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# Lecture 10

## Induction Motor II

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# 1 Frequency of The Induced Voltage (or Current) in Rotor Windings $f_r$

We know that

$$f_s = \frac{PN_s}{120}$$

How can we get  $f_r$ ? a common mistake to assume that  $f_r = \frac{PN_r}{120}$  (wrong), Why? because it doesn't depend on rotor speed  $N_r$ , it depends on the relative speed, therefore:

$$f_r = \frac{P(N_s - N_r)}{120}$$

$$f_r = \frac{P(N_s - N_r)}{120} \times \frac{N_s}{N_s}$$

$$\boxed{f_r = sf_s}$$

## 2 Equivalent Circuit of Three Phase Induction Motor

Remember that the rotor windings are shorted through brushes riding on the slip rings.

What if we removed the brushes? in other words if we removed the short circuit.

1. Motor will not start.
2. Motor will act as a transformer Why? there are winding in rotor and stator and there are electromagnetic induction between them.

In most cases, turns ration in induction motor is 1:1

Therefore, the equivalent circuit of three phase induction motor is similar to the equivalent circuit of the transformer, with two main differences:

1. There are no electrical load on rotor circuit(short circuit)<sup>1</sup>
2. The primary and secondary circuit don't have the same frequency, instead, there are  $f_s$  in stator circuit and  $f_r$  in rotor circuit.

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<sup>1</sup>There are mechanical load, but not considered, there are no electric load e.g. a resistor

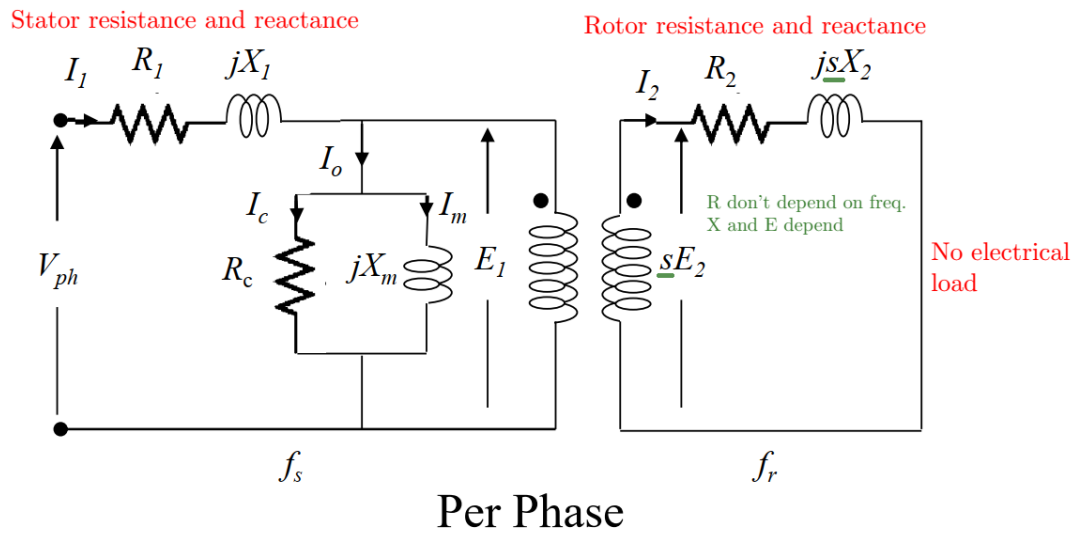


Figure 1

Note: referring is not valid between two circuits with different frequencies.

because  $f_r = sf_s$ , we divide the rotor circuit by  $s$

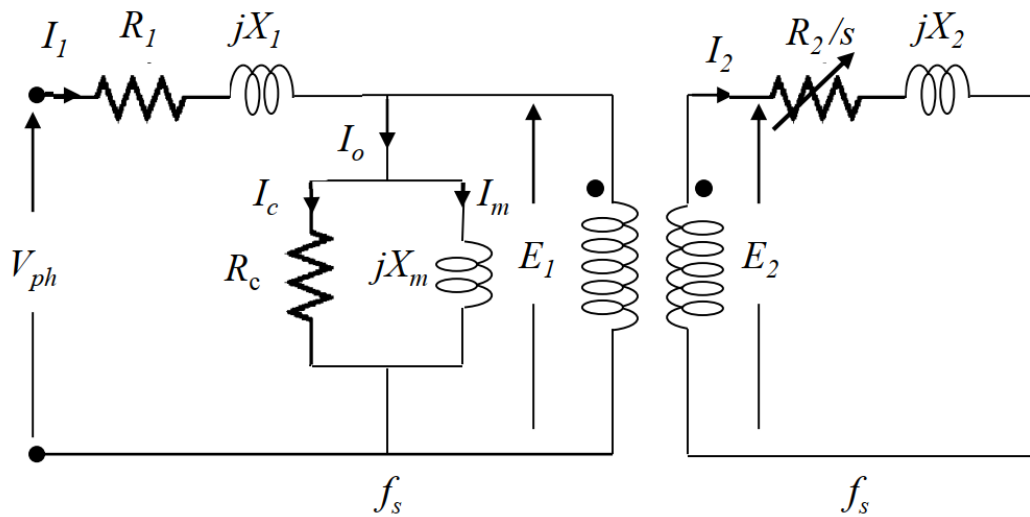


Figure 2

Because turns ratio is 1:1, referring is very simple, no need to multiply or divide.

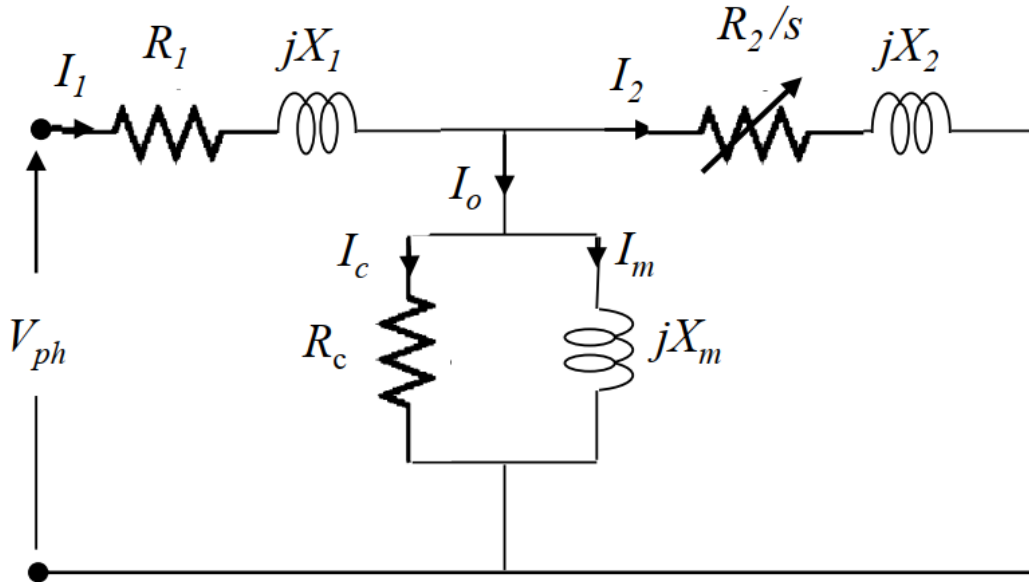


Figure 3

Analysis of this circuit is not easy, we apply two approximations.

**First approximation:**

Voltage drop on stator is very low such that supply voltage  $V_{ph}$  is the same on magnetizing branch. (we can move the magnetizing branch as shown in fig 4)

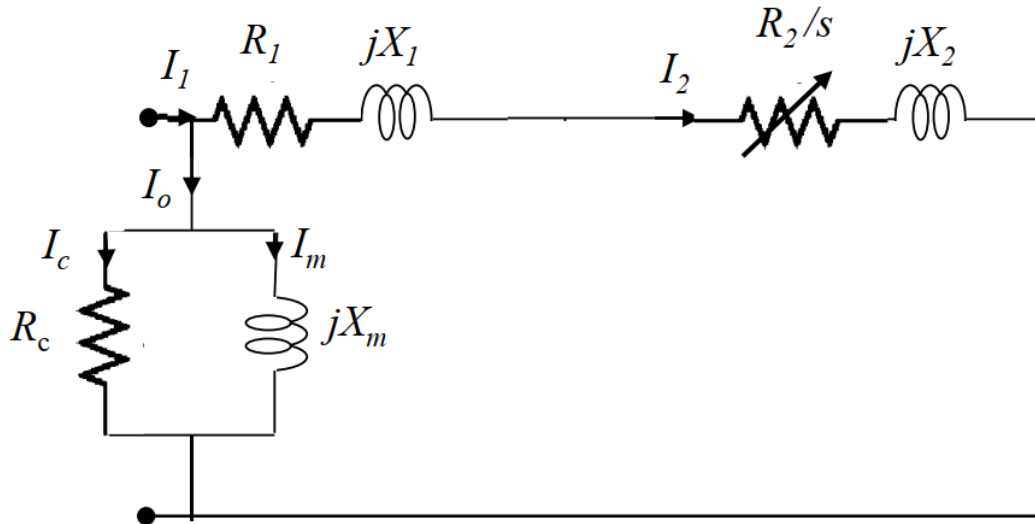


Figure 4

**Second approximation:**  $I_0$  is very small relative to  $I_1$  (nearly 5%) (magnetizing branch can be ignored)

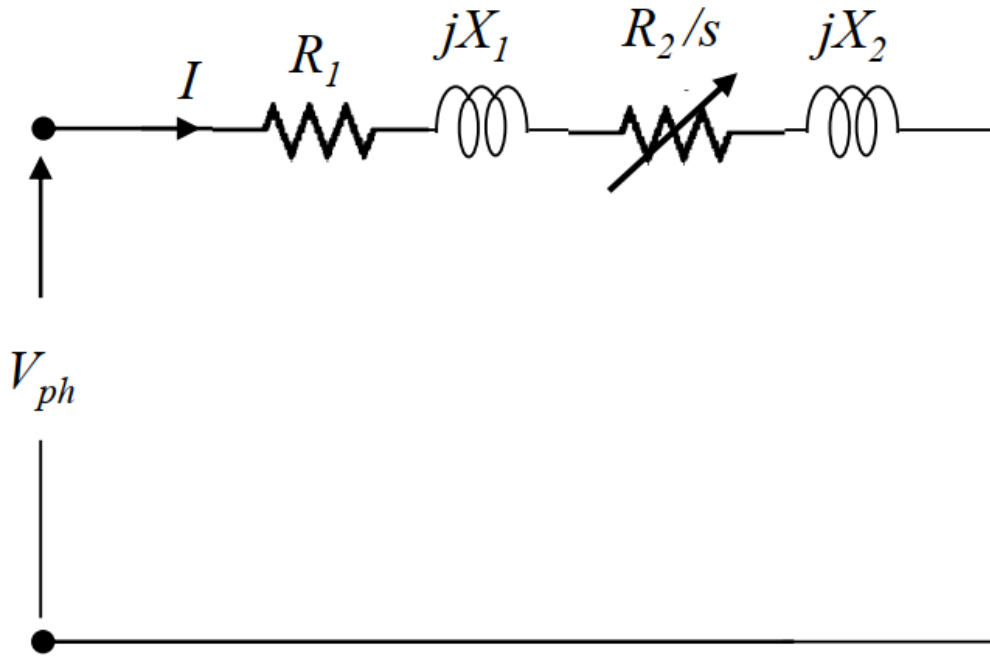


Figure 5: Equivalent Circuit Per Phase.

We divide  $\frac{R_2}{s}$  to  $\left(R_2 + R_2 \frac{1-s}{s}\right)$

**Why?** to separate  $R_2$  which is a physical resistance and  $R_2 \frac{1-s}{s}$  on which the developed power is noticed (called dynamic resistance)

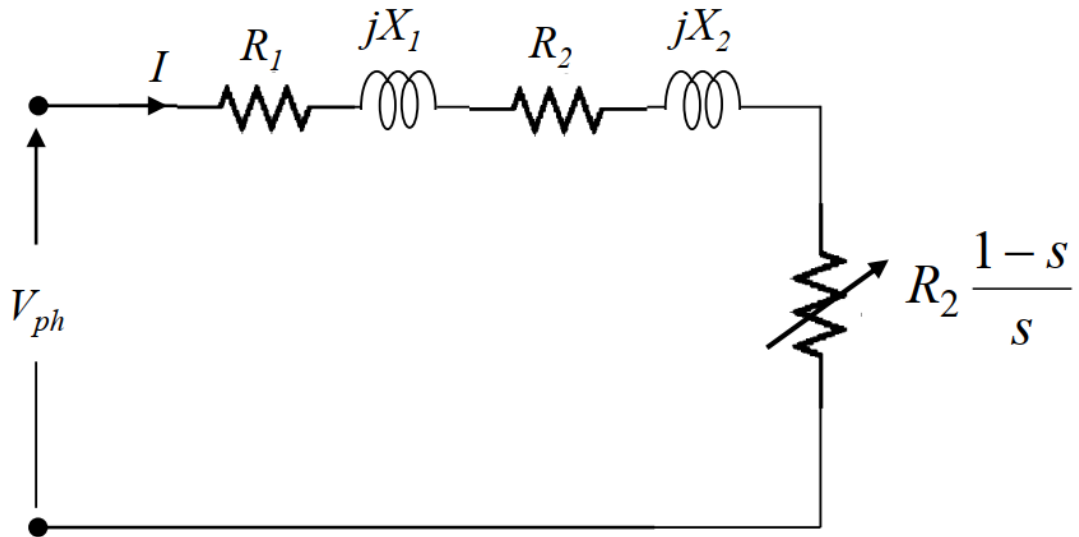


Figure 6: Equivalent Circuit Per Phase.

### 3 Power Flow and Losses

From ohm's law:

$$\overline{I_{\text{ph}}} = \frac{V_{\text{ph}}}{\left(R_1 + \frac{R_2}{s}\right) + j(X_1 + X_2)} = |I_{\text{ph}}| \angle \phi$$

Note the power factor in 3 phase induction motor is **always lagging**.

We know that input power:

$$P_{\text{in}} = 3V_{\text{ph}}I_{\text{ph}}\cos(\phi)$$

Air gap power = input power – stator losses

it is the power that transfers from stator to rotor.

$$P_g = P_{\text{in}} - S.L. = 3I^2\frac{R_2}{s}$$

Developed power = Air gap power - rotor copper loss

$$R.C.L = 3I^2R_2$$

$$P_d = P_g - R.C.L = 3I^2R_2\frac{1-s}{s}$$

Output power = developed power - rotational losses

$$P_{\text{out}} = P_d - P_r$$

$$\boxed{P_g : R.C.L : P_d = 1 : s : 1 - s}$$

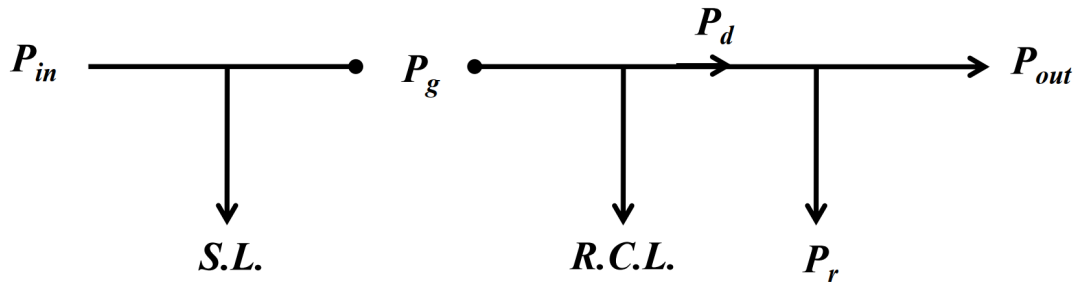


Figure 7: Power flow and losses in induction motor.

### Is there poles in induction motor?

Yes, but not similar to synchronous or DC machines, in synchronous or DC machines the poles are real, physical poles, in induction motor, poles depends on the way of winding.

In examples: the efficiency  $\eta$  of induction motor is around 85%.

1. Q — A 10 hp, 6-pole, 440 V, 60 Hz, Y-connected, three-phase induction motor is designed to operate at 3% slip on full load. The stator loss is 250 W and the rotational loss is 4% of the power output. Determine :

1. Air gap power.
2. Rotor copper loss R.C.L.
3. Input power
4. Efficiency.
5. Output torque.
6. Developed torque

A —

$$P_{\text{out}} = 10\text{hp} \times 746 = 7460W$$

$$P_r = 0.04 \times 7460 = 298.4W$$

$$(1) \quad P_d = P_{\text{out}} + P_r = 7460 + 298.4 = 7758.4W$$

$$(2) \quad P_g = \frac{P_d}{1-s} = \frac{7758.4}{1-0.03} = 7998.3W$$

$$R.C.L. = s \times P_g = 0.03 \times 7998.3 = 239.9W$$

or

$$(3) \quad R.C.L. = P_g - P_d = 7998.3 - 7758.4 = 239.9W$$

$$(4) \quad P_{\text{in}} = P_g + \text{stator losses S.L.} = 7998.3 + 250 = 8248.3W$$

$$(5) \quad \eta = \frac{P_{\text{out}}}{P_{\text{in}}} = \frac{7460}{8248.3} = 90.44\%$$

$$T_{\text{out}} = \frac{P_{\text{out}}}{\omega_r}$$

$$\omega_r = 2\pi \frac{N_r}{60}$$

$$N_r = N_s(1-s)$$

$$N_s = 120 \times \frac{f}{P} = 120 \times \frac{60}{6} = 1200 \text{ rpm}$$

$$\therefore N_r = 1200(1-0.03) = 1164 \text{ rpm}$$



$$\therefore \omega_r = 2\pi \frac{1164}{60} = 121.83 \text{ rad/sec}$$

$$\therefore T_{\text{out}} = \frac{7460}{121.83} = 61.23 \text{ N.m.}$$

(6)

$$T_d = \frac{P_d}{\omega_r} = \frac{7758.4}{121.83} = 63.68 \text{ N.m.}$$

## 4 Torque-Slip Relationship

We know

$$T_{\text{out}} = \frac{P_{\text{out}}}{\omega_r}$$

Therefore :

$$T_d = \frac{P_d}{\omega_r} = \frac{P_g(1-s)}{\omega_s(1-s)} = \frac{P_g}{\omega_s}$$

Why we did that? because  $\omega_s$  is constant, unlike  $\omega_r$

Remember:

$$P_g = 3I^2 \frac{R_2}{s} = 3R_s \times \frac{1}{s} \times \underbrace{\frac{V_{\text{ph}}^2}{\left(R_1 + \frac{R_2}{s}\right)^2 + (X_1 + X_2)^2}}_{I^2}$$

$$\therefore T_d = \frac{3R_2}{\omega_s \times s} \times \frac{V_{\text{ph}}^2}{\left(R_1 + \frac{R_2}{s}\right)^2 + (X_1 + X_2)^2}$$

Everything in this equation is constant, except ( $s$ )

Note that ( $s$ ) varies from 1 to 0, not from 0 to 1, because start occur first.

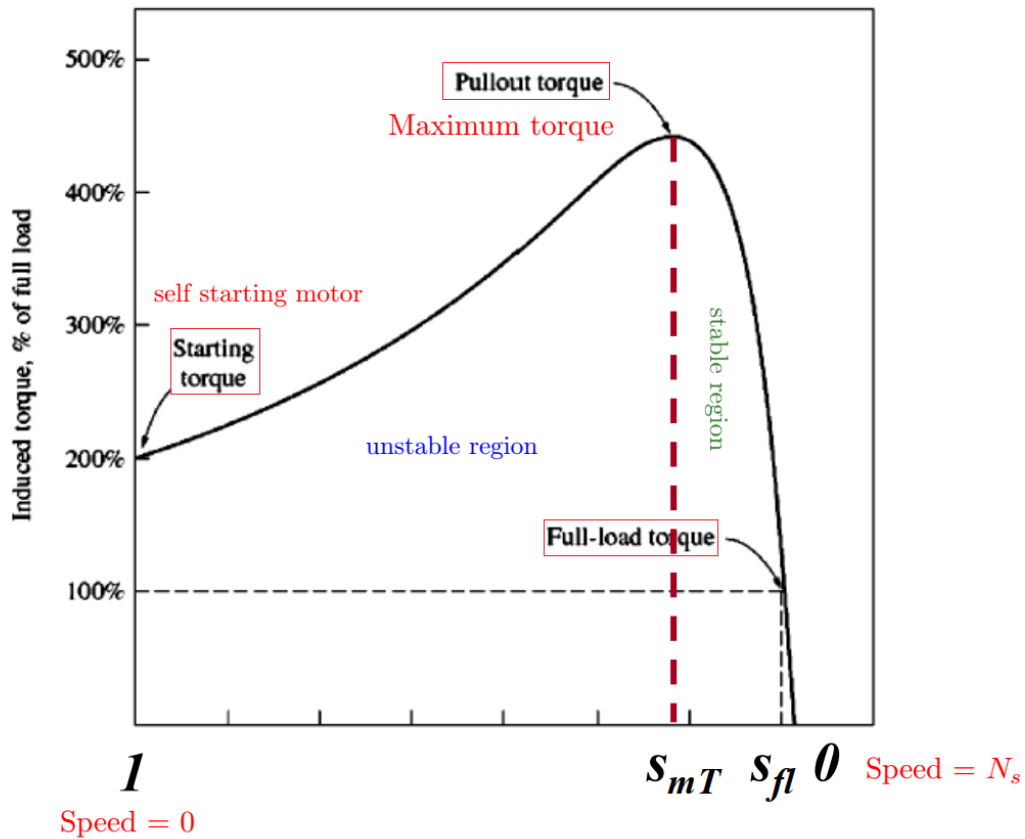


Figure 8: Torque-Slip relationship - To get stable operation in any motor,  $\frac{dT}{d\omega}$  = **negative**, when one increases the other decreases.

$s_{\text{starting}} = 1$ .

$s$  at full load will be given.

What about  $s_{mT}$  ?

We want to find  $s_{mT}$  (slip corresponding to maximum torque)

It can be obtained by differentiation, but there are easier way.

The maximum possible torque occurs when the air-gap power  $P_g$  is maximum.

Using maximum power transfer theorem, the maximum power transfer to the load resistor  $\frac{R_2}{s}$  occurs when **the magnitude of that resistance = the magnitude of the impedance** seen by the source.

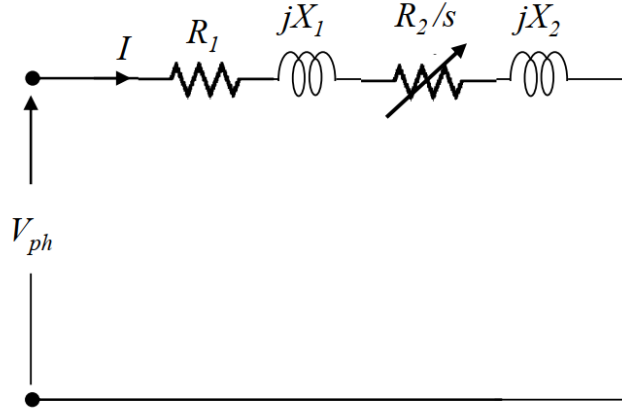


Figure 9

$$\left| R_1 + \frac{R_2}{s_{mT}} \right| = |j(X_1 + X_2)|$$

$$\therefore \frac{R_2}{s_{mT}} = |R_1 + j(X_1 + X_2)| = \sqrt{R_1^2 + (X_1 + X_2)^2}$$

$$s_{mT} = \frac{R_2}{\sqrt{R_1^2 + (X_1 + X_2)^2}}$$

In some examples, stator impedance is small and can be ignored, then  $s_{mT} = \frac{R_2}{X_2}$ .  $N_{r_{mt}}$  is the rotor speed corresponding to maximum torque, and is the minimum speed motor operate at.

$$N_{r_{mT}} = N_s(1 - s_{mT})$$

To get the maximum torque, substitute by  $s_{mT}$  in the general torque equation

$$T_{\max} = \frac{3R_2}{\omega_s \times s_{mT}} \times \frac{V_{ph}^2}{\left( R_1 + \frac{R_2}{s_{mT}} \right)^2 + (X_1 + X_2)^2}$$

## 5 Maximum Developed Power

The maximum developed power occurs when the power consumed in the  $R_2 \frac{1-s}{s}$  resistor is maximum

It can be obtained by differentiation, but there are easier way.

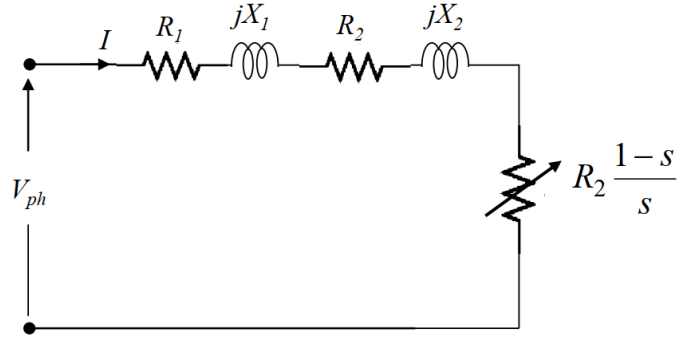


Figure 10

the maximum power transfer to the load resistor occurs when the **magnitude of that resistance = the magnitude of the impedance ( $R_l$ )** seen by the source.

$$R_L = \sqrt{(R_1 + R_2)^2 + (X_1 + X_2)^2} = R_2 \frac{1 - s_{mP}}{s_{mP}}$$

$s_{mP}$  is the slip corresponding to maximum power.

$$\therefore s_{mP} = \frac{R_2}{R_2 + R_L}$$

$$N_{r_{mP}} = N_s(1 - s_{mP})$$

$$P_{d_{\max}} = 3I_{mP}^2 R_L$$

$$P_{d_{\max}} = 3I_{mP}^2 R_L$$

$$P_{d_{\max}} = \frac{3V_{mP}^2}{2(R_1 + R_2 + R_L)}$$

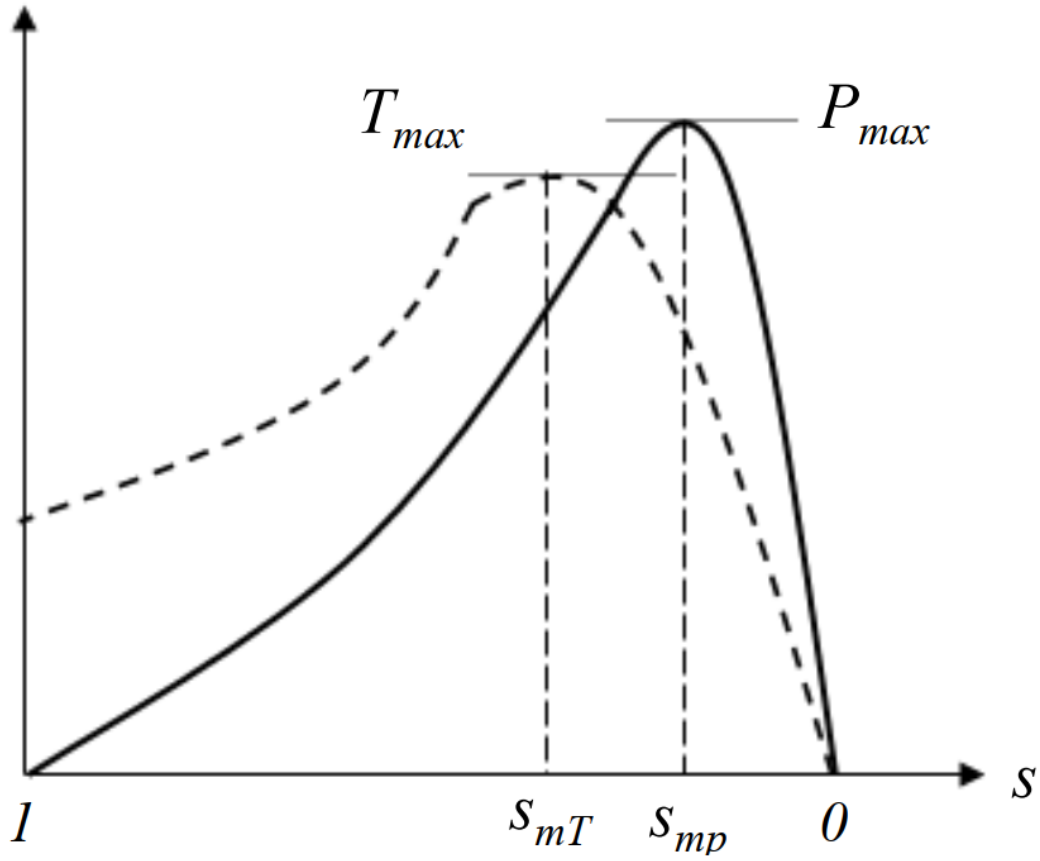


Figure 11: Maximum power and maximum torque occur at two different points

Note that:

$$s_{mT} > s_{mP}$$

## 6 Effect of Varying Rotor Resistance

$$\therefore T_d = \frac{3R_2}{\omega_s \times s} \times \frac{V_{ph}^2}{\left(R_1 + \frac{R_2}{s}\right)^2 + (X_1 + X_2)^2}$$

We want to change rotor resistance, substitute:

$$R_2 = R_2 + R_{ext}$$

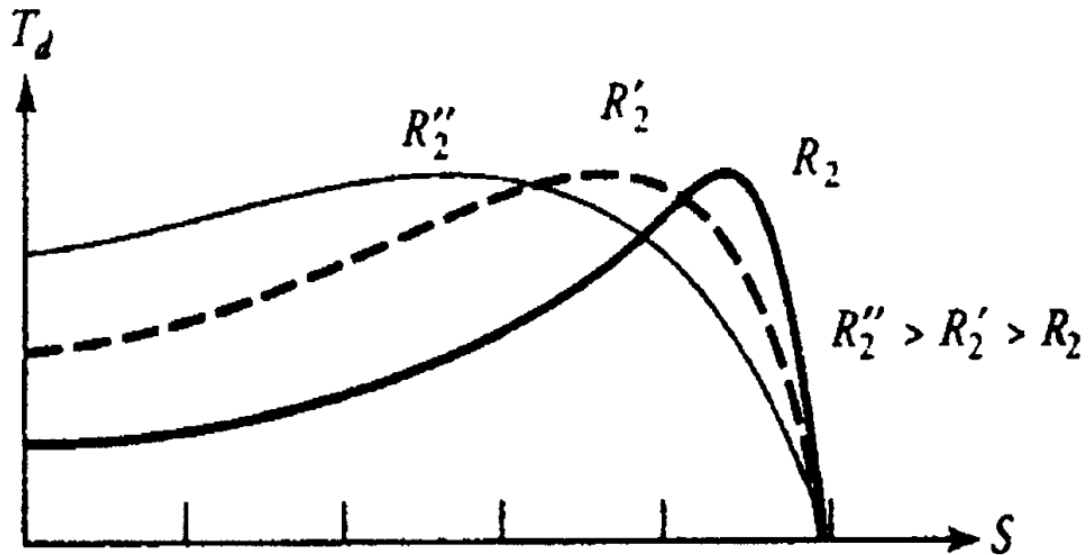


Figure 12: Effect of varying rotor resistance

Note that **the maximum torque doesn't depend on  $R_2$**

When  $R_2$  increases  $\uparrow$ , starting torque increases  $\uparrow$ , which is good, to be able to start,  
**motor torque must be greater than load torque + the friction**

At start:

$$s_{mT} = 1$$

We want to get  $R_{\text{external}}$  to obtain Maximum torque at start:

$$s_{mT} = \frac{R_2 + R_{\text{ext}}}{\sqrt{R_1^2 + (X_1 + X_2)^2}} = 1$$

$$R_{\text{ext}} = \sqrt{R_1^2 + (X_1 + X_2)^2} - R_2$$

Note that this method can be used only in **wound rotor**, not the squirrel cage rotor.