

**OPA337**  
**OPA2337**  
**OPA338**  
**OPA2338**

For most current data sheet and other product information, visit [www.burr-brown.com](http://www.burr-brown.com)

## MicroSIZE, Single-Supply CMOS OPERATIONAL AMPLIFIERS

### MicroAmplifier™ Series

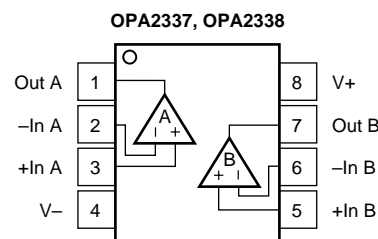
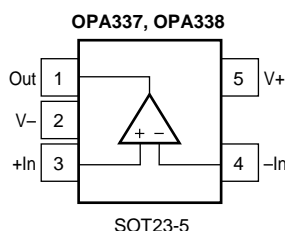
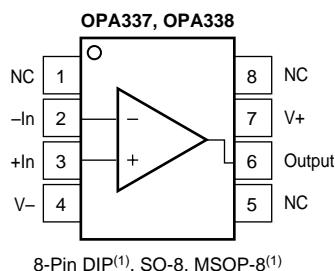
## FEATURES

- **MicroSIZE PACKAGES:**
  - SOT23-5
  - SOT23-8
- **SINGLE-SUPPLY OPERATION**
- **RAIL-TO-RAIL OUTPUT SWING**
- **FET-INPUT:  $I_B = 10\text{pA max}$**
- **HIGH SPEED:**
  - OPA337: 3MHz,  $1.2\text{V}/\mu\text{s}$  ( $G = 1$ )
  - OPA338: 12.5MHz,  $4.6\text{V}/\mu\text{s}$  ( $G = 5$ )
- **OPERATION FROM 2.5V to 5.5V**
- **HIGH OPEN-LOOP GAIN: 120dB**
- **LOW QUIESCENT CURRENT:  $525\mu\text{A/amp}$**
- **SINGLE AND DUAL VERSIONS**

## APPLICATIONS

- **BATTERY-POWERED INSTRUMENTS**
- **PHOTODIODE PRE-AMPS**
- **MEDICAL INSTRUMENTS**
- **TEST EQUIPMENT**
- **AUDIO SYSTEMS**
- **DRIVING ADCs**
- **CONSUMER PRODUCTS**

SPICE Model available at [www.burr-brown.com](http://www.burr-brown.com)



NOTE: (1) DIP AND MSOP-8 versions for OPA337, OPA2337 only.

## DESCRIPTION

The OPA337 and OPA338 series rail-to-rail output CMOS operational amplifiers are designed for low cost and miniature applications. Packaged in the new SOT23-8, the OPA2337EA and OPA2338EA are Burr-Brown's *smallest* dual op amps. At 1/4 the size of a conventional SO-8 surface mount, they are ideal for space-sensitive applications.

Utilizing advanced CMOS technology, OPA337 and OPA338 op amps provide low bias current, high-speed operation, high open-loop gain, and rail-to-rail output swing. They operate on a single supply with operation as low as 2.5V while drawing only  $525\mu\text{A}$  quiescent current. In addition, the input common-mode voltage range includes ground—ideal for single-supply operation.

The OPA337 series is unity-gain stable. The OPA338 series is optimized for gains greater than or equal to five. They are easy to use and free from phase inversion and overload problems found in some other op amps. Excellent performance is maintained as the amplifiers swing to their specified limits. The dual versions feature completely independent circuitry for lowest crosstalk and freedom from interaction, even when overdriven or overloaded.

PACKAGE	G = 1 STABLE		G ≥ 5 STABLE	
	SINGLE OPA337	DUAL OPA2337	SINGLE OPA338	DUAL OPA2338
SOT23-5	✓		✓	
SOT23-8		✓		✓
MSOP-8	✓			
SO-8	✓	✓	✓	✓
DIP-8	✓	✓		

# SPECIFICATIONS: $V_S = 2.7V$ to $5.5V$

At  $T_A = +25^\circ C$ , and  $R_L = 25k\Omega$  connected to  $V_S/2$ , unless otherwise noted.

**Boldface** limits apply over the specified temperature range,  $-40^\circ C$  to  $+85^\circ C$ ,  $V_S = 5V$ .

PARAMETER	CONDITION	OPA337NA, EA, UA, PA OPA2337EA, UA, PA OPA338NA, UA OPA2338EA, UA			UNITS
		MIN	TYP <sup>(1)</sup>	MAX	
<b>OFFSET VOLTAGE</b> Input Offset Voltage $V_{OS}$ $T_A = -40^\circ C$ to $+85^\circ C$ vs Temperature $dV_{OS}/dT$ vs Power Supply Rejection Ratio PSRR $T_A = -40^\circ C$ to $+85^\circ C$ Channel Separation (dual versions)	$V_S = 2.7V$ to $5.5V$ $V_S = 2.7V$ to $5.5V$ dc		$\pm 0.5$ $\pm 2$ 25 0.3	$\pm 3$ <b><math>\pm 3.5</math></b> 125 <b>125</b>	mV mV $\mu V/^\circ C$ $\mu V/V$ $\mu V/V$
<b>INPUT BIAS CURRENT</b> Input Bias Current $I_B$ $T_A = -40^\circ C$ to $+85^\circ C$ Input Offset Current $I_{OS}$			$\pm 0.2$ See Typical Curve $\pm 0.2$	$\pm 10$ $\pm 10$	pA pA
<b>NOISE</b> Input Voltage Noise, $f = 0.1Hz$ to $10Hz$ Input Voltage Noise Density, $f = 1kHz$ $e_n$ Current Noise Density, $f = 1kHz$ $i_n$			6 26 0.6		$\mu Vp-p$ $nV/\sqrt{Hz}$ $fA/\sqrt{Hz}$
<b>INPUT VOLTAGE RANGE</b> Common-Mode Voltage Range $V_{CM}$ Common-Mode Rejection Ratio CMRR $T_A = -40^\circ C$ to $+85^\circ C$	$T_A = -40^\circ C$ to $+85^\circ C$ $-0.2V < V_{CM} < (V+) - 1.2V$ $-0.2V < V_{CM} < (V+) - 1.2V$	<b>-0.2</b> 74 <b>74</b>	90	<b>(V+) - 1.2</b>	V dB dB
<b>INPUT IMPEDANCE</b> Differential Common-Mode			$10^{13} \parallel 2$ $10^{13} \parallel 4$		$\Omega \parallel pF$ $\Omega \parallel pF$
<b>OPEN-LOOP GAIN</b> Open-Loop Voltage Gain $A_{OL}$ $T_A = -40^\circ C$ to $+85^\circ C$ $T_A = -40^\circ C$ to $+85^\circ C$	$R_L = 25k\Omega$ , $125mV < V_O < (V+) - 125mV$ $R_L = 25k\Omega$ , $125mV < V_O < (V+) - 125mV$ $R_L = 5k\Omega$ , $500mV < V_O < (V+) - 500mV$ $R_L = 5k\Omega$ , $500mV < V_O < (V+) - 500mV$	100 <b>100</b> 100 <b>100</b>	120  114		dB dB dB dB
<b>OPA337 FREQUENCY RESPONSE</b> Gain-Bandwidth Product GBW Slew Rate SR Settling Time: 0.1% 0.01% Overload Recovery Time Total Harmonic Distortion + Noise THD+N	$V_S = 5V$ , $G = 1$ $V_S = 5V$ , $G = 1$ $V_S = 5V$ , 2V Step, $C_L = 100pF$ , $G = 1$ $V_S = 5V$ , 2V Step, $C_L = 100pF$ , $G = 1$ $V_{IN} \cdot G = V_S$ $V_S = 5V$ , $V_O = 3Vp-p$ , $G = 1$ , $f = 1kHz$		3 1.2 2 2.5 2 0.001		MHz V/ $\mu s$ $\mu s$ $\mu s$ $\mu s$ %
<b>OPA338 FREQUENCY RESPONSE</b> Gain-Bandwidth Product GBW Slew Rate SR Settling Time: 0.1% 0.01% Overload Recovery Time Total Harmonic Distortion + Noise THD+N	$V_S = 5V$ , $G = 5$ $V_S = 5V$ , $G = 5$ $V_S = 5V$ , 2V Step, $C_L = 100pF$ , $G = 5$ $V_S = 5V$ , 2V Step, $C_L = 100pF$ , $G = 5$ $V_{IN} \cdot G = V_S$ $V_S = 5V$ , $V_O = 3Vp-p$ , $G = 5$ , $f = 1kHz$		12.5 4.6 1.4 1.9 0.5 0.0035		MHz V/ $\mu s$ $\mu s$ $\mu s$ $\mu s$ %
<b>OUTPUT</b> Voltage Output Swing from Rail <sup>(2)</sup> $T_A = -40^\circ C$ to $+85^\circ C$ $T_A = -40^\circ C$ to $+85^\circ C$ Short-Circuit Current $I_{SC}$ Capacitive Load Drive $C_{LOAD}$	$R_L = 25k\Omega$ , $A_{OL} \geq 100dB$ $R_L = 25k\Omega$ , $A_{OL} \geq 100dB$ $R_L = 5k\Omega$ , $A_{OL} \geq 100dB$ $R_L = 5k\Omega$ , $A_{OL} \geq 100dB$		40  150  $\pm 9$	125 <b>125</b> 500 <b>500</b>	mV mV mV mV mA
<b>POWER SUPPLY</b> Specified Voltage Range $V_S$ Minimum Operating Voltage Quiescent Current (per amplifier) $I_Q$ $T_A = -40^\circ C$ to $+85^\circ C$	$T_A = -40^\circ C$ to $+85^\circ C$  $I_O = 0$ $I_O = 0$	<b>2.7</b>	2.5 0.525	<b>5.5</b> 1 <b>1.2</b>	V V mA mA

The information provided herein is believed to be reliable; however, BURR-BROWN assumes no responsibility for inaccuracies or omissions. BURR-BROWN assumes no responsibility for the use of this information, and all use of such information shall be entirely at the user's own risk. Prices and specifications are subject to change without notice. No patent rights or licenses to any of the circuits described herein are implied or granted to any third party. BURR-BROWN does not authorize or warrant any BURR-BROWN product for use in life support devices and/or systems.

# SPECIFICATIONS: $V_S = 2.7V$ to $5.5V$ (Cont.)

At  $T_A = +25^\circ C$ , and  $R_L = 25k\Omega$  connected to  $V_S/2$ , unless otherwise noted.

**Boldface** limits apply over the specified temperature range,  $-40^\circ C$  to  $+85^\circ C$ ,  $V_S = 5V$ .

PARAMETER	CONDITION	OPA337NA, EA, UA, PA OPA2337EA, UA, PA OPA338NA, UA OPA2338EA, UA			UNITS
		MIN	TYP	MAX	
<b>TEMPERATURE RANGE</b>					
Specified Range		-40		+85	$^\circ C$
Operating Range		-55		+125	$^\circ C$
Storage Range		-55		+125	$^\circ C$
Thermal Resistance	$\theta_{JA}$				
SOT23-5 Surface Mount			200		$^\circ C/W$
SOT23-8 Surface Mount			200		$^\circ C/W$
MSOP-8			150		$^\circ C/W$
SO-8 Surface Mount			150		$^\circ C/W$
8-Pin DIP			100		$^\circ C/W$

NOTES: (1)  $V_S = 5V$ . (2) Output voltage swings are measured between the output and negative and positive power supply rails.

## ABSOLUTE MAXIMUM RATINGS<sup>(1)</sup>

Supply Voltage .....	5.5V
Input Voltage <sup>(2)</sup> .....	(V-) -0.5V to (V+) +0.5V
Input Current <sup>(2)</sup> .....	10mA
Output Short Circuit <sup>(3)</sup> .....	Continuous
Operating Temperature .....	$-55^\circ C$ to $+125^\circ C$
Storage Temperature .....	$-55^\circ C$ to $+125^\circ C$
Junction Temperature .....	$150^\circ C$
Lead Temperature (soldering, 10s) .....	$300^\circ C$

NOTES: (1) Stresses above these ratings may cause permanent damage. Exposure to absolute maximum ratings for extended periods may degrade device reliability. (2) Input signal voltage is limited by internal diodes connected to power supplies. See text. (3) Short circuit to ground, one amplifier per package.



## ELECTROSTATIC DISCHARGE SENSITIVITY

This integrated circuit can be damaged by ESD. Burr-Brown recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

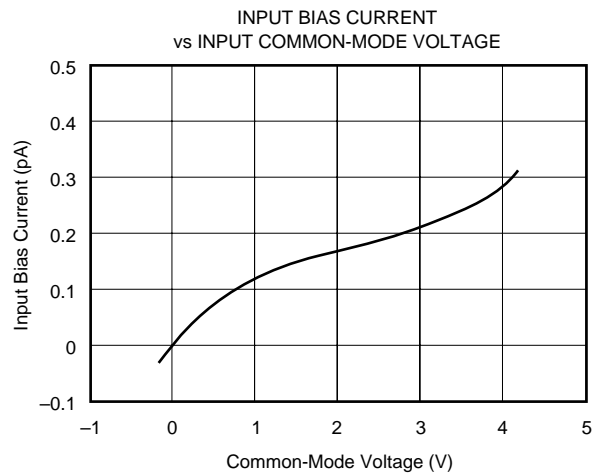
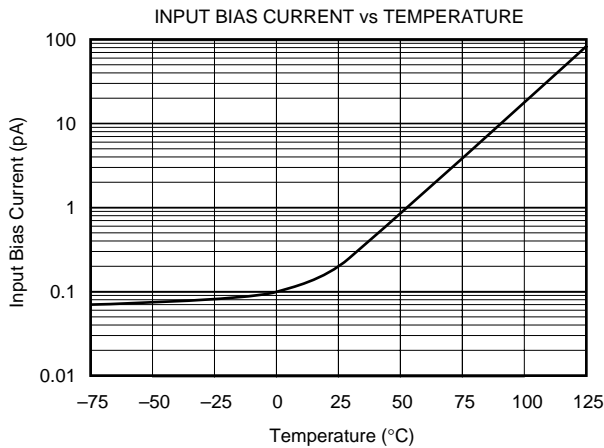
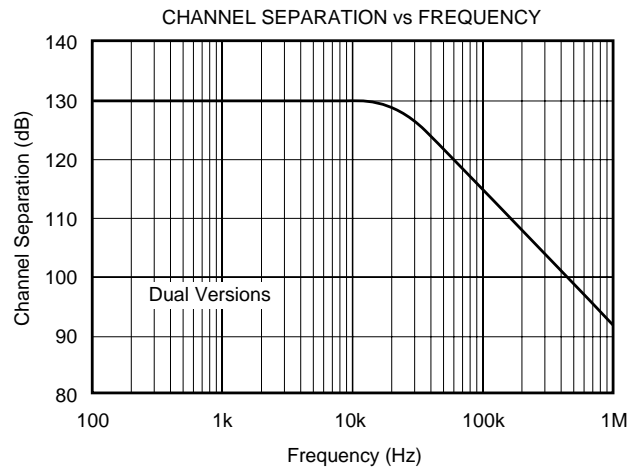
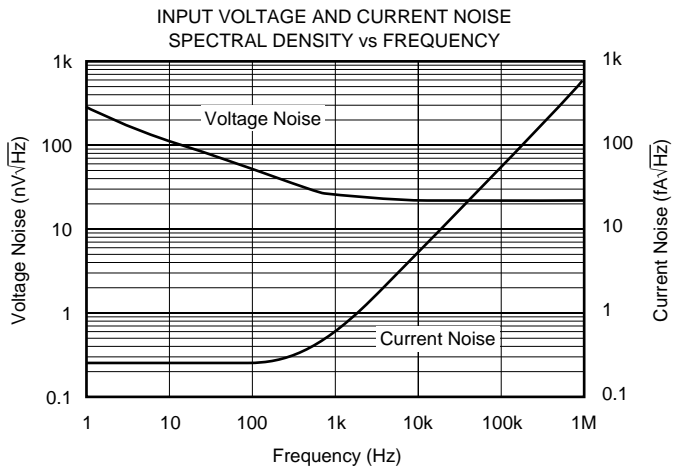
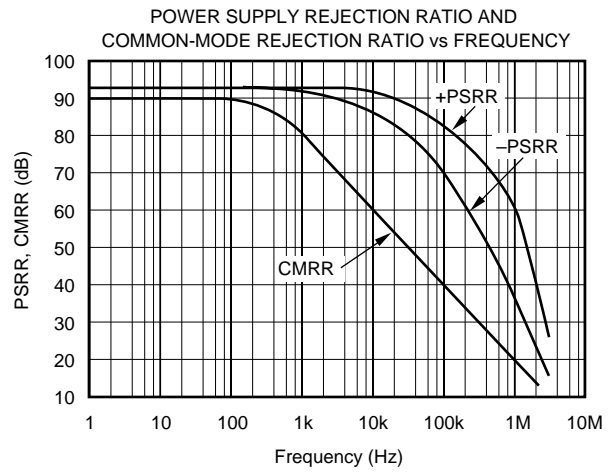
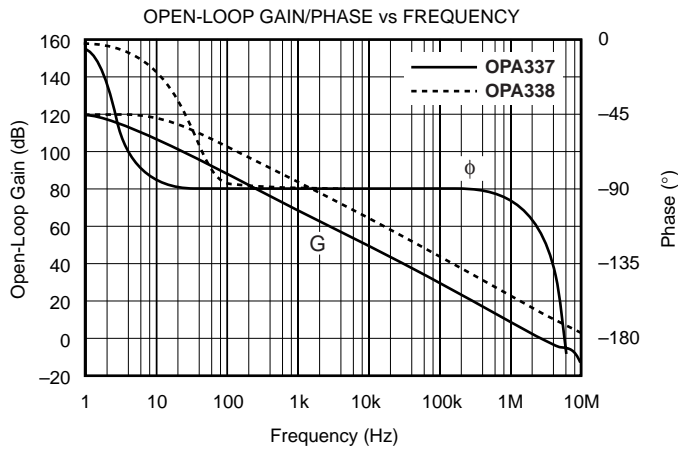
## PACKAGE/ORDERING INFORMATION

PRODUCT	DESCRIPTION	PACKAGE	PACKAGE DRAWING NUMBER	SPECIFIED TEMPERATURE RANGE	PACKAGE MARKING	ORDERING NUMBER <sup>(1)</sup>	TRANSPORT MEDIA
<b>OPA337 Series</b>							
OPA337NA	Single, G = 1 Stable	5-Lead SOT23-5	331	$-40^\circ C$ to $+85^\circ C$	C37	OPA337NA/250	Tape and Reel
"	"	"	"	"	"	OPA337NA/3K	Tape and Reel
OPA337EA	Single, G = 1 Stable	MSOP-8	337	$-40^\circ C$ to $+85^\circ C$	G37	OPA337EA/250	Tape and Reel
"	"	"	"	"	"	OPA337EA/2K5	Tape and Reel
OPA337PA	Single, G = 1 Stable	8-Pin DIP	006	$-40^\circ C$ to $+85^\circ C$	OPA337PA	OPA337PA	Rails
OPA337UA	Single, G = 1 Stable	SO-8 Surface Mount	182	$-40^\circ C$ to $+85^\circ C$	OPA337UA	OPA337UA	Rails
"	"	"	"	"	"	OPA337UA/2K5	Tape and Reel
OPA2337EA	Dual, G = 1 Stable	8-Lead SOT23-8	348	$-40^\circ C$ to $+85^\circ C$	A7	OPA2337EA/250	Tape and Reel
"	"	"	"	"	"	OPA2337EA/3K	Tape and Reel
OPA2337PA	Dual, G = 1 Stable	8-Pin DIP	006	$-40^\circ C$ to $+85^\circ C$	OPA2337PA	OPA2337PA	Rails
OPA2337UA	Dual, G = 1 Stable	SO-8 Surface Mount	182	$-40^\circ C$ to $+85^\circ C$	OPA2337UA	OPA2337UA	Rails
"	"	"	"	"	"	OPA2337UA/2K5	Tape and Reel
<b>OPA338 Series</b>							
OPA338NA	Single, G $\geq 5$ Stable	5-Lead SOT23-5	331	$-40^\circ C$ to $+85^\circ C$	A38	OPA338NA/250	Tape and Reel
"	"	"	"	"	"	OPA338NA/3K	Tape and Reel
OPA338UA	Single, G $\geq 5$ Stable	SO-8 Surface Mount	182	$-40^\circ C$ to $+85^\circ C$	OPA338UA	OPA338UA	Rails
"	"	"	"	"	"	OPA338UA/2K5	Tape and Reel
OPA2338EA	Dual, G $\geq 5$ Stable	8-Lead SOT23-8	348	$-40^\circ C$ to $+85^\circ C$	A8	OPA2338EA/250	Tape and Reel
"	"	"	"	"	"	OPA2338EA/3K	Tape and Reel
OPA2338UA	Dual, G $\geq 5$ Stable	SO-8 Surface Mount	182	$-40^\circ C$ to $+85^\circ C$	OPA2338UA	OPA2338UA	Rails
"	"	"	"	"	"	OPA2338UA/2K5	Tape and Reel

NOTES: (1) Models with a slash (/) are available only in Tape and Reel in the quantities indicated (e.g., /2K5 indicates 2500 devices per reel). Ordering 2500 pieces of "OPA2337UA/2K5" will get a single 2500-piece Tape and Reel.

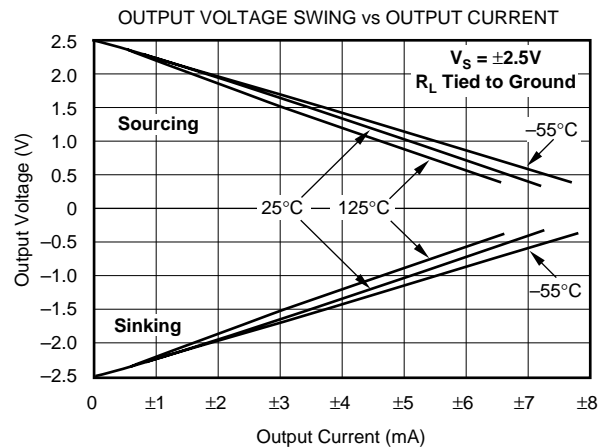
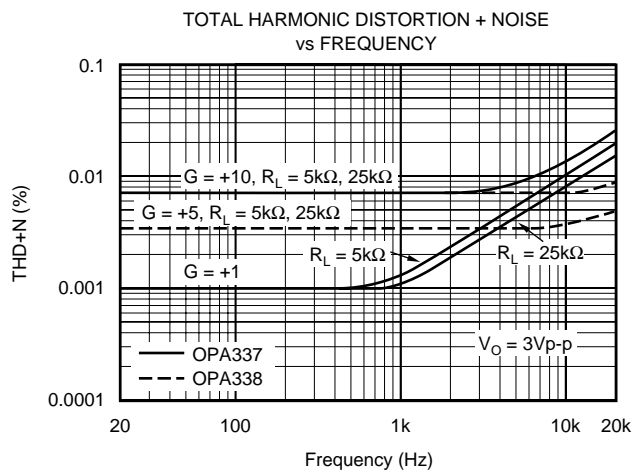
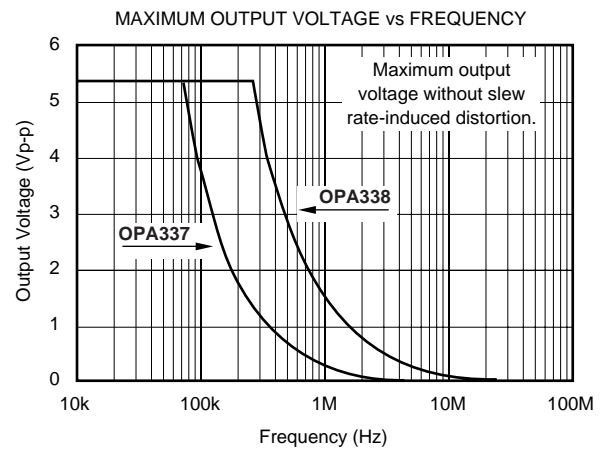
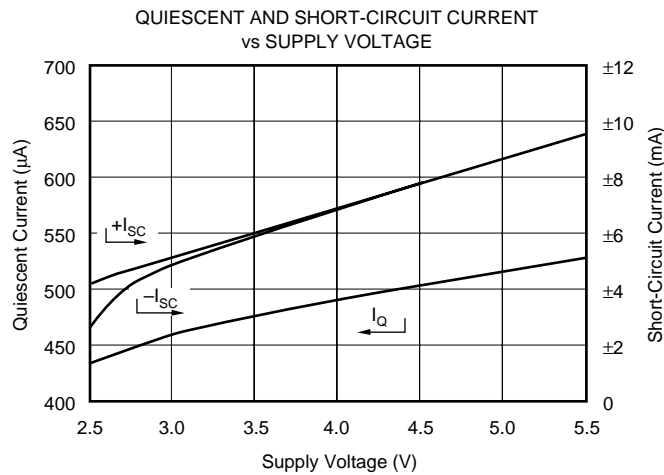
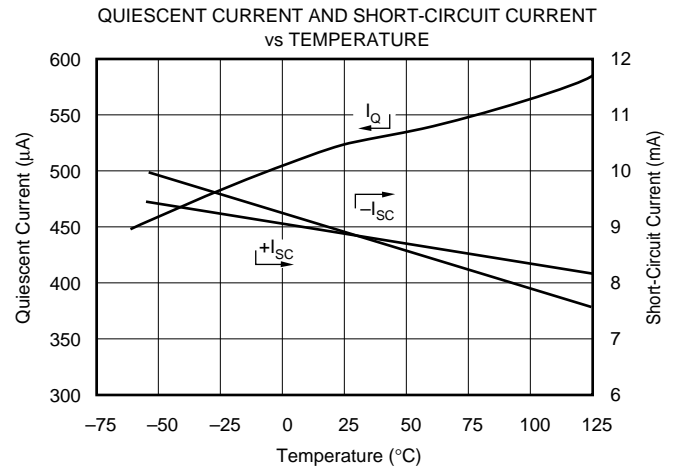
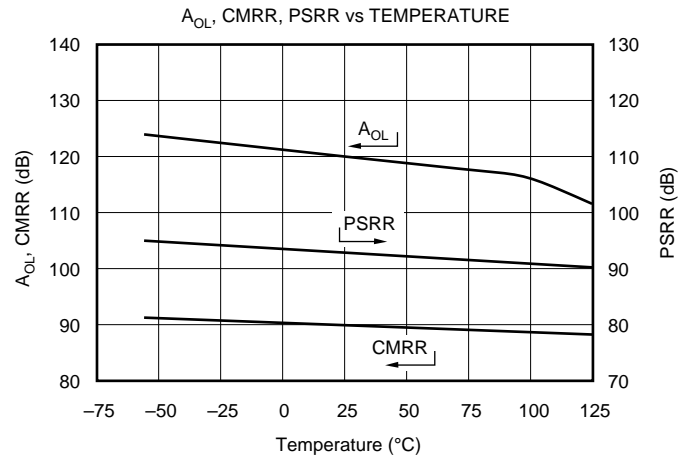
# TYPICAL PERFORMANCE CURVES

At  $T_A = +25^\circ\text{C}$ ,  $V_S = +5\text{V}$ , and  $R_L = 25\text{k}\Omega$  connected to  $V_S/2$ , unless otherwise noted.



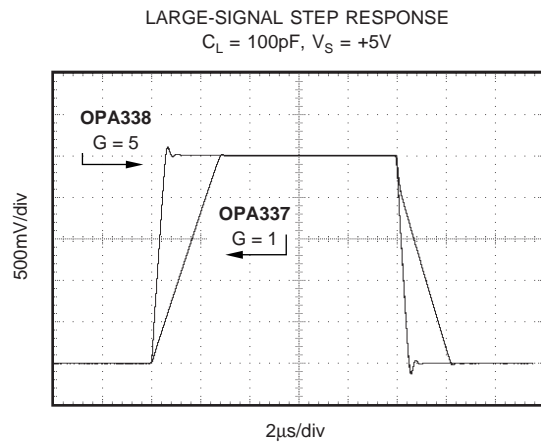
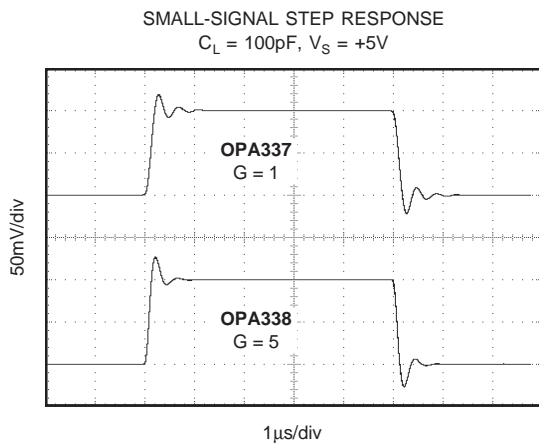
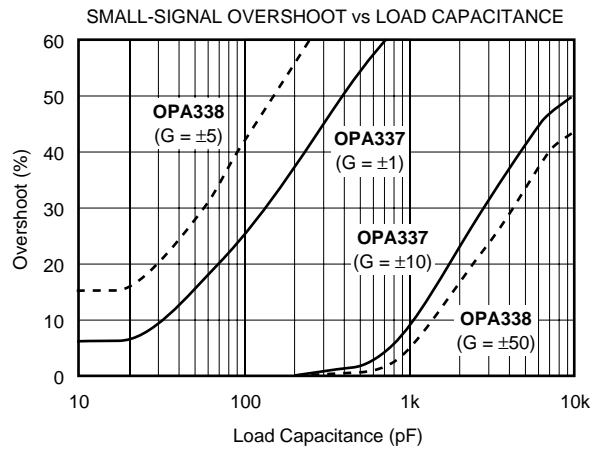
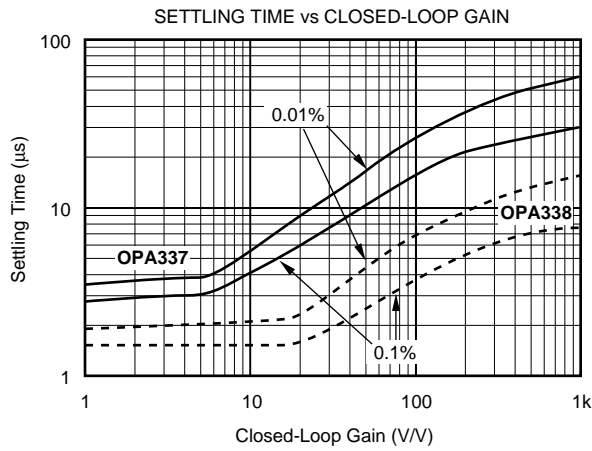
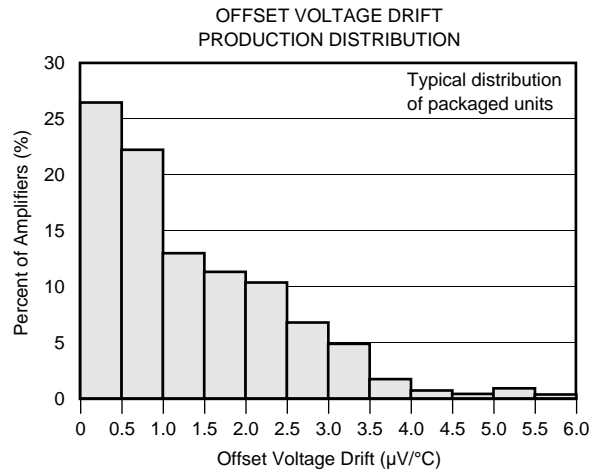
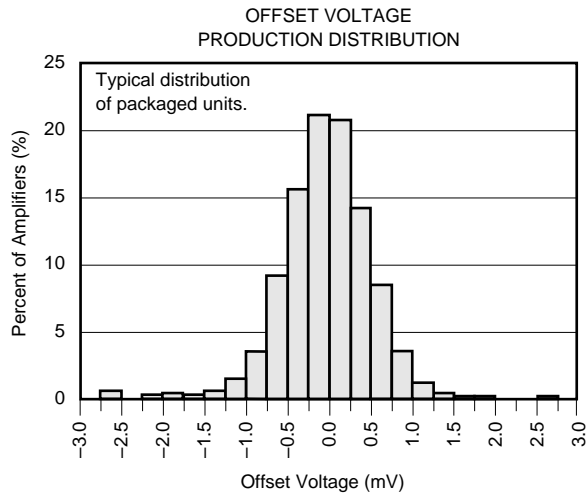
# TYPICAL PERFORMANCE CURVES (Cont.)

At  $T_A = +25^\circ\text{C}$ ,  $V_S = +5\text{V}$ , and  $R_L = 25\text{k}\Omega$  connected to  $V_S/2$ , unless otherwise noted.



# TYPICAL PERFORMANCE CURVES (Cont.)

At  $T_A = +25^\circ\text{C}$ ,  $V_S = +5\text{V}$ , and  $R_L = 25\text{k}\Omega$  connected to  $V_S/2$ , unless otherwise noted.



## APPLICATIONS INFORMATION

The OPA337 series and OPA338 series are fabricated on a state-of-the-art CMOS process. The OPA337 series is unity-gain stable. The OPA338 series is optimized for gains greater than or equal to five. Both are suitable for a wide range of general purpose applications. Power supply pins should be bypassed with 0.01μF ceramic capacitors.

### OPERATING VOLTAGE

The OPA337 series and OPA338 series can operate from a +2.5V to +5.5V single supply with excellent performance. Unlike most op amps which are specified at only one supply voltage, these op amps are specified for real-world applications; a single limit applies throughout the +2.7V to +5.5V supply range. This allows a designer to have the same assured performance at any supply voltage within the specified voltage range. Most behavior remains unchanged throughout the full operating voltage range. Parameters which vary significantly with operating voltage are shown in typical performance curves.

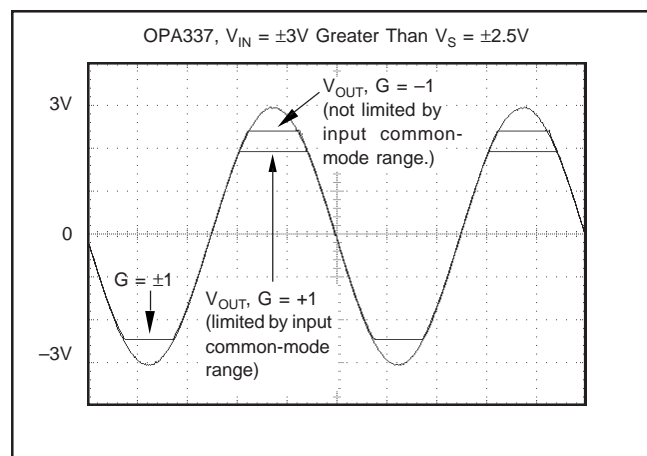


FIGURE 1. OPA337—No Phase Inversion with Inputs Greater than the Power Supply Voltage.

### INPUT VOLTAGE

The input common-mode range extends from (V<sub>-</sub>) - 0.2V to (V<sub>+</sub>) - 1.2V. For normal operation, inputs should be limited to this range. The absolute maximum input voltage is 500mV beyond the supplies. Inputs greater than the input common-mode range but less than maximum input voltage, while not valid, will not cause any damage to the op amp. Furthermore, if input current is limited the inputs may go beyond the power supplies without phase inversion (Figure 1) unlike some other op amps.

Normally, input currents are 0.2pA. However, large inputs (greater than 500mV beyond the supply rails) can cause excessive current to flow in or out of the input pins. Therefore, as well as keeping the input voltage below the maximum rating, it is also important to limit the input current to less than 10mA. This is easily accomplished with an input resistor as shown in Figure 2.

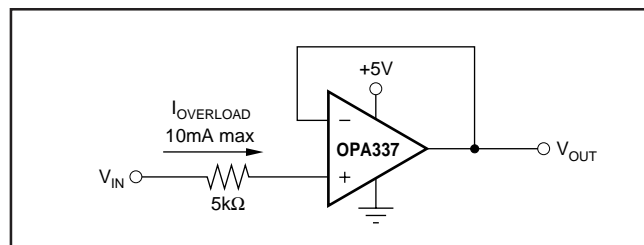


FIGURE 2. Input Current Protection for Voltages Exceeding the Supply Voltage.

### USING THE OPA338 IN LOW GAINS

The OPA338 series is optimized for gains greater than or equal to five. It has significantly wider bandwidth (12.5MHz) and faster slew rate (4.6V/μs) when compared to the OPA337 series. The OPA338 series can be used in lower gain configurations at low frequencies while maintaining its high slew rate with the proper compensation.

Figure 3 shows the OPA338 in a unity-gain buffer configuration. At dc, the compensation capacitor C<sub>1</sub> is effectively “open” resulting in 100% feedback (closed-loop gain = 1). As frequency increases, C<sub>1</sub> becomes lower impedance and closed-loop gain increases, eventually becoming 1 + R<sub>2</sub>/R<sub>1</sub> (in this case five, which is equal to the minimum gain required for stability).

The required compensation capacitor value can be determined from the following equation:

$$C_1 = 1/(2\pi f_C R_1)$$

Since f<sub>C</sub> may shift with process variations, it is recommended that a value less than f<sub>C</sub> be used for determining C<sub>1</sub>. With f<sub>C</sub> = 1MHz and R<sub>1</sub> = 2.5kΩ, the compensation capacitor is about 68pF.

The selection of the compensation capacitor C<sub>1</sub> is important. A proper value ensures that the closed-loop circuit gain is greater than or equal to five at high frequencies. Referring to the “Open-Loop Gain vs Frequency” plot in the Typical Performance Curves section, the OPA338 gain line (dashed in the curve) has a constant slope (-20dB/decade) up to approximately 3MHz. This frequency is referred to as f<sub>C</sub>. Beyond f<sub>C</sub> the slope of the curve increases, suggesting that closed-loop gains less than 5 are not appropriate.

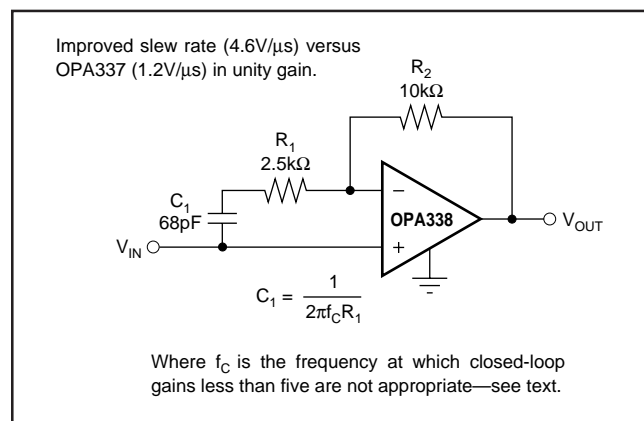


FIGURE 3. Compensation of OPA338 for Unity-Gain Buffer.



Figure 4 shows a compensation technique using an inverting configuration. The low frequency gain is set by the resistor ratio while the high frequency gain is set by the capacitor ratio. As with the noninverting circuit, for frequencies above  $f_C$  the gain must be greater than the recommended minimum stable gain for the op amp.

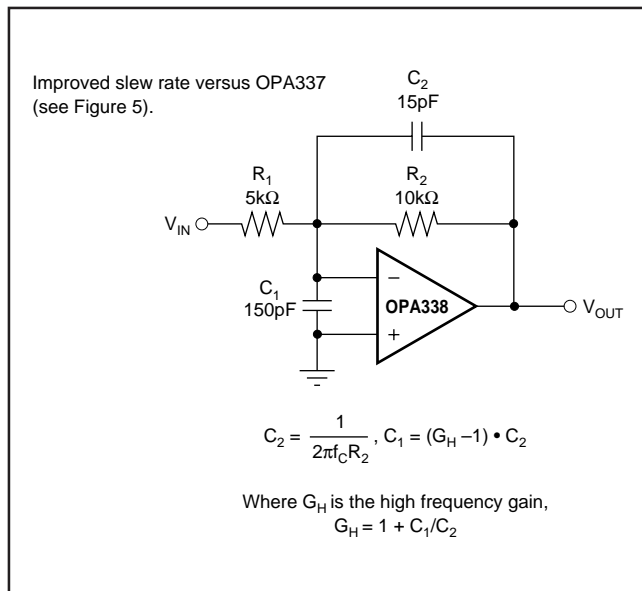


FIGURE 4. Inverting Compensation Circuit of OPA338 for Low Gain.

Resistors  $R_1$  and  $R_2$  are chosen to set the desired dc signal gain. Then the value for  $C_2$  is determined as follows:

$$C_2 = 1/(2\pi f_C R_2)$$

$C_1$  is determined from the desired high frequency gain ( $G_H$ ):

$$C_1 = (G_H - 1) \cdot C_2$$

For a desired dc gain of 2 and high frequency gain of 10, the following resistor and capacitor values result:

$$\begin{aligned} R_1 &= 10k\Omega & C_1 &= 150pF \\ R_2 &= 5k\Omega & C_2 &= 15pF \end{aligned}$$

The capacitor values shown are the nearest standard values. Capacitor values may need to be adjusted slightly to optimize performance. For more detailed information, consult the OPA686 product data sheet.

Figure 5 shows the large-signal transient response using the circuit given in Figure 4. As shown, the OPA338 is stable in low gain applications and provides improved slew rate performance when compared to the OPA337.

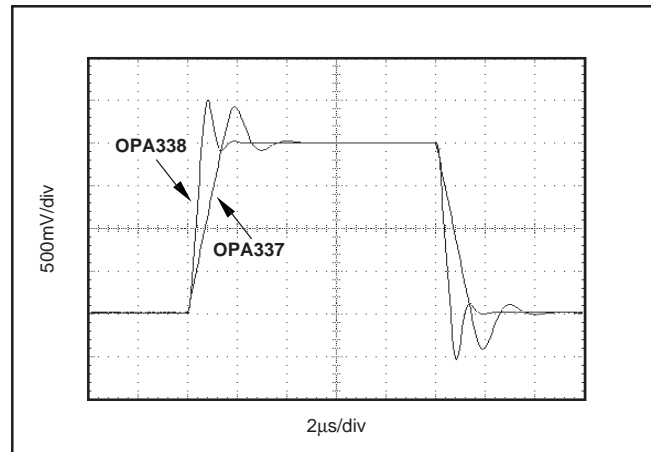


FIGURE 5.  $G = 2$ , Slew-Rate Comparison of OPA338 and OPA337.

## TYPICAL APPLICATION

Figure 6 shows the OPA2337 in a typical application. The ADS7822 is a 12-bit, micro-power sampling analog-to-digital converter available in the tiny MSOP-8 package. As with the OPA2337, it operates with a supply voltage as low as +2.7V. When used with the miniature SOT23-8 package of the OPA2337, the circuit is ideal for space-limited and low power applications. In addition, OPA2337's high input impedance allows large value resistors to be used which results in small physical capacitors, further reducing circuit size. For further information, consult the ADS7822 product data sheet.



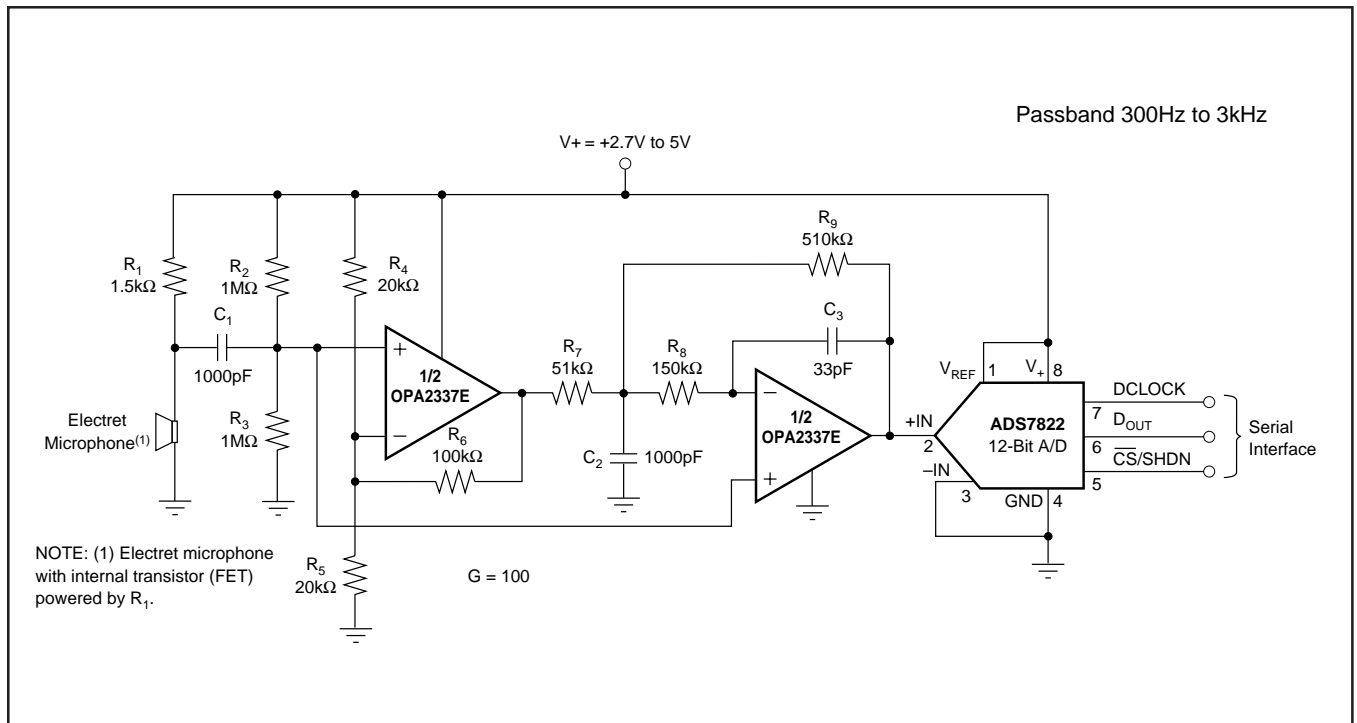


FIGURE 6. Low Power, Single-Supply, Speech Bandpass Filtered Data Acquisition System.

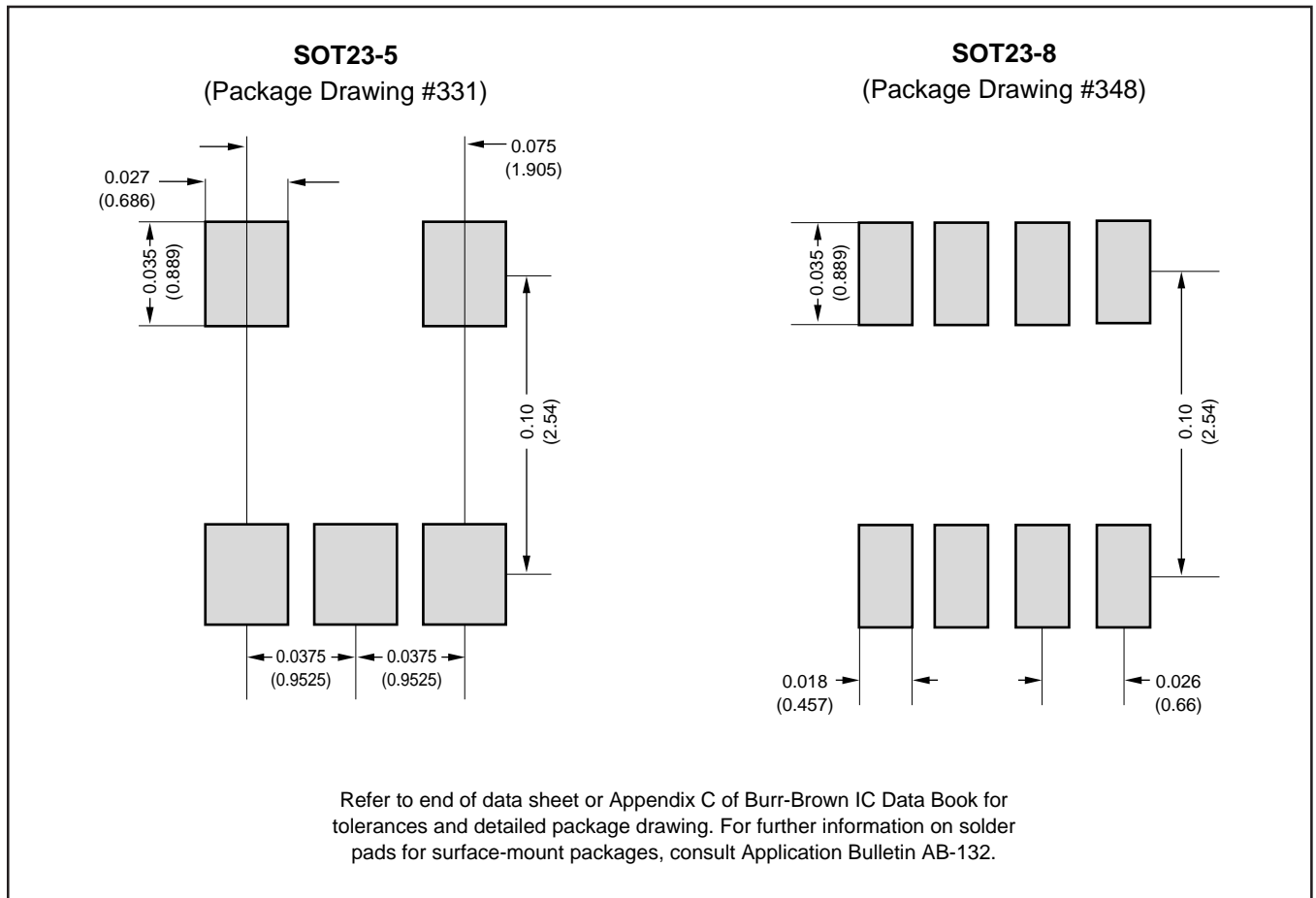


FIGURE 7. Recommended SOT23-5 and SOT23-8 Solder Footprints.

## **IMPORTANT NOTICE**

Texas Instruments and its subsidiaries (TI) reserve the right to make changes to their products or to discontinue any product or service without notice, and advise customers to obtain the latest version of relevant information to verify, before placing orders, that information being relied on is current and complete. All products are sold subject to the terms and conditions of sale supplied at the time of order acknowledgment, including those pertaining to warranty, patent infringement, and limitation of liability.

TI warrants performance of its semiconductor products to the specifications applicable at the time of sale in accordance with TI's standard warranty. Testing and other quality control techniques are utilized to the extent TI deems necessary to support this warranty. Specific testing of all parameters of each device is not necessarily performed, except those mandated by government requirements.

Customers are responsible for their applications using TI components.

In order to minimize risks associated with the customer's applications, adequate design and operating safeguards must be provided by the customer to minimize inherent or procedural hazards.

TI assumes no liability for applications assistance or customer product design. TI does not warrant or represent that any license, either express or implied, is granted under any patent right, copyright, mask work right, or other intellectual property right of TI covering or relating to any combination, machine, or process in which such semiconductor products or services might be or are used. TI's publication of information regarding any third party's products or services does not constitute TI's approval, warranty or endorsement thereof.