



CC2650 SimpleLink™ Multistandard Wireless MCU

1 Device Overview

1.1 Features

- Microcontroller
 - Powerful ARM® Cortex®-M3
 - EEMBC CoreMark[®] Score: 142
 - Up to 48-MHz Clock Speed
 - 128KB of In-System Programmable Flash
 - 8KB of SRAM for Cache
 - 20KB of Ultralow-Leakage SRAM
 - 2-Pin cJTAG and JTAG Debugging
 - Supports Over-The-Air Upgrade (OTA)
- Ultralow-Power Sensor Controller
 - Can Run Autonomous From the Rest of the System
 - 16-Bit Architecture
 - 2KB of Ultralow-Leakage SRAM for Code and Data
- Efficient Code Size Architecture, Placing Drivers, Bluetooth® Low Energy Controller, IEEE 802.15.4 MAC, and Bootloader in ROM
- · RoHS-Compliant Packages
 - 4-mm x 4-mm RSM VQFN32 (10 GPIOs)
 - 5-mm x 5-mm RHB VQFN32 (15 GPIOs)
 - 7-mm × 7-mm RGZ VQFN48 (31 GPIOs)
- Peripherals
 - All Digital Peripheral Pins Can Be Routed to Any GPIO
 - Four General-Purpose Timer Modules (Eight 16-Bit or Four 32-Bit Timers, PWM Each)
 - 12-Bit ADC, 200-ksamples/s, 8-Channel Analog MUX
 - Continuous Time Comparator
 - Ultralow-Power Analog Comparator
 - Programmable Current Source
 - UART
 - 2× SSI (SPI, MICROWIRE, TI)
 - I2C
 - 12S
 - Real-Time Clock (RTC)
 - AES-128 Security Module
 - True Random Number Generator (TRNG)
 - 10, 15, or 31 GPIOs, Depending on Package Ontion
 - Support for Eight Capacitive-Sensing Buttons
 - Integrated Temperature Sensor
- External System
 - On-Chip internal DC-DC Converter

- Very Few External Components
- Seamless Integration With the SimpleLink™ CC2590 and CC2592 Range Extenders
- Pin Compatible With the SimpleLink CC13xx in 4-mm x 4-mm and 5-mm x 5-mm VQFN Packages
- Low Power
 - Wide Supply Voltage Range
 - Normal Operation: 1.8 to 3.8 V
 - External Regulator Mode: 1.7 to 1.95 V
 - Active-Mode RX: 5.9 mA
 - Active-Mode TX at 0 dBm: 6.1 mA
 - Active-Mode TX at +5 dBm: 9.1 mA
 - Active-Mode MCU: 61 μA/MHz
 - Active-Mode MCU: 48.5 CoreMark/mA
 - Active-Mode Sensor Controller: 8.2 µA/MHz
 - Standby: 1 μA (RTC Running and RAM/CPU Retention)
 - Shutdown: 100 nA (Wake Up on External Events)
- RF Section
 - 2.4-GHz RF Transceiver Compatible With Bluetooth Low Energy (BLE) 4.1 Specification and IEEE 802.15.4 PHY and MAC
 - Excellent Receiver Sensitivity (–97 dBm for BLE and –100 dBm for 802.15.4), Selectivity, and Blocking Performance
 - Link budget of 102 dB/105 dB (BLE/802.15.4)
 - Programmable Output Power up to +5 dBm
 - Single-Ended or Differential RF Interface
 - Suitable for Systems Targeting Compliance With Worldwide Radio Frequency Regulations
 - ETSI EN 300 328 (Europe)
 - EN 300 440 Class 2 (Europe)
 - FCC CFR47 Part 15 (US)
 - ARIB STD-T66 (Japan)
- Tools and Development Environment
 - Full-Feature and Low-Cost Development Kits
 - Multiple Reference Designs for Different RF Configurations
 - Packet Sniffer PC Software
 - Sensor Controller Studio
 - SmartRF™ Studio
 - SmartRF Flash Programmer 2
 - IAR Embedded Workbench[®] for ARM
 - Code Composer Studio™



1.2 Applications

- Consumer Electronics
- Mobile Phone Accessories
- Sports and Fitness Equipment
- HID Applications
- Home and Building Automation
- Lighting Control

- Alarm and Security
- Electronic Shelf Labeling
- Proximity Tags
- Medical
- Remote Controls
- · Wireless Sensor Networks

1.3 Description

The CC2650 device is a wireless MCU targeting *Bluetooth* Smart, ZigBee[®] and 6LoWPAN, and ZigBee RF4CE remote control applications.

The device is a member of the CC26xx family of cost-effective, ultralow power, 2.4-GHz RF devices. Very low active RF and MCU current and low-power mode current consumption provide excellent battery lifetime and allow for operation on small coin cell batteries and in energy-harvesting applications.

The CC2650 device contains a 32-bit ARM Cortex-M3 processor that runs at 48 MHz as the main processor and a rich peripheral feature set that includes a unique ultralow power sensor controller. This sensor controller is ideal for interfacing external sensors and for collecting analog and digital data autonomously while the rest of the system is in sleep mode. Thus, the CC2650 device is ideal for applications within a whole range of products including industrial, consumer electronics, and medical.

The *Bluetooth* Low Energy controller and the IEEE 802.15.4 MAC are embedded into ROM and are partly running on a separate ARM Cortex-M0 processor. This architecture improves overall system performance and power consumption and frees up flash memory for the application.

The Bluetooth Smart and ZigBee stacks are available free of charge from www.ti.com.

Device Information (1)

PART NUMBER	PACKAGE	BODY SIZE (NOM)
CC2650F128RGZ	VQFN (48)	7.00 mm × 7.00 mm
CC2650F128RHB	VQFN (32)	5.00 mm × 5.00 mm
CC2650F128RSM	VQFN (32)	4.00 mm × 4.00 mm

(1) For more information, see Section 9, Mechanical Packaging and Orderable Information.



1.4 Functional Block Diagram

Figure 1-1 shows a block diagram for the CC2650.

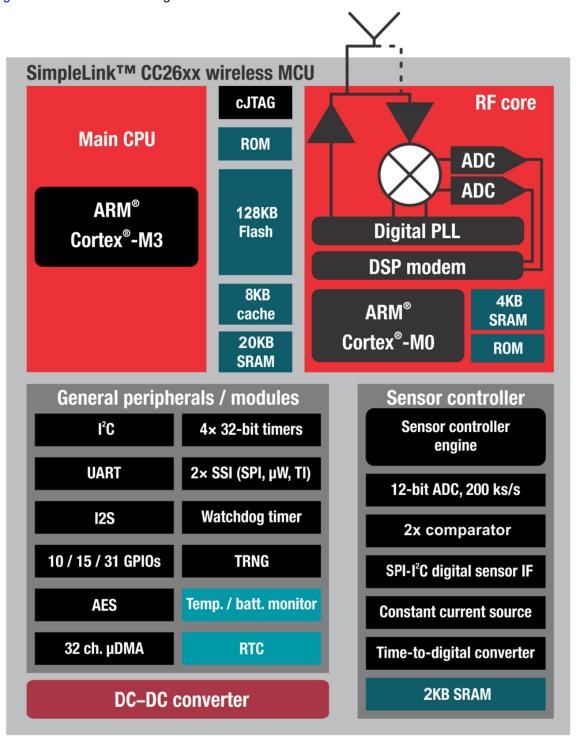


Figure 1-1. Block Diagram



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2 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Ch	hanges from February 21, 2015 to October 15, 2015	Page
	Removed RHB package option from CC2620	6
	Added motional inductance recommendation to the 24-MHz XOSC table	17
	Added VOH and VOL min and max values for 4-mA and 8-mA load	20
	Added min and max values for VIH and VIL	21
	Added SPI timing parameters	22
	Added IEEE 802.15.4 Sensitivity vs Channel Frequency	24
	Added BLE Sensitivity vs Channel Frequency	24
	Added RF Output Power vs Channel Frequency	24
	Added Receive Mode Current vs Supply Voltage (VDDS) graph	24
	• Changed SoC ADC ENOB vs Sampling Frequency (Input Frequency = FS / 10) graph	26
	Clarified Brown Out Detector status and functionality in the Power Modes table	33
	Added application circuit schematics and layout for 5XD and 4XS	36



Device Comparison

Table 3-1. Device Family Overview

DEVICE	PHY SUPPORT	FLASH (KB)	RAM (KB)	GPIO	PACKAGE ⁽¹⁾
CC2650F128xxx	Multi-Protocol (2)	128	20	31, 15, 10	RGZ, RHB, RSM
CC2640F128xxx	Bluetooth low energy	128	20	31, 15, 10	RGZ, RHB, RSM
CC2630F128xxx	IEEE 802.15.4 Zigbee(/6LoWPAN)	128	20	31, 15, 10	RGZ, RHB, RSM
CC2620F128xxx	IEEE 802.15.4 (RF4CE)	128	20	31, 10	RGZ, RSM

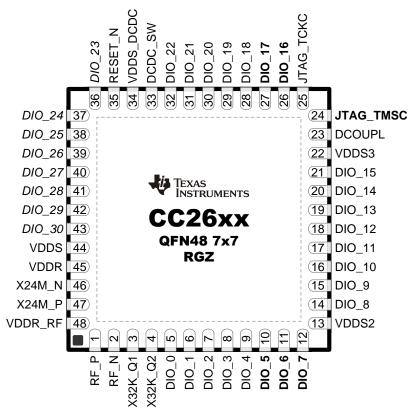
Package designator replaces the xxx in device name to form a complete device name, RGZ is 7-mm x 7-mm VQFN48, RHB is 5-mm x 5-mm VQFN32, and RSM is 4-mm x 4-mm VQFN32.

The CC2650 device supports all PHYs and can be reflashed to run all the supported standards.



4 Terminal Configuration and Functions

4.1 Pin Diagram – RGZ Package



Note: I/O pins marked in **bold** have high drive capabilities. I/O pins marked in *italics* have analog capabilities.

Figure 4-1. RGZ (7-mm × 7-mm) Pinout, 0.5-mm Pitch

4.2 Signal Descriptions – RGZ Package

Table 4-1. Signal Descriptions – RGZ Package

NAME	NO.	TYPE	DESCRIPTION
DCDC_SW	33	Power	Output from internal DC-DC ⁽¹⁾
DCOUPL	23	Power	1.27-V regulated digital-supply decoupling capacitor ⁽²⁾
DIO_0	5	Digital I/O	GPIO, Sensor Controller
DIO_1	6	Digital I/O	GPIO, Sensor Controller
DIO_2	7	Digital I/O	GPIO, Sensor Controller
DIO_3	8	Digital I/O	GPIO, Sensor Controller
DIO_4	9	Digital I/O	GPIO, Sensor Controller
DIO_5	10	Digital I/O	GPIO, Sensor Controller, high-drive capability
DIO_6	11	Digital I/O	GPIO, Sensor Controller, high-drive capability
DIO_7	12	Digital I/O	GPIO, Sensor Controller, high-drive capability
DIO_8	14	Digital I/O	GPIO
DIO_9	15	Digital I/O	GPIO
DIO_10	16	Digital I/O	GPIO
DIO_11	17	Digital I/O	GPIO

⁽¹⁾ See Section 8.2, technical reference manual for more details.

⁽²⁾ Do not supply external circuitry from this pin.



Table 4-1. Signal Descriptions – RGZ Package (continued)

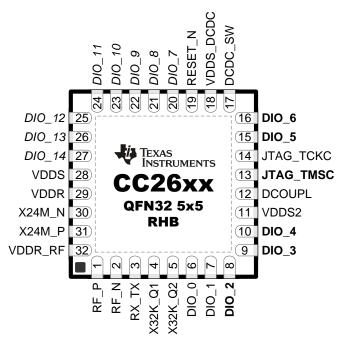
NAME	NO.	TYPE	DESCRIPTION
DIO_12	18	Digital I/O	GPIO
DIO_13	19	Digital I/O	GPIO
DIO_14	20	Digital I/O	GPIO
DIO_15	21	Digital I/O	GPIO
DIO_16	26	Digital I/O	GPIO, JTAG_TDO, high-drive capability
DIO_17	27	Digital I/O	GPIO, JTAG_TDI, high-drive capability
DIO_18	28	Digital I/O	GPIO
DIO_19	29	Digital I/O	GPIO
DIO_20	30	Digital I/O	GPIO
DIO_21	31	Digital I/O	GPIO
DIO_22	32	Digital I/O	GPIO
DIO_23	36	Digital/Analog I/O	GPIO, Sensor Controller, Analog
DIO_24	37	Digital/Analog I/O	GPIO, Sensor Controller, Analog
DIO_25	38	Digital/Analog I/O	GPIO, Sensor Controller, Analog
DIO_26	39	Digital/Analog I/O	GPIO, Sensor Controller, Analog
DIO_27	40	Digital/Analog I/O	GPIO, Sensor Controller, Analog
DIO_28	41	Digital/Analog I/O	GPIO, Sensor Controller, Analog
DIO_29	42	Digital/Analog I/O	GPIO, Sensor Controller, Analog
DIO_30	43	Digital/Analog I/O	GPIO, Sensor Controller, Analog
JTAG_TMSC	24	Digital I/O	JTAG TMSC, high-drive capability
JTAG_TCKC	25	Digital I/O	JTAG TCKC
RESET_N	35	Digital input	Reset, active-low. No internal pullup.
RF_P	1	RF I/O	Positive RF input signal to LNA during RX Positive RF output signal to PA during TX
RF_N	2	RF I/O	Negative RF input signal to LNA during RX Negative RF output signal to PA during TX
VDDR	45	Power	1.7-V to 1.95-V supply, typically connect to output of internal DC-DC (3) (2)
VDDR_RF	48	Power	1.7-V to 1.95-V supply, typically connect to output of internal DC-DC (4)(2)
VDDS	44	Power	1.8-V to 3.8-V main chip supply ⁽¹⁾
VDDS2	13	Power	1.8-V to 3.8-V DIO supply ⁽¹⁾
VDDS3	22	Power	1.8-V to 3.8-V DIO supply ⁽¹⁾
VDDS_DCDC	34	Power	1.8-V to 3.8-V DC-DC supply
X32K_Q1	3	Analog I/O	32-kHz crystal oscillator pin 1
X32K_Q2	4	Analog I/O	32-kHz crystal oscillator pin 2
X24M_N	46	Analog I/O	24-MHz crystal oscillator pin 1
X24M_P	47	Analog I/O	24-MHz crystal oscillator pin 2
EGP		Power	Ground – Exposed Ground Pad

⁽³⁾ If internal DC-DC is not used, this pin is supplied internally from the main LDO.

⁽⁴⁾ If internal DC-DC is not used, this pin must be connected to VDDR for supply from the main LDO.



4.3 Pin Diagram – RHB Package



Note: I/O pins marked in **bold** have high drive capabilities. I/O pins marked in *italics* have analog capabilities.

Figure 4-2. RHB (5-mm × 5-mm) Pinout, 0.5-mm Pitch

4.4 Signal Descriptions - RHB Package

Table 4-2. Signal Descriptions – RHB Package

NAME	NO.	TYPE	DESCRIPTION
DCDC_SW	17	Power	Output from internal DC-DC ⁽¹⁾
DCOUPL	12	Power	1.27-V regulated digital-supply decoupling ⁽²⁾
DIO_0	6	Digital I/O	GPIO, Sensor Controller
DIO_1	7	Digital I/O	GPIO, Sensor Controller
DIO_2	8	Digital I/O	GPIO, Sensor Controller, high-drive capability
DIO_3	9	Digital I/O	GPIO, Sensor Controller, high-drive capability
DIO_4	10	Digital I/O	GPIO, Sensor Controller, high-drive capability
DIO_5	15	Digital I/O	GPIO, High drive capability, JTAG_TDO
DIO_6	16	Digital I/O	GPIO, High drive capability, JTAG_TDI
DIO_7	20	Digital/Analog I/O	GPIO, Sensor Controller, Analog
DIO_8	21	Digital/Analog I/O	GPIO, Sensor Controller, Analog
DIO_9	22	Digital/Analog I/O	GPIO, Sensor Controller, Analog
DIO_10	23	Digital/Analog I/O	GPIO, Sensor Controller, Analog
DIO_11	24	Digital/Analog I/O	GPIO, Sensor Controller, Analog
DIO_12	25	Digital/Analog I/O	GPIO, Sensor Controller, Analog
DIO_13	26	Digital/Analog I/O	GPIO, Sensor Controller, Analog
DIO_14	27	Digital/Analog I/O	GPIO, Sensor Controller, Analog
JTAG_TMSC	13	Digital I/O	JTAG TMSC, high-drive capability
JTAG_TCKC	14	Digital I/O	JTAG TCKC
RESET_N	19	Digital input	Reset, active-low. No internal pullup.

⁽¹⁾ See Section 8.2, technical reference manual for more details.

⁽²⁾ Do not supply external circuitry from this pin.



Table 4-2. Signal Descriptions – RHB Package (continued)

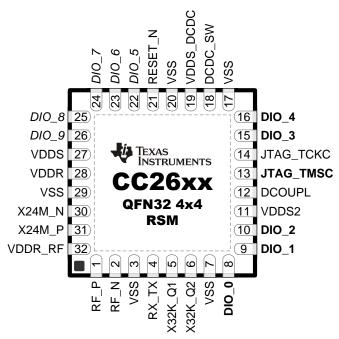
NAME	NO.	TYPE	DESCRIPTION	
RF_N	2	RF I/O	Negative RF input signal to LNA during RX Negative RF output signal to PA during TX	
RF_P	1	RF I/O	Positive RF input signal to LNA during RX Positive RF output signal to PA during TX	
RX_TX	3	RF I/O	Optional bias pin for the RF LNA	
VDDR	29	Power	1.7-V to 1.95-V supply, typically connect to output of internal DC-DC ⁽³⁾⁽²⁾	
VDDR_RF	32	Power	1.7-V to 1.95-V supply, typically connect to output of internal DC-DC ⁽⁴⁾⁽²⁾	
VDDS	28	Power	1.8-V to 3.8-V main chip supply ⁽¹⁾	
VDDS2	11	Power	1.8-V to 3.8-V GPIO supply ⁽¹⁾	
VDDS_DCDC	18	Power	1.8-V to 3.8-V DC-DC supply	
X32K_Q1	4	Analog I/O	32-kHz crystal oscillator pin 1	
X32K_Q2	5	Analog I/O	32-kHz crystal oscillator pin 2	
X24M_N	30	Analog I/O	24-MHz crystal oscillator pin 1	
X24M_P	31	Analog I/O	24-MHz crystal oscillator pin 2	
EGP		Power	Ground – Exposed Ground Pad	

⁽³⁾ If internal DC-DC is not used, this pin is supplied internally from the main LDO.

⁽⁴⁾ If internal DC-DC is not used, this pin must be connected to VDDR for supply from the main LDO.



4.5 Pin Diagram – RSM Package



Note: I/O pins marked in **bold** have high drive capabilities. I/O pins marked in *italics* have analog capabilities.

Figure 4-3. RSM (4-mm × 4-mm) Pinout, 0.4-mm Pitch

4.6 Signal Descriptions – RSM Package

Table 4-3. Signal Descriptions - RSM Package

NAME	NO.	TYPE	DESCRIPTION
DCDC_SW	18	Power	Output from internal DC-DC. ⁽¹⁾ . Tie to ground for external regulator mode (1.7-V to 1.95-V operation)
DCOUPL	12	Power	1.27-V regulated digital-supply decoupling capacitor (2)
DIO_0	8	Digital I/O	GPIO, Sensor Controller, high-drive capability
DIO_1	9	Digital I/O	GPIO, Sensor Controller, high-drive capability
DIO_2	10	Digital I/O	GPIO, Sensor Controller, high-drive capability
DIO_3	15	Digital I/O	GPIO, High drive capability, JTAG_TDO
DIO_4	16	Digital I/O	GPIO, High drive capability, JTAG_TDI
DIO_5	22	Digital/Analog I/O	GPIO, Sensor Controller, Analog
DIO_6	23	Digital/Analog I/O	GPIO, Sensor Controller, Analog
DIO_7	24	Digital/Analog I/O	GPIO, Sensor Controller, Analog
DIO_8	25	Digital/Analog I/O	GPIO, Sensor Controller, Analog
DIO_9	26	Digital/Analog I/O	GPIO, Sensor Controller, Analog
JTAG_TMSC	13	Digital I/O	JTAG TMSC
JTAG_TCKC	14	Digital I/O	JTAG TCKC
RESET_N	21	Digital Input	Reset, active-low. No internal pullup.
RF_N	2	RF I/O	Negative RF input signal to LNA during RX Negative RF output signal to PA during TX
RF_P	1	RF I/O	Positive RF input signal to LNA during RX Positive RF output signal to PA during TX
RX_TX	4	RF I/O	Optional bias pin for the RF LNA

⁽¹⁾ See Section 8.2, technical reference manual for more details.

⁽²⁾ Do not supply external circuitry from this pin.



Table 4-3. Signal Descriptions – RSM Package (continued)

NAME	NO.	TYPE	DESCRIPTION
VDDR	28	Power	1.7-V to 1.95-V supply, typically connect to output of internal DC-DC. (3)(2)
VDDR_RF	32	Power	1.7-V to 1.95-V supply, typically connect to output of internal DC-DC (4)(2)
VDDS	27	Power	1.8-V to 3.8-V main chip supply ⁽¹⁾
VDDS2	11	Power	1.8-V to 3.8-V GPIO supply ⁽¹⁾
VDDS_DCDC	19	Power	1.8-V to 3.8-V DC-DC supply. Tie to ground for external regulator mode (1.7-V to 1.95-V operation).
vss	3, 7, 17, 20, 29	Power	Ground
X32K_Q1	5	Analog I/O	32-kHz crystal oscillator pin 1
X32K_Q2	6	Analog I/O	32-kHz crystal oscillator pin 2
X24M_N	30	Analog I/O	24-MHz crystal oscillator pin 1
X24M_P	31	Analog I/O	24-MHz crystal oscillator pin 2
EGP	•	Power	Ground – Exposed Ground Pad

⁽³⁾ If internal DC-DC is not used, this pin is supplied internally from the main LDO.

⁽⁴⁾ If internal DC-DC is not used, this pin must be connected to VDDR for supply from the main LDO.



5 Specifications

5.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)(1)(2)

		MIN	MAX	UNIT
Supply voltage, VDDS ⁽³⁾ VDDR supplied by internal DC-DC regulator or internal GLDO		-0.3	4.1	V
Supply voltage, VDDS ⁽³⁾ and External regulator mode (VDDS and VDDR pins connected on PCB)		-0.3	2.25	V
Voltage on any digital pin (4)		-0.3	VDDS + 0.3, max 4.1	V
Voltage on crystal oscillator pins	, X32K_Q1, X32K_Q2, X24M_N and X24M_P	-0.3	VDDR + 0.3, max 2.25	V
	Voltage scaling enabled	-0.3	VDDS	
Voltage on ADC input (Vin)	Voltage scaling disabled, internal reference	-0.3	1.49	V
	Voltage scaling disabled, VDDS as reference	-0.3	VDDS / 2.9	
Input RF level		5	dBm	
T _{stg}	Storage temperature	-40	150	°C

⁽¹⁾ All voltage values are with respect to ground, unless otherwise noted.

5.2 ESD Ratings

				VALUE	UNIT
	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS001 (1)	All pins	±2500	
V _{ESD}	(ESD) performance	Charged device model (CDM), per JESD22-C101 ⁽²⁾	RF pins	±750	V
		Charged device model (CDM), per JESD22-C101	Non-RF pins	±750	

⁽¹⁾ JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

5.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	MAX	UNIT
Ambient temperature range		-40	85	ů
Operating supply voltage (VDDS and VDDR), external regulator mode	For operation in 1.8-V systems (VDDS and VDDR pins connected on PCB, internal DC-DC cannot be used)	1.7	1.95	V
Operating supply voltage (VDDS)	For operation in battery-powered and 3.3-V systems (internal DC-DC can be used to minimize power consumption)	1.8	3.8	V

²⁾ Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

⁽³⁾ VDDS2 and VDDS3 must be at the same potential as VDDS.

⁽⁴⁾ Including analog-capable DIO.

⁽²⁾ JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.



Power Consumption Summary

Measured on the TI CC2650EM-5XD reference design with T_c = 25°C, V_{DDS} = 3.0 V with internal DC-DC converter, unless otherwise noted.

	PARAMETER	TEST CONDITIONS	MIN TYP	MAX	UNIT
		Reset. RESET_N pin asserted or VDDS below Power-on-Reset threshold	100		nA
		Shutdown. No clocks running, no retention	150		
		Standby. With RTC, CPU, RAM and (partial) register retention. RCOSC_LF	1		
		Standby. With RTC, CPU, RAM and (partial) register retention. XOSC_LF	1.2		
		Standby. With Cache, RTC, CPU, RAM and (partial) register retention. RCOSC_LF	2.5		μΑ
I _{core}	Core current consumption	Standby. With Cache, RTC, CPU, RAM and (partial) register retention. XOSC_LF	2.7		
		Idle. Supply Systems and RAM powered.	550		
		Active. Core running CoreMark	1.45 mA + 31 μA/MHz		
		Radio RX ⁽¹⁾	5.9		
		Radio RX ⁽²⁾	6.1		
		Radio TX, 0-dBm output power ⁽¹⁾	6.1		mA
		Radio TX, 5-dBm output power ⁽²⁾	9.1		
Periph	eral Current Consumption (Ac	lds to core current I _{core} for each peripheral unit ac	tivated) ⁽³⁾		
	Peripheral power domain	Delta current with domain enabled	20		μΑ
	Serial power domain	Delta current with domain enabled	13		μΑ
	RF Core	Delta current with power domain enabled, clock enabled, RF core idle	237		μΑ
	μDMA	Delta current with clock enabled, module idle	130		μΑ
I _{peri}	Timers	Delta current with clock enabled, module idle	113		μΑ
	I ² C	Delta current with clock enabled, module idle	12		μΑ
	128	Delta current with clock enabled, module idle	36		μΑ
	SSI	Delta current with clock enabled, module idle	93		μΑ
	UART	Delta current with clock enabled, module idle	164		μA

- Single-ended RF mode is optimized for size and power consumption. Measured on CC2650EM-4XS.
- Differential RF mode is optimized for RF performance. Measured on CC2650EM-5XD. I_{peri} is not supported in Standby or Shutdown.

General Characteristics

Measured on the TI CC2650EM-5XD reference design with T_c = 25°C, V_{DDS} = 3.0 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT				
FLASH MEMORY									
Supported flash erase cycles before failure		100			k Cycles				
Flash page/sector erase current	Average delta current		12.6		mA				
Flash page/sector size			4		KB				
Flash write current	Average delta current, 4 bytes at a time		8.15		mA				
Flash page/sector erase time ⁽¹⁾			8		ms				
Flash write time ⁽¹⁾	4 bytes at a time		8		μs				

(1) This number is dependent on Flash aging and will increase over time and erase cycles.



1-Mbps GFSK (Bluetooth Low Energy) - RX 5.6

Measured on the TI CC2650EM-5XD reference design with $T_c = 25^{\circ}C$, $V_{DDS} = 3.0 \text{ V}$, $f_{RF} = 2440 \text{ MHz}$, unless otherwise noted.

Selectivity Simple-ended mode. Measured at 10°-2 (2650EM-4XS), at the SMA connector, BER = 10°-3 (2650EM-5XD) A denoted mode. Measured at the CC2650EM-5XD A denoted mode. Measured at the CC2650EM-4XS, at the SMA connector, BER = 10°-3 Difference between the incoming carrier frequency and the internally generated data rate A denoted mode. Measured at a rate A denoted mode. Measured and rate A denoted mode. Measured and rate A denoted mode. Measured and rate A denoted mode. Measured mode. Measured and rate A denoted mode. Measured mode. Measur	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
At the SMA connector, BER = 10 ⁻³ General Content of the SMA connector, BER =	Receiver sensitivity	Differential mode. Measured at the CC2650EM-5XD SMA connector, BER = 10 ⁻³		-97		dBm
SMA connector, BER = 10 ⁻³ Receiver saturation SMA connector, BER = 10 ⁻³ Single-ended mode. Measured on CC2650EM-4XS, at the SMA connector, BER = 10 ⁻³ Difference between the incoming carrier frequency and the internally generated data rate and the internal generated	Receiver sensitivity	Single-ended mode. Measured on CC2650EM-4XS, at the SMA connector, BER = 10^{-3}		-96		dBm
at the SMA connector, BER = 10 ⁻³ 0 0 0 0 0 0 0 0 0	Receiver saturation			4		dBm
And the internally generated carrier frequency y -390 390 872 Data rate error tolerance Difference between incoming data rate and the internally generated at a 1 MHz (1) generated at 22 MHz, generated at 34 (25 generated data data data data data data data	Receiver saturation			0		dBm
Delta rate error tolerance Internally generated data rate -750 750 ppr	Frequency error tolerance		-350		350	kHz
Co-channel rejection (1)	Data rate error tolerance		-750		750	ppm
Selectivity, ±1 MHz (¹) selectivity, ±2 MHz (¹) Selectivity, ±2 MHz (¹) Selectivity, ±2 MHz (¹) Selectivity, ±3 MHz (¹) Selectivity, ±3 MHz (¹) Selectivity, ±3 MHz (¹) Selectivity, ±4 MHz (¹) Selectivity, ±5 MHz or more (¹) Selectivity, ±5 MHz or more (¹) Selectivity, Image frequency (¹) Selectivity,	Co-channel rejection (1)	channel.		-6		dB
Selectivity, ±2 MHz (1) 22 MHz, BER = 10 ⁻³ Wanted signal at −67 dBm, modulated interferer at ±3 MHz, BER = 10 ⁻³ Wanted signal at −67 dBm, modulated interferer at ±4 MHz, BER = 10 ⁻³ Wanted signal at −67 dBm, modulated interferer at ±4 MHz, BER = 10 ⁻³ Selectivity, ±5 MHz or more(1) Wanted signal at −67 dBm, modulated interferer at ±5 MHz, BER = 10 ⁻³ Selectivity, lmage frequency(1) Selectivity, lmage frequency(1) BER = 10 ⁻³ Selectivity, lmage frequency(1) Wanted signal at −67 dBm, modulated interferer at ±1 MHz (1) EFR = 10 ⁻³ Selectivity, lmage frequency(1) BER = 10 ⁻³ Selectivity, lmage frequency(1) BER = 10 ⁻³ Selectivity, lmage frequency(1) BER = 10 ⁻³ 30 MHz to 2000 MHz ±1 MHz from image frequency, BER = 10 ⁻³ 31 / 26(2) dB Out-of-band blocking 2003 MHz to 2399 MHz Out-of-band blocking 2003 MHz to 2399 MHz Out-of-band blocking 3000 MHz to 2399 MHz Out-of-band blocking 3000 MHz to 12.75 GHz Intermodulation interferer at 2405 and 2408 MHz, −64 dBm. Two interferers at 2405 and 2408 MHz respectively, at the given power level Conducted measurement in a 50-Ω single-ended load. Suitable for systems targeting compliance with EN 300 328, EN 300 440 class 2, FCC CFR47, Part 15 and ARIB STD-T-66 RSSI dynamic range 70 dB	Selectivity, ±1 MHz (1)	±1 MHz,		7 / 3 ⁽²⁾		dB
Selectivity, ±3 MHz (¹) ±3 MHz, BER = 10⁻³ Selectivity, ±4 MHz (¹) Selectivity, ±4 MHz (¹) Selectivity, ±5 MHz or more(¹) Selectivity, ±5 MHz or more(¹) Wanted signal at −67 dBm, modulated interferer at ±5 MHz, BER = 10⁻³ Selectivity, lmage frequency(¹) Wanted signal at −67 dBm, modulated interferer at image frequency, BER = 10⁻³ Selectivity, lmage frequency Wanted signal at −67 dBm, modulated interferer at image frequency, BER = 10⁻³ Selectivity, lmage frequency #1 MHz (¹) Wanted signal at −67 dBm, modulated interferer at ±1 MHz (¹) #2 Mare signal at −67 dBm, modulated interferer at ±1 MHz (¹) #3 / 26(²) #3 / 26(²) #4 MB #4 MHz (¹) #5 MHz, 10 minage frequency, BER = 10⁻³ #5 MHz, 10 minage frequency, BER = 10⁻³ #6 Bin #6 Dut-of-band blocking (³) #6 Bin #6 Dut-of-band blocking	Selectivity, ±2 MHz ⁽¹⁾	±2 MHz,		34 / 25 ⁽²⁾		dB
Selectivity, ±4 MHz (1) ±4 MHz, BER = 10 ⁻³ 42 / 29 ⁽²⁾ dB Selectivity, ±5 MHz or more (1) ½5 MHz, BER = 10 ⁻³ 32 dB Selectivity, Image frequency (1) Wanted signal at −67 dBm, modulated interferer at image frequency, BER = 10 ⁻³ 25 dB Selectivity, Image frequency ±1 MHz (1) Wanted signal at −67 dBm, modulated interferer at ±1 MHz from image frequency, BER = 10 ⁻³ 3 / 26 ⁽²⁾ dB Out-of-band blocking (3) 30 MHz to 2000 MHz −20 dBn Out-of-band blocking 2003 MHz to 2399 MHz −5 dBn Out-of-band blocking 2484 MHz to 2997 MHz −8 dBn Out-of-band blocking 3000 MHz to 12.75 GHz −8 dBn Intermodulation Wanted signal at 2402 MHz, −64 dBm. Two interferers at 2405 and 2408 MHz respectively, at the given power level −34 dBn Spurious emissions, 30 to 1000 MHz Conducted measurement in a 50 Ω single-ended load. Suitable for systems targeting compliance with EN 300 328, Bn 300 440 class 2, FCC CFR47, Part 15 and ARIB STD-T-66 −71 dBn RSSI dynamic range To ARIB STD-T-66 To ARIB STD-T-66 ARIB STD-T-66 To ARIB STD-T-66 To ARIB STD-T-66	Selectivity, ±3 MHz ⁽¹⁾	±3 MHz,		38 / 26 ⁽²⁾		dB
#5 MHz, BER = 10 ⁻³ Wanted signal at -67 dBm, modulated interferer at image frequency, ±1 MHz (1) Selectivity, Image frequency ±1 MHz (1) Wanted signal at -67 dBm, modulated interferer at ±1 MHz from image frequency, BER = 10 ⁻³ Selectivity, Image frequency ±1 MHz from image frequency, BER = 10 ⁻³ 3 / 26 ⁽²⁾ dB Out-of-band blocking (3) 30 MHz to 2000 MHz 2003 MHz to 2399 MHz -20 dBn Out-of-band blocking 2048 MHz to 2997 MHz -8 dBn Out-of-band blocking 3000 MHz to 12.75 GHz Wanted signal at 2402 MHz, -64 dBm. Two interferers at 2405 and 2408 MHz respectively, at the given power level Conducted measurement in a 50-Ω single-ended load. Suitable for systems targeting compliance with EN 300 328, EN 300 440 class 2, FCC CFR47, Part 15 and ARIB STD-T-66 RSSI dynamic range RSSI dynamic range	Selectivity, ±4 MHz ⁽¹⁾	±4 MHz,		42 / 29 ⁽²⁾		dB
Selectivity, Image frequency (1) image frequency, BER = 10 ⁻³	Selectivity, ±5 MHz or more ⁽¹⁾	Wanted signal at –67 dBm, modulated interferer at ≥ ±5 MHz, BER = 10 ⁻³		32		dB
±1 MHz from image frequency, BER = 10 ⁻³ Out-of-band blocking (3) 30 MHz to 2000 MHz -20 dBm Out-of-band blocking 2003 MHz to 2399 MHz -5 dBm Out-of-band blocking 2484 MHz to 2997 MHz -8 dBm Out-of-band blocking 3000 MHz to 12.75 GHz -8 dBm Intermodulation was a 2402 MHz, -64 dBm. Two interferers at 2405 and 2408 MHz respectively, at the given power level Conducted measurement in a 50-Ω single-ended load. Suitable for systems targeting compliance with EN 300 328, EN 300 440 class 2, FCC CFR47, Part 15 and ARIB STD-T-66 Conducted measurement in a 50 Ω single-ended load. Suitable for systems targeting compliance with EN 300 328, EN 300 440 class 2, FCC CFR47, Part 15 and ARIB STD-T-66 RSSI dynamic range 70 dBm	Selectivity, Image frequency ⁽¹⁾	image frequency.		25		dB
Out-of-band blocking2003 MHz to 2399 MHz-5dBmOut-of-band blocking2484 MHz to 2997 MHz-8dBmOut-of-band blocking3000 MHz to 12.75 GHz-8dBmIntermodulationWanted signal at 2402 MHz, -64 dBm. Two interferers at 2405 and 2408 MHz respectively, at the given power level-34dBmSpurious emissions, 30 to 1000 MHzConducted measurement in a 50-Ω single-ended load. Suitable for systems targeting compliance with EN 300 328, EN 300 440 class 2, FCC CFR47, Part 15 and ARIB STD-T-66-71dBmSpurious emissions, 1 to 12.75 GHzConducted measurement in a 50 Ω single-ended load. Suitable for systems targeting compliance with EN 300 328, EN 300 440 class 2, FCC CFR47, Part 15 and ARIB STD-T-66-62dBmRSSI dynamic range70dB	Selectivity, Image frequency ±1 MHz ⁽¹⁾	Wanted signal at -67 dBm, modulated interferer at ± 1 MHz from image frequency, BER = 10^{-3}		3 / 26 ⁽²⁾		dB
Out-of-band blocking2484 MHz to 2997 MHz-8dBmOut-of-band blocking3000 MHz to 12.75 GHz-8dBmIntermodulationWanted signal at 2402 MHz, -64 dBm. Two interferers at 2405 and 2408 MHz respectively, at the given power level-34dBmSpurious emissions, 30 to 1000 MHzConducted measurement in a 50-Ω single-ended load. Suitable for systems targeting compliance with EN 300 328, EN 300 440 class 2, FCC CFR47, Part 15 and ARIB STD-T-66-71dBmSpurious emissions, 1 to 12.75 GHzConducted measurement in a 50 Ω single-ended load. Suitable for systems targeting compliance with EN 300 328, EN 300 440 class 2, FCC CFR47, Part 15 and ARIB STD-T-66-62dBmRSSI dynamic range70dB	Out-of-band blocking (3)	30 MHz to 2000 MHz		-20		dBm
Out-of-band blocking 3000 MHz to 12.75 GHz Wanted signal at 2402 MHz, -64 dBm. Two interferers at 2405 and 2408 MHz respectively, at the given power level Conducted measurement in a 50-Ω single-ended load. Suitable for systems targeting compliance with EN 300 328, EN 300 440 class 2, FCC CFR47, Part 15 and ARIB STD-T-66 Spurious emissions, 1 to 12.75 GHz Conducted measurement in a 50 Ω single-ended load. Suitable for systems targeting compliance with EN 300 328, EN 300 440 class 2, FCC CFR47, Part 15 and ARIB STD-T-66 RSSI dynamic range 70 dB	Out-of-band blocking	2003 MHz to 2399 MHz		- 5		dBm
IntermodulationWanted signal at 2402 MHz, -64 dBm. Two interferers at 2405 and 2408 MHz respectively, at the given power level-34dBmSpurious emissions, 30 to 1000 MHzConducted measurement in a 50-Ω single-ended load. Suitable for systems targeting compliance with EN 300 328, EN 300 440 class 2, FCC CFR47, Part 15 and ARIB STD-T-66-71dBmSpurious emissions, 1 to 12.75 GHzConducted measurement in a 50 Ω single-ended load. Suitable for systems targeting compliance with EN 300 328, EN 300 440 class 2, FCC CFR47, Part 15 and ARIB STD-T-66-62dBmRSSI dynamic range70dB	Out-of-band blocking	2484 MHz to 2997 MHz		-8		dBm
Intermodulation interferers at 2405 and 2408 MHz respectively, at the given power level Conducted measurement in a 50-Ω single-ended load. Suitable for systems targeting compliance with EN 300 328, EN 300 440 class 2, FCC CFR47, Part 15 and ARIB STD-T-66 Spurious emissions, 1 Conducted measurement in a 50 Ω single-ended load. Suitable for systems targeting compliance with EN 300 328, EN 300 440 class 2, FCC CFR47, Part 15 and ARIB STD-T-66 RSSI dynamic range To dBm	Out-of-band blocking	3000 MHz to 12.75 GHz		-8		dBm
Spurious emissions, 30 to 1000 MHzload. Suitable for systems targeting compliance with EN 300 328, EN 300 440 class 2, FCC CFR47, Part 15 and ARIB STD-T-66-71dBmSpurious emissions, 1 to 12.75 GHzConducted measurement in a 50 Ω single-ended 	Intermodulation	interferers at 2405 and 2408 MHz respectively, at		-34		dBm
Spurious emissions, 1 to 12.75 GHz load. Suitable for systems targeting compliance with EN 300 328, EN 300 440 class 2, FCC CFR47, Part 15 and ARIB STD-T-66 RSSI dynamic range 70 dB	Spurious emissions, 30 to 1000 MHz	load. Suitable for systems targeting compliance with EN 300 328, EN 300 440 class 2, FCC CFR47, Part		–7 1		dBm
	Spurious emissions, 1 to 12.75 GHz	load. Suitable for systems targeting compliance with EN 300 328, EN 300 440 class 2, FCC CFR47, Part		-62		dBm
RSSI accuracy +4 dB	RSSI dynamic range			70		dB
ub ub	RSSI accuracy			±4		dB

⁽¹⁾ Numbers given as I/C dB.

X / Y, where X is +N MHz and Y is -N MHz.

Excluding one exception at F_{wanted} / 2, per *Bluetooth* Specification.



5.7 1-Mbps GFSK (Bluetooth Low Energy) – TX

Measured on the TI CC2650EM-5XD reference design with $T_c = 25$ °C, $V_{DDS} = 3.0$ V, $f_{RF} = 2440$ MHz, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Output power, highest setting	Differential mode, delivered to a single-ended 50- $\!\Omega$ load through a balun		5		dBm
Output power, highest setting	Measured on CC2650EM-4XS, delivered to a single-ended 50- Ω load		2		dBm
Output power, lowest setting	Delivered to a single-ended 50-Ω load through a balun		-21		dBm
	f < 1 GHz, outside restricted bands		-43		dBm
Spurious emission conducted	f < 1 GHz, restricted bands ETSI		-65		dBm
measurement ⁽¹⁾	f < 1 GHz, restricted bands FCC		-76		dBm
	f > 1 GHz, including harmonics		-46		dBm

⁽¹⁾ Suitable for systems targeting compliance with worldwide radio-frequency regulations ETSI EN 300 328 and EN 300 440 Class 2 (Europe), FCC CFR47 Part 15 (US), and ARIB STD-T66 (Japan).

5.8 IEEE 802.15.4 (Offset Q-PSK DSSS, 250 kbps) - RX

Measured on the TI CC2650EM-5XD reference design with T_c = 25°C, V_{DDS} = 3.0 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Receiver sensitivity	Differential mode. Measured at the CC2650EM-5XD SMA connector, PER = 1%		-100		dBm
Receiver sensitivity	Single-ended mode. Measured on CC2650EM-4XS, at the SMA connector, PER = 1%		-97		dBm
Receiver saturation	Measured at the CC2650EM-5XD SMA connector, PER = 1%		+4		dBm
Adjacent channel rejection	Wanted signal at –82 dBm, modulated interferer at ±5 MHz, PER = 1%		39		dB
Alternate channel rejection	Wanted signal at –82 dBm, modulated interferer at ±10 MHz, PER = 1%		52		dB
Channel rejection, ±15 MHz or more	Wanted signal at –82 dBm, undesired signal is IEEE 802.15.4 modulated channel, stepped through all channels 2405 to 2480 MHz, PER = 1%		57		dB
Blocking and desensitization, 5 MHz from upper band edge	Wanted signal at –97 dBm (3 dB above the sensitivity level), CW jammer, PER = 1%		64		dB
Blocking and desensitization, 10 MHz from upper band edge	Wanted signal at –97 dBm (3 dB above the sensitivity level), CW jammer, PER = 1%		64		dB
Blocking and desensitization, 20 MHz from upper band edge	Wanted signal at –97 dBm (3 dB above the sensitivity level), CW jammer, PER = 1%		65		dB
Blocking and desensitization, 50 MHz from upper band edge	Wanted signal at –97 dBm (3 dB above the sensitivity level), CW jammer, PER = 1%		68		dB
Blocking and desensitization, –5 MHz from lower band edge	Wanted signal at –97 dBm (3 dB above the sensitivity level), CW jammer, PER = 1%		63		dB
Blocking and desensitization, –10 MHz from lower band edge	Wanted signal at –97 dBm (3 dB above the sensitivity level), CW jammer, PER = 1%		63		dB
Blocking and desensitization, –20 MHz from lower band edge	Wanted signal at –97 dBm (3 dB above the sensitivity level), CW jammer, PER = 1%		65		dB
Blocking and desensitization, –50 MHz from lower band edge	Wanted signal at –97 dBm (3 dB above the sensitivity level), CW jammer, PER = 1%		67		dB
Spurious emissions, 30 MHz to 1000 MHz	Conducted measurement in a 50 Ω single-ended load. Suitable for systems targeting compliance with EN 300 328, EN 300 440 class 2, FCC CFR47, Part 15 and ARIB STD-T-66		-71		dBm
Spurious emissions, 1 GHz to 12.75 GHz	Conducted measurement in a 50 Ω single-ended load. Suitable for systems targeting compliance with EN 300 328, EN 300 440 class 2, FCC CFR47, Part 15 and ARIB STD-T-66		-62		dBm
Frequency error tolerance	Difference between the incoming carrier frequency and the internally generated carrier frequency		>200		ppm



IEEE 802.15.4 (Offset Q-PSK DSSS, 250 kbps) - RX (continued)

Measured on the TI CC2650EM-5XD reference design with $T_c = 25$ °C, $V_{DDS} = 3.0$ V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Symbol rate error tolerance	Difference between incoming symbol rate and the internally generated symbol rate		>1000		ppm
RSSI dynamic range			100		dB
RSSI accuracy			±4		dB

5.9 IEEE 802.15.4 (Offset Q-PSK DSSS, 250 kbps) – TX

Measured on the TI CC2650EM-5XD reference design with T_c = 25°C, V_{DDS} = 3.0 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Output power, highest setting	Delivered to a single-ended $50-\Omega$ load through a balun		5		dBm
Output power, highest setting	Measured on CC2650EM-4XS, delivered to a single-ended 50- Ω load			dBm	
Output power, lowest setting	Delivered to a single-ended 50-Ω load through a balun		-21		dBm
Error vector magnitude	At maximum output power	2%			
	f < 1 GHz, outside restricted bands		-43		
	f < 1 GHz, restricted bands ETSI		-65		alD.ee
Spurious emission conducted	f < 1 GHz, restricted bands FCC		-76		dBm
measurement	f > 1 GHz, including harmonics		-46		
	Suitable for systems targeting compliance with worldwide radio-frequency regulations ETSI EN 300 328 and EN 300 440 Class 2 (Europe), FCC CFR47 Part 15 (US), and ARIB STD-T66 (Japan)				

5.10 24-MHz Crystal Oscillator (XOSC_HF)

Measured on the TI CC2650EM-5XD reference design with $T_c = 25$ °C, $V_{DDS} = 3.0$ V, unless otherwise noted. (1)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
ESR Equivalent series resistance			20	60	Ω
L _M Motional inductance	Relates to load capacitance (C _L in Farads)		$< 1.6 \times 10^{-24} / C_L^2$		н
C _L Crystal load capacitance		5		9	pF
Crystal frequency			24		MHz
Crystal frequency tolerance (2)		-40		40	ppm
Start-up time ⁽³⁾			150		μs

⁽¹⁾ Probing or otherwise stopping the XTAL while the DC-DC converter is enabled may cause permanent damage to the device.

5.11 32.768-kHz Crystal Oscillator (XOSC_LF)

Measured on the TI CC2650EM-5XD reference design with $T_c = 25^{\circ}\text{C}$, $V_{DDS} = 3.0 \text{ V}$, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Crystal frequency			32.768		kHz
Crystal frequency tolerance, <i>Bluetooth</i> low-energy applications (1)		-500		500	ppm
ESR Equivalent series resistance			30	100	kΩ
C _L Crystal load capacitance		6		12	pF

Includes initial tolerance of the crystal, drift over temperature, ageing and frequency pulling due to incorrect load capacitance. As per Bluetooth and IEEE 802.15.4 specification.

⁽²⁾ Includes initial tolerance of the crystal, drift over temperature, ageing and frequency pulling due to incorrect load capacitance. As per Bluetooth and IEEE 802.15.4 specification.

⁽³⁾ Kick-started based on a temperature and aging compensated RCOSC_HF using precharge injection.



5.12 48-MHz RC Oscillator (RCOSC_HF)

Measured on the TI CC2650EM-5XD reference design with $T_c = 25$ °C, $V_{DDS} = 3.0$ V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Frequency			48		MHz
Uncalibrated frequency accuracy			±1%		
Calibrated frequency accuracy ⁽¹⁾			±0.25%		
Start-up time			5		μs

⁽¹⁾ Accuracy relative to the calibration source (XOSC_HF).

5.13 32-kHz RC Oscillator (RCOSC_LF)

Measured on the TI CC2650EM-5XD reference design with $T_c = 25$ °C, $V_{DDS} = 3.0$ V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Calibrated frequency			32.8		kHz
Temperature coefficient			50		ppm/°C

5.14 ADC Characteristics

 $T_c = 25$ °C, $V_{DDS} = 3.0$ V and voltage scaling enabled, unless otherwise noted. (1)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Input voltage range		0		VDDS	V
	Resolution			12		Bits
	Sample rate				200	ksps
	Offset	Internal 4.3-V equivalent reference (2)		2		LSB
	Gain error	Internal 4.3-V equivalent reference ⁽²⁾		2.4		LSB
DNL ⁽³⁾	Differential nonlinearity			>–1		LSB
INL ⁽⁴⁾	Integral nonlinearity			±3		LSB
		Internal 4.3-V equivalent reference (2), 200 ksps, 9.6-kHz input tone		9.8		
ENOB	Effective number of bits	VDDS as reference, 200 ksps, 9.6-kHz input tone		10		Bits
		Internal 1.44-V reference, voltage scaling disabled, 32 samples average, 200 ksps, 300-Hz input tone		11.1		
		Internal 4.3-V equivalent reference (2), 200 ksps, 9.6-kHz input tone		-65		
THD	Total harmonic distortion	VDDS as reference, 200 ksps, 9.6-kHz input tone		-69		dB
		Internal 1.44-V reference, voltage scaling disabled, 32 samples average, 200 ksps, 300-Hz input tone		-71		
	Signal-to-noise	Internal 4.3-V equivalent reference (2), 200 ksps, 9.6-kHz input tone		60		
SINAD, SNDR	and	VDDS as reference, 200 ksps, 9.6-kHz input tone		63		dB
ONDIK	Distortion ratio	Internal 1.44-V reference, voltage scaling disabled, 32 samples average, 200 ksps, 300-Hz input tone		69		
		Internal 4.3-V equivalent reference ⁽²⁾ , 200 ksps, 9.6-kHz input tone		67		
SFDR	Spurious-free dynamic range	VDDS as reference, 200 ksps, 9.6-kHz input tone		72		dB
	Tange	Internal 1.44-V reference, voltage scaling disabled, 32 samples average, 200 ksps, 300-Hz input tone		73		
	Conversion time	Serial conversion, time-to-output, 24-MHz clock		50		clock- cycles
	Current consumption	Internal 4.3-V equivalent reference (2)		0.66		mA
	Current consumption	VDDS as reference		0.75		mA

Using IEEE Std 1241™-2010 for terminology and test methods.

Input signal scaled down internally before conversion, as if voltage range was 0 to 4.3 V.

No missing codes. Positive DNL typically varies from +0.3 to +3.5, depending on device (see Figure 5-24).

⁽⁴⁾ For a typical example, see Figure 5-25.



ADC Characteristics (continued)

 $T_c = 25$ °C, $V_{DDS} = 3.0$ V and voltage scaling enabled, unless otherwise noted. (1)

PARAMETER	TEST CONDITIONS	MIN TYP	MAX	UNIT
Reference voltage	Equivalent fixed internal reference (input voltage scaling enabled)	4.3 ⁽²⁾⁽⁵⁾		V
Reference voltage	Fixed internal reference (input voltage scaling disabled)	1.44 ±1%		V
Reference voltage	VDDS as reference (Also known as <i>RELATIVE</i>) (input voltage scaling enabled)	VDDS		V
Reference voltage	VDDS as reference (Also known as <i>RELATIVE</i>) (input voltage scaling disabled)	VDDS / 2.82 ⁽⁵⁾		V
Input Impedance	200 ksps, voltage scaling enabled. Capacitive input, Input impedance depends on sampling frequency and sampling time	>1		ΜΩ

⁽⁵⁾ Applied voltage must be within absolute maximum ratings (Section 5.1) at all times.

5.15 Temperature Sensor

Measured on the TI CC2650EM-5XD reference design with T_c = 25°C, V_{DDS} = 3.0 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Resolution			4		°C
Range		-40	·	85	°C
Accuracy			±5		°C
Supply voltage coefficient ⁽¹⁾			3.2		°C/V

⁽¹⁾ Automatically compensated when using supplied driver libraries.

5.16 Battery Monitor

Measured on the TI CC2650EM-5XD reference design with $T_c = 25$ °C, $V_{DDS} = 3.0$ V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Resolution			50		mV
Range		1.8	·	3.8	V
Accuracy			13		mV

5.17 Continuous Time Comparator

 $T_c = 25$ °C, $V_{DDS} = 3.0$ V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Input voltage range		0		VDDS	V
External reference voltage		0		VDDS	V
Internal reference voltage	DCOUPL as reference		1.27		V
Offset			3		mV
Hysteresis			<2		mV
Decision time	Step from -10 mV to 10 mV		0.72		μs
Current consumption when enabled ⁽¹⁾			8.6		μΑ

⁽¹⁾ Additionally, the bias module must be enabled when running in standby mode.



5.18 Low-Power Clocked Comparator

 $T_c = 25$ °C, $V_{DDS} = 3.0$ V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Input voltage range		0		VDDS	V
Clock frequency			32		kHz
Internal reference voltage, VDDS / 2			1.49 – 1.51		V
Internal reference voltage, VDDS / 3			1.01 – 1.03		V
Internal reference voltage, VDDS / 4			0.78 - 0.79		V
Internal reference voltage, DCOUPL / 1			1.25 – 1.28		V
Internal reference voltage, DCOUPL / 2			0.63 - 0.65		V
Internal reference voltage, DCOUPL / 3			0.42 - 0.44		V
Internal reference voltage, DCOUPL / 4			0.33 - 0.34		V
Offset			<2		mV
Hysteresis			<5		mV
Decision time	Step from -50 mV to 50 mV		<1		clock-cycle
Current consumption when enabled			362		nA

5.19 Programmable Current Source

 $T_c = 25$ °C, $V_{DDS} = 3.0$ V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Current source programmable output range			0.25 - 20		μΑ
Resolution			0.25		μΑ
Current consumption ⁽¹⁾	Including current source at maximum programmable output		23		μΑ

⁽¹⁾ Additionally, the bias module must be enabled when running in standby mode.

5.20 DC Characteristics

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
	$T_A = 25^{\circ}C, V_{DDS} = 1.8 V$	·		·	
GPIO VOH at 8-mA load	IOCURR = 2, high-drive GPIOs only	1.32	1.54		V
GPIO VOL at 8-mA load	IOCURR = 2, high-drive GPIOs only		0.26	0.32	V
GPIO VOH at 4-mA load	IOCURR = 1	1.32	1.58		V
GPIO VOL at 4-mA load	IOCURR = 1		0.21	0.32	V
GPIO pullup current	Input mode, pullup enabled, Vpad = 0 V		71.7		μΑ
GPIO pulldown current	Input mode, pulldown enabled, Vpad = VDDS		21.1		μΑ
GPIO high/low input transition, no hysteresis	IH = 0, transition between reading 0 and reading 1		0.88		V
GPIO low-to-high input transition, with hysteresis	IH = 1, transition voltage for input read as $0 \rightarrow 1$		1.07		V
GPIO high-to-low input transition, with hysteresis	IH = 1, transition voltage for input read as $1 \rightarrow 0$		0.74		V
GPIO input hysteresis	IH = 1, difference between $0 \rightarrow 1$ and $1 \rightarrow 0$ points		0.33		V



DC Characteristics (continued)

PARAMETER	TEST CONDITIONS	MIN TYP MAX	UNIT
	$T_A = 25^{\circ}C, V_{DDS} = 3.0 V$	•	•
GPIO VOH at 8-mA load	IOCURR = 2, high-drive GPIOs only	2.68	V
GPIO VOL at 8-mA load	IOCURR = 2, high-drive GPIOs only	0.33	V
GPIO VOH at 4-mA load	IOCURR = 1	2.72	V
GPIO VOL at 4-mA load	IOCURR = 1	0.28	V
	T _A = 25°C, V _{DDS} = 3.8 V	•	
GPIO pullup current	Input mode, pullup enabled, Vpad = 0 V	277	μΑ
GPIO pulldown current	Input mode, pulldown enabled, Vpad = VDDS	113	μΑ
GPIO high/low input transition, no hysteresis	IH = 0, transition between reading 0 and reading 1	1.67	V
GPIO low-to-high input transition, with hysteresis	IH = 1, transition voltage for input read as $0 \rightarrow 1$	1.94	V
GPIO high-to-low input transition, with hysteresis	IH = 1, transition voltage for input read as $1 \rightarrow 0$	1.54	V
GPIO input hysteresis	IH = 1, difference between $0 \rightarrow 1$ and $1 \rightarrow 0$ points	0.4	V
	T _A = 25°C		
VIH	Lowest GPIO input voltage reliably interpreted as a «High»	0.8	VDDS ⁽¹⁾
VIL	Highest GPIO input voltage reliably interpreted as a «Low»	0.2	VDDS ⁽¹⁾

⁽¹⁾ Each GPIO is referenced to a specific VDDS pin. See the technical reference manual listed in Section 8.2 for more details.

5.21 Thermal Characteristics

NAME	DESCRIPTION	RSM (°C/W) ⁽¹⁾ (2)	RHB (°C/W) ⁽¹⁾ (2)	RGZ (°C/W) ⁽¹⁾ (2)
$R\theta_{JA}$	Junction-to-ambient thermal resistance	36.9	32.8	29.6
$R\theta_{JC(top)}$	Junction-to-case (top) thermal resistance	30.3	24.0	15.7
$R\theta_{JB}$	Junction-to-board thermal resistance	7.6	6.8	6.2
Psi _{JT}	Junction-to-top characterization parameter	0.4	0.3	0.3
Psi _{JB}	Junction-to-board characterization parameter	7.4	6.8	6.2
Rθ _{JC(bot)}	Junction-to-case (bottom) thermal resistance	2.1	1.9	1.9

[°]C/W = degrees Celsius per watt.

- JESD51-2, Integrated Circuits Thermal Test Method Environmental Conditions Natural Convection (Still Air).
- JESD51-3, Low Effective Thermal Conductivity Test Board for Leaded Surface Mount Packages.
- JESD51-7, High Effective Thermal Conductivity Test Board for Leaded Surface Mount Packages.
- JESD51-9, Test Boards for Area Array Surface Mount Package Thermal Measurements.

Power dissipation of 2 W and an ambient temperature of 70°C is assumed.

⁽²⁾ These values are based on a JEDEC-defined 2S2P system (with the exception of the Theta JC [Rθ_{JC}] value, which is based on a JEDEC-defined 1S0P system) and will change based on environment as well as application. For more information, see these EIA/JEDEC standards:



5.22 Timing Requirements

			MIN	NOM	MAX	UNIT
Rising supply-volta	ge slew rate		0		100	mV/μs
Falling supply-volta	age slew rate		0		20	mV/μs
Falling supply-volta	age slew rate, with low-power flash	settings ⁽¹⁾			3	mV/μs
Positive temperature gradient in standby (2) No limitation for negative temperature gradient, or outside standby mode				5	°C/s	
CONTROL INPUT	AC CHARACTERISTICS(3)				,	
RESET_N low dura	ation		1			μs
SYNCHRONOUS	SERIAL INTERFACE (SSI) (4)					
S1 (SLAVE) (5)	t _{clk_per}	SSICIk period	12		65024	system clocks
S2 ⁽⁵⁾	t _{clk_high}	SSICIk high time		0.5		t _{clk_per}
S3 ⁽⁵⁾	t _{clk_low}	SSICIk low time		0.5		t _{clk_per}

- (1) For smaller coin cell batteries, with high worst-case end-of-life equivalent source resistance, a 22-μF VDDS input capacitor (see Figure 7-1) must be used to ensure compliance with this slew rate.
- (2) Applications using RCOSC_LF as sleep timer must also consider the drift in frequency caused by a change in temperature (see Section 5.13).
- (3) $T_A = -40$ °C to 85°C, $V_{DDS} = 1.7$ V to 3.8 V, unless otherwise noted.
- (4) T_c = 25°C, V_{DDS} = 3.0 V, unless otherwise noted. Device operating as SLAVE. For SSI MASTER operation, see Section 5.23.
- (5) Refer to SSI timing diagrams Figure 5-1, Figure 5-2, and Figure 5-3.

5.23 Switching Characteristics

Measured on the TI CC2650EM-5XD reference design with T_c = 25°C, V_{DDS} = 3.0 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
WAKEUP AND TIMING					
Idle → Active			14		μs
$Standby \rightarrow Active$			151		μs
Shutdown → Active			1015		μs
SYNCHRONOUS SERIAL INTERFACE (SSI)	(1)				
S1 (TX only) $^{(2)}$ t_{clk_per} (SSICIk period)	One-way communication to SLAVE	4		65024	system clocks
S1 (TX and RX) ⁽²⁾ t _{clk_per} (SSIClk period)	Normal duplex operation	8		65024	system clocks
S2 (2) t _{clk_high} (SSIClk high time)			0.5		t _{clk_per}
S3 (2) t _{clk_low} (SSIClk low time)			0.5		t _{clk_per}

- (1) Device operating as MASTER. For SSI SLAVE operation, see Section 5.22.
- (2) Refer to SSI timing diagrams Figure 5-1, Figure 5-2, and Figure 5-3.

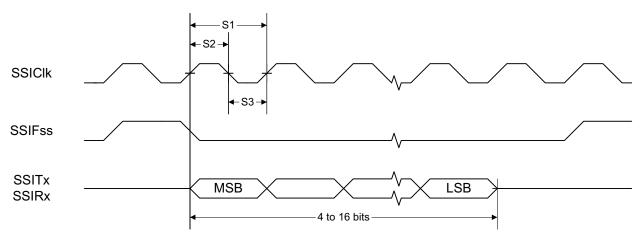


Figure 5-1. SSI Timing for TI Frame Format (FRF = 01), Single Transfer Timing Measurement



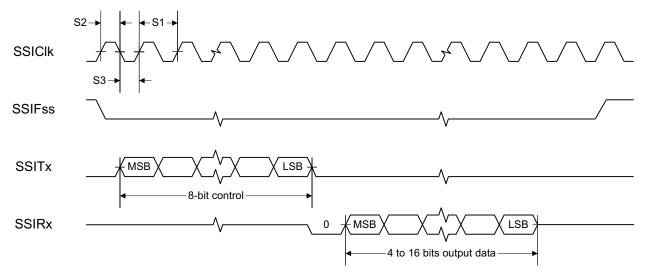


Figure 5-2. SSI Timing for MICROWIRE Frame Format (FRF = 10), Single Transfer

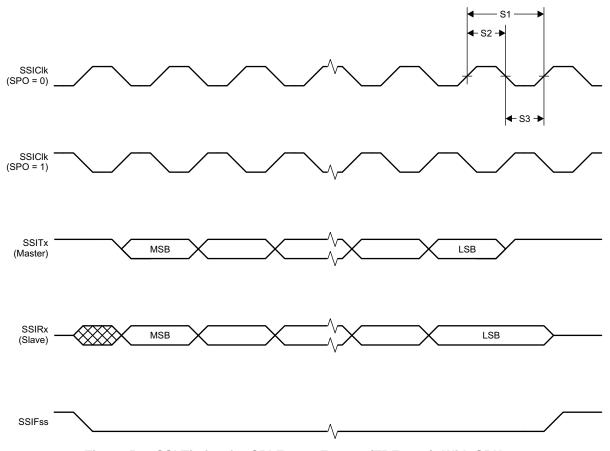
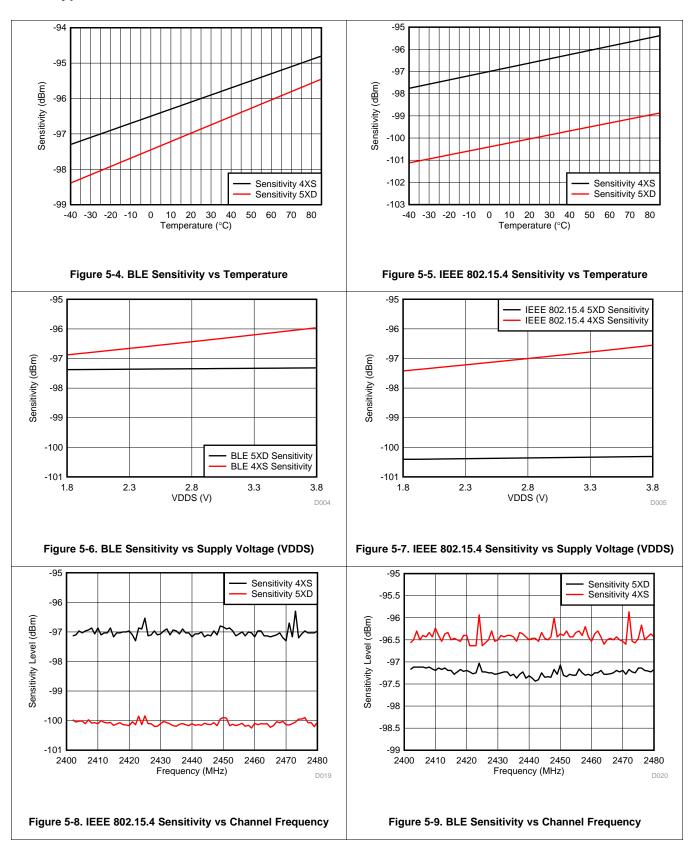
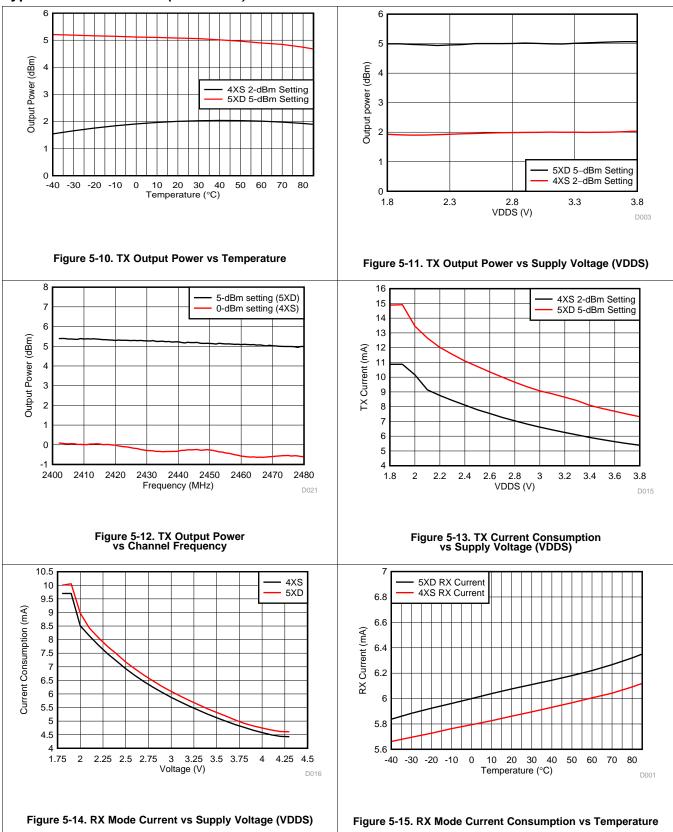


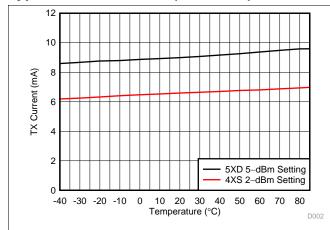
Figure 5-3. SSI Timing for SPI Frame Format (FRF = 00), With SPH = 1

5.24 Typical Characteristics





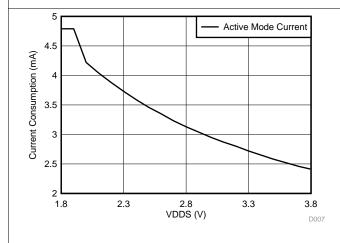




(mA) Active Mode Current Active Mode Current Consumpstion 3.05 3 2.95 2.9 -40 -30 -20 -10 0 20 30 40 50 60 70 80 10 Temperature (°C)

Figure 5-16. TX Mode Current Consumption vs Temperature

Figure 5-17. Active Mode (MCU Running, No Peripherals) Current Consumption vs Temperature



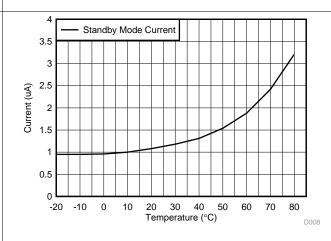


Figure 5-18. Active Mode (MCU Running, No Peripherals) Current Consumption vs Supply Voltage (VDDS)

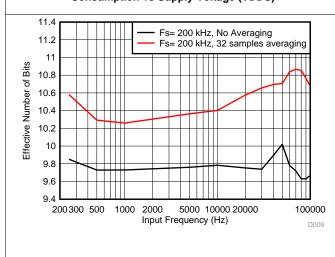


Figure 5-19. Standby Mode Current Consumption With RCOSC RTC vs Temperature

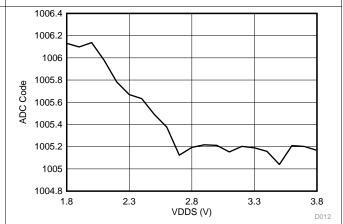


Figure 5-20. SoC ADC Effective Number of Bits vs Input Frequency (Internal Reference, No Scaling)

Figure 5-21. SoC ADC Output vs Supply Voltage (Fixed Input, Internal Reference, No Scaling)



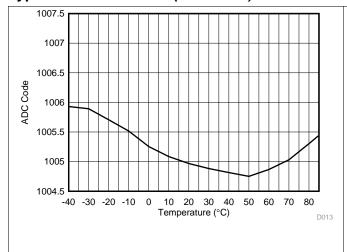


Figure 5-22. SoC ADC Output vs Temperature (Fixed Input, Internal Reference, No Scaling)

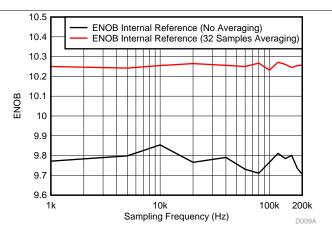
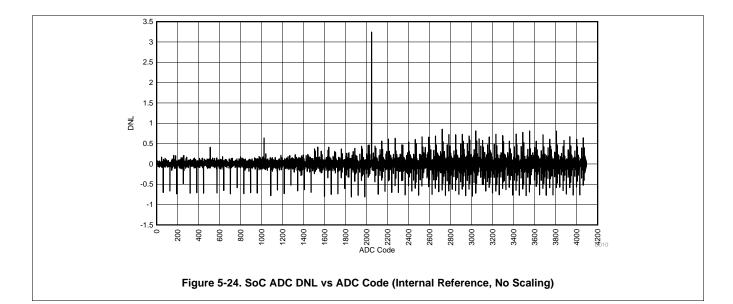
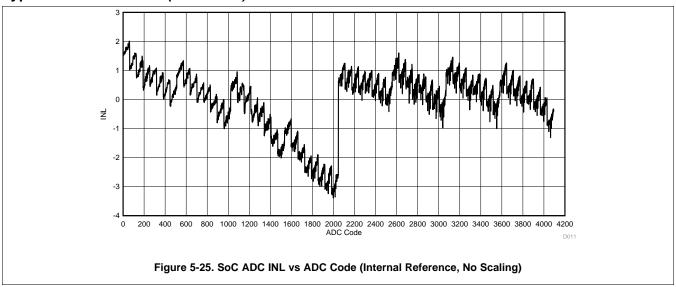


Figure 5-23. SoC ADC ENOB vs Sampling Frequency (Input Frequency = FS / 10)





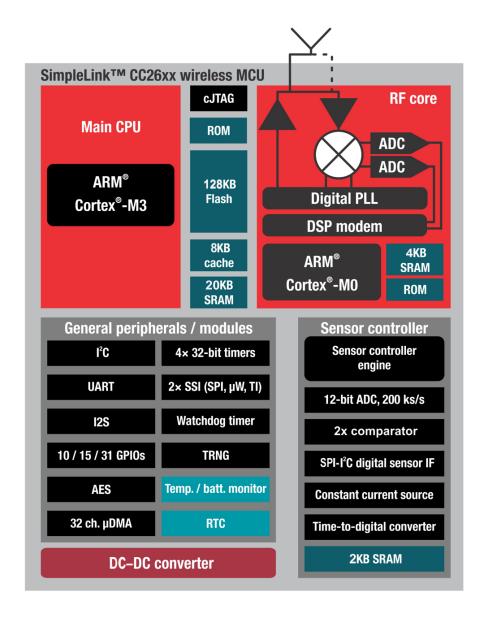


6 Detailed Description

6.1 Overview

The core modules of the CC26xx product family are shown in the Section 6.2.

6.2 Functional Block Diagram



6.3 Main CPU

The SimpleLink CC2650 Wireless MCU contains an ARM Cortex-M3 (CM3) 32-bit CPU, which runs the application and the higher layers of the protocol stack.

The CM3 processor provides a high-performance, low-cost platform that meets the system requirements of minimal memory implementation, and low-power consumption, while delivering outstanding computational performance and exceptional system response to interrupts.

CM3 features include the following:

- 32-bit ARM Cortex-M3 architecture optimized for small-footprint embedded applications
- Outstanding processing performance combined with fast interrupt handling
- ARM Thumb[®]-2 mixed 16- and 32-bit instruction set delivers the high performance expected of a 32-bit ARM core in a compact memory size usually associated with 8- and 16-bit devices, typically in the range of a few kilobytes of memory for microcontroller-class applications:
 - Single-cycle multiply instruction and hardware divide
 - Atomic bit manipulation (bit-banding), delivering maximum memory use and streamlined peripheral control
 - Unaligned data access, enabling data to be efficiently packed into memory
- Fast code execution permits slower processor clock or increases sleep mode time
- · Harvard architecture characterized by separate buses for instruction and data
- Efficient processor core, system, and memories
- Hardware division and fast digital-signal-processing oriented multiply accumulate
- · Saturating arithmetic for signal processing
- Deterministic, high-performance interrupt handling for time-critical applications
- · Enhanced system debug with extensive breakpoint and trace capabilities
- Serial wire trace reduces the number of pins required for debugging and tracing
- Migration from the ARM7[™] processor family for better performance and power efficiency
- · Optimized for single-cycle flash memory use
- Ultralow-power consumption with integrated sleep modes
- 1.25 DMIPS per MHz

6.4 RF Core

The RF Core contains an ARM Cortex-M0 processor that interfaces the analog RF and base-band circuitries, handles data to and from the system side, and assembles the information bits in a given packet structure. The RF core offers a high level, command-based API to the main CPU.

The RF core is capable of autonomously handling the time-critical aspects of the radio protocols (802.15.4 RF4CE and ZigBee, *Bluetooth* Low Energy) thus offloading the main CPU and leaving more resources for the user application.

The RF core has a dedicated 4-KB SRAM block and runs initially from separate ROM memory. The ARM Cortex-M0 processor is not programmable by customers.



6.5 Sensor Controller

The Sensor Controller contains circuitry that can be selectively enabled in standby mode. The peripherals in this domain may be controlled by the Sensor Controller Engine which is a proprietary power-optimized CPU. This CPU can read and monitor sensors or perform other tasks autonomously, thereby significantly reducing power consumption and offloading the main CM3 CPU.

The Sensor Controller is set up using a PC-based configuration tool, called Sensor Controller Studio, and potential use cases may be (but are not limited to):

- · Analog sensors using integrated ADC
- Digital sensors using GPIOs, bit-banged I²C, and SPI
- UART communication for sensor reading or debugging
- Capacitive sensing
- · Waveform generation
- Pulse counting
- Keyboard scan
- Quadrature decoder for polling rotation sensors
- Oscillator calibration

NOTE

Texas Instruments provides application examples for some of these use cases, but not for all of them.

The peripherals in the Sensor Controller include the following:

- The low-power clocked comparator can be used to wake the device from any state in which the comparator is active. A configurable internal reference can be used in conjunction with the comparator. The output of the comparator can also be used to trigger an interrupt or the ADC.
- Capacitive sensing functionality is implemented through the use of a constant current source, a time-to-digital converter, and a comparator. The continuous time comparator in this block can also be used as a higher-accuracy alternative to the low-power clocked comparator. The Sensor Controller will take care of baseline tracking, hysteresis, filtering and other related functions.
- The ADC is a 12-bit, 200-ksamples/s ADC with eight inputs and a built-in voltage reference. The ADC can be triggered by many different sources, including timers, I/O pins, software, the analog comparator, and the RTC.
- The Sensor Controller also includes a SPI–I²C digital interface.
- The analog modules can be connected to up to eight different GPIOs.

The peripherals in the Sensor Controller can also be controlled from the main application processor.

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Table 6-1. GPIOs Connected to the Sensor Controller (1)

ANALOG CAPABLE	7 × 7 RGZ DIO NUMBER	5 × 5 RHB DIO NUMBER	4 × 4 RSM DIO NUMBER
Υ	30	14	
Y	29	13	
Y	28	12	
Y	27	11	9
Υ	26	9	8
Υ	25	10	7
Y	24	8	6
Y	23	7	5
N	7	4	2
N	6	3	1
N	5	2	0
N	4	1	
N	3	0	
N	2		
N	1		
N	0		

⁽¹⁾ Depending on the package size, up to 16 pins can be connected to the Sensor Controller. Up to 8 of these pins can be connected to analog modules.

6.6 Memory

The flash memory provides nonvolatile storage for code and data. The flash memory is in-system programmable.

The SRAM (static RAM) can be used for both storage of data and execution of code and is split into two 4-KB blocks and two 6-KB blocks. Retention of the RAM contents in standby mode can be enabled or disabled individually for each block to minimize power consumption. In addition, if flash cache is disabled, the 8-KB cache can be used as a general-purpose RAM.

The ROM provides preprogrammed embedded TI RTOS kernel, Driverlib and lower layer protocol stack software (802.15.4 MAC and *Bluetooth* Low Energy Controller). It also contains a bootloader that can be used to reprogram the device using SPI or UART.

6.7 Debug

The on-chip debug support is done through a dedicated cJTAG (IEEE 1149.7) or JTAG (IEEE 1149.1) interface.



6.8 Power Management

To minimize power consumption, the CC2650 device supports a number of power modes and power management features (see Table 6-2).

Table 6-2. Power Modes

MODE	SOFTWARE CONFIGURABLE POWER MODES				
	ACTIVE	IDLE	STANDBY	SHUTDOWN	RESET PIN HELD
CPU	Active	Off	Off	Off	Off
Flash	On	Available	Off	Off	Off
SRAM	On	On	On	Off	Off
Radio	Available	Available	Off	Off	Off
Supply System	On	On	Duty Cycled	Off	Off
Current	1.45 mA + 31 μA/MHz	550 µA	1 μΑ	0.15 μΑ	0.1 μΑ
Wake-up Time to CPU Active (1)	-	14 µs	151 µs	1015 μs	1015 µs
Register Retention	Full	Full	Partial	No	No
SRAM Retention	Full	Full	Full	No	No
High-Speed Clock	XOSC_HF or RCOSC_HF	XOSC_HF or RCOSC_HF	Off	Off	Off
Low-Speed Clock	XOSC_LF or RCOSC_LF	XOSC_LF or RCOSC_LF	XOSC_LF or RCOSC_LF	Off	Off
Peripherals	Available	Available	Off	Off	Off
Sensor Controller	Available	Available	Available	Off	Off
Wake up on RTC	Available	Available	Available	Off	Off
Wake up on Pin Edge	Available	Available	Available	Available	Off
Wake up on Reset Pin	Available	Available	Available	Available	Available
Brown Out Detector (BOD)	Active	Active	Duty Cycled ⁽²⁾	Off	N/A
Power On Reset (POR)	Active	Active	Active	Active	N/A

⁽¹⁾ Not including RTOS overhead

In active mode, the application CM3 CPU is actively executing code. Active mode provides normal operation of the processor and all of the peripherals that are currently enabled. The system clock can be any available clock source (see Table 6-2).

In idle mode, all active peripherals can be clocked, but the Application CPU core and memory are not clocked and no code is executed. Any interrupt event will bring the processor back into active mode.

In standby mode, only the always-on domain (AON) is active. An external wake event, RTC event, or sensor-controller event is required to bring the device back to active mode. MCU peripherals with retention do not need to be reconfigured when waking up again, and the CPU continues execution from where it went into standby mode. All GPIOs are latched in standby mode.

In shutdown mode, the device is turned off entirely, including the AON domain and the Sensor Controller. The I/Os are latched with the value they had before entering shutdown mode. A change of state on any I/O pin defined as a *wake from Shutdown pin* wakes up the device and functions as a reset trigger. The CPU can differentiate between a reset in this way, a reset-by-reset pin, or a power-on-reset by reading the reset status register. The only state retained in this mode is the latched I/O state and the Flash memory contents.

The Brown Out Detector is disabled between recharge periods in STANDBY. Lowering the supply voltage below the BOD threshold between two recharge periods while in STANDBY may cause the BOD to lock the device upon wake-up until a Reset/POR releases it. To avoid this, it is recommended that STANDBY mode is avoided if there is a risk that the supply voltage (VDDS) may drop below the specified operating voltage range. For the same reason, it is also good practice to ensure that a power cycling operation, such as a battery replacement, triggers a Power-on-reset by ensuring that the VDDS decoupling network is fully depleted before applying supply voltage again (for example, inserting new batteries).



The Sensor Controller is an autonomous processor that can control the peripherals in the Sensor Controller independently of the main CPU, which means that the main CPU does not have to wake up, for example, to execute an ADC sample or poll a digital sensor over SPI. The main CPU saves both current and wake-up time that would otherwise be wasted. The Sensor Controller Studio enables the user to configure the sensor controller and choose which peripherals are controlled and which conditions wake up the main CPU.

6.9 Clock Systems

The CC2650 supports two external and two internal clock sources.

A 24-MHz crystal is required as the frequency reference for the radio. This signal is doubled internally to create a 48-MHz clock.

The 32-kHz crystal is optional. *Bluetooth* low energy requires a slow-speed clock with better than ±500 ppm accuracy if the device is to enter any sleep mode while maintaining a connection. The internal 32-kHz RC oscillator can in some use cases be compensated to meet the requirements. The low-speed crystal oscillator is designed for use with a 32-kHz watch-type crystal.

The internal high-speed oscillator (48-MHz) can be used as a clock source for the CPU subsystem.

The internal low-speed oscillator (32.768-kHz) can be used as a reference if the low-power crystal oscillator is not used.

The 32-kHz clock source can be used as external clocking reference through GPIO.

6.10 General Peripherals and Modules

The I/O controller controls the digital I/O pins and contains multiplexer circuitry to allow a set of peripherals to be assigned to I/O pins in a flexible manner. All digital I/Os are interrupt and wake-up capable, have a programmable pullup and pulldown function and can generate an interrupt on a negative or positive edge (configurable). When configured as an output, pins can function as either push-pull or open-drain. Five GPIOs have high drive capabilities (marked in **bold** in Section 4).

The SSIs are synchronous serial interfaces that are compatible with SPI, MICROWIRE, and Texas Instruments synchronous serial interfaces. The SSIs support both SPI master and slave up to 4 MHz.

The UART implements a universal asynchronous receiver/transmitter function. It supports flexible baudrate generation up to a maximum of 3 Mbps and is compatible with the *Bluetooth* HCl specifications.

Timer 0 is a general-purpose timer module (GPTM), which provides two 16-bit timers. The GPTM can be configured to operate as a single 32-bit timer, dual 16-bit timers or as a PWM module.

Timer 1, Timer 2, and Timer 3 are also GPTMs. Each of these timers is functionally equivalent to Timer 0.

In addition to these four timers, the RF core has its own timer to handle timing for RF protocols; the RF timer can be synchronized to the RTC.

The I²C interface is used to communicate with devices compatible with the I²C standard. The I²C interface is capable of 100-kHz and 400-kHz operation, and can serve as both I²C master and I²C slave.

The TRNG module provides a true, nondeterministic noise source for the purpose of generating keys, initialization vectors (IVs), and other random number requirements. The TRNG is built on 24 ring oscillators that create unpredictable output to feed a complex nonlinear combinatorial circuit.

The watchdog timer is used to regain control if the system fails due to a software error after an external device fails to respond as expected. The watchdog timer can generate an interrupt or a reset when a predefined time-out value is reached.

The device includes a direct memory access (µDMA) controller. The µDMA controller provides a way to offload data transfer tasks from the CM3 CPU, allowing for more efficient use of the processor and the available bus bandwidth. The µDMA controller can perform transfer between memory and peripherals. The µDMA controller has dedicated channels for each supported on-chip module and can be programmed to automatically perform transfers between peripherals and memory as the peripheral is ready to transfer more data. Some features of the µDMA controller include the following (this is not an exhaustive list):

- Highly flexible and configurable channel operation of up to 32 channels
- Transfer modes:
 - Memory-to-memory
 - Memory-to-peripheral
 - Peripheral-to-memory
 - Peripheral-to-peripheral
- Data sizes of 8, 16, and 32 bits

The AON domain contains circuitry that is always enabled, except for in Shutdown (where the digital supply is off). This circuitry includes the following:

- The RTC can be used to wake the device from any state where it is active. The RTC contains three compare and one capture registers. With software support, the RTC can be used for clock and calendar operation. The RTC is clocked from the 32-kHz RC oscillator or crystal. The RTC can also be compensated to tick at the correct frequency even when the internal 32-kHz RC oscillator is used instead of a crystal.
- The battery monitor and temperature sensor are accessible by software and give a battery status indication as well as a coarse temperature measure.

6.11 System Architecture

Depending on the product configuration, CC26xx can function either as a Wireless Network Processor (WNP—an IC running the wireless protocol stack, with the application running on a separate MCU), or as a System-on-Chip (SoC), with the application and protocol stack running on the ARM CM3 core inside the device.

In the first case, the external host MCU communicates with the device using SPI or UART. In the second case, the application must be written according to the application framework supplied with the wireless protocol stack.

7 Application, Implementation, and Layout

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

7.1 Application Information

Very few external components are required for the operation of the CC2650 device. This section provides some general information about the various configuration options when using the CC2650 in an application, and then shows two examples of application circuits with schematics and layout. Complete reference designs are available in the product folder on www.ti.com. Figure 7-1 shows the various RF front-end configuration options. The RF front end can be used in differential- or single-ended configurations with the options of having internal or external biasing. These options allow for various trade-offs between cost, board space, and RF performance. Differential operation with external bias gives the best performance while single-ended operation with internal bias gives the least amount of external components and the lowest power consumption. Reference designs exist for each of these options.

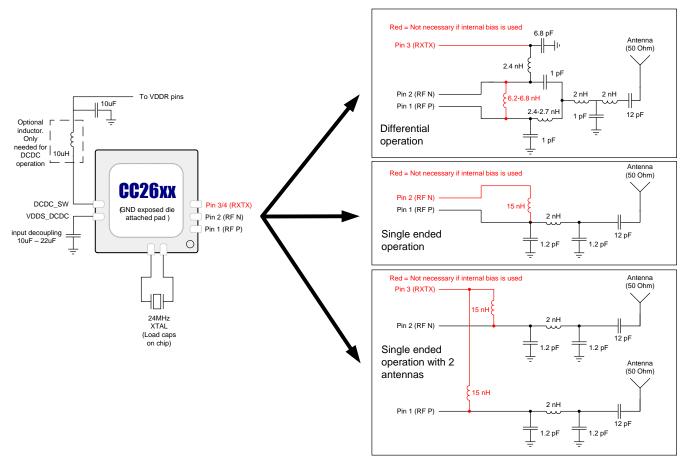


Figure 7-1. CC2650 Application Circuit



Figure 7-2 shows the various supply voltage configuration options. Not all power supply decoupling capacitors or digital I/Os are shown. Exact pin positions will vary between the different package options. For a detailed overview of power supply decoupling and wiring, see the TI reference designs and the CC26xx technical reference manual (Section 8.2).

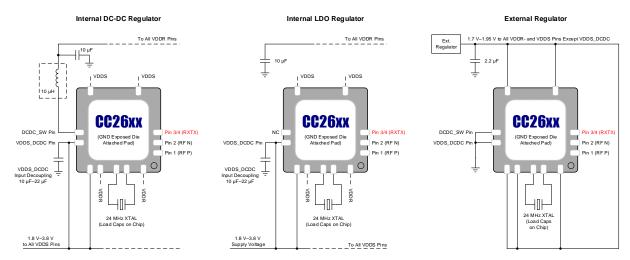


Figure 7-2. Supply Voltage Configurations

7.2 5 x 5 External Differential (5XD) Application Circuit

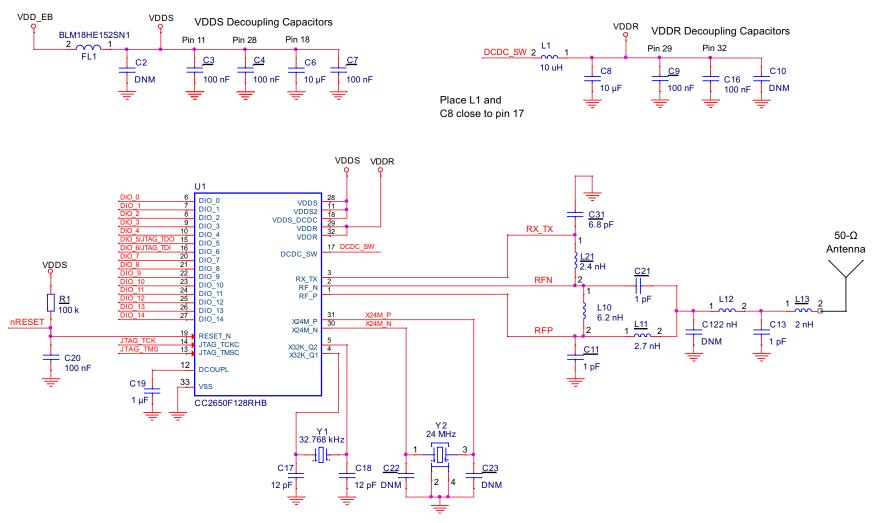


Figure 7-3. 5 x 5 External Differential (5XD) Application Circuit



7.2.1 Layout

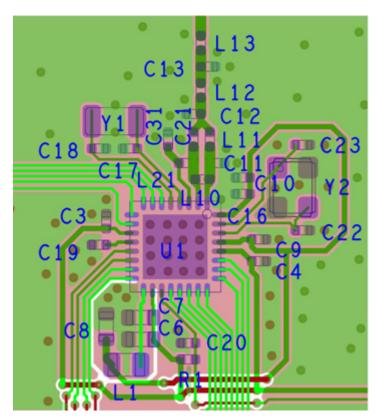


Figure 7-4. 5 × 5 External Differential (5XD) Layout

7.3 4 x 4 External Single-ended (4XS) Application Circuit

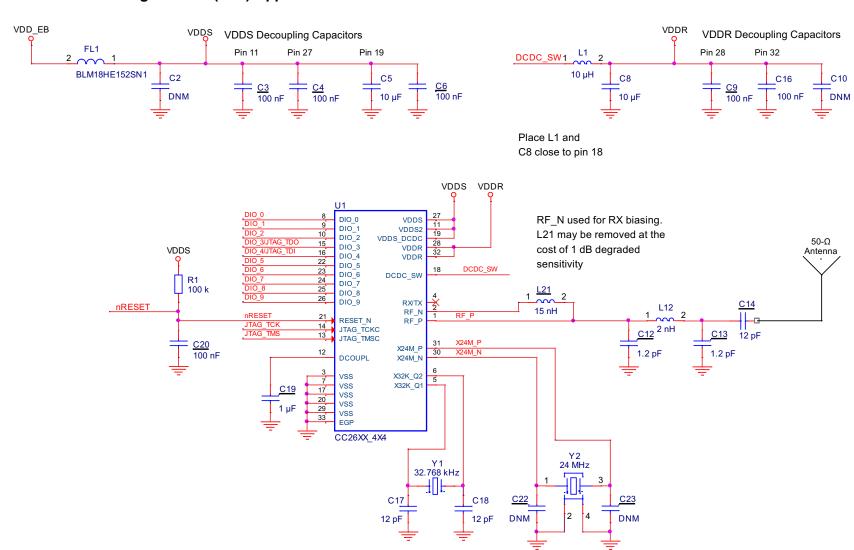


Figure 7-5. 4 × 4 External Single-ended (4XS) Application Circuit



7.3.1 Layout

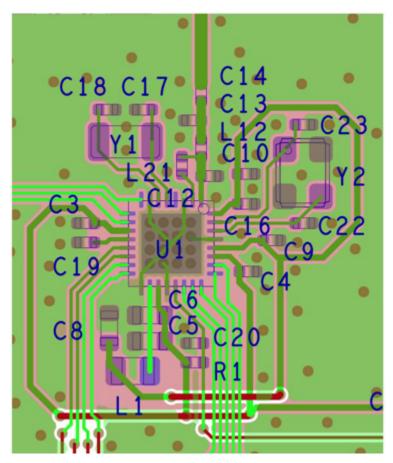


Figure 7-6. 4 x 4 External Single-ended (4XS) Layout



8 Device and Documentation Support

8.1 Device Support

8.1.1 Development Support

TI offers an extensive line of development tools, including tools to evaluate the performance of the processors, generate code, develop algorithm implementations, and fully integrate and debug software and hardware modules.

The following products support development of the CC2650 device applications:

Software Tools:

SmartRF Studio 7:

SmartRF Studio is a PC application that helps designers of radio systems to easily evaluate the RF-IC at an early stage in the design process.

- Test functions for sending and receiving radio packets, continuous wave transmit and receive
- Evaluate RF performance on custom boards by wiring it to a supported evaluation board or debugger
- Can also be used without any hardware, but then only to generate, edit and export radio configuration settings
- Can be used in combination with several development kits for Texas Instruments' CCxxxx RF-ICs

Sensor Controller Studio:

Sensor Controller Studio provides a development environment for the CC26xx Sensor Controller. The Sensor Controller is a proprietary, power-optimized CPU in the CC26xx, which can perform simple background tasks autonomously and independent of the System CPU state.

- Allows for Sensor Controller task algorithms to be implemented using a C-like programming language
- Outputs a Sensor Controller Interface driver, which incorporates the generated Sensor Controller machine code and associated definitions
- Allows for rapid development by using the integrated Sensor Controller task testing and debugging functionality. This allows for live visualization of sensor data and algorithm verification.

IDEs and Compilers:

Code Composer Studio:

- · Integrated development environment with project management tools and editor
- Code Composer Studio (CCS) 6.1 and later has built-in support for the CC26xx device family
- Best support for XDS debuggers; XDS100v3, XDS110 and XDS200
- High integration with TI-RTOS with support for TI-RTOS Object View

IAR Embedded Workbench for ARM

- Integrated development environment with project management tools and editor
- IAR EWARM 7.30.3 and later has built-in support for the CC26xx device family
- Broad debugger support, supporting XDS100v3, XDS200, IAR I-Jet and Segger J-Link
- Integrated development environment with project management tools and editor
- RTOS plugin available for TI-RTOS

For a complete listing of development-support tools for the CC2650 platform, visit the Texas Instruments website at www.ti.com. For information on pricing and availability, contact the nearest TI field sales office or authorized distributor.



8.1.2 Device Nomenclature

To designate the stages in the product development cycle, TI assigns prefixes to all part numbers and date-code. Each device has one of three prefixes/identifications: X, P, or null (no prefix) (for example, CC2650 is in production; therefore, no prefix/identification is assigned).

Device development evolutionary flow:

- **X** Experimental device that is not necessarily representative of the final device's electrical specifications and may not use production assembly flow.
- P Prototype device that is not necessarily the final silicon die and may not necessarily meet final electrical specifications.
- **null** Production version of the silicon die that is fully qualified.

Production devices have been characterized fully, and the quality and reliability of the device have been demonstrated fully. Tl's standard warranty applies.

Predictions show that prototype devices (X or P) have a greater failure rate than the standard production devices. Texas Instruments recommends that these devices not be used in any production system because their expected end-use failure rate still is undefined. Only qualified production devices are to be used.

TI device nomenclature also includes a suffix with the device family name. This suffix indicates the package type (for example, RSM).

For orderable part numbers of the CC2650 device in the RSM, RHB or RGZ package types, see the Package Option Addendum of this document, the TI website (www.ti.com), or contact your TI sales representative.



8.2 Documentation Support

The following documents describe the CC2650. Copies of these documents are available on the Internet at www.ti.com.

SWCU117 CC26xx SimpleLink™ Wireless MCU Technical Reference Manual

SWRZ058 CC26xx SimpleLink™ Wireless MCU Errata

8.3 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

TI E2E™ Online Community TI's Engineer-to-Engineer (E2E) Community. Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

TI Embedded Processors Wiki Texas Instruments Embedded Processors Wiki. Established to help developers get started with Embedded Processors from Texas Instruments and to foster innovation and growth of general knowledge about the hardware and software surrounding these devices.

8.4 Texas Instruments Low-Power RF Website

Texas Instruments' Low-Power RF website has all the latest products, application and design notes, FAQ section, news and events updates. Go to www.ti.com/lprf.

8.5 Low-Power RF Online Community

- Forums, videos, and blogs
- · RF design help
- E2E interaction

Join at: www.ti.com/lprf-forum.

8.6 Texas Instruments Low-Power RF Developer Network

Texas Instruments has launched an extensive network of low-power RF development partners to help customers speed up their application development. The network consists of recommended companies, RF consultants, and independent design houses that provide a series of hardware module products and design services, including:

- RF circuit, low-power RF, and ZigBee design services
- Low-power RF and ZigBee module solutions and development tools
- RF certification services and RF circuit manufacturing

For help with modules, engineering services or development tools:

Search the Low-Power RF Developer Network to find a suitable partner. www.ti.com/lprfnetwork

8.7 Low-Power RF eNewsletter

The Low-Power RF eNewsletter is up-to-date on new products, news releases, developers' news, and other news and events associated with low-power RF products from TI. The Low-Power RF eNewsletter articles include links to get more online information.

Sign up at: www.ti.com/lprfnewsletter



8.8 Additional Information

Texas Instruments offers a wide selection of cost-effective, low-power RF solutions for proprietary and standard-based wireless applications for use in industrial and consumer applications. The selection includes RF transceivers, RF transmitters, RF front ends, and Systems-on-Chips as well as various software solutions for the sub-1-GHz and 2.4-GHz frequency bands.

In addition, Texas Instruments provides a large selection of support collateral such as development tools, technical documentation, reference designs, application expertise, customer support, third-party and university programs.

The Low-Power RF E2E Online Community provides technical support forums, videos and blogs, and the chance to interact with engineers from all over the world.

With a broad selection of product solutions, end-application possibilities, and a range of technical support, Texas Instruments offers the broadest low-power RF portfolio.

8.9 Trademarks

SimpleLink, SmartRF, Code Composer Studio, E2E are trademarks of Texas Instruments.

ARM7 is a trademark of ARM Limited (or its subsidiaries).

ARM, Cortex, ARM Thumb are registered trademarks of ARM Limited (or its subsidiaries).

Bluetooth is a registered trademark of Bluetooth SIG, Inc.

CoreMark is a registered trademark of Embedded Microprocessor Benchmark Consortium.

IAR Embedded Workbench is a registered trademark of IAR Systems AB.

IEEE Std 1241 is a trademark of Institute of Electrical and Electronics Engineers, Incorporated.

ZigBee is a registered trademark of ZigBee Alliance, Inc.

8.10 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

8.11 Export Control Notice

Recipient agrees to not knowingly export or re-export, directly or indirectly, any product or technical data (as defined by the U.S., EU, and other Export Administration Regulations) including software, or any controlled product restricted by other applicable national regulations, received from Disclosing party under this Agreement, or any direct product of such technology, to any destination to which such export or re-export is restricted or prohibited by U.S. or other applicable laws, without obtaining prior authorization from U.S. Department of Commerce and other competent Government authorities to the extent required by those laws.

8.12 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms and definitions.

9 Mechanical Packaging and Orderable Information

9.1 Packaging Information

The following pages include mechanical packaging and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.





5-Sep-2015

PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package	Pins	Package	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Device Marking	Samples
	(1)		Drawing		Qty	(2)	(6)	(3)		(4/5)	
CC2650F128RGZR	ACTIVE	VQFN	RGZ	48	2500	Green (RoHS & no Sb/Br)	CU NIPDAU Call TI	Level-3-260C-168 HR	-40 to 85	CC2650 F128	Samples
CC2650F128RGZT	ACTIVE	VQFN	RGZ	48	250	Green (RoHS & no Sb/Br)	CU NIPDAU Call TI	Level-3-260C-168 HR	-40 to 85	CC2650 F128	Samples
CC2650F128RHBR	ACTIVE	VQFN	RHB	32	3000	Green (RoHS & no Sb/Br)	CU NIPDAU Call TI	Level-3-260C-168 HR	-40 to 85	CC2650 F128	Samples
CC2650F128RHBT	ACTIVE	VQFN	RHB	32	250	Green (RoHS & no Sb/Br)	CU NIPDAU Call TI	Level-3-260C-168 HR	-40 to 85	CC2650 F128	Samples
CC2650F128RSMR	ACTIVE	VQFN	RSM	32	3000	Green (RoHS & no Sb/Br)	CU NIPDAU Call TI	Level-3-260C-168 HR	-40 to 85	CC2650 F128	Samples
CC2650F128RSMT	ACTIVE	VQFN	RSM	32	250	Green (RoHS & no Sb/Br)	CU NIPDAU Call TI	Level-3-260C-168 HR	-40 to 85	CC2650 F128	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes. **Pb-Free** (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between

the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.



PACKAGE OPTION ADDENDUM

5-Sep-2015

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

PACKAGE MATERIALS INFORMATION

www.ti.com 3-Aug-2015

TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
CC2650F128RGZR	VQFN	RGZ	48	2500	330.0	16.4	7.3	7.3	1.1	12.0	16.0	Q2
CC2650F128RGZT	VQFN	RGZ	48	250	180.0	16.4	7.3	7.3	1.1	12.0	16.0	Q2
CC2650F128RHBR	VQFN	RHB	32	3000	330.0	12.4	5.3	5.3	1.1	8.0	12.0	Q2
CC2650F128RHBT	VQFN	RHB	32	250	180.0	12.4	5.3	5.3	1.1	8.0	12.0	Q2
CC2650F128RSMR	VQFN	RSM	32	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
CC2650F128RSMT	VQFN	RSM	32	250	180.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2

www.ti.com 3-Aug-2015



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
CC2650F128RGZR	VQFN	RGZ	48	2500	367.0	367.0	38.0
CC2650F128RGZT	VQFN	RGZ	48	250	210.0	185.0	35.0
CC2650F128RHBR	VQFN	RHB	32	3000	367.0	367.0	35.0
CC2650F128RHBT	VQFN	RHB	32	250	210.0	185.0	35.0
CC2650F128RSMR	VQFN	RSM	32	3000	367.0	367.0	35.0
CC2650F128RSMT	VQFN	RSM	32	250	210.0	185.0	35.0



NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.

- B. This drawing is subject to change without notice.
- C. Quad Flatpack, No-leads (QFN) package configuration.
- D. The package thermal pad must be soldered to the board for thermal and mechanical performance.
- E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
- F. Falls within JEDEC MO-220.



RGZ (S-PVQFN-N48)

PLASTIC QUAD FLATPACK NO-LEAD

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No—Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



NOTE: All linear dimensions are in millimeters



RGZ (S-PVQFN-N48)

PLASTIC QUAD FLATPACK NO-LEAD



NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat—Pack Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com http://www.ti.com>.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
- F. Customers should contact their board fabrication site for recommended solder mask tolerances and via tenting recommendations for vias placed in the thermal pad.



PLASTIC QUAD FLATPACK NO-LEAD



NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M—1994.

- B. This drawing is subject to change without notice.
- C. QFN (Quad Flatpack No-Lead) Package configuration.
- D. The package thermal pad must be soldered to the board for thermal and mechanical performance.
- E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
- F. Falls within JEDEC MO-220.



PLASTIC QUAD FLATPACK NO-LEAD

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No—Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



NOTE: A. All linear dimensions are in millimeters



PLASTIC QUAD FLATPACK NO-LEAD



NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat—Pack Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com www.ti.com.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
- E. Customers should contact their board fabrication site for recommended solder mask tolerances and via tenting recommendations for any larger diameter vias placed in the thermal pad.





- NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
 - B. This drawing is subject to change without notice.
 - C. QFN (Quad Flatpack No-Lead) Package configuration.
 - The package thermal pad must be soldered to the board for thermal and mechanical performance. See the Product Data Sheet for details regarding the exposed thermal pad dimensions.



RSM (S-PVQFN-N32)

PLASTIC QUAD FLATPACK NO-LEAD

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No—Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



Bottom View

Exposed Thermal Pad Dimensions

4207868-2/1 07/14

NOTE: All linear dimensions are in millimeters



RSM (S-PVQFN-N32)

PLASTIC QUAD FLATPACK NO-LEAD



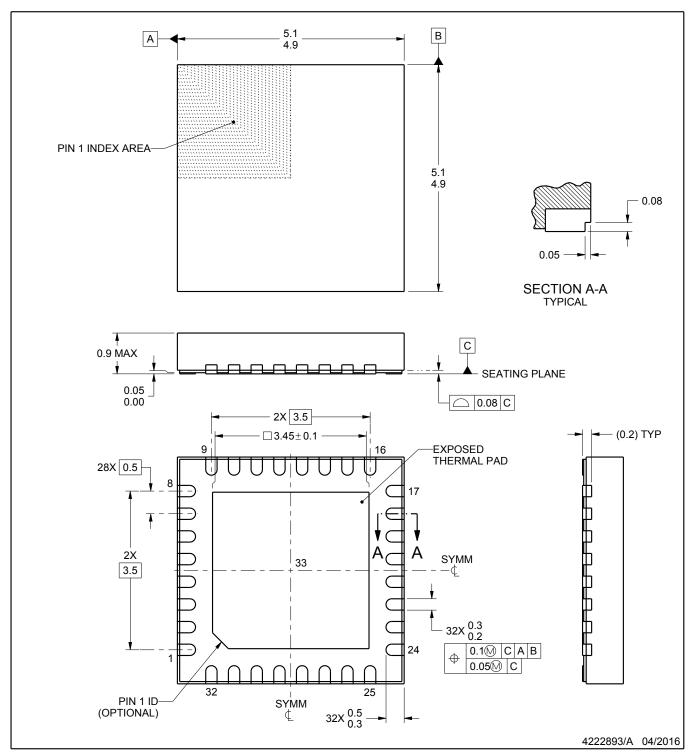
NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat—Pack Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com http://www.ti.com.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
- F. Customers should contact their board fabrication site for recommended solder mask tolerances and via tenting recommendations for vias placed in the thermal pad.





PLASTIC QUAD FLATPACK - NO LEAD

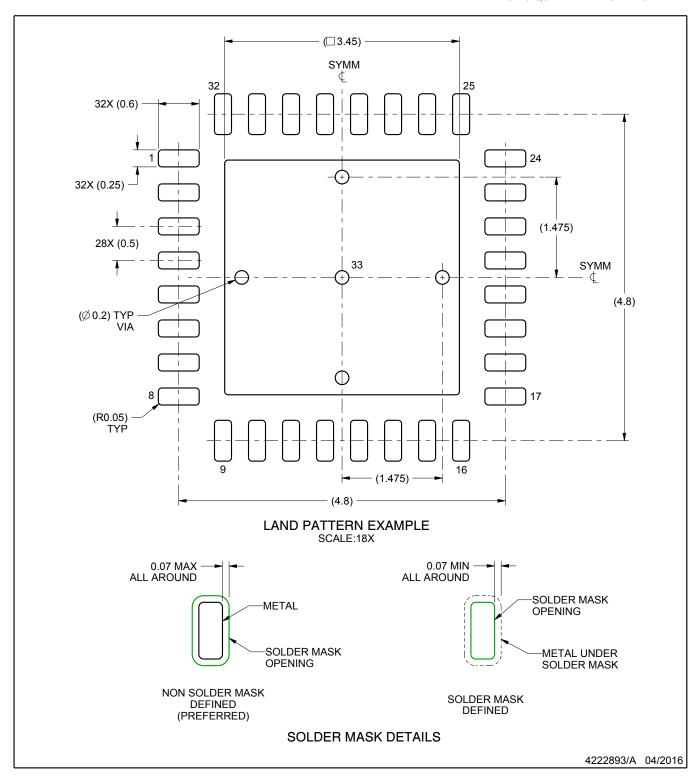


NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
 2. This drawing is subject to change without notice.
- 3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.



PLASTIC QUAD FLATPACK - NO LEAD

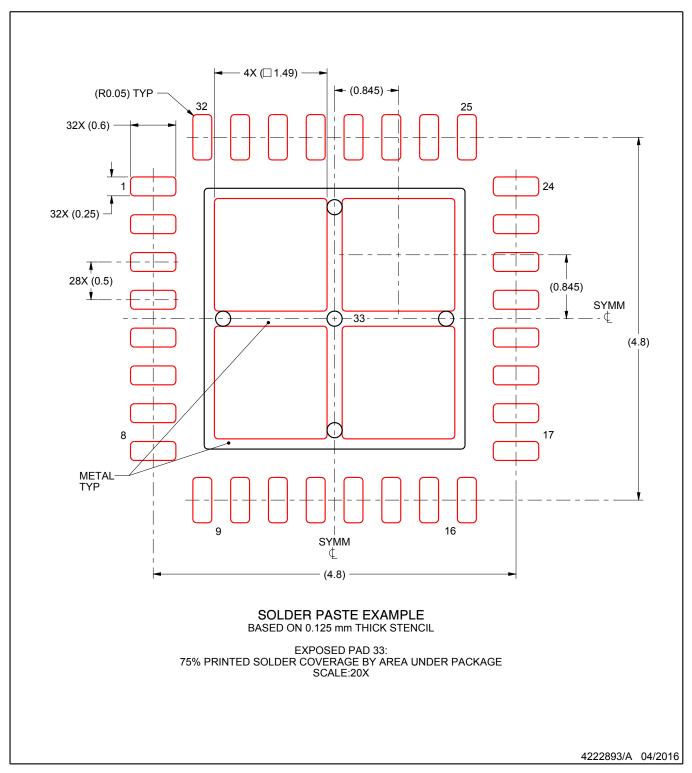


NOTES: (continued)

- 4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
- Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.



PLASTIC QUAD FLATPACK - NO LEAD



NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.



PLASTIC QUAD FLATPACK NO-LEAD

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No—Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



NOTE: A. All linear dimensions are in millimeters



PLASTIC QUAD FLATPACK NO-LEAD



NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat—Pack Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com www.ti.com.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
- E. Customers should contact their board fabrication site for recommended solder mask tolerances and via tenting recommendations for any larger diameter vias placed in the thermal pad.



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