

## FIR Filter

A Finite Impulse Response filter is a type of digital filter used in signal processing to perform various filtering operation on digital signals. Unlike IIR filters, FIR filters are stable and have linear phase characteristics. It has min phase distortion.

General equation,  $y(n) = b_0 x(n) + b_1 x(n-1) + b_2 x(n-2) + \dots + b_N x(n-N)$

↓

o/p at time n      N - Filter order  
in index n       $b_0, b_1, \dots, b_N$  filter coefficients

Designing an FIR filter involves determining the values of the filter coefficient to achieve the desired filter characteristics. There are various methods to design FIR filter including window methods, freq. sampling methods, and optimal linear phase method etc.

Some common applications

LPF, HPF, BPF for signal conditionis

Signal interpolation and decimation

Equalization of audio signal

Image processing such as image smoothing or edge detection

N.B. Interpolation increases the number of samples in signal to obtain a higher sampling rate, whereas decimation reduces the sampling rate

## FIR Filter

### Disadvantage:

### Advantages

- Complex implementation
- required more filter coefficients to be stored
- Long duration impulse response
- narrow transition band
- FIR filter requires more arithmetic operation
- Sable in design with A
- can be realized in both recursive or non-recursive
- exact linear phase
- flexible
- low sensitivity to quantization noise
- efficiently realized in hardware

# Impulse Response

representation

Symmetric Phase FIR Filter

$$h(n) = h(M-1-n) ; 0 \leq n \leq M-1 \quad M \rightarrow \text{filter order}$$

Let,  $M=8$

$$h(n) = h(8-1-n), 0 \leq n \leq 7$$

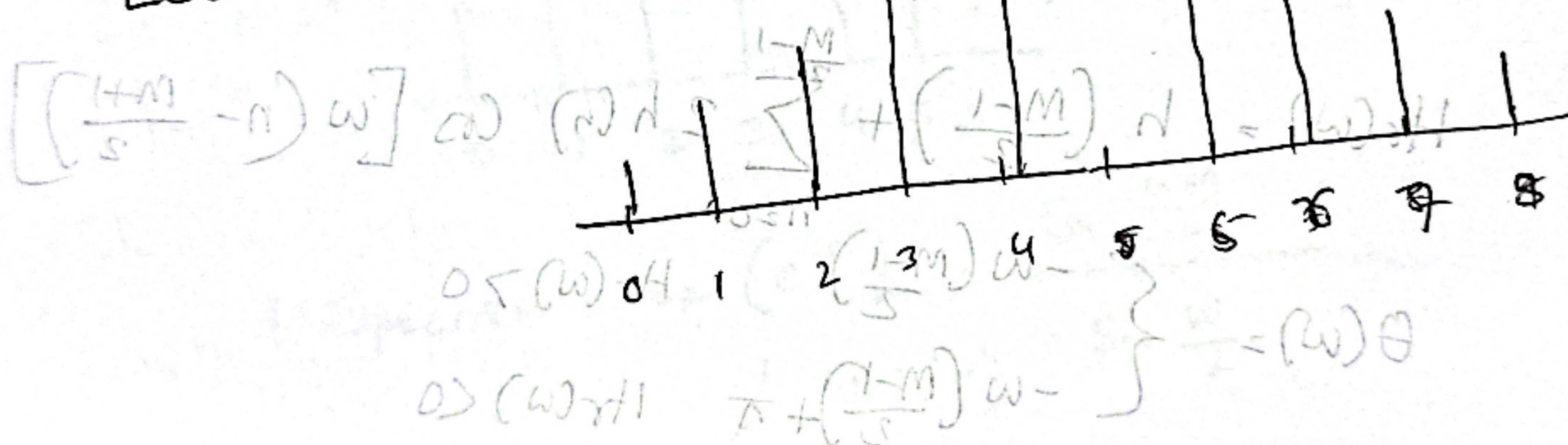
$$n=0 \quad h(0) = h(7)$$

$$n=1 \quad h(1) \stackrel{\text{WVS}}{=} h(7) \quad \text{Hence symmetric signal obtained} \quad (1)$$

$$n=2 \quad h(2) = \frac{1-M}{2} h(5) \quad \begin{array}{c} \text{[Diagram showing a sum of terms from } 0 \text{ to } M-1 \text{ with coefficients } \frac{1-M}{2}, 1, \dots, 1 \text{ and a final term } h(5) \text{ with coefficient } \frac{1-M}{2}.] \\ = (\omega)_r H \end{array}$$

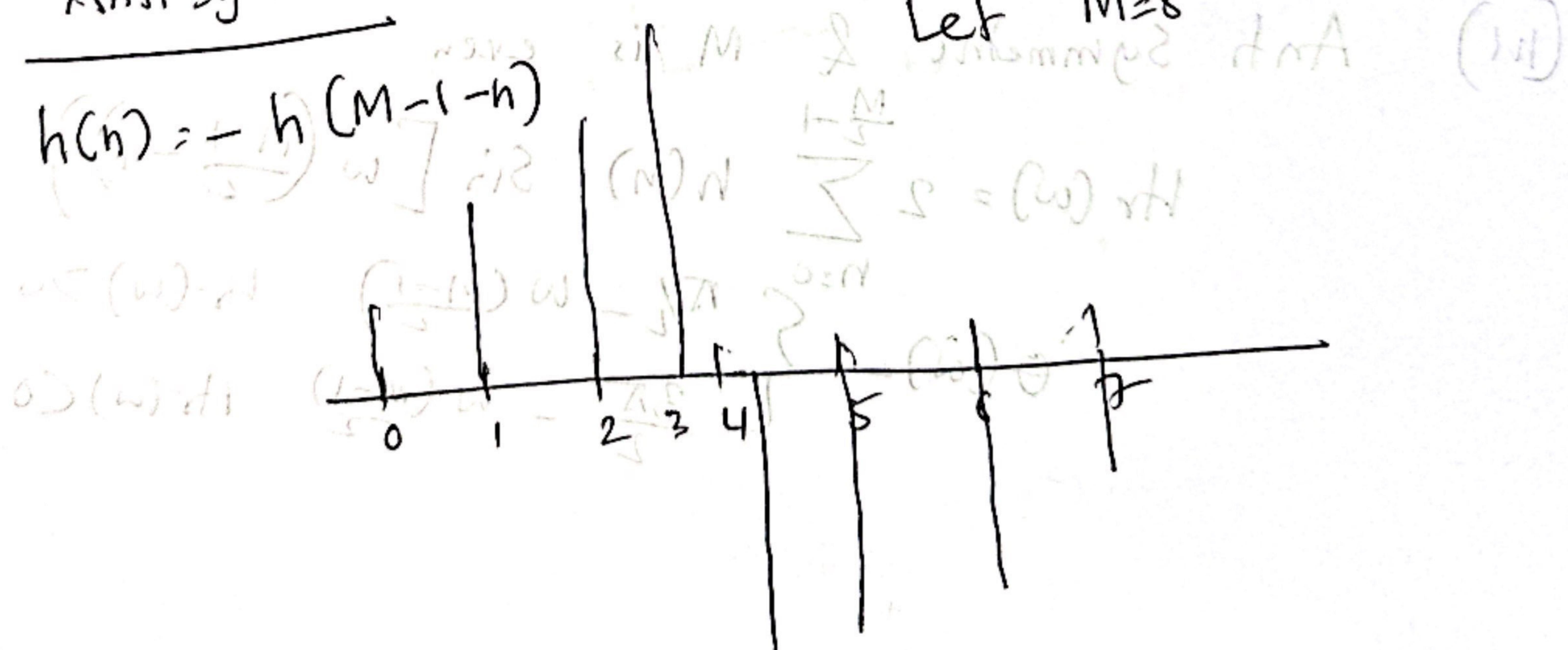
Symmetric representation of FIR filter

Let  $M=9$   $\Rightarrow$  M odd  $\Rightarrow$  symmetric signal obtained  $\quad (1)$



Anti-symmetric

$$h(n) = -h(M-1-n)$$



## Freq response

$$\text{given } h(n) \xrightarrow{\text{DTFT}} H(\omega) \quad \text{where } n \geq 0; \quad (n-1-M)d = (n)d$$

freq response of FIR filter can be given as

$$H(\omega) = \sum_{n=0}^{\infty} h_n e^{-jn\theta(\omega)} \quad \text{real part of } H(\omega)$$

(i) Symmetric Impulse response with  $(M)_N$  even

$$H_r(\omega) = \sum_{n=0}^{\frac{M-1}{2}} 2h(n) \cos \left[ \omega \left( n - \frac{M-1}{2} \right) \right]$$

$$\theta(\omega) = \begin{cases} -\omega \left( \frac{M-1}{2} \right), & H_r(\omega) > 0 \\ -\omega \left( \frac{M-1}{2} \right) + \pi, & H_r(\omega) < 0 \end{cases}$$

(ii) Symmetric Impulse response when  $M$  is odd

$$H_r(\omega) = h \left( \frac{M-1}{2} \right) + \sum_{n=0}^{\frac{M-1}{2}} 2h(n) \cos \left[ \omega \left( n - \frac{M+1}{2} \right) \right]$$

$$\theta(\omega) = \begin{cases} -\omega \left( \frac{M-1}{2} \right), & H_r(\omega) > 0 \\ -\omega \left( \frac{M-1}{2} \right) + \pi, & H_r(\omega) < 0 \end{cases}$$

(iii) Anti Symmetric &  $M$  is even

$$H_r(\omega) = 2 \sum_{n=0}^{\frac{M}{2}-1} h(n) \sin \left[ \omega \left( \frac{M-1}{2} - n \right) \right]$$

$$\theta(\omega) = \begin{cases} \pi - \omega \left( \frac{M-1}{2} \right), & H_r(\omega) > 0 \\ \frac{3\pi}{2} - \omega \left( \frac{M-1}{2} \right), & H_r(\omega) < 0 \end{cases}$$

Anti-symmetric with  $M$  odd

$$H_r(\omega) = 2 \sum_{n=0}^{\frac{M-1}{2}} h(n) \sin \left( \omega \left[ \frac{M-1}{2} - n \right] \right)$$

$$\theta(\omega) = \begin{cases} \pi - \omega \left( \frac{M-1}{2} \right) & H_r(\omega) > 0 \\ \frac{3\pi}{2} - \omega \left( \frac{M-1}{2} \right) & H_r(\omega) < 0 \end{cases}$$

Designing of Linear Phase FIR Filter using windows

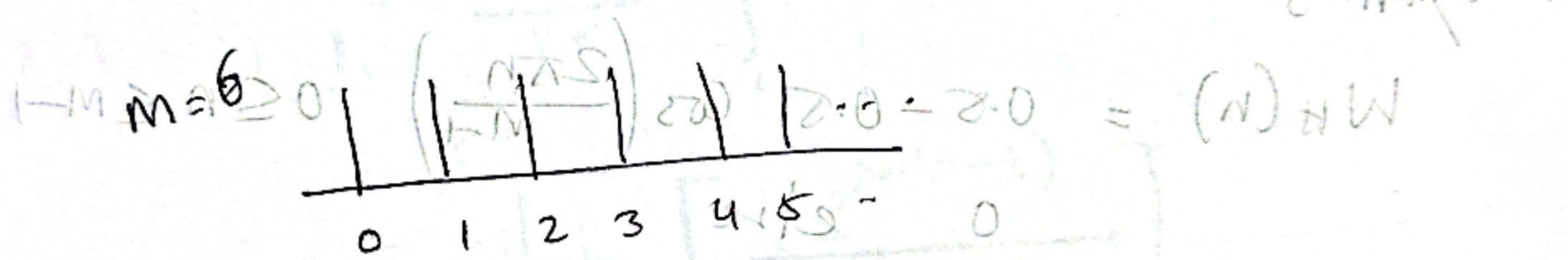
Method

$$\left( \frac{\omega \left( \frac{1-M}{2} \right)}{2\pi} \right) = (\omega_s)_{FW}$$

a) Rectangular window

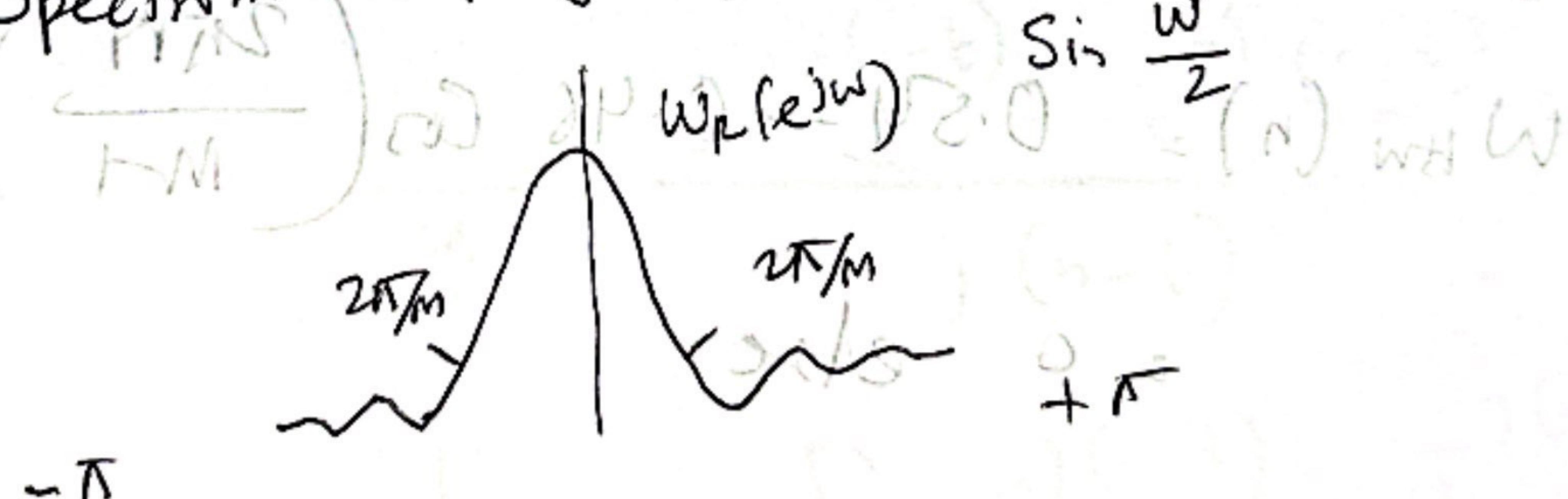
$$w_R(n) = 1 \quad 0 \leq n \leq M-1$$

else



$$W_R(e^{j\omega}) = \frac{\sin \frac{\omega M}{2}}{\sin \frac{\omega}{2}}$$

Spectrum

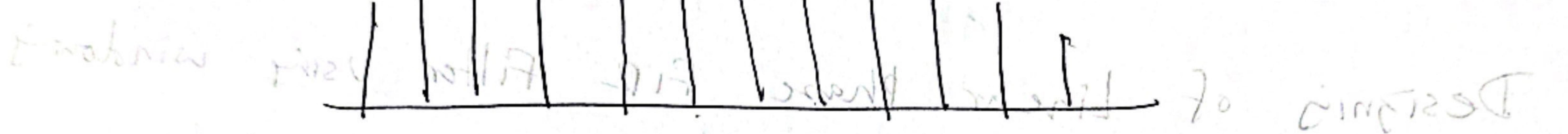


Bartlett window: (Translational window)  $w$  is symmetric about  $\frac{M}{2}$

$$w_B(n) = 1 - 2 \left[ n^2 \left( \frac{m-1}{2} \right) \right] / (m-1) \quad |0 \leq n \leq m-1$$

$$\text{at } n=0 \quad w_B(0) = 1 \quad (\frac{m-1}{2})w = \frac{\pi}{2}$$

$$\text{at } n=m-1 \quad w_B(m-1) = 0 \quad (\frac{m-1}{2})w = \frac{3\pi}{2}$$



Spectrum  $W_B(e^{j\omega}) = \left( \frac{\sin((m-1)\omega)}{\sin(\omega)} \right)^2$

bottom

Hann/Hanning window

$$w_H(n) = 0.5 - 0.5 \left| \cos \left( \frac{2\pi n}{M-1} \right) \right| \quad |0 \leq n \leq M-1|$$

0 else

Hann/Hanning window

$$w_{HH}(n) = 0.54 - 0.46 \cos \left( \frac{2\pi n}{M-1} \right) \quad |0 \leq n \leq M-1|$$

2, 0 else

Design the symmetric FIR LPR whose

$$H_d(\omega) = \begin{cases} e^{-j\omega t} & |\omega| \leq \omega_c \\ 0 & \text{otherwise} \end{cases} \quad \text{with } N=7 \\ \omega_c = 1 \text{ rad/sec}$$

$$h_d(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} H_d(\omega) e^{j\omega n} d\omega \quad \text{--- (1)}$$

$$\begin{aligned} H_d(\omega) &\rightarrow \int_0^1 e^{-j\omega t} e^{j\omega n} d\omega \\ &= \frac{1}{2\pi} \int_{-1}^1 e^{-j\omega t} e^{j\omega n} d\omega \\ &= \frac{1}{2\pi} \int_{-1}^1 e^{j\omega(n-t)} d\omega \\ &= \frac{1}{2\pi} \left[ \frac{e^{j\omega(n-t)}}{j(n-t)} \right] \Big|_{-1}^1 \\ &= \frac{1}{2\pi} \left[ \frac{e^{j\omega(n-t)}}{j(n-t)} - \frac{e^{-j\omega(n-t)}}{j(n-t)} \right] \end{aligned}$$

$$\begin{aligned} \frac{(t-n+1-M)}{(t-n+1-M)\pi} &\leq \frac{R}{2\pi} \\ \frac{e^{j(n-t)} - e^{-j(n-t)}}{j(n-t)\pi} &= \frac{e^{j(n-t)} - e^{-j(n-t)}}{2j(n-t)\pi} \end{aligned}$$

$$\begin{aligned} \frac{1}{2\pi(n-t)} \left[ \frac{e^{j(n-t)} - e^{-j(n-t)}}{2j(n-t)\pi} \right] &= \frac{1}{2\pi(n-t)} \left[ \frac{(e^{j(n-t)} - e^{-j(n-t)})}{(t-n)\pi} \right] \\ &= \frac{\sin(n-t)}{(t-n)\pi} \end{aligned}$$

$$= \frac{\sin(n-t)}{(t-n)(n-t)} \quad n \neq t$$

$$t-n+1-M = t+n-$$

If  $n = t$   $\Rightarrow H(t)$  determine with help of

$$h_d(t) = \frac{1}{2\pi} \int_{-\pi}^{\pi} h(\omega) e^{j\omega t} d\omega = \frac{1}{\pi}$$

$$\textcircled{1} \quad h_d(n) = \begin{cases} \frac{\sin(n-t)}{\pi(n-t)} & n \neq t \\ \frac{1}{\pi} & n = t \end{cases} \quad \text{with } H(t)$$

determine the value of  $t$

$$h(n) = h(M-1-t) \frac{1}{\pi}$$

$$h(n) = h_d(n) \cdot w(n)$$

$$h_d(n) w(n) = h_d(M-1-t) w(n)$$

$$\left[ \frac{h_d(n)}{(t-n)\pi} = \frac{h_d(M-1-t)}{(M-1-n)\pi} \right]$$

$$\frac{-\sin((n-t)(t-n)\pi)}{\pi(n-t)} = \frac{\sin((M-1-t)(M-1-n-t)\pi)}{\pi(M-1-n-t)}$$

$$\left[ \frac{\sin(-(t-n)\pi)}{\pi(-n-t)} \right] = \frac{\sin((M-1-n-t)\pi)}{\pi(M-1-n-t)}$$

$$-\pi(n-t) \stackrel{(t=n)}{=} M-1-n-t$$

$$-n+t = M-1-n-t \quad t = \frac{M-1}{2}$$

$$h_d(n) = \begin{cases} \frac{\sin\left(n - \frac{M-1}{2}\right)}{\pi \cdot \left(n - \frac{M-1}{2}\right)} & n \neq \frac{M-1}{2} \\ \frac{1}{\pi} & n = \frac{M-1}{2} \end{cases}$$

Since  $M=2$

$$h_d(n) = \begin{cases} \frac{\sin\left(n - 3\right)}{\pi \cdot \left(n - 3\right)} & n \neq 3 \\ \frac{1}{\pi} & n = 3 \end{cases}$$

$$n = 0 \rightarrow 6$$