

Evaluation of Interference Cancellation Architectures for Heterogeneous Cellular Networks

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Outline

1 Motivation

- From Homogeneous to Heterogeneous Cellular Networks
- Challenge - Inter-Channel Interference Cancellation

2 Basic Concepts

3 Signal & System Model

- Transceiver System Model
- Interference Model
- Signal Model

4 Metric Computing Device

5 Simulation Results

6 Summary

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From Homogeneous to Heterogeneous Cellular Networks

- Current homogeneous cellular networks are reaching their breaking point
- Introducing pico/femto base-stations and relay nodes in a macro cellular networks:
 - improves spatial reuse and coverage
 - allows to achieve higher data rates
- Totally random base stations (BS) placement
- Significant imbalance in the downlink transmit powers: Home eNode B (HeNB) vs. macro eNode B

Inter-Channel Interference (ICI) Management

Conventional ICI management approaches need to be revised:

- ICI Avoidance:

Complexity of cell coordination based systems increases drastically

- ICI Mitigation:

Signal randomization at the transmitter → Interference is just noise

Interference cancellation at the user equipment (UE)

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2 **Basic Concepts**

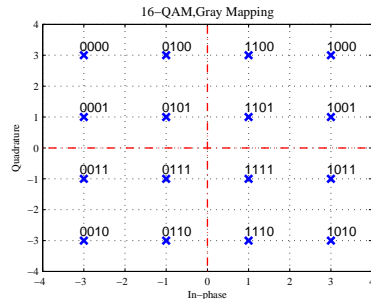
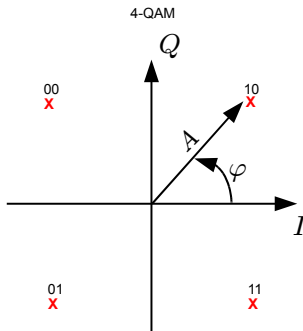
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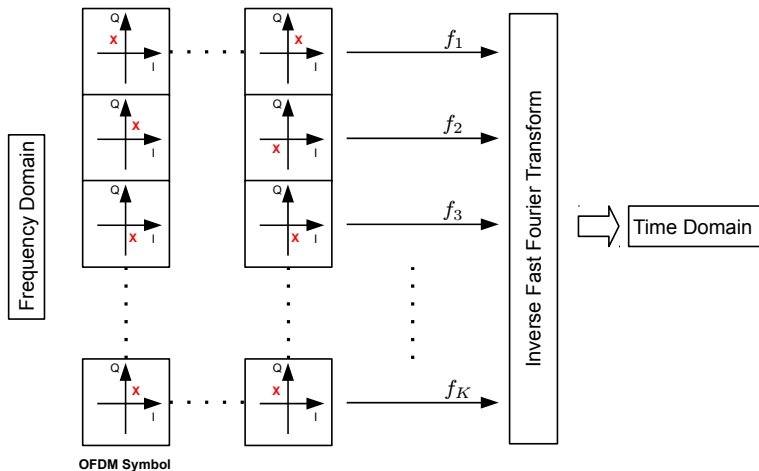
6 Summary

Quadrature Amplitude Modulation (QAM)



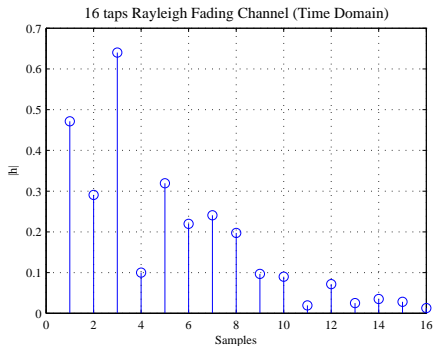
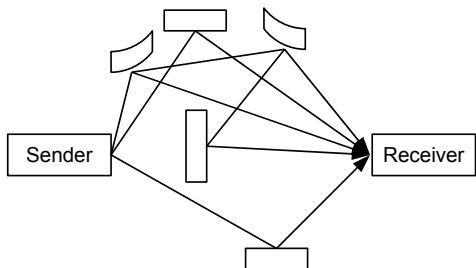
- Possibility to transport n -bits using one symbol
- Information is embedded in both: amplitude A and phase φ

Orthogonal Frequency Division Multiplexing (OFDM)



OFDM Symbol

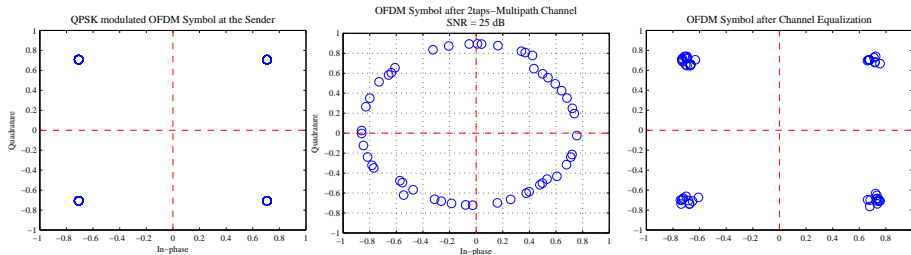
Propagation Channel: Rayleigh Multipath Fading



- Non line of sight (NLOS) multipath propagation channel.

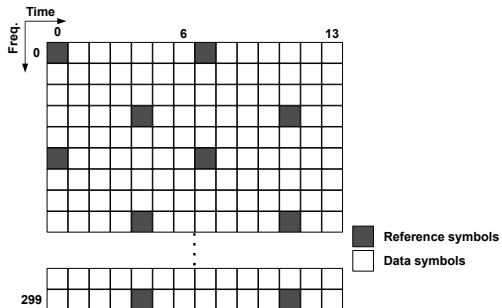
- Each tap = delayed and attenuated signal copy

Multipath Fading's Effect on the Signal



- Multipath channel effect needs to be corrected in the receiver
- Equalization needs knowledge about the propagation channel

Pilot-Based Least Squares (LS) Channel Estimation



- Pilots are reference symbols.
- Each cell has a specific pilot pattern.

- Position and modulation scheme are well known by the receiver.

and more Definitions

- Additive White Gaussian Noise (AWGN):

Normally distributed random signal with equal power over all frequencies

- Signal to Noise Ratio:

$\text{SNR} = \text{signal power} / \text{noise power} \text{ [dB]}$

- Signal to Interference Ratio:

$\text{SIR} = \text{signal of interest power} / \text{co-channel interferer power [dB]}$

- Mean squared Error:

$\text{MSE} = \text{average squared difference between a value } v \text{ and its estimation } \hat{v}.$

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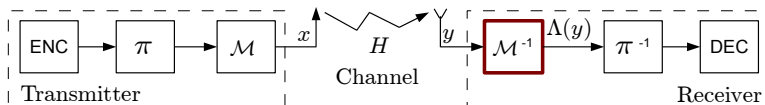
4 Metric Computing Device

5 Simulation Results

6 Summary

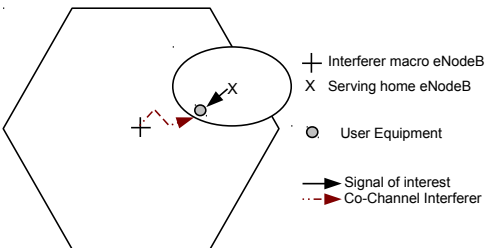
Transceiver System Model

Bit interleaved coded modulation (BICM) transceiver



- Sender: encoder (ENC), bit interleaver π , modulator \mathcal{M}
- Receiver: metric computing device \mathcal{M}^{-1} , deinterleaver π^{-1} and decoder (DEC)

Interference Model



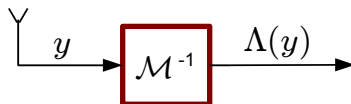
System Assumptions

- 1 x_1 : symbol of interest (SOI), 4-QAM
- 2 x_2 : Co-Channel interferer symbol, 4-QAM
- 3 H_1 : SOI's channel
- 4 H_2 : interferer's channel

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Metric as Log-Likelihood Ratio (LLR) (1)

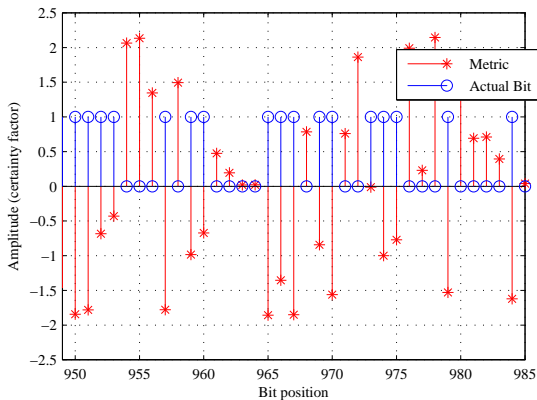


$\Lambda(y)$ is the log-likelihood ratio of a bit

- 1 $\Lambda(y) > 0 \longrightarrow$ Bit is more likely to be a 0.
- 2 $\Lambda(y) < 0 \longrightarrow$ Bit is more likely to be a 1.

Data is passed from one component to another in form of LLR.
The main physical difference between II-R and IA-R is \mathcal{M}^{-1}

Metric as Log-Likelihood Ratio (LLR) (2)



Amplitude = Certainty factor

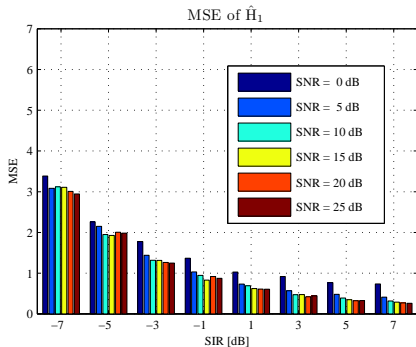
IA-LLR, an optimized II-LLR

$$\Lambda^{\text{IA-R}}(y) = \Lambda^{\text{II-R}}(y) + \Psi(\hat{H}_1, \hat{H}_2), \quad (3)$$

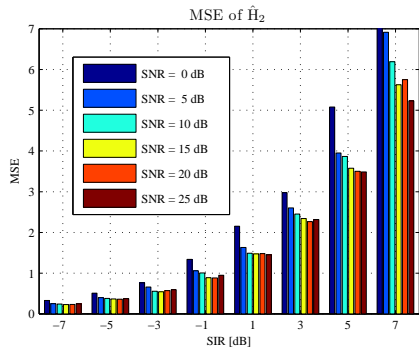
$\Psi()$ is a cost function depending on:

- \hat{H}_1 : channel estimation of the signal of interest
- \hat{H}_2 : Interferer-channel estimation.

Mean Squared Error, Quality Factor for Estimations (1)

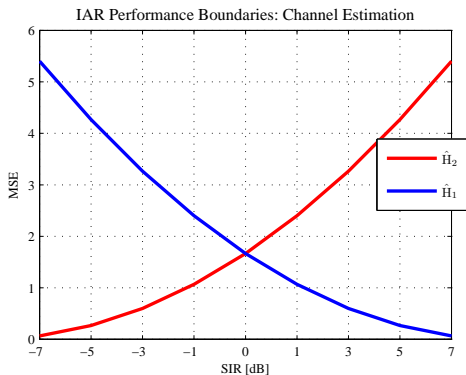


$SIR < 0 \rightarrow \text{MSE of } \hat{H}_1 \nearrow$



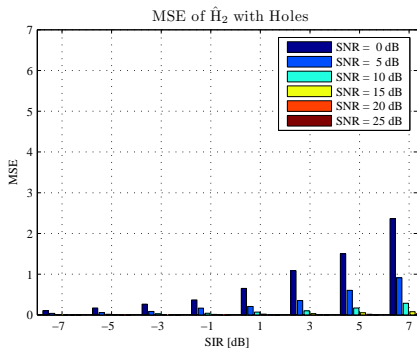
$SIR > 0 \rightarrow \text{MSE of } \hat{H}_2 \nearrow$

Mean Squared Error, Quality Factor for Estimations (2)

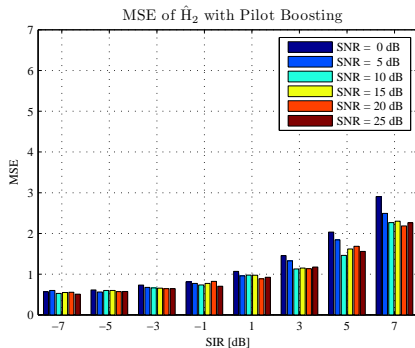


IAR Performance Boundaries: \hat{H}_1 & \hat{H}_2

Base Station Channel Estimation Enhancement



Serving BS: Holes



Interferer BS: Pilot Boosting

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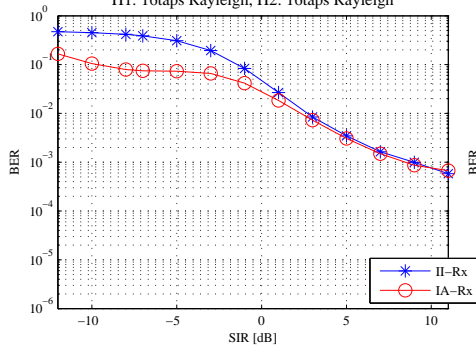
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Theory: Perfect Channel Knowledge (1)

Performance of IA-R vs II-R by SNR = 5 dB

H1: perfect, H2: perfect

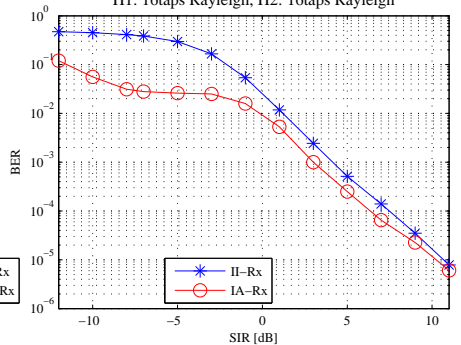
H1: 16taps Rayleigh, H2: 16taps Rayleigh



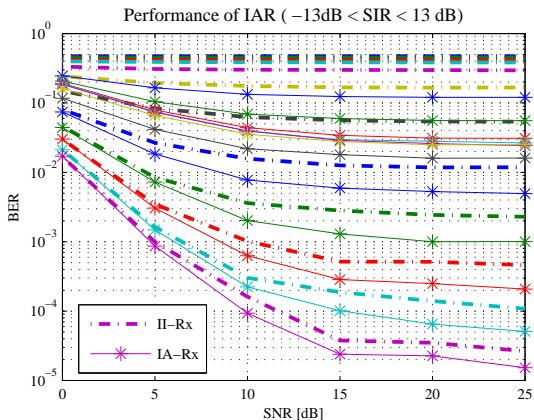
Performance of IA-R vs II-R by SNR = 20 dB

H1: perfect, H2: perfect

H1: 16taps Rayleigh, H2: 16taps Rayleigh



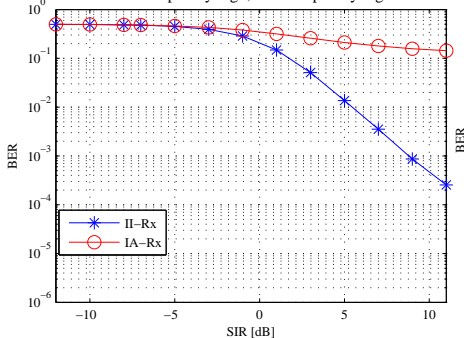
Theory: Perfect Channel Knowledge (2)



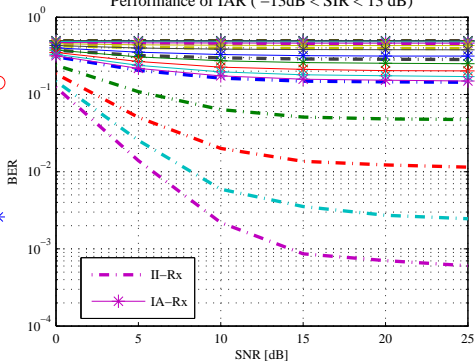
Under the assumption perfect channel knowledge, IA-R always out performs II-R

The Hard Reality: Channel Estimation

Performance of IA-R vs II-R by SNR = 15 dB
H1: LS, H2: LS
H1: 16taps Rayleigh, H2: 16taps Rayleigh

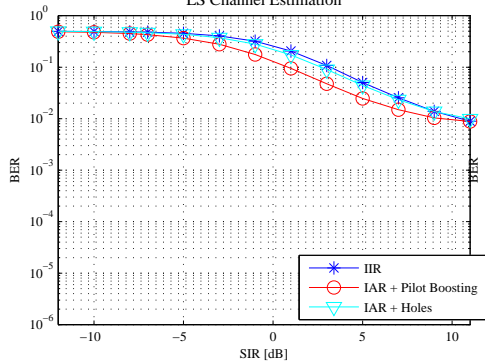


Performance of IAR ($-13\text{dB} < \text{SIR} < 13\text{dB}$)

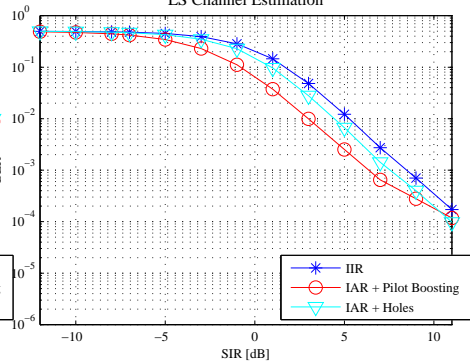


Proposed Solution: Base Station Enhancement (1)

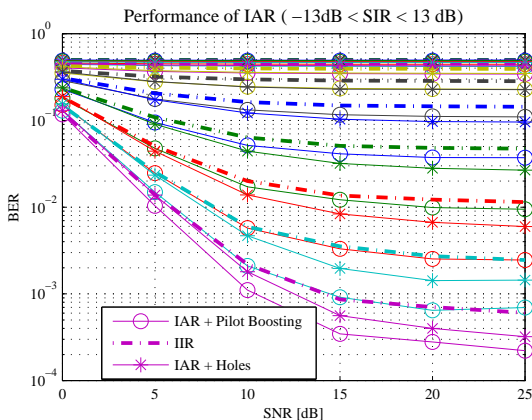
Performance of different IC Methods by SNR = 5
LS Channel Estimation



Performance of different IC Methods by SNR = 20
LS Channel Estimation



Proposed Solution: Base Station Enhancement (2)



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- Interference cancellation at the receiver is a very promising technology, but still at the begin
- Sole limitation of the IA-R is the quality of the channel estimation
- Base Stations need to be involved in the channel estimation process
- Worst case, switch off the cost function $\Psi()$ and IA-R becomes II-R

Reference

- [1] Ghaffar R. and Knopp R., Low complexity metrics for BICM SISO and MIMO systems. In *71st IEEE VTC, 2010*.
- [2] Ghaffar R. and Knopp R., Interference-aware receiver structure for multi-user MIMO and LTE. *Eurasip Journal 2011*.
- [3] Knisely, D. and Yoshizawa, T., Standardization of femtocells in 3GPP, *Communications Magazine, IEEE vol.47, no.9 2009*.
- [4] Stefania Sesia, Issam Toufik, and Matthew Baker. LTE, The UMTS Long Term Evolution. *Wiley Publishing, 2009*.

Questions?

Thank You!