

Solutions to Quiz-2 Paper (Group-B)

IEC103

Q1. Analyse the phase shift oscillator shown in Fig. Q1

Obtain the values of C and R_F for oscillation frequency of 10 KHz using $R = 10\text{ K}\Omega = R_1$.

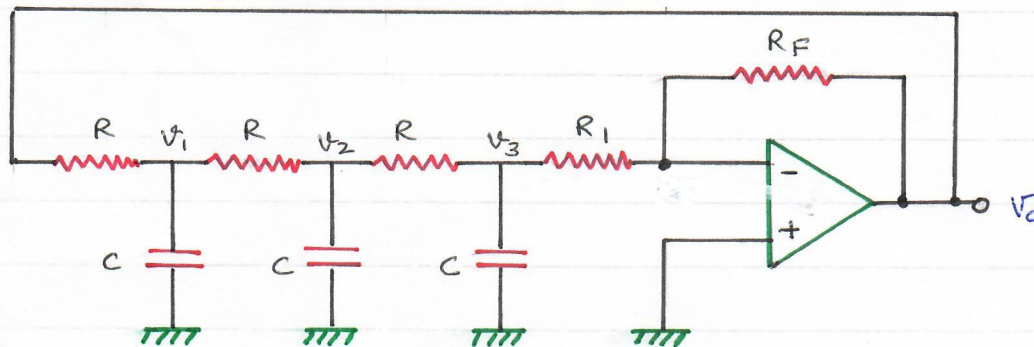
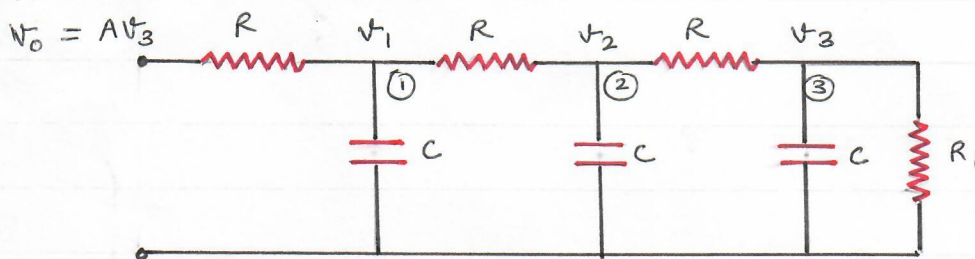


Fig. Q1

Sol.



Applying KCL at node

$$\frac{V_1 - AV_3}{R} + \frac{V_1}{(1/sC)} + \frac{V_1 - V_2}{R} = 0 \quad \dots (i)$$

$$\frac{V_2 - V_1}{R} + \frac{V_2}{(1/sC)} + \frac{V_2 - V_3}{R} = 0 \quad \dots (ii)$$

$$\frac{V_3 - V_2}{R} + \frac{V_3}{(1/sC)} + \frac{V_3}{R_1} = 0 \quad \dots (iii)$$

$$V_1 \left(\frac{1}{R} + sC + \frac{1}{R} \right) - \frac{1}{R} V_2 - \frac{A}{R} V_3 = 0 \quad \dots (i)$$

$$-\frac{1}{R} V_1 + \left(\frac{1}{R} + sC + \frac{1}{R} \right) V_2 - \frac{1}{R} V_3 = 0 \quad \dots (ii)$$

$$-\frac{1}{R} V_2 + \left(\frac{1}{R} + sC + \frac{1}{R_1} \right) V_3 = 0 \quad \dots (iii) \quad R_1 = R$$

Let $\frac{1}{R} = G$

$$V_1(2G+3C) - GV_2 - AGV_3 = 0 \quad \dots (i)$$

$$-GV_1 + (2G+5C)V_2 - GV_3 = 0 \quad \dots (ii)$$

$$-GV_2 + (2G+5C)V_3 = 0 \quad \dots (iii)$$

Eq(iii) $GV_2 = (2G+5C)V_3$

$$V_2 = R(2G+5C)V_3 \quad \dots (iv)$$

Substituting

the value of V_2 from the above equation in (ii)

$$-GV_1 + (2G+5C)R(2G+5C)V_3 - GV_3 = 0$$

$$\Rightarrow GV_1 = [(2G+5C)R(2G+5C) - G]V_3$$

$$\Rightarrow V_1 = R[(2G+5C)^2R - G]V_3 \quad \dots (v)$$

Substituting the expression for V_1 in eq (v) & V_2 in eq (iv) in eqn.(i)

$$(2G+5C)R[(2G+5C)^2R - G]V_3 - GR(2G+5C)V_3 - AGV_3 = 0$$

$$\Rightarrow (2G+5C)R[(2G+5C)^2R - G] - (2G+5C) - AG = 0$$

$$(2G+5C)[R^2(2G+5C)^2 - 1] - (2G+5C) - AG = 0$$

$$(2G+5C)[R^2(2G+5C)^2 - 1 - 1] - AG = 0$$

$$(2G+5C)[R^2(4G^2 + 5^2C^2 + 4G5C) - 2] - AG = 0$$

$$\Rightarrow (2G+5C)[4 + 5^2R^2C^2 + 45RC - 2] - AG = 0$$

$$\Rightarrow (2G+5C)(5^2R^2C^2 + 45RC + 2) - AG = 0$$

$$\Rightarrow (2S^2RC^2 + 95C + 4G + 5^3R^2C^3 + 45^2RC^2 + 25C) - AG = 0$$

put $S = j\omega$

$$-2\omega^2RC^2 + j8\omega C + 4G - j\omega^3R^2C^3 - 4\omega^2RC^2 + j2\omega C - AG = 0$$

$$\Rightarrow (4G - AG - 2\omega^2RC^2 - 4\omega^2RC^2) + j(8\omega C - \omega^3R^2C^3 + 2\omega C) = 0$$

$$\Rightarrow (4G - AG - 6\omega^2RC^2) + j(10\omega C - \omega^3R^2C^3) = 0$$

$$4G - AG - 6W^2 RC^2 = 0$$

$$\& \quad 10WC - W^3 R^2 C^3 = 0$$

$$\Rightarrow W^2 R^2 C^2 = 10$$

$$\Rightarrow \boxed{W = \frac{\sqrt{10}}{RC}}$$

$$4G - AG - 6RC^2 W^2 = 0$$

$$\Rightarrow 4G - AG - 6RC^2 \times \frac{10}{R^2 C^2} = 0$$

$$\Rightarrow 4G - AG - 60G = 0$$

$$\Rightarrow -56G = AG$$

$$\Rightarrow A = -56 = -\frac{R_F}{R_1} = \frac{R_F}{R}$$

$$\therefore \boxed{R_F = 56R}$$

$$R = 1K\Omega \Rightarrow R_F = 56K\Omega$$

$$\begin{aligned} \text{frequency of oscillation required} &= 10 \text{ KHz} = 10 \times 10^3 \text{ Hz} \\ &= 1591.5 \text{ rad/s} \end{aligned}$$

$$\therefore W = 1591.5 = \frac{\sqrt{10}}{RC}$$

$$\Rightarrow C = \frac{\sqrt{10}}{1591.5 \times R} =$$

$$\Rightarrow C = \frac{\sqrt{10}}{1591.5 \times 1000} = 1.987 \mu\text{F}$$

Q2) Determine the output waveform (V_{out}) of the circuit shown in Fig. Q2 shown below. Use simplified model of the diode. The diode is a silicon diode.

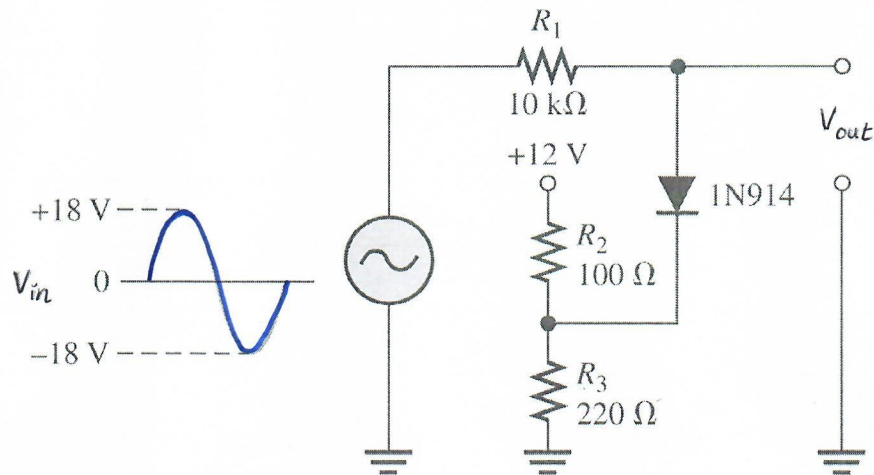
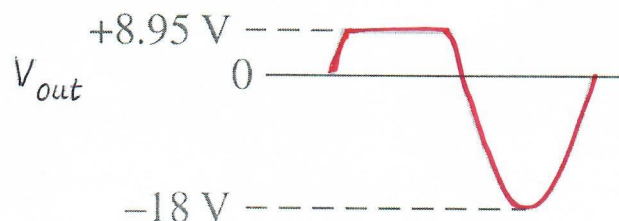


Fig. Q2

Sol. The circuit is a +ve limiter. Use voltage divider formula to determine the bias voltage.

$$V_{BIAS} = \text{Voltage at cathode terminal of diode} = \frac{R_3}{R_2 + R_3} \times 12 = \frac{220}{100 + 220} \times 12 = 8.25 \text{ V}$$

The output voltage waveform is shown in Fig. below. The +ve part of the voltage waveform is limited to $V_{BIAS} + 0.7 = 8.95 \text{ V}$.



Q3. Derive expression for ripple factor of the unregulated power supply using capacitor filter shown in Fig. Q3 below.

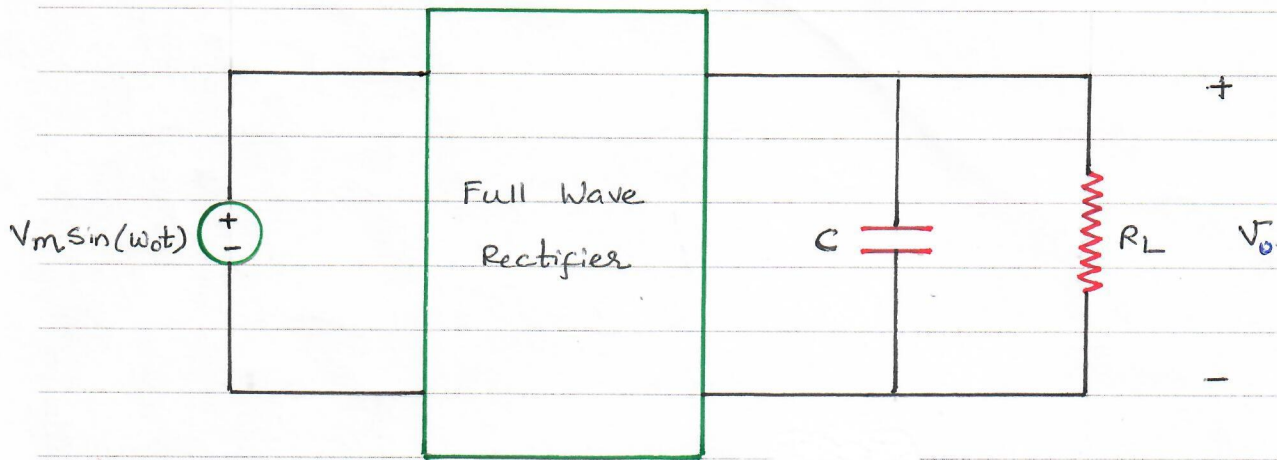


Fig. Q3

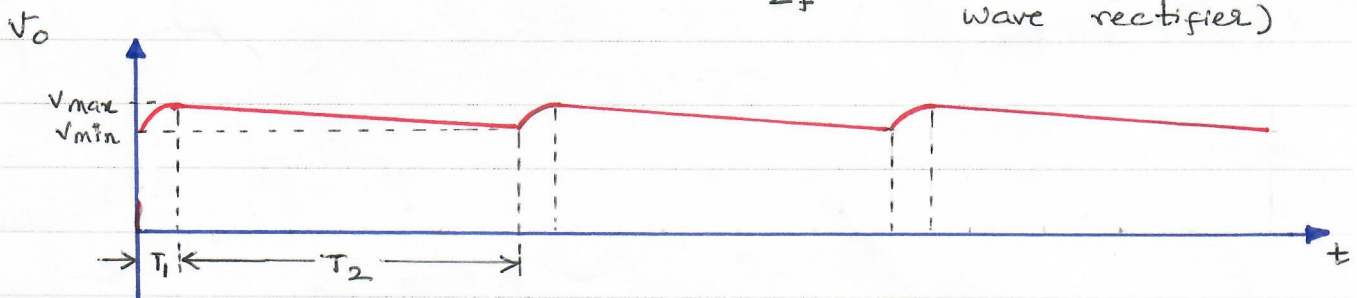
Assume the time constant $R_L C$ is much greater than the time period of the source voltage.

Sol.

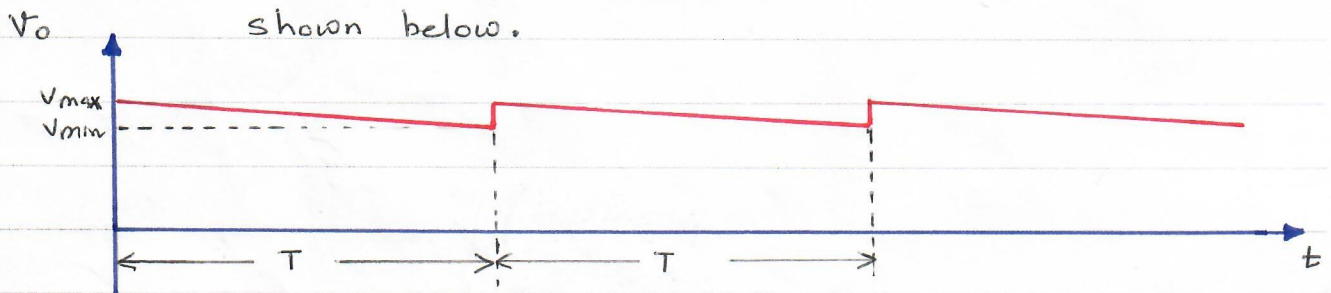
$$\omega = 2\pi f$$

The output waveform will be as shown below.

$$T = T_1 + T_2 = \frac{1}{2f} \quad (\text{since rectifier is a full wave rectifier})$$



$T = T_1 + T_2$ The above waveform can be approximated to waveform shown below.



$$R_L C \gg T$$

$$R_L C \gg T$$

$$V_{\min} = V_{\max} e^{-T/R_L C}$$

$$= V_{\max} \left(1 - \frac{T}{R_L C} + \frac{T^2}{2R_L^2 C^2} - \dots \right)$$

$$\approx V_{\max} \left(1 - \frac{T}{R_L C} \right) \quad \because R_L C \gg T$$

$$\text{Ripple factor} = \frac{\text{RMS value of ripple voltage}}{\text{DC value of the voltage}}$$

$$\begin{aligned} V_{dc} &= \frac{(V_{\max} + V_{\min})}{2} = \frac{1}{2} \left[V_{\max} + V_{\max} \left(1 - \frac{T}{R_L C} \right) \right] \\ &= \frac{1}{2} \left[V_{\max} + V_{\max} - V_{\max} \frac{T}{R_L C} \right] \\ &= \frac{1}{2} \left[2V_{\max} - V_{\max} \frac{T}{R_L C} \right] \\ &= V_{\max} \left(1 - \frac{T}{2R_L C} \right) \end{aligned}$$

$$\begin{aligned} \text{Peak-to-peak to ripple voltage} &= V_{\max} - V_{\min} \quad \dots = V_{p-p} \\ &= V_{\max} - V_{\max} \left(1 - \frac{T}{R_L C} \right) \end{aligned}$$

$$= V_{\max} \frac{T}{R_L C} = V_{p-p}$$

$$\text{RMS value of ripple voltage} = \frac{V_{p-p}}{2\sqrt{3}} = \frac{V_{\max} T}{2\sqrt{3} R_L C}$$

$$\begin{aligned} \therefore \gamma &= \frac{V_{\max} T}{2\sqrt{3} R_L C} \\ \therefore \gamma &= \frac{V_{\max} T}{V_{\max} \left(1 - \frac{T}{R_L C} \right)} = \frac{T}{2\sqrt{3} R_L C \left(1 - \frac{T}{R_L C} \right)} = \frac{T}{2\sqrt{3} (R_L C - T)} \\ &\approx \frac{T}{2\sqrt{3} R_L C} = \frac{1}{4\sqrt{3} f R_L C} \\ &\because T = \frac{1}{2f} \end{aligned}$$

$$\therefore T = \frac{1}{2f}$$