#### 1

# Digital Signal Processing

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Abstract—This manual provides a simple introduction to digital signal processing.

#### 1 Software Installation

Run the following commands

sudo apt-get update sudo apt-get install libffi-dev libsndfile1 python3 -scipy python3-numpy python3-matplotlib sudo pip install cffi pysoundfile

#### 2 DIGITAL FILTER

2.1 Download the sound file from

wget https://github.com/samar2605/EE3900/ blob/master/Assignment%201/codes/ Sound Noise.way

2.2 You will find a spectrogram at https: //academo.org/demos/spectrum-analyzer. Upload the sound file that you downloaded in Problem 2.1 in the spectrogram and play. Observe the spectrogram. What do you find? Solution: There are a lot of yellow lines between 440 Hz to 5.1 KHz. These represent the

- synthesizer key tones. Also, the key strokes are audible along with background noise.
- 2.3 Write the python code for removal of out of band noise and execute the code.

## **Solution:**

```
import soundfile as sf
from scipy import signal
#read .wav file
input signal,fs = sf.read('Sound Noise.wav'
#sampling frequency of Input signal
sampl freq=fs
#order of the filter
order=4
#cutoff frquency 4kHz
cutoff freq=4000.0
#digital frequency
Wn=2*cutoff freq/sampl freq
# b and a are numerator and denominator
   polynomials respectively
b, a = signal.butter(order, Wn, 'low')
#filter the input signal with butterworth filter
output signal = signal.filtfilt(b, a,
   input signal)
\#output \ signal = signal.lfilter(b, a,
   input signal)
#write the output signal into .wav file
sf.write('Sound With ReducedNoise.wav',
   output signal, fs)
```

2.4 The output of the python script in Problem 2.3 is the audio file Sound\_With\_ReducedNoise.wav. Play the file in the spectrogram in Problem 2.2.

What do you observe?

**Solution:** The key strokes as well as background noise is subdued in the audio. Also, the signal is blank for frequencies above 5.1 kHz.

3 Difference Equation

3.1 Let

$$x(n) = \left\{ \begin{array}{l} 1, 2, 3, 4, 2, 1 \\ 1 \end{array} \right\} \tag{3.1}$$

Sketch x(n).

3.2 Let

$$y(n) + \frac{1}{2}y(n-1) = x(n) + x(n-2),$$
  
$$y(n) = 0, n < 0 \quad (3.2)$$

Sketch y(n).

**Solution:** The following code yields Fig. 3.2.

```
import numpy as np
import matplotlib.pyplot as plt
#If using termux
import subprocess
import shlex
#end if
x=np.array([1.0,2.0,3.0,4.0,2.0,1.0])
k = 20
y = np.zeros(20)
y[0] = x[0]
y[1] = -0.5*y[0]+x[1]
for n in range(2,k-1):
        if n < 6:
                 y[n] = -0.5*y[n-1]+x[n]+x
                     [n-2]
        elif n > 5 and n < 8:
                 y[n] = -0.5*y[n-1]+x[n-2]
        else:
                 y[n] = -0.5*y[n-1]
print(y)
#subplots
plt.subplot(2, 1, 1)
plt.stem(range(0,6),x)
plt.title('Digital_Filter_Input-Output')
```

```
plt.ylabel('$x(n)$')
plt.grid()# minor

plt.subplot(2, 1, 2)
plt.stem(range(0,k),y)
plt.xlabel('$n$')
plt.ylabel('$y(n)$')
plt.grid()# minor

#If using termux
plt.savefig('../figs/A1_3.pdf')
plt.savefig('../figs/A1_3.eps')
#subprocess.run(shlex.split("termux-open ../
figs/A1_3.pdf"))
#else
plt.show()
```

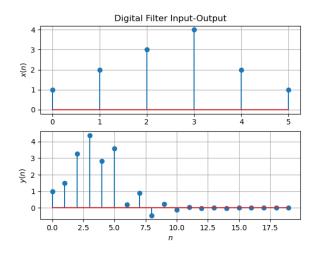


Fig. 3.2

3.3 Repeat the above exercise using a C code. **Solution:** The following code yields Fig. 3.3A

```
#include<stdio.h>

int main(){
    FILE *fptr;
    float x[]={1.0,2.0,3.0,4.0,2.0,1.0};
    int k = 20;
    float y[20] = {0};

y[0] = x[0];
    y[0] = x[0];
    y[1] = -0.5*y[0]+x[1];

for(int n=2;n<k-1;n++){
```

```
if (n < 6)
        y[n] = -0.5*y[n-1]+x[n]+x[n
        -2];
else if (n > 5 && n < 8)
        y[n] = -0.5*y[n-1]+x[n-2];
else
        y[n] = -0.5*y[n-1];
}
fptr=fopen("Y.dat","w");
for(int i=0;i<20;i++){
        fprintf(fptr,"%f\n",y[i]);
}
fclose(fptr);
return 0;
}</pre>
```

```
import matplotlib.pyplot as plt
import numpy as np
y = np.loadtxt("Y.dat",dtype = "double")
x = np.array([1,2,3,4,2,1])
# ploting graph
#subplot for x(n)
plt.subplot(211)
plt.stem(np.arange(len(x)),x)
plt.xlabel("n")
plt.ylabel("$x(n)$")
plt.grid()
#subplot for y(n)
plt.subplot(212)
plt.stem(np.arange(len(y)),y)
plt.xlabel("n")
plt.ylabel("$y(n)$")
plt.grid()
#plot fig output
plt.show()
```

## 4 Z-TRANSFORM

4.1 The Z-transform of x(n) is defined as

$$X(z) = \mathcal{Z}\{x(n)\} = \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$
 (4.1)

Show that

$$Z\{x(n-1)\} = z^{-1}X(z)$$
 (4.2)

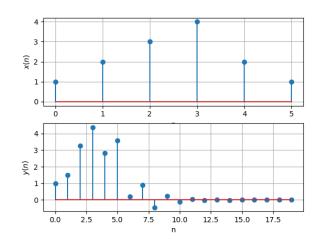


Fig. 3.3

and find

$$\mathcal{Z}\{x(n-k)\}\tag{4.3}$$

**Solution:** From (4.1),

$$\mathcal{Z}\{x(n-k)\} = \sum_{n=-\infty}^{\infty} x(n-1)z^{-n}$$

$$= \sum_{n=-\infty}^{\infty} x(n)z^{-n-1} = z^{-1} \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$
(4.4)
$$(4.5)$$

resulting in (4.2). Similarly, it can be shown that

$$Z\{x(n-k)\} = z^{-k}X(z)$$
 (4.6)

4.2 Obtain X(z) for x(n) defined in problem 3.1. **Solution:** 

$$Z(x(n)) = \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$

$$= x(0)z^{0} + x(1)z^{-1} + x(2)z^{-2} + x(3)z^{-3} +$$

$$(4.8)$$

$$x(4)z^{-4} + x(5)z^{-5}$$

$$= 1 + 2z^{-1} + 3z^{-2} + 4z^{-3} + 2z^{-4} + z^{-5}$$

$$(4.9)$$

4.3 Find

$$H(z) = \frac{Y(z)}{X(z)} \tag{4.10}$$

from (3.2) assuming that the Z-transform is a linear operation.

**Solution:** Applying (4.6) in (3.2),

$$Y(z) + \frac{1}{2}z^{-1}Y(z) = X(z) + z^{-2}X(z)$$
 (4.11)

$$\implies \frac{Y(z)}{X(z)} = \frac{1 + z^{-2}}{1 + \frac{1}{2}z^{-1}} \tag{4.12}$$

## 4.4 Find the Z transform of

$$\delta(n) = \begin{cases} 1 & n = 0 \\ 0 & \text{otherwise} \end{cases}$$
 (4.13)

and show that the Z-transform of

$$u(n) = \begin{cases} 1 & n \ge 0 \\ 0 & \text{otherwise} \end{cases}$$
 (4.14)

is

$$U(z) = \frac{1}{1 - z^{-1}}, \quad |z| > 1$$
 (4.15)

**Solution:** It is easy to show that

$$\delta(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} 1 \tag{4.16}$$

and from (4.14),

$$U(z) = \sum_{n=0}^{\infty} z^{-n}$$
 (4.17)

$$=\frac{1}{1-z^{-1}}, \quad |z| > 1 \tag{4.18}$$

using the fomula for the sum of an infinite geometric progression.

## 4.5 Show that

$$a^{n}u(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} \frac{1}{1 - az^{-1}} \quad |z| > |a| \tag{4.19}$$

## **Solution:**

$$Z(a^{n}u(n)) = \sum_{n=-\infty}^{\infty} a^{n}u(n)z^{-n}$$
 (4.20)

$$=\sum_{n=0}^{\infty} a^n z^{-n}$$
 (4.21)

$$= \frac{1}{1 - az^{-1}}, \quad \left| az^{-1} \right| < 1 \qquad (4.22)$$

$$= \frac{1}{1 - az^{-1}}, \quad |a| < |z| \tag{4.23}$$

using the fomula for the sum of an infinite geometric progression.

## 4.6 Let

$$H(e^{j\omega}) = H(z = e^{j\omega}).$$
 (4.24)

Plot  $|H(e^{j\omega})|$ . Is it periodic? If so, find the

period.  $H(e^{j\omega})$  is known as the *Discret Time* Fourier Transform (DTFT) of x(n).

**Solution:** The following code plots Fig. 4.6.

wget https://raw.githubusercontent.com/ gadepall/EE1310/master/filter/codes/dtft. py

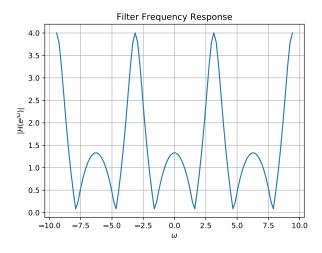


Fig. 4.6:  $|H(e^{j\omega})|$ 

**Solution:** The following code plots Fig. 4.6.

wget https://raw.githubusercontent.com/ samar2605/EE3900/master/Assignment %201/codes/dtft.py

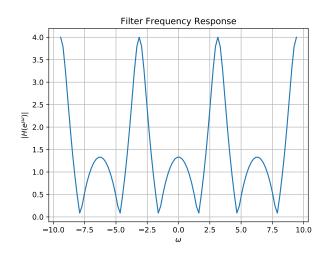


Fig. 4.6:  $|H(e^{j\omega})|$ 

It is bounded between (0,4) and periodic with period  $(2\pi)$ 

$$H\left(e^{j\omega}\right) = \frac{1 + e^{-2j\omega}}{1 + \frac{e^{-j\omega}}{2}}\tag{4.25}$$

$$\Rightarrow \left| H\left(e^{j\omega}\right) \right| = \frac{\left| 1 + e^{-2j\omega} \right|}{\left| 1 + \frac{e^{-j\omega}}{2} \right|}$$

$$= \frac{\left| 1 + e^{2j\omega} \right|}{\left| e^{2j\omega} + \frac{e^{j\omega}}{2} \right|}$$

$$= \frac{\left| 1 + \cos 2\omega + j \sin 2\omega \right|}{\left| e^{j\omega} + \frac{1}{2} \right|}$$

$$= \frac{\left| 4\cos^{2}\left(\omega\right) + 4j \sin\left(\omega\right)\cos\left(\omega\right) \right|}{\left| 2e^{j\omega} + 1 \right|}$$

$$= \frac{\left| 4\cos\left(\omega\right) \right| \left| \cos\left(\omega\right) + j \sin\left(\omega\right) \right|}{\left| 2\cos\left(\omega\right) + 1 + 2j \sin\left(\omega\right) \right|}$$

$$= \frac{\left| 4\cos\left(\omega\right) \right| \left| \cos\left(\omega\right) + 1 + 2j \sin\left(\omega\right) \right|}{\left| 2\cos\left(\omega\right) + 1 + 2j \sin\left(\omega\right) \right|}$$

$$= \frac{\left| 4\cos\left(\omega\right) \right| \left| \cos\left(\omega\right) + 1 + 2j \sin\left(\omega\right) \right|}{\left| 2\cos\left(\omega\right) + 1 + 2j \sin\left(\omega\right) \right|}$$

$$= \frac{\left| 4\cos\left(\omega\right) \right| \left| \cos\left(\omega\right) + 1 + 2j \sin\left(\omega\right) \right|}{\left| 2\cos\left(\omega\right) + 1 + 2j \sin\left(\omega\right) \right|}$$

$$= \frac{\left| 4\cos\left(\omega\right) \right| \left| \cos\left(\omega\right) + 1 + 2j \sin\left(\omega\right) \right|}{\left| 2\cos\left(\omega\right) + 1 + 2j \sin\left(\omega\right) \right|}$$

$$= \frac{\left| 4\cos\left(\omega\right) \right| \left| \cos\left(\omega\right) + 1 + 2j \sin\left(\omega\right) \right|}{\left| 2\cos\left(\omega\right) + 1 + 2j \sin\left(\omega\right) \right|}$$

$$\therefore \left| H\left(e^{j\omega}\right) \right| = \frac{|4\cos(\omega)|}{\sqrt{5 + 4\cos(\omega)}} \tag{4.31}$$

4.7 Express x(n) in terms of  $H(e^{j\omega})$ .

#### **Solution:**

$$H(e^{j\omega}) = \sum_{k=-\infty}^{\infty} h(k)e^{-j\omega k}$$
 (4.32)

and

$$h(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} H(e^{j\omega}) e^{j\omega n} d\omega \qquad (4.33)$$

Now,

$$\frac{1}{2\pi} \int_{-\pi}^{\pi} H(e^{j\omega}) e^{j\omega n} d\omega \qquad (4.34)$$

$$= \frac{1}{2\pi} \int_{-\pi}^{\pi} \sum_{k=-\infty}^{\infty} h(k) e^{-j\omega k} e^{j\omega n} d\omega \qquad (4.35)$$

$$= \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} h(k) \int_{-\pi}^{\pi} e^{j\omega(n-k)} d\omega \qquad (4.36)$$

$$= \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} h(k) \int_{-\pi}^{\pi} \cos w(n-k) \qquad (4.37)$$

$$d\omega + \int_{-\pi}^{\pi} \sin w(n-k)d\omega$$

$$= \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} h(k) \int_{-\pi}^{\pi} \cos w(n-k)$$
 (4.38)

$$= \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} h(k) \frac{\sin w(n-k)}{n-k} \bigg|_{-\pi}^{\pi}$$
 (4.39)

$$= \frac{1}{2\pi} \sum_{k \neq n} h(n) \frac{\sin \pi (n-k)}{n-k} + \sum_{k=n} h(n) \frac{\sin \pi (n-k)}{n-k}$$
(4.40)

$$=\frac{0+2\pi h(n)}{2\pi}$$
 (4.41)

$$= h(n) \tag{4.42}$$

## 5 Impulse Response

5.1 Using long division, find

$$h(n), \quad n < 5 \tag{5.1}$$

for H(z) in (4.12).

5.2 Find an expression for h(n) using H(z), given that

$$h(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} H(z) \tag{5.2}$$

and there is a one to one relationship between h(n) and H(z). h(n) is known as the *impulse response* of the system defined by (3.2).

**Solution:** From (4.12),

$$H(z) = \frac{1}{1 + \frac{1}{2}z^{-1}} + \frac{z^{-2}}{1 + \frac{1}{2}z^{-1}}$$
 (5.3)

$$\implies h(n) = \left(-\frac{1}{2}\right)^{n} u(n) + \left(-\frac{1}{2}\right)^{n-2} u(n-2)$$
(5.4)

using (4.19) and (4.6).

5.3 Sketch h(n). Is it bounded? Justify theoretically.

**Solution:** The following code plots Fig. 5.3.

wget https://raw.githubusercontent.com/gadepall/EE1310/master/filter/codes/hn.py

- 5.4 Convergent? Justify using the ratio test.
- 5.5 The system with h(n) is defined to be stable if

$$\sum_{n=-\infty}^{\infty} h(n) < \infty \tag{5.5}$$

Is the system defined by (3.2) stable for the impulse response in (5.2)?

5.6 Verify the above result using a python code.

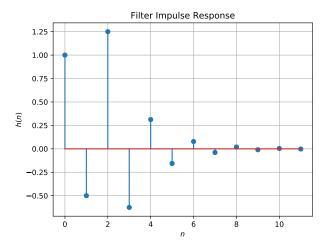


Fig. 5.3: h(n) as the inverse of H(z)

5.7 Compute and sketch h(n) using

$$h(n) + \frac{1}{2}h(n-1) = \delta(n) + \delta(n-2), \quad (5.6)$$

This is the definition of h(n).

**Solution:** The following code plots Fig. 5.7. Note that this is the same as Fig. 5.3.

wget https://raw.githubusercontent.com/ gadepall/EE1310/master/**filter**/codes/hndef .py

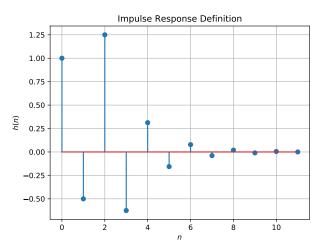


Fig. 5.7: h(n) from the definition

## 5.8 Compute

$$y(n) = x(n) * h(n) = \sum_{n=-\infty}^{\infty} x(k)h(n-k)$$
 (5.7)

Comment. The operation in (5.7) is known as *convolution*.

**Solution:** The following code plots Fig. 5.8. Note that this is the same as y(n) in Fig. 3.2.

wget https://raw.githubusercontent.com/ gadepall/EE1310/master/**filter**/codes/ ynconv.py

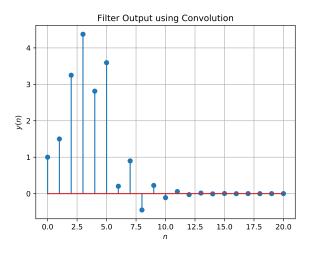


Fig. 5.8: y(n) from the definition of convolution

- 5.9 Express the above convolution using a Teoplitz matrix.
- 5.10 Show that

$$y(n) = \sum_{n=-\infty}^{\infty} x(n-k)h(k)$$
 (5.8)

## 6 DFT AND FFT

6.1 Compute

$$X(k) \stackrel{\triangle}{=} \sum_{n=0}^{N-1} x(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1$$
(6.1)

and H(k) using h(n).

6.2 Compute

$$Y(k) = X(k)H(k) \tag{6.2}$$

6.3 Compute

$$y(n) = \frac{1}{N} \sum_{k=0}^{N-1} Y(k) \cdot e^{j2\pi kn/N}, \quad n = 0, 1, \dots, N-1$$
(6.3)

**Solution:** The following code plots Fig. 5.8. Note that this is the same as y(n) in Fig. 3.2.

wget https://raw.githubusercontent.com/ gadepall/EE1310/master/**filter**/codes/yndft. py

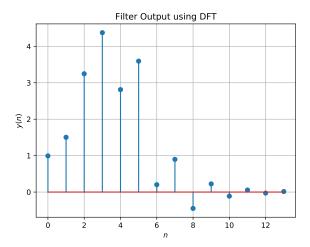


Fig. 6.3: y(n) from the DFT

- 6.4 Repeat the previous exercise by computing X(k), H(k) and y(n) through FFT and IFFT.
- 6.5 Wherever possible, express all the above equations as matrix equations.
- 6.6 Verify the above equations by generating the DFT matrix in python.

#### 7 Exercises

Answer the following questions by looking at the python code in Problem 2.3.

7.1 The command

in Problem 2.3 is executed through the following difference equation

$$\sum_{m=0}^{M} a(m) y(n-m) = \sum_{k=0}^{N} b(k) x(n-k) \quad (7.1)$$

where the input signal is x(n) and the output signal is y(n) with initial values all 0. Replace **signal.filtfilt** with your own routine and verify.

- 7.2 Repeat all the exercises in the previous sections for the above *a* and *b*.
- 7.3 What is the sampling frequency of the input signal?

**Solution:** Sampling frequency(fs)=44.1kHZ.

7.4 What is type, order and cutoff-frequency of the above butterworth filter

**Solution:** The given butterworth filter is low pass with order=2 and cutoff-frequency=4kHz.

7.5 Modifying the code with different input parameters and to get the best possible output.