

# Chisel3 Cheat Sheet

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## Notation In This Document:

### For Functions and Constructors:

Arguments given as `kwd:type` (name and type(s))

Arguments in brackets (`[...]`) are optional.

### For Operators:

`c`, `x`, `y` are Chisel Data; `n`, `m` are Scala Int

`w(x)`, `w(y)` are the widths of `x`, `y` (respectively)

`minVal(x)`, `maxVal(x)` are the minimum or maximum possible values of `x`

## Basic Chisel Constructs

### Chisel Wire Operators:

```
// Allocate a as wire of type UInt()
val x = Wire(UInt())
x := y // Connect wire y to wire x
```

**When** executes blocks conditionally by Bool, and is equivalent to Verilog if

```
when(condition1) {
  // run if condition1 true and skip rest
} .elsewhen(condition2) {
  // run if condition2 true and skip rest
} .otherwise {
  // run if none of the above ran
}
```

**Switch** executes blocks conditionally by data

```
switch(x) {
  is(value1) {
    // run if x === value1
  }
  is(value2) {
    // run if x === value2
  }
}
```

**Enum** generates value literals for enumerations

```
val s1::s2::...::sn::Nil
  = Enum(nodeType:UInt, n:Int)
s1, s2, ..., sn will be created as nodeType literals
with distinct values
nodeType    type of s1, s2, ..., sn
n            element count
```

### Math Helpers:

`log2Ceil(in:Int): Int`  $\log_2(\text{in})$  rounded up

`log2Floor(in:Int): Int`  $\log_2(\text{in})$  rounded down

`isPow2(in:Int): Boolean` True if `in` is a power of 2

## Basic Data Types

### Constructors:

<code>Bool()</code>	type, boolean value
<code>true.B</code> or <code>false.B</code>	literal values
<code>UInt(32.W)</code>	type 32-bit unsigned
<code>UInt()</code>	type, width inferred
<code>77.U</code> or <code>"hdead".U</code>	unsigned literals
<code>1.U(16.W)</code>	literal with forced width
<code>SInt()</code> or <code>SInt(64.W)</code>	like UInt
<code>-3.S</code>	signed literals
<code>3.S(2.W)</code>	signed 2-bits wide <i>value -1</i>

**Bits, UInt, SInt Casts:** reinterpret cast except for:

`UInt → SInt` Zero-extend to SInt

## State Elements

**Registers** retain state until updated

```
val my_reg = Reg(UInt(32.W))
```

Flavors

`RegInit(7.U(32.w))` reg with initial value 7

`RegNext(next_val)` update each clock, no init

`RegNext(next, init)` update, with init

`RegEnable(next, enable)` update, with enable gate

**Updating:** assign to latch new value on next clock:

```
my_reg := next_val
```

**Read-Write Memory** provide addressable memories

```
val my_mem = Mem(n:Int, out:Data)
```

out memory element type

n memory depth (elements)

**Using:** access elements by indexing:

```
val readVal = my_mem(addr:UInt/Int)
```

for synchronous read: assign output to Reg

```
my_mem(addr:UInt/Int) := y
```

## Modules

**Defining:** subclass Module with elements, code:

```
class Accum(width:Int) extends Module {
  val io = IO(new Bundle {
    val in = Input(UInt(width.W))
    val out = Output(UInt(width.W))
  })
  val sum = Reg(UInt())
  sum := sum + io.in
  io.out := sum
}
```

**Usage:** access elements using dot notation:

(code inside a Module is always running)

```
val my_module = Module(new Accum(32))
my_module.io.in := some_data
val sum := my_module.io.out
```

### Operators:

Chisel	Explanation	Width
<code>!x</code>	Logical NOT	1
<code>x &amp;&amp; y</code>	Logical AND	1
<code>x    y</code>	Logical OR	1
<code>x(n)</code>	Extract bit, 0 is LSB	1
<code>x(n, m)</code>	Extract bitfield	$n - m + 1$
<code>x &lt;&lt; y</code>	Dynamic left shift	$w(x) + \text{maxVal}(y)$
<code>x &gt;&gt; y</code>	Dynamic right shift	$w(x) - \text{minVal}(y)$
<code>x &lt;&lt; n</code>	Static left shift	$w(x) + n$
<code>x &gt;&gt; n</code>	Static right shift	$w(x) - n$
<code>Fill(n, x)</code>	Replicate x, n times	$n * w(x)$
<code>Cat(x, y)</code>	Concatenate bits	$w(x) + w(y)$
<code>Mux(c, x, y)</code>	If c, then x; else y	$\max(w(x), w(y))$
<code>~x</code>	Bitwise NOT	$w(x)$
<code>x &amp; y</code>	Bitwise AND	$\max(w(x), w(y))$
<code>x   y</code>	Bitwise OR	$\max(w(x), w(y))$
<code>x ^ y</code>	Bitwise XOR	$\max(w(x), w(y))$
<code>x === y</code>	Equality (triple equals)	1
<code>x != y</code>	Inequality	1
<code>x + y</code>	Addition	$\max(w(x), w(y))$
<code>x +% y</code>	Addition	$\max(w(x), w(y))$
<code>x +% y</code>	Addition	$\max(w(x), w(y)) + 1$
<code>x - y</code>	Subtraction	$\max(w(x), w(y))$
<code>x -% y</code>	Subtraction	$\max(w(x), w(y))$
<code>x -&amp; y</code>	Subtraction	$\max(w(x), w(y)) + 1$
<code>x * y</code>	Multiplication	$w(x) + w(y)$
<code>x / y</code>	Division	$w(x)$
<code>x % y</code>	Modulus	$\text{bits}(\text{maxVal}(y)) - 1$
<code>x &gt; y</code>	Greater than	1
<code>x &gt;= y</code>	Greater than or equal	1
<code>x &lt; y</code>	Less than	1
<code>x &lt;= y</code>	Less than or equal	1
<code>x &gt;&gt; y</code>	Arithmetic right shift	$w(x) - \text{minVal}(y)$
<code>x &gt;&gt; n</code>	Arithmetic right shift	$w(x) - n$

### UInt bit-reduction methods:

Chisel	Explanation	Width
<code>x.andR</code>	AND-reduce	1
<code>x.orR</code>	OR-reduce	1
<code>x.xorR</code>	XOR-reduce	1

As an example to apply the `andR` method to an `SInt` use `x.asUInt.andR`

## Hardware Generation

**Functions** provide block abstractions for code. Scala functions that instantiate or return Chisel types are code generators.

**Also:** Scala's `if` and `for` can be used to control hardware generation and are equivalent to Verilog `generate if/for`

```
val number = Reg(if(can_be_negative) SInt()
                  else UInt())
```

will create a Register of type `SInt` or `UInt` depending on the value of a Scala variable

## Aggregate Types

**Bundle** contains `Data` types indexed by name

**Defining:** subclass `Bundle`, define components:

```
class MyBundle extends Bundle {
  val a = Bool()
  val b = UInt(32.W)
}
```

**Constructor:** instantiate `Bundle` subclass:

```
val my_bundle = new MyBundle()
```

**Inline defining:** define a `Bundle` type:

```
val my_bundle = new Bundle {
  val a = Bool()
  val b = UInt(32.W)
}
```

**Using:** access elements through dot notation:

```
val bundleVal = my_bundle.a
my_bundle.a := true.B
```

**Vec** is an indexable vector of `Data` types

```
val myVec = Vec(elts:Iterable[Data])
  elts initial element Data (vector depth inferred)
val myVec = Vec.fill(n:Int) {gen:Data}
  n vector depth (elements)
  gen initial element Data, called once per element
```

**Using:** access elements by dynamic or static indexing:

```
readVal := myVec(ind: UInt / idx: Int)
myVec(ind: UInt / idx: Int) := writeVal
```

**Functions:** (`T` is the `Vec` element's type)

```
.forall(p:T=>Bool): Bool AND-reduce p on all elts
.exists(p:T=>Bool): Bool OR-reduce p on all elts
.contains(x:T): Bool True if this contains x
.count(p:T=>Bool): UInt count elts where p is True
```

```
.indexWhere(p:T=>Bool): UInt
.lastIndexWhere(p:T=>Bool): UInt
.onlyIndexWhere(p:T=>Bool): UInt
```

## Standard Library: Function Blocks

**Stateless:**

`PopCount(in:Bits/Seq[Bool]): UInt`  
Returns number of hot (= 1) bits in `in`

`Reverse(in:UInt): UInt`  
Reverses the bit order of `in`

`UIntToOH(in:UInt, [width:Int]): Bits`  
Returns the one-hot encoding of `in`  
`width` (*optional, else inferred*) output width

`OHToUInt(in:Bits/Seq[Bool]): UInt`  
Returns the `UInt` representation of one-hot in

`PriorityEncoder(in:Bits/Iterable[Bool]): UInt`  
Returns the position the least significant 1 in `in`

`PriorityEncoderOH(in:Bits): UInt`  
Returns the position of the hot bit in `in`

`Mux1H(in:Iterable[(Data, Bool]): Data`

`Mux1H(sel:Bits/Iterable[Bool],  
in:Iterable[Data]): Data`

`PriorityMux(in:Iterable[(Bool, Bits]): Bits`  
`PriorityMux(sel:Bits/Iterable[Bool],  
in:Iterable[Bits]): Bits`

A mux tree with either a one-hot select or multiple selects (where the first inputs are prioritized)  
`in` iterable of combined input and select (`Bool, Bits`)  
tuples or just mux input `Bits`  
`sel` select signals or bitvector, one per input

**Stateful:**

`Counter(n:Int): UInt`  
.inc() bumps counter returning true when `n` reached  
.value returns current value

`LFSR16([increment:Bool]): UInt`  
16-bit LFSR (to generate pseudorandom numbers)  
`increment` (*optional, default True*) shift on next clock  
`ShiftRegister(in:Data, n:Int, [en:Bool]): Data`  
Shift register, returns `n`-cycle delayed input in  
`en` (*optional, default True*) enable

## Standard Library: Interfaces

**DecoupledIO** is a `Bundle` with a ready-valid interface

**Constructor:**

`Decoupled(gen:Data)`

`gen Chisel Data` to wrap ready-valid protocol around

**Interface:**

```
(in) .ready ready Bool
(out) .valid valid Bool
(out) .bits data
```

**ValidIO** is a `Bundle` with a valid interface

**Constructor:**

`Valid(gen:Data)`

`gen Chisel Data` to wrap valid protocol around

**Interface:**

```
(out) .valid valid Bool
(out) .bits data
```

**Queue** is a `Module` providing a hardware queue

**Constructor:**

`Queue(enq:DecoupledIO, entries:Int)`  
`enq` `DecoupledIO` source for the queue  
`entries` size of queue

**Interface:**

```
.io.enq DecoupledIO source (flipped)
.io.deq DecoupledIO sink
.io.count UInt count of elements in the queue
```

**Pipe** is a `Module` delaying input data

**Constructor:**

`Pipe(enqValid:Bool, enqBits:Data, [latency:Int])`  
`Pipe(enq:ValidIO, [latency:Int])`  
`enqValid` input data, valid component  
`enqBits` input data, data component  
`enq` input data as `ValidIO`  
`latency` (*optional, default 1*) cycles to delay data by

**Interface:**

```
.io.enq ValidIO source (flipped)
.io.deq ValidIO sink
```

**Arbiters** are `Modules` connecting multiple producers to one consumer

**Arbiter** prioritizes lower producers

**RRArbiter** runs in round-robin order

**Constructor:**

`Arbiter(gen:Data, n:Int)`  
`gen` data type  
`n` number of producers

**Interface:**

```
.io.in Vec of DecoupledIO inputs (flipped)
.io.out DecoupledIO output
.io.chosen UInt input index on .io.out,
does not imply output is valid
```