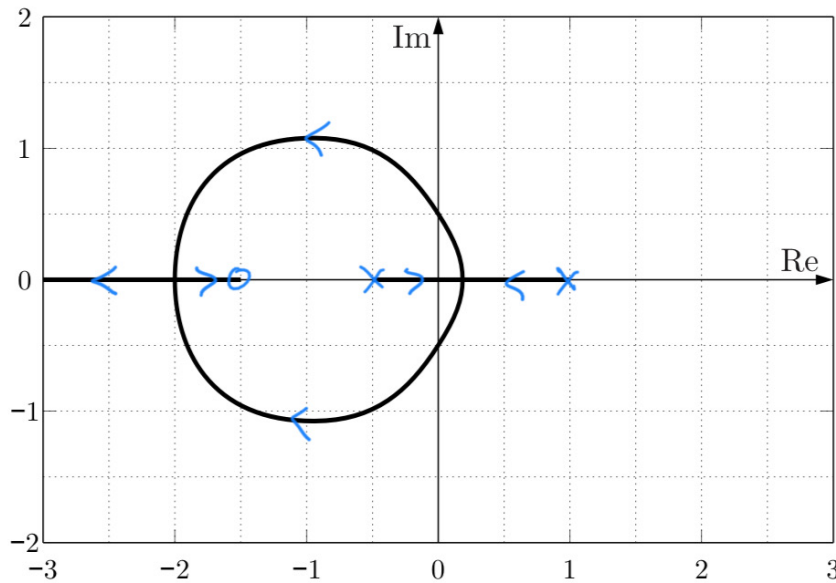


# Exam 02-Control Engineering

Monday, 10. February 2020 15:45

1.

a.



b.

From figure,

$$s_{N_1} = -1.5$$

$$s_{p_1} = -0.5$$

$$s_{p_2} = 1$$

Now,

$$G_0(s) = 1 \cdot \frac{(s + 1.5)}{(s + 0.5)(s - 1)}$$

$$= \frac{1}{2} \left\{ \frac{2s + 3}{(s + 0.5)(s - 1)} \right\}$$

$$= \frac{1}{2} \frac{(s + 0.5) + (s - 1) + 3.5}{(s + 0.5)(s - 1)}$$

$$= \frac{1}{2} \left\{ \frac{1}{s - 1} + \frac{1}{s + 0.5} + \frac{3.5}{(s + 0.5)(s - 1)} \right\}$$

$$= \frac{1}{2} \left\{ \frac{-1}{1 - s} + \frac{2}{1 + 2s} + \frac{-7}{(1 + 2s)(1 - s)} \right\}$$

$$g(t) = \mathcal{L}^{-1}\{G_0(s)\}(t)$$

$$= \frac{1}{2} \left\{ +1 e^t + \frac{2}{2} e^{-\frac{t}{2}} + \frac{-7}{3} (e^{-\frac{t}{2}} - e^t) \right\}$$

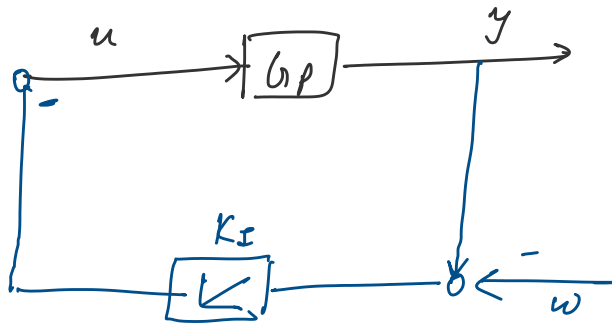
↓ , -t/2 ,

$$= \frac{1}{2} \left\{ e^t \left( \frac{10}{3} \right) + e^{-2} \left( \frac{-4}{3} \right) \right\}$$

$$g(t) = \frac{5}{3} e^t - \frac{2}{3} e^{-\frac{t}{2}}$$

2.

a



- b The  $G_I(j\omega)$  controller always show integrating behaviours.  $\therefore$  It is preferable for stationary accuracy

From Bode plot:

$\alpha_{PD} > \alpha_I$   
 $\therefore G_{PD}(j\omega)$  is preferable w.r.t phase margin

c)  $\therefore$  General form of  $G_p = \frac{1}{T_1 T_2 s^2 + (T_2 + T_1)s + 1}$

with slope changes of  $-1$  at  $\frac{1}{T_1}$  and  $\frac{1}{T_2}$

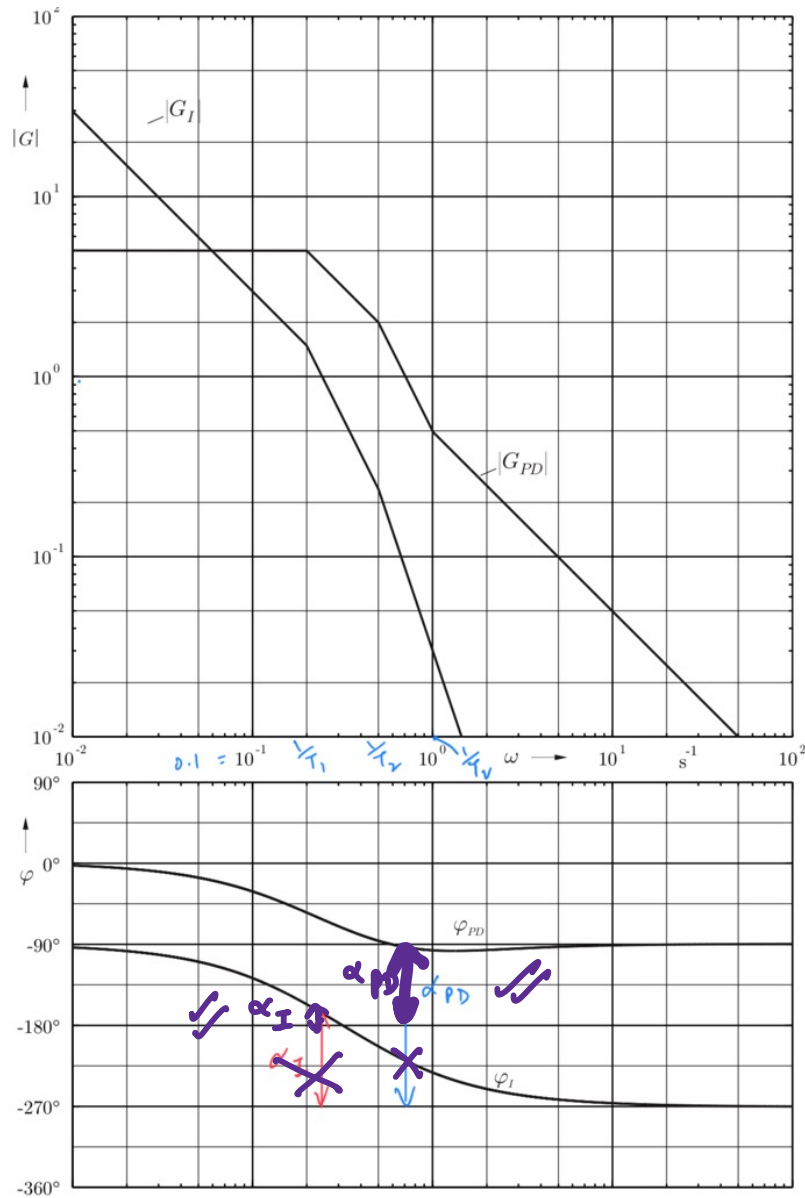
From BODE Plot and comparing it w/ given  $G_p(j\omega)$

$$\frac{1}{T_1} = 0.2 \text{ sec}^{-1} \Rightarrow T_1 = 5 \text{ sec}$$

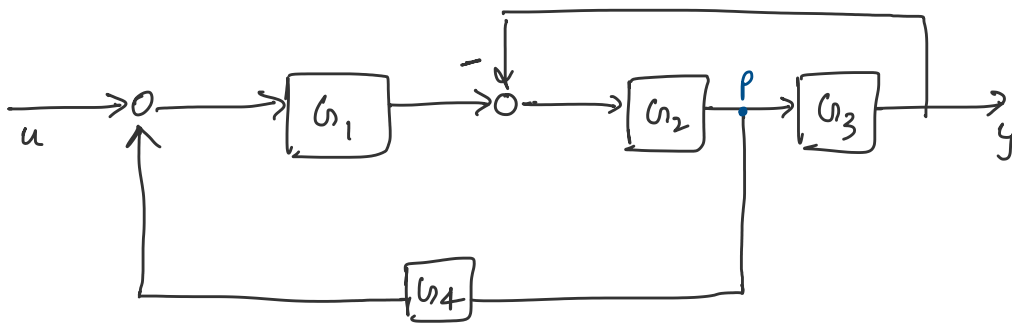
$$\frac{1}{T_2} = 0.5 \text{ sec}^{-1} \Rightarrow T_2 = 2 \text{ sec}$$

$$\therefore a_2 = T_1 T_2 = 10 \text{ sec}^2$$

$$u_1 = (1 + 1.2) = 1.2 \text{ sec}$$



3



$$\{(U + G_4 P) G_1 - Y\} G_2 G_3 = Y$$

$$\text{also, } Y = G_3 P \Rightarrow P = \frac{Y}{G_3}$$

$$\text{Now, } \left\{ U + \frac{G_4 Y}{G_3} \right\} G_1 - Y \} G_2 G_3 = Y$$

$$((G_3 \quad / \quad )$$

$$Y(1 + G_2 G_3 - G_1 G_2 G_4) = U G_2 G_3 G_1$$

$$G(s) = \frac{Y(s)}{U(s)} = \frac{G_2 G_3 G_1}{1 + G_2 G_3 - G_1 G_2 G_4}$$

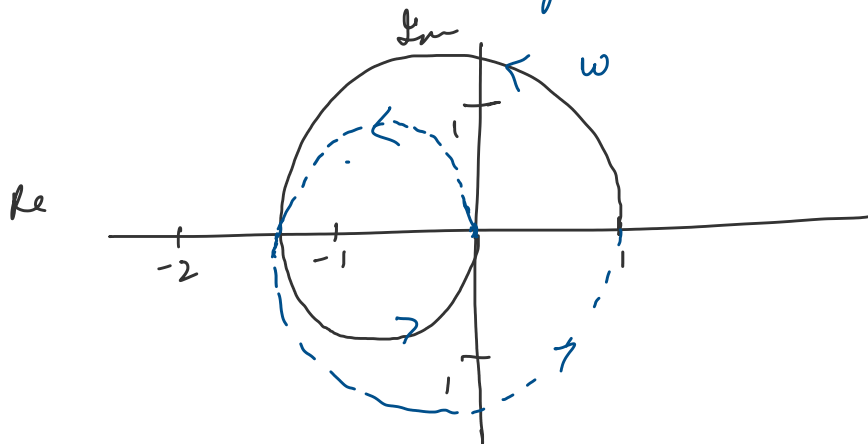
$$b. \quad G(s) = \frac{G_2 G_3}{1 + G_2 G_3}$$

$$\text{here, } G_0(s) = G_2 G_3. \quad (\text{say})$$

Now,

$\therefore G_2$  and  $G_3$  both have one pole in right half plane.

$\therefore$  Total pole in right half plane  $= 2 = p$



$\therefore$  2 revolves anticlockwise about -1

$\therefore m = -2$ , no. of revolutions of  $C''$  around -1 (opposite mathematically to direction i.e. clockwise)

$$\text{Now, } m = n - p$$

$$n = m + p = -2 + 2 = 0$$

$\therefore$  there are 0 zeros in  $N(s)$  inside  $C$  (right s halfplane)  
or 0 no. of poles of  $G_2(s)$  in right s half plane. or  $C''$

and,  $G_2(s) \approx$  stable if  $n \rightarrow 0$

$\therefore$  In our case  $G(s)$  is stable

4)

a)  $|sI - A| = 0$  For characteristic eq<sup>n</sup>

$$\Rightarrow \begin{vmatrix} s+1 & 5 \\ -1 & s-3 \end{vmatrix} = 0$$

$$\Rightarrow s^2 + s - 3s - 3 + 5 = 0$$

$$s^2 - 2s + 2 = 0$$

$$\cancel{(s-1)^2 = 0} \quad s = \frac{-2 \pm \sqrt{4-0}}{2} = -1 \pm i$$

$\therefore$  poles lie in left half plane  
 $\therefore$  system is stable.

b)  $s_1 = -1 + j \quad s_2 = -1 - j$

Characteristic eq<sup>n</sup>

$$(s-s_1)(s-s_2) = 0$$

$$\{s - (-1+j)\} \{s + 1+j\} = 0$$

$$\{s + 1 - j\} \{s + 1 + j\} = 0$$

$$(s+1)^2 - j^2 = 0$$

$$s^2 + 1 + 2s + 1 = 0$$

$$s^2 + 2s + 2 = 0 \quad \text{————— (1)}$$

Now, For given value of  $U = -KX$

$$\dot{X} = AX + B - KX$$

$$= (A - BK)X$$

$$= \left\{ \begin{pmatrix} -1 & -5 \\ 1 & 3 \end{pmatrix} - \begin{pmatrix} k_1 & k_2 \\ 0 & 0 \end{pmatrix} \right\} X$$

$$= \underbrace{\begin{pmatrix} -1-k_1 & -5-k_2 \\ 1 & 3 \end{pmatrix}}_{A_k} X$$

Now, Characteristic eq<sup>n</sup>

$$|sI - A_k| = 0$$

$$\begin{vmatrix} s+1+k_1 & +5+k_2 \\ -1 & s-3 \end{vmatrix} = 0$$

$$s^2 + s(1+k_1) - 3s - 3(1+k_1) + (5+k_2) = 0$$

$$s^2 + s(k_1 - 2) + (2 - 3k_1 + k_2) = 0 \quad \text{--- (2)}$$

Comparing eq (1) and (2)

$$2 = k_1 - 2$$

$$\boxed{4 = k_1}$$

$$2 - 3 \cdot 4 + k_2 = 2$$

$$\boxed{k_2 = 12}$$