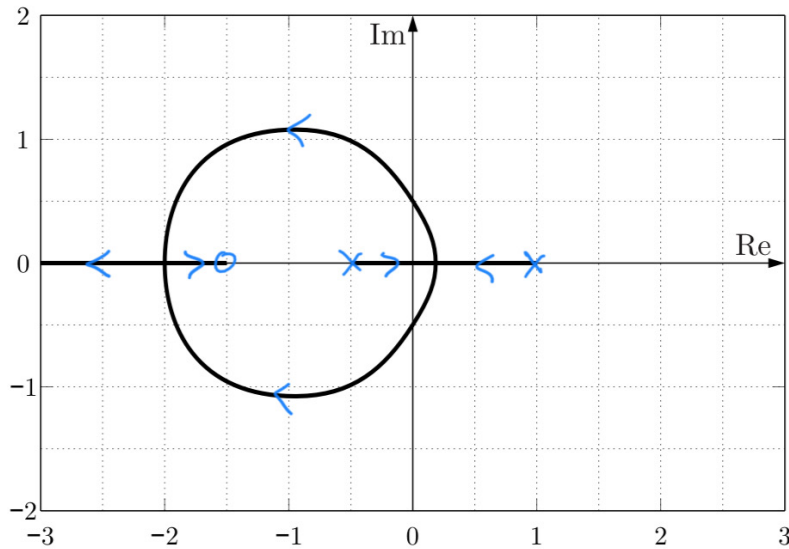


Exam 02-Control Engineering

Monday, 10. February 2020 15:45

1.

a



b

From figure,

$$s_{N_1} = -1.5$$

$$s_{p_1} = -0.5 \quad s_{p_2} = 1$$

Now,

$$G_0(s) = 1 \cdot \frac{(s + 1.5)}{(s + 0.5)(s - 1)}$$

$$= \frac{1}{2} \left\{ \frac{2s + 3}{(s + 0.5)(s - 1)} \right\}$$

$$= \frac{1}{2} \frac{(s + 0.5) + (s - 1) + 3.5}{(s + 0.5)(s - 1)}$$

$$= \frac{1}{2} \left\{ \frac{1}{s - 1} + \frac{1}{s + 0.5} + \frac{3.5}{(s + 0.5)(s - 1)} \right\}$$

$$= \frac{1}{2} \left\{ \frac{-1}{1 - s} + \frac{2}{1 + 2s} + \frac{-7}{(1 + 2s)(1 - s)} \right\}$$

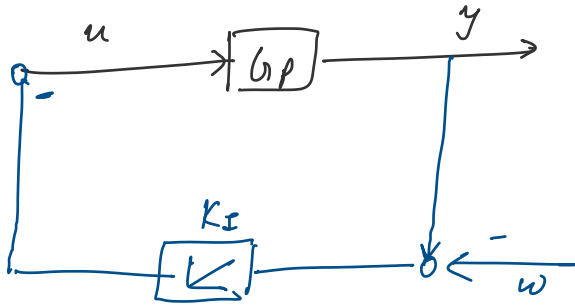
$$g(t) = \mathcal{L}^{-1}\{G_0(s)\}(t)$$

$$= \frac{1}{2} \left\{ +1 e^t + \frac{2}{2} e^{-\frac{t}{2}} + \frac{-7}{3} (e^{-\frac{t}{2}} - e^t) \right\}$$

$$= \frac{1}{2} \left\{ e^t \left(\frac{10}{3} \right) + e^{-\frac{t}{2}} \left(\frac{-4}{3} \right) \right\}$$

$$g(t) = \frac{5}{3} e^{-t} - \frac{2}{3} e^{-2t}$$

2.
a



- b The $G_I(j\omega)$ controller always show integrating behaviour. \therefore It is preferable for stationary accuracy

From Bode plot:

$\alpha_{PD} > \alpha_I$
 $\therefore G_{PD}(j\omega)$ is preferable w.r.t phase margin

c) \therefore General form of $G_p = \frac{1}{T_1 T_2 s^2 + (T_2 + T_1)s + 1}$

with slope changes of -1 at $\frac{1}{T_1}$ and $\frac{1}{T_2}$

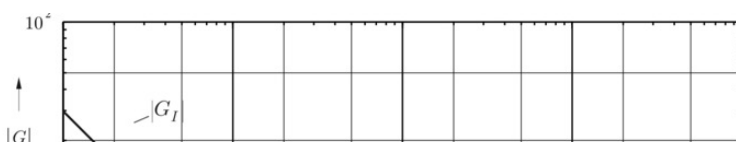
From BODE Plot and comparing it w/ given $G_p(j\omega)$

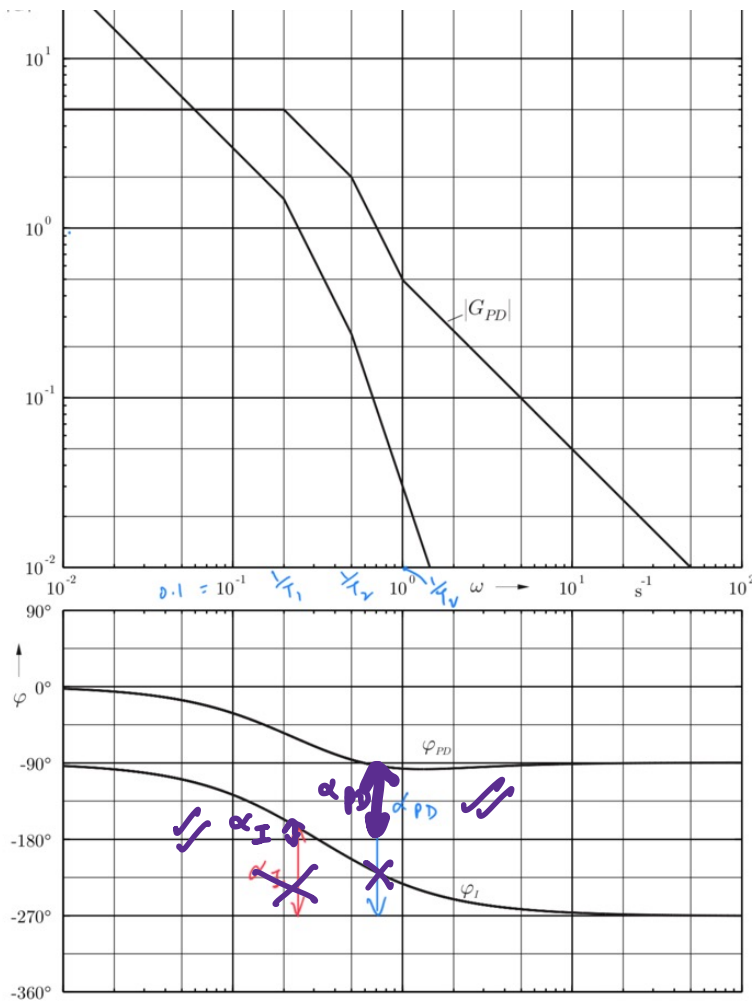
$$\frac{1}{T_1} = 0.2 \text{ sec}^{-1} \Rightarrow T_1 = 5 \text{ sec}$$

$$\frac{1}{T_2} = 0.5 \text{ sec}^{-1} \Rightarrow T_2 = 2 \text{ sec}$$

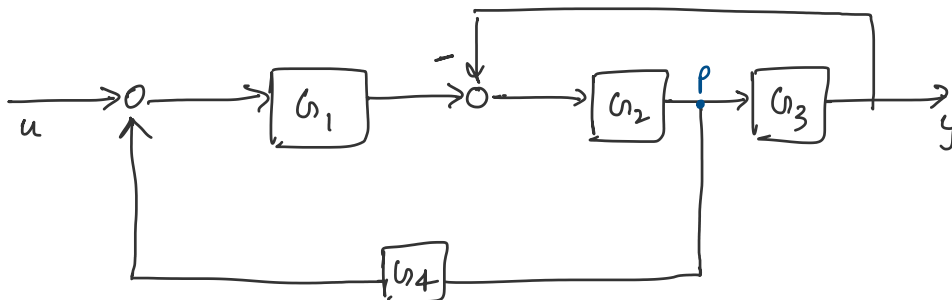
$$\therefore a_2 = T_1 T_2 = 10 \text{ sec}^2 \quad \checkmark$$

$$a_1 = (T_1 + T_2) = 7 \text{ sec} \quad \checkmark$$





3



$$\{(U + G_4 P) G_1 - Y\} G_2 G_3 = Y$$

$$\text{also, } Y = G_3 P \Rightarrow P = \frac{Y}{G_3}$$

$$\text{Now, } \left\{ \left(U + \frac{G_4}{G_3} Y \right) G_1 - Y \right\} G_2 G_3 = Y$$

$$Y (1 + G_2 G_3 - G_1 G_2 G_4) = U G_2 G_3 G_1$$

$$G(s) = \frac{Y(s)}{U(s)} = \frac{G_2 G_3 G_1}{1 + G_2 G_3 - G_1 G_2 G_4}$$

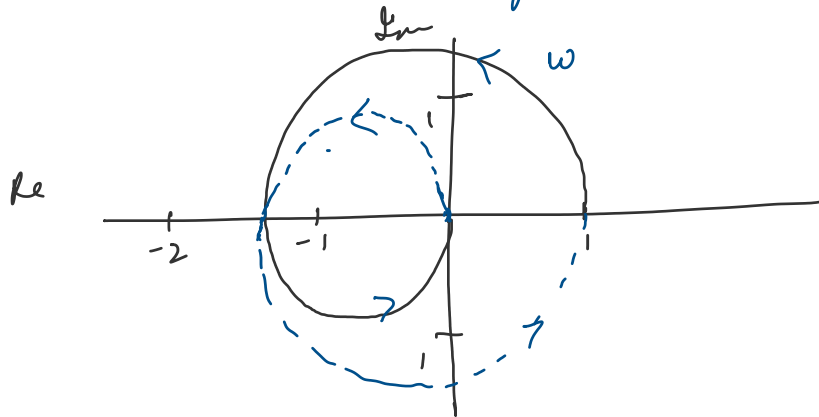
$$b. \quad G(s) = \frac{G_2 G_3}{1 + G_2 G_3}$$

here, $G_0(s) = G_2 G_3$. (say)

Now,

$\therefore G_2$ and G_3 both have one pole in right half plane.

\therefore Total poles in right half plane $= 2 = p$



\therefore 2 rounds anticlockwise about -1

$\therefore m = -2$, no. of revolutions of C'' around -1 (opposite mathematically to direction i.e. clockwise)

Now, $m = n - p$

$$n = m + p = -2 + 2 = 0$$

\therefore there are 0 zeros in $N(s)$ inside C (right s halfplane)

or 0 no. of poles of $G_2(s)$ in right s half plane. or C''

and, $G_2(s)$ is stable if $n \rightarrow 0$

\therefore In our case $G(s)$ is stable

4)

a) $|sI - A| = 0$ For characteristic eqⁿ

$$\Rightarrow \begin{vmatrix} s+1 & 5 \\ -1 & s-3 \end{vmatrix} = 0$$

$$\Rightarrow s^2 + s - 3s - 3 + 5 = 0$$

$$s^2 - 2s + 2 = 0$$

$$\cancel{(s-1)^2 = 0} \quad s = \frac{-2 \pm \sqrt{4-8}}{2} = -1 \pm i$$

\therefore poles lie in left half plane
 \therefore system is stable.

$$b) \quad s_1 = -1 + j \quad s_2 = -1 - j$$

characteristic eqⁿ

$$(s-s_1)(s-s_2) = 0$$

$$\{s - (-1+j)\} \{s + 1+j\} = 0$$

$$\{s+1-j\} \{s+1+j\} = 0$$

$$(s+1)^2 - j^2 = 0$$

$$s^2 + 1 + 2s + 1 = 0$$

$$s^2 + 2s + 2 = 0 \quad \text{————— (1)}$$

Now, For given value of $U = -KX$

$$\dot{X} = AX + B - KX$$

$$= (A - BK)X$$

$$= \left\{ \begin{pmatrix} -1 & -5 \\ 1 & 3 \end{pmatrix} - \begin{pmatrix} k_1 & k_2 \\ 0 & 0 \end{pmatrix} \right\} X$$

$$= \underbrace{\begin{pmatrix} -1-k_1 & -5-k_2 \\ 1 & 3 \end{pmatrix}}_{A_k} X$$

Now, characteristic eqⁿ

$$|sI - A_k| = 0$$

$$\begin{vmatrix} s+1+k_1 & +5+k_2 \\ -1 & s-3 \end{vmatrix} = 0$$

$$s^2 + s(1+k_1) - 3s - 3(1+k_1) + (5+k_2) = 0$$

$$s^2 + s(k_1 - 2) + (2 - 3k_1 + k_2) = 0 \quad \text{————— (2)}$$

Comparing eq ① and ②

$$2 = k_1 - 2$$
$$\boxed{4 = k_1}$$

$$2 - 3 \cdot 4 + k_2 = 2$$

$$\boxed{k_2 = 12}$$