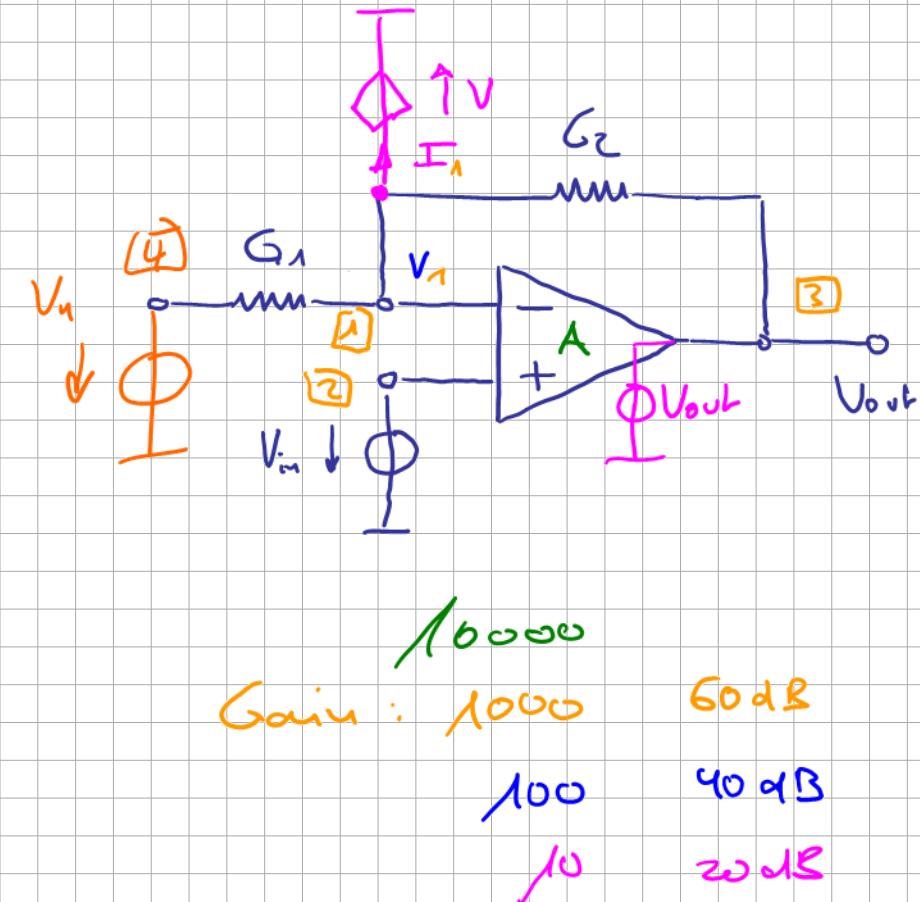


Figure 16. Open-Loop Gain and Phase Margin vs. Frequency

# Phase Margin

Loop gain: find the phase at gain 1 and check the distance to the point  $-180^\circ$

$$\text{Loop Gain} = \frac{\text{Open-loop gain}}{\text{Closed-loop gain}}$$

$$= \frac{A}{\frac{G_1 + G_2}{G_2}} = 1 \quad (0 \text{ dB})$$

$$= \frac{G_2 A}{G_1 + G_2}$$

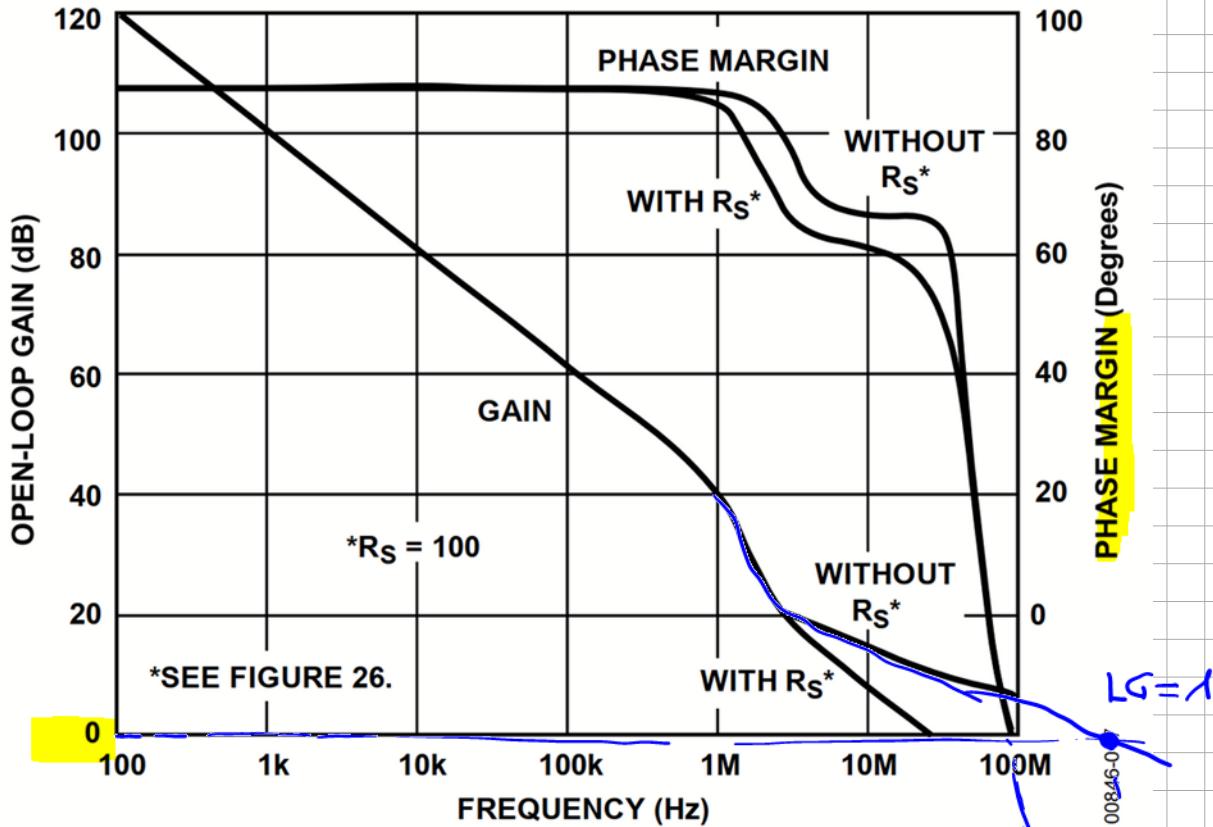


Figure 16. Open-Loop Gain and Phase Margin vs. Frequency

(for this circuit)

unstable

# Phase Margin

Loop gain: find the phase at gain 1 and check the distance to the point  $-180^\circ$

$$\text{Loop Gain} = \frac{\text{Open-loop gain}}{\text{Closed-loop gain}}$$

$$= \frac{A}{G_1 + G_2}$$

$$\frac{G_1 + G_2}{G_2} = 10$$

(20dB)

$$= \frac{G_2 A}{G_1 + G_2}$$

(for this circuit)

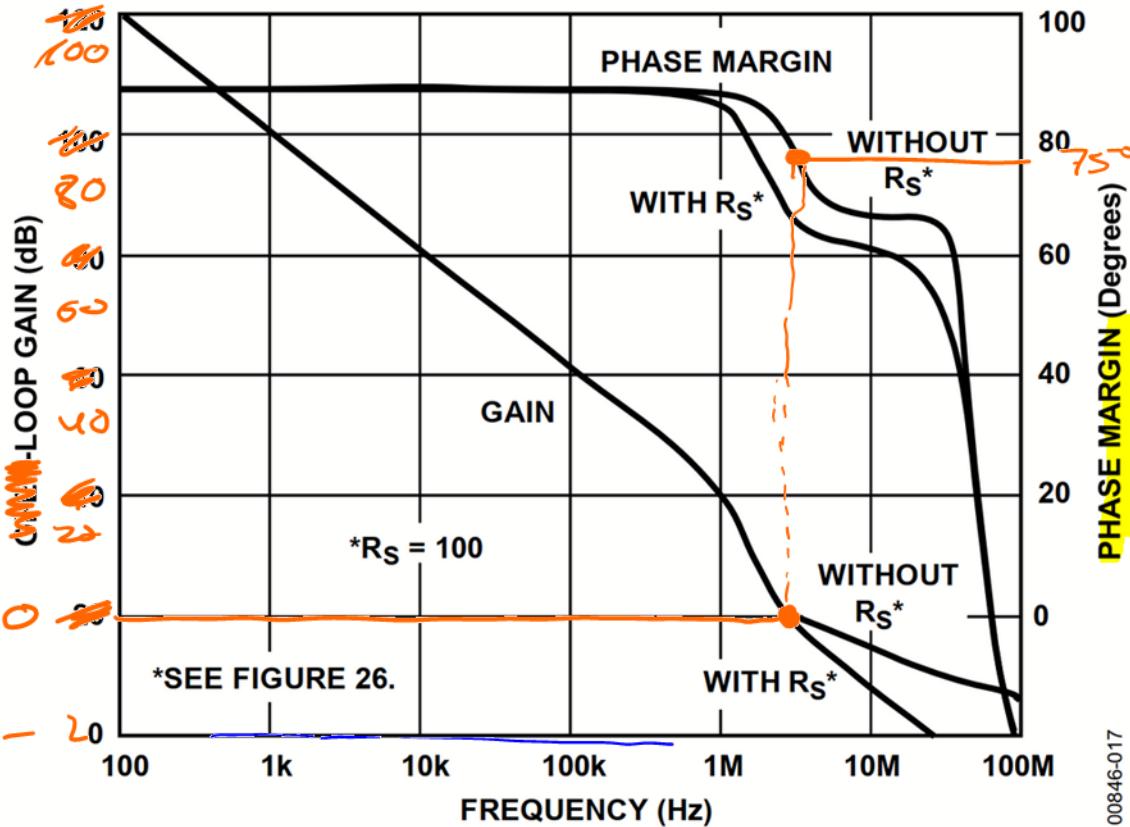


Figure 16. Open-Loop Gain and Phase Margin vs. Frequency

# Phase Margin

Loop gain: find the phase at gain 1 and check the distance to the point  $-180^\circ$

$$\text{Loop Gain} = \frac{\text{Open-loop gain}}{\text{Closed-loop gain}}$$

$$= \frac{A}{G_1 + G_2}$$

$$\frac{G_1 + G_2}{C_2}$$

50 dB?

$$= \frac{G_2 A}{G_1 + G_2}$$

(for this circuit)

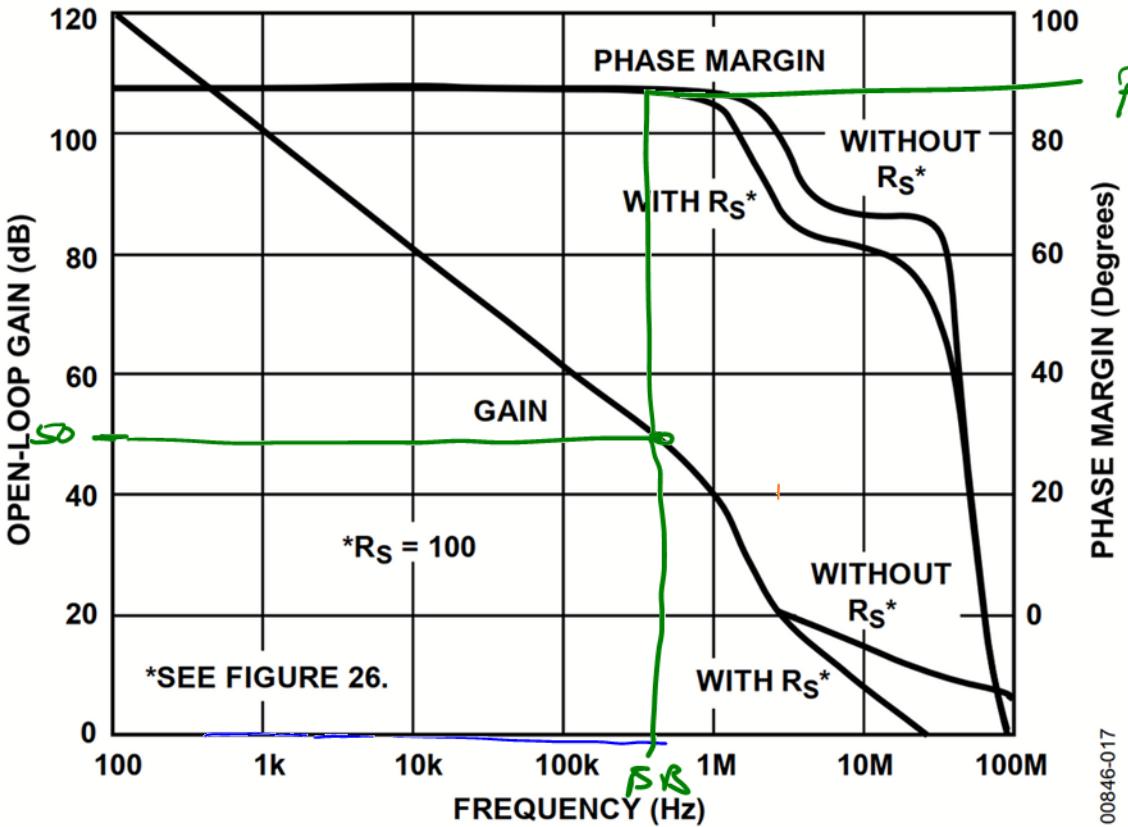


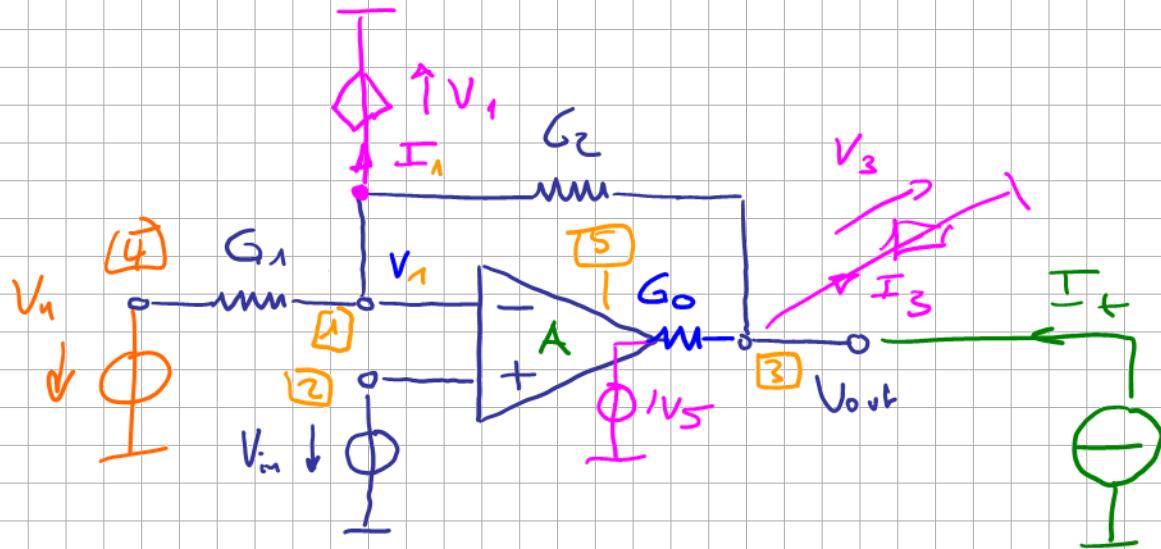
Figure 16. Open-Loop Gain and Phase Margin vs. Frequency

A

00846-017

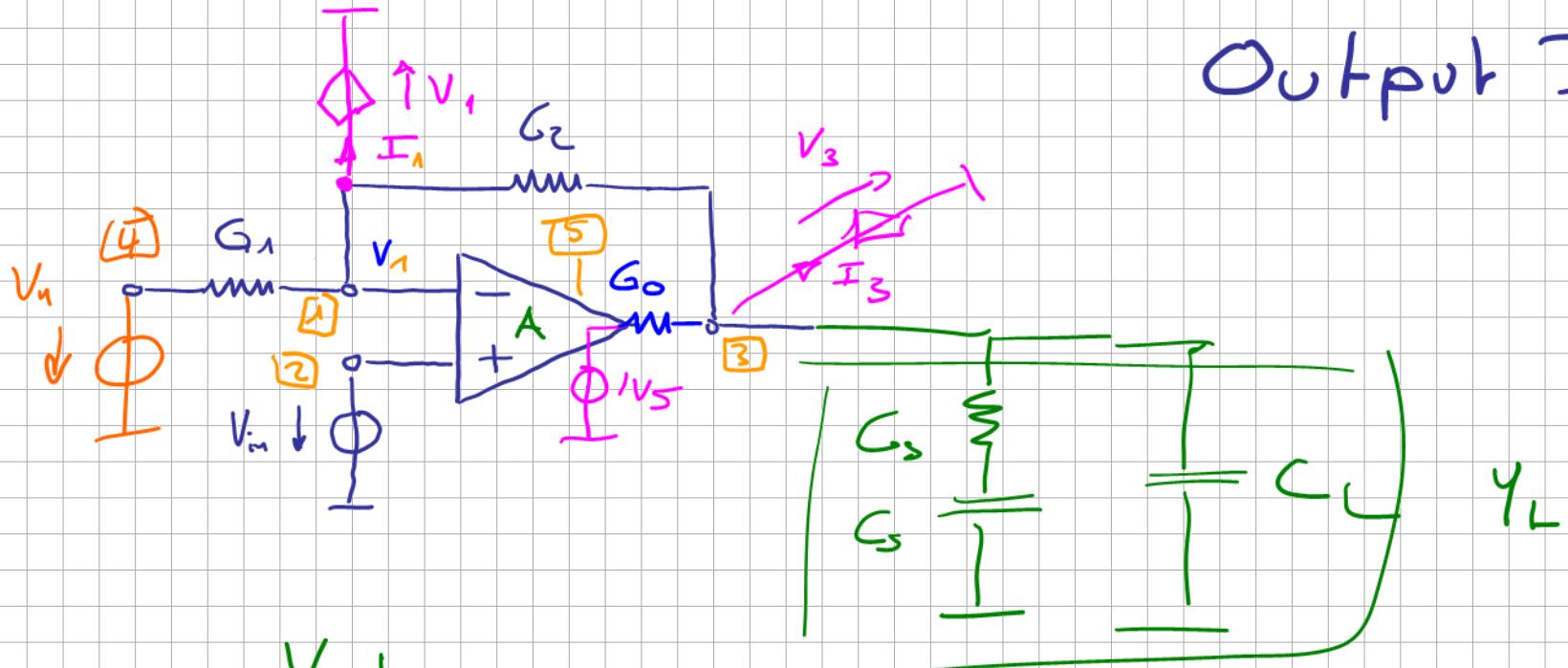
PM

Output Impedance?



$$\frac{V_{out}}{I_L} = Z_{out}$$

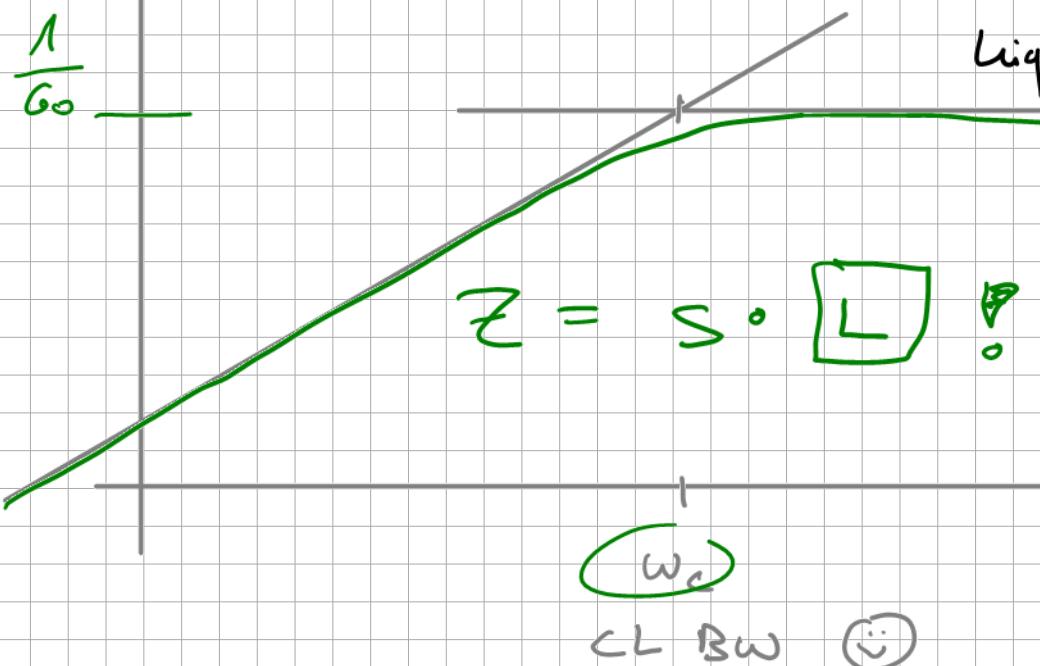
Output Impedance?



$$\frac{V_{out}}{I_L} = Z_{out}$$

Sinker Network

$|Z_{out}|$  (log)



low  $s$

$$Z_{out} = \frac{s(G_1 + G_2)}{G_0 G_2 \omega_1 + s(G_0 + G_2)(G_1 + G_2)}$$

$$\approx \frac{s(G_1 + G_2)}{\omega_1 G_2} \cdot \frac{1}{G_0}$$

closed-loop gain

$$\approx \frac{s(G_1 + G_2)}{s(G_0 + G_2)(G_1 + G_2)} = \frac{1}{G_0 + G_2}$$

$$\approx \frac{1}{G_0}$$

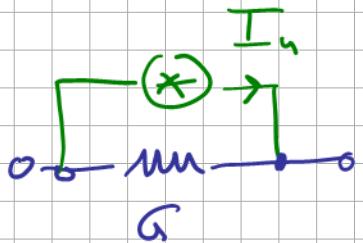
(Snubber Network)



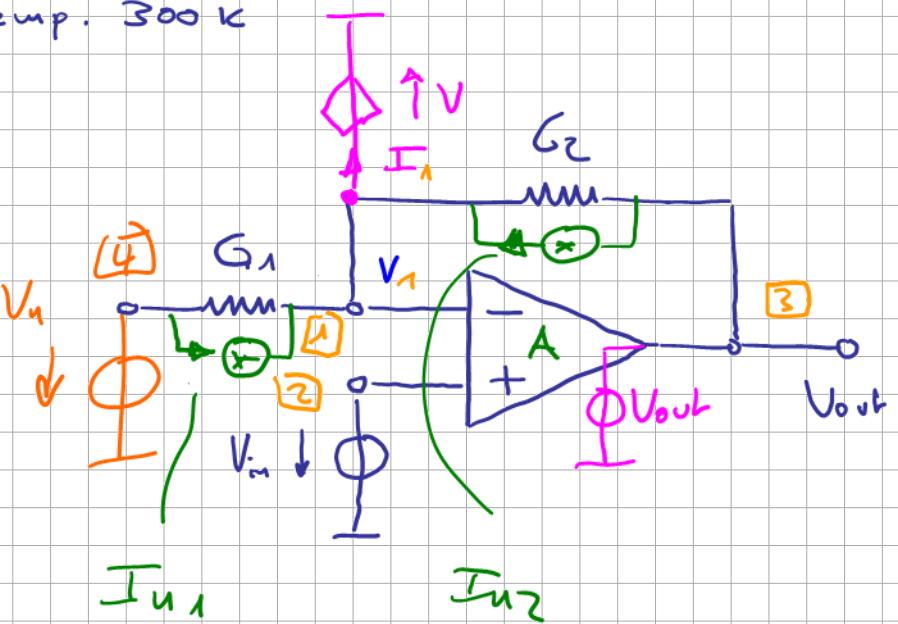
$$V_u^2 = 4kTR \quad \begin{matrix} \nearrow \text{Boltzmann Constant} \\ \searrow \text{Temp. } 300 \text{ K} \end{matrix}$$

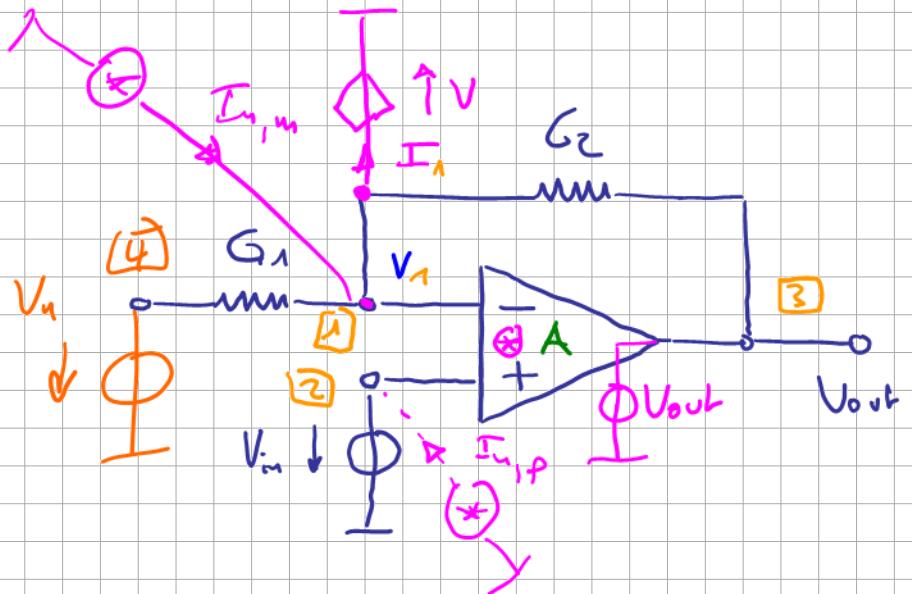
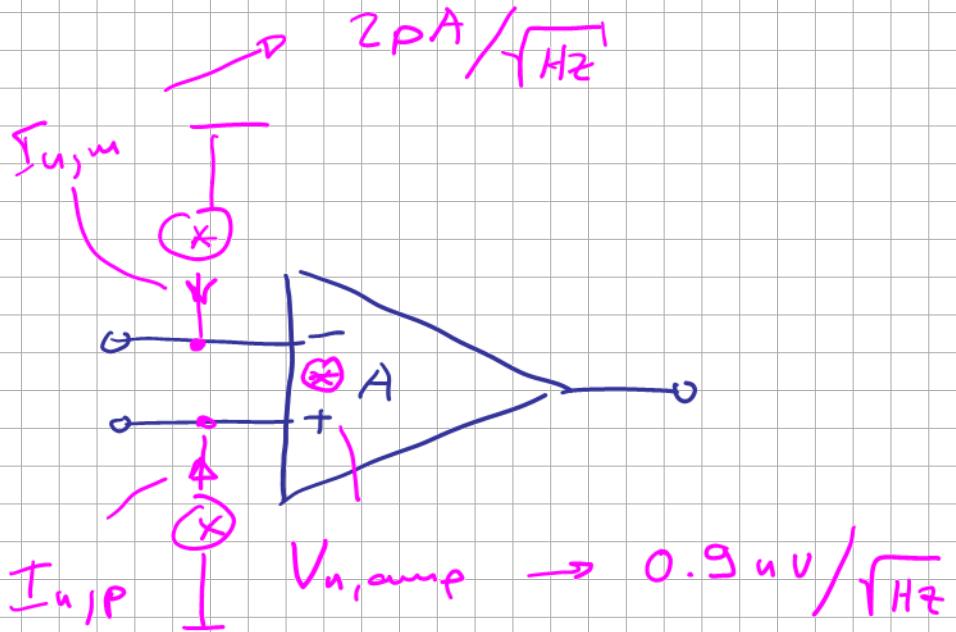
$300 \text{ K}$   
 $1 \text{ k}\Omega$

$$\sqrt{V_u^2} = \sqrt{4kTR} \approx 4 \text{ uV}/\sqrt{\text{Hz}}$$



$$I_u^2 = 4kT G$$



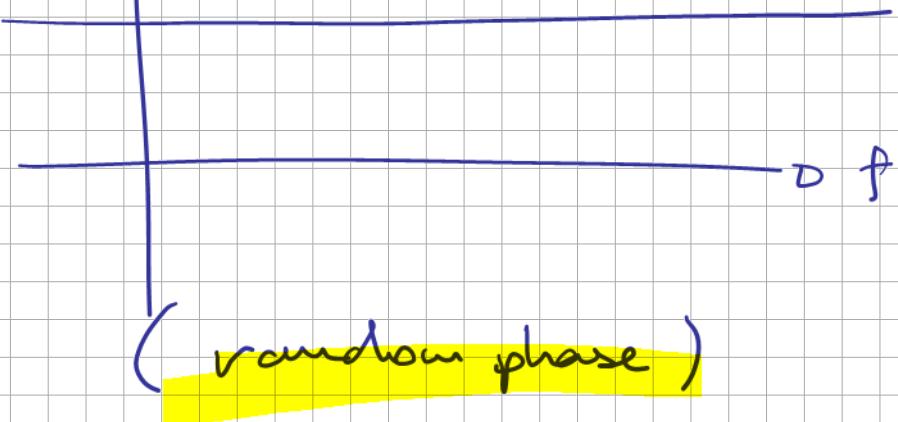


$$\begin{aligned}
 V_{n,out}^2 &= V_{n,amp}^2 \cdot |T_{n,amp}|^2 + I_{n,m}^2 |T_{n,m}|^2 \\
 (\frac{V_n}{\sqrt{Hz}})^2 &+ I_{n1}^2 |T_{n1}|^2 + I_{n2}^2 |T_{n2}|^2
 \end{aligned}$$

$$\begin{aligned}
 T_{n,amp} &\approx \frac{G_1 + G_2}{G_2} \\
 T_{n,m} &= T_{n1} = T_{n2} = -\frac{1}{G_2}
 \end{aligned}$$

White noise

$$S \propto (k^2)$$



add power !

$X, Y$  not correlated

$$\begin{aligned} \text{Var}[X+Y] &= \text{Var}[X] + \text{Var}[Y] \\ &\quad + 2 \text{Cov}(X, Y) \end{aligned}$$

Dirac Impulse

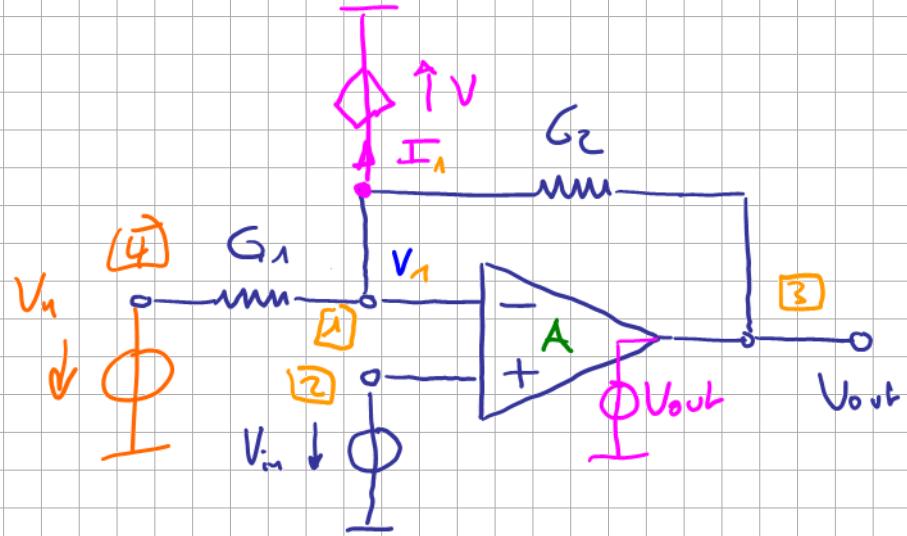
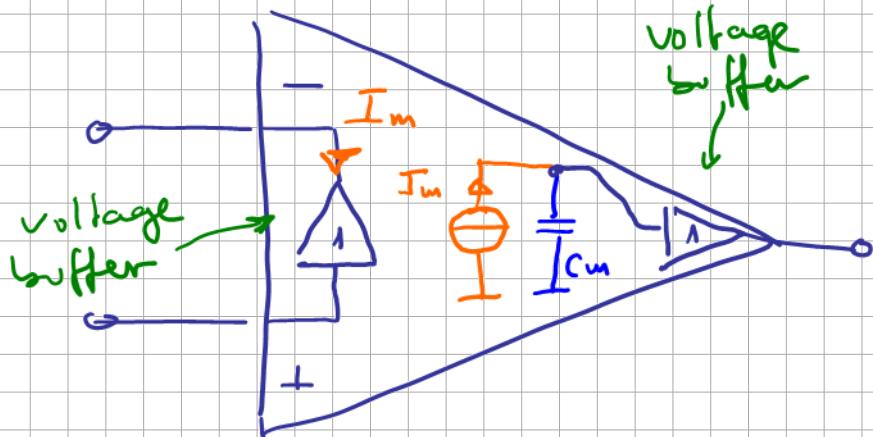
$$S \propto (k^2)$$



(deterministic phase)

add amplitude !

# AD 812 CFB OpAmp

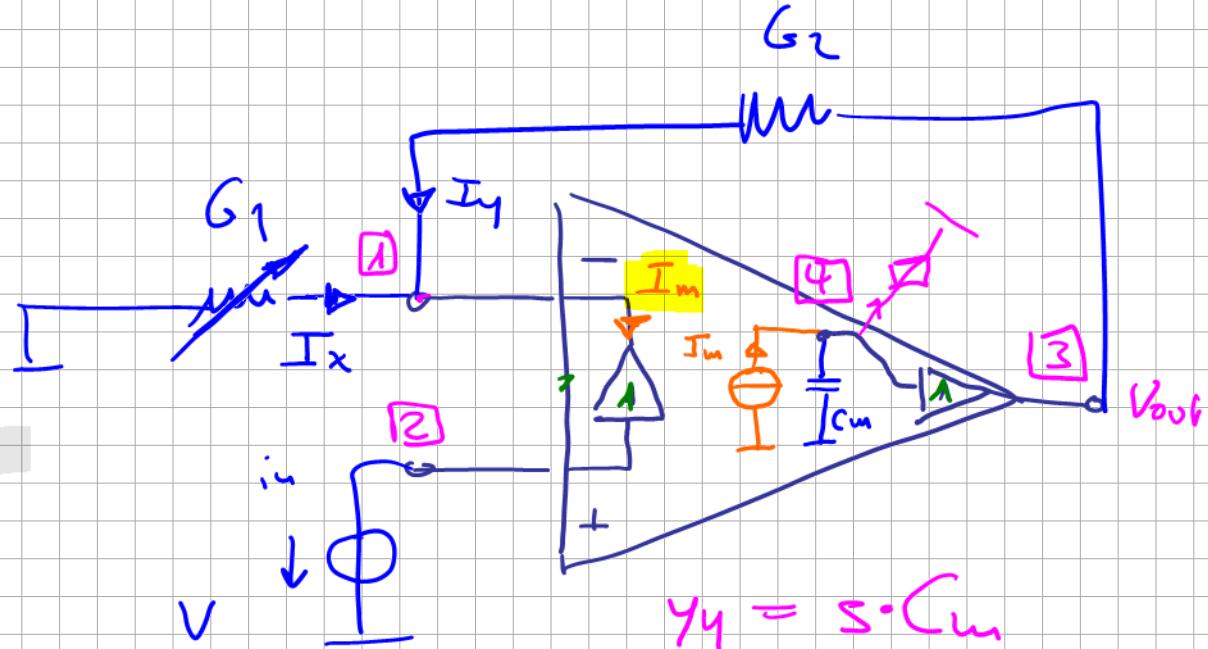
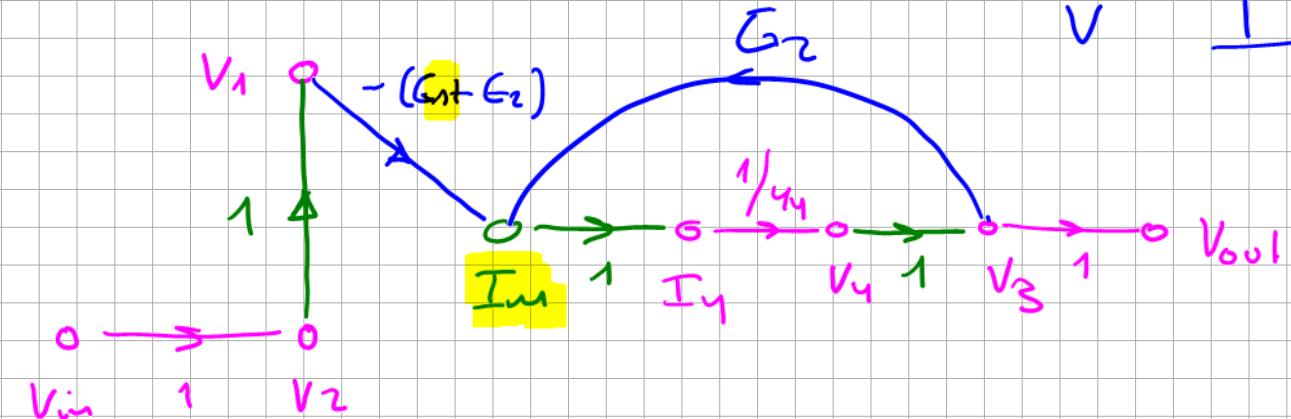


# AD 812 CFB OpAmp

$$I_{in} = I_x + I_y \quad (\text{Kirchhoff})$$

$$= -V_1 G_1 + (V_3 - V_1) \cdot G_2$$

$$= V_1 \cdot (-G_1 - G_2) + V_3 \cdot G_2$$



# Noise Density

$$\Delta V_{n,out}^2$$

(linear)

$$V_{n,out}^2$$

Unit  $\frac{V^2}{Hz}$

$$V_{noise,rms}^2 = \int V_{n,out}^2 df$$



$$SNR = \frac{V_{sig,rms}^2}{V_{noise,rms}^2}$$