

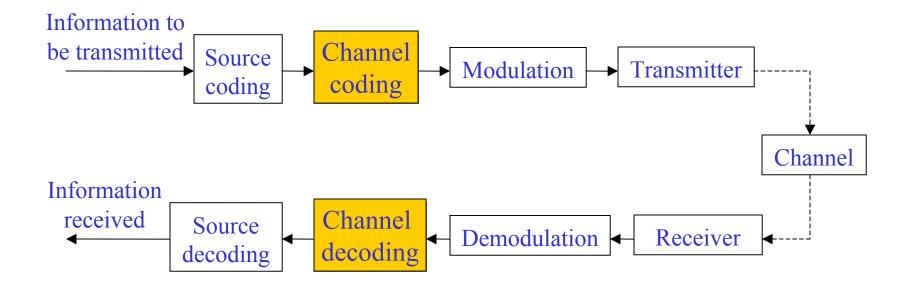
Chapter 4

Channel Coding

Outline

- FEC (Forward Error Correction)
- Block Codes
- Convolutional Codes
- Interleaving
- Information Capacity Theorem
- Turbo Codes
- CRC (Cyclic Redundancy Check)
- ARQ (Automatic Repeat Request)
 - Stop-and-wait ARQ
 - Go-back-N ARQ
 - Selective-repeat ARQ

Channel Coding in Digital Communication Systems





Forward Error Correction (FEC)

- The key idea of FEC is to transmit enough redundant data to allow receiver to recover from errors all by itself. No sender retransmission required.
- The major categories of FEC codes are
 - Block codes,
 - Cyclic codes,
 - Reed-Solomon codes (Not covered here),
 - Convolutional codes, and
 - Turbo codes, etc.



Block Codes

- Information is divided into blocks of length *k*
- r parity bits or check bits are added to each block (total length n = k + r),.
- Code rate R = k/n
- Decoder looks for codeword closest to received vector (code vector + error vector)
- Tradeoffs between
 - Efficiency
 - Reliability
 - Encoding/Decoding complexity



Block Codes: Linear Block Codes

Linear Block Code

The block length c(x) or C of the Linear Block Code is

$$c(x) = m(x) g(x)$$
 or $C = m G$

where m(x) or m is the information codeword block length, g(x) is the generator polynomial, G is the generator matrix.

$$G = [p \mid I],$$

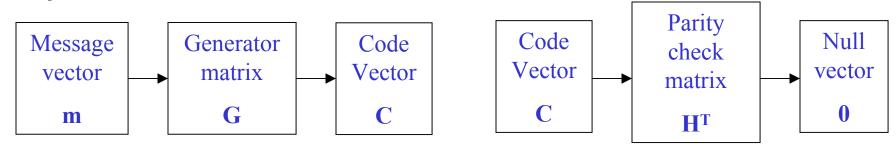
where p_i = Remainder of $[x^{n-k+i-1}/g(x)]$ for i=1, 2, ..., k, and I is unit matrix.

The parity check matrix

 $H = [p^T | I]$, where p^T is the transpose of the matrix p.



Block Codes: Linear Block Codes



Operations of the generator matrix and the parity check matrix

The parity check matrix H is used to detect errors in the received code by using the fact that $c * H^T = \mathbf{0}$ (null vector)

Let $x = c \oplus e$ be the received message where c is the correct code and e is the error

Compute
$$S = x * H^T = (c \oplus e) * H^T = c H^T \oplus e H^T = e H^T$$

If S is 0 then message is correct else there are errors in it, from common known error patterns the correct message can be decoded.



Block Codes: Example

Example: Find linear block code encoder **G** if code generator polynomial $g(x)=1+x+x^3$ for a (7, 4) code.

We have n = Total number of bits = 7, k = Number of information bits = 4, r = Number of parity bits = n - k = 3.

$$G = [P \mid I] = \begin{bmatrix} p_1 & 1 & 0 & \cdots & 0 \\ p_2 & 0 & 1 & \cdots & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots \\ p_k & 0 & 0 & \cdots & 1 \end{bmatrix},$$

where

$$p_i = \text{Re } mainder of \left[\frac{x^{n-k+i-1}}{g(x)} \right], \quad i = 1, 2, \dots, k$$



Block Codes: Example (Continued)

$$p_1 = \text{Re}\left[\frac{x^3}{1+x+x^3}\right] = 1+x \to [110]$$

$$p_2 = \text{Re}\left[\frac{x^4}{1 + x + x^3}\right] = x + x^2 \to [011]$$

$$p_3 = \text{Re}\left[\frac{x^5}{1+x+x^3}\right] = 1+x+x^2 \to [111]$$

$$p_4 = \text{Re}\left[\frac{x^6}{1 + x + x^3}\right] = 1 + x^2 \rightarrow [101]$$

$$p_{2} = \operatorname{Re}\left[\frac{x^{4}}{1+x+x^{3}}\right] = x+x^{2} \to [011]$$

$$p_{3} = \operatorname{Re}\left[\frac{x^{5}}{1+x+x^{3}}\right] = 1+x+x^{2} \to [111]$$

$$G = \begin{bmatrix} 1101000 \\ 0110100 \\ 11110010 \\ 1010001 \end{bmatrix}$$



Cyclic Codes

It is a block code which uses a shift register to perform encoding and decoding

The code word with n bits is expressed as

$$c(x)=c_1x^{n-1}+c_2x^{n-2}....+c_n$$

where each c_i is either a 1 or 0.

$$c(x) = m(x) x^{n-k} + c_p(x)$$

where $c_p(x)$ = remainder from dividing $m(x) x^{n-k}$ by generator g(x) if the received signal is c(x) + e(x) where e(x) is the error.

To check if received signal is error free, the remainder from dividing c(x) + e(x) by g(x) is obtained(syndrome). If this is 0 then the received signal is considered error free else error pattern is detected from known error syndromes.



Cyclic Redundancy Check (CRC)

- Using parity, some errors are masked careful choice of bit combinations can lead to better detection.
- Binary (*n*, *k*) CRC codes can detect the following error patterns
 - 1. All error bursts of length *n-k* or less.
 - 2. All combinations of minimum Hamming distance d_{min} 1 or fewer errors.
 - 3. All error patters with an odd number of errors if the generator polynomial g(x) has an even number of nonzero coefficients.



Common CRC Codes

Code	Generator polynomial $g(x)$	Parity check bits
CRC-12	$1+x+x^2+x^3+x^{11}+x^{12}$	12
CRC-16	$1+x^2+x^{15}+x^{16}$	16
CRC-CCITT	$1+x^5+x^{15}+x^{16}$	16

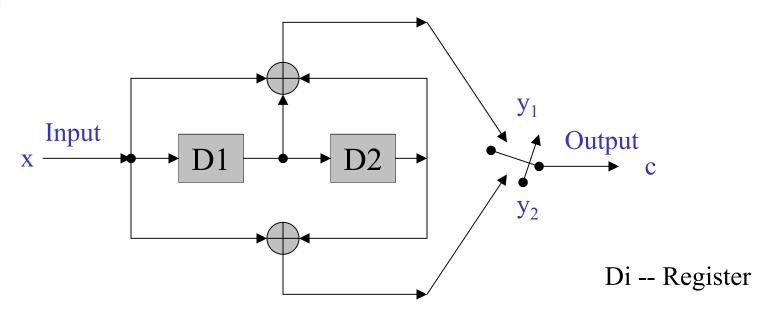


Convolutional Codes

- Encoding of information stream rather than information blocks
- Value of certain information symbol also affects the encoding of next *M* information symbols, i.e., memory *M*
- Easy implementation using shift register
 - \rightarrow Assuming k inputs and n outputs
- Decoding is mostly performed by the <u>Viterbi</u>
 <u>Algorithm</u> (not covered here)



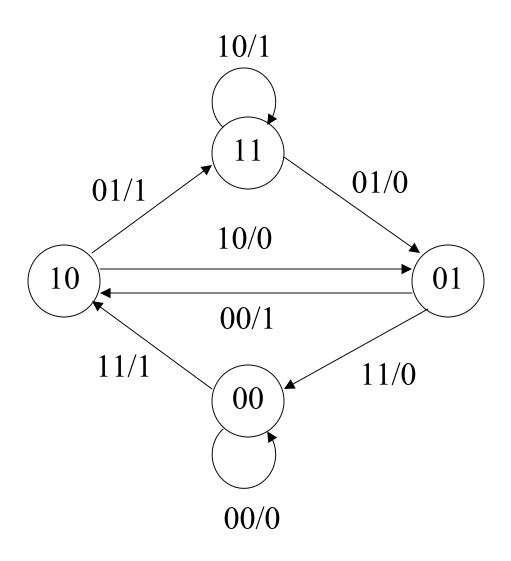
Convolutional Codes: (n=2, k=1, M=2) Encoder



Input:	1	1	1	0	0	0	• • •
Output:	11	01	10	01	11	00	• • •
Input:	1	0	1	0	0	0	• • •
Output	11	10	00	10	11	00	

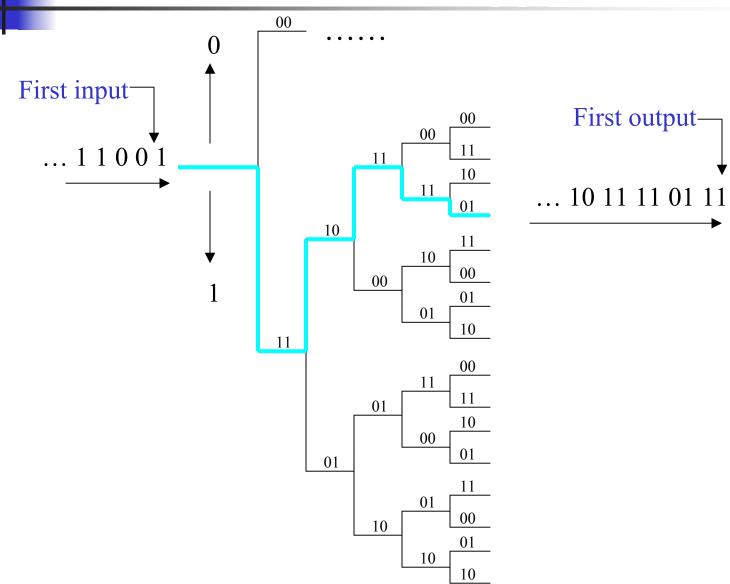


State Diagram



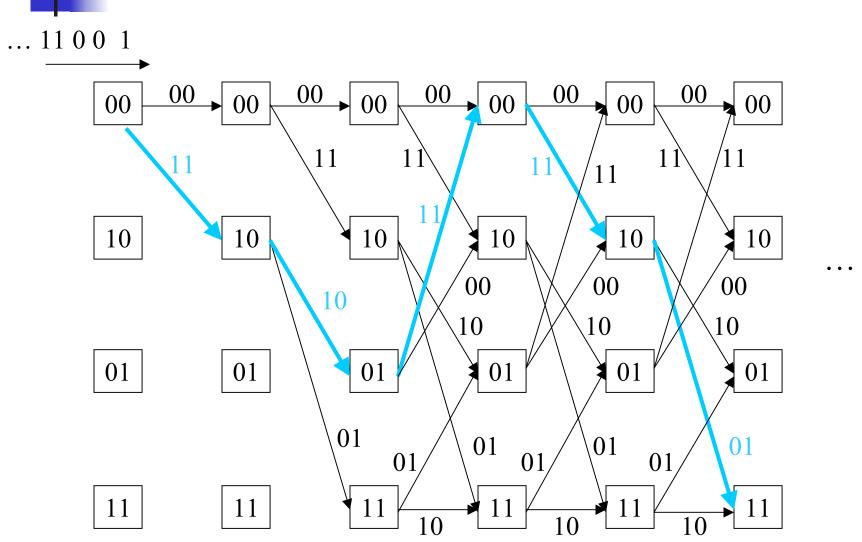


Tree Diagram



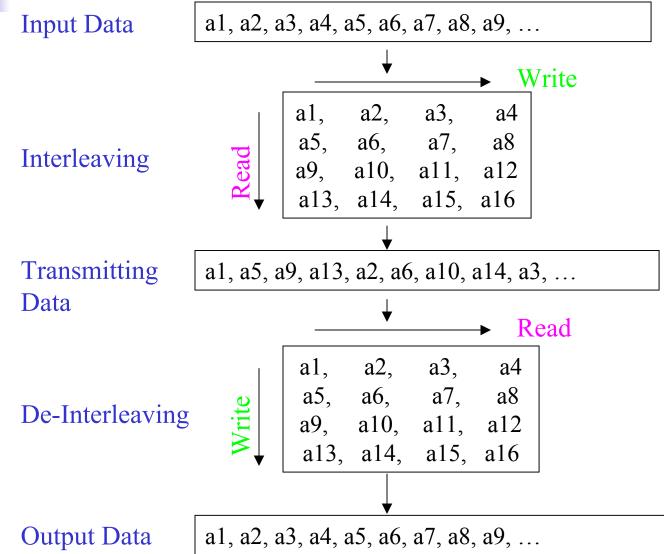


Trellis



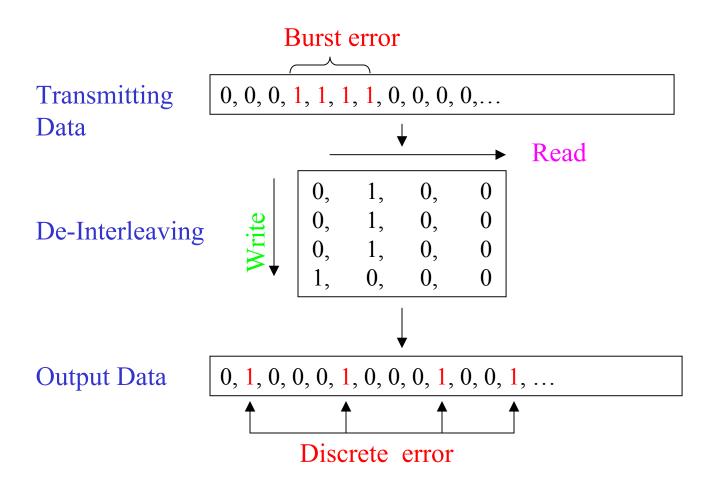


Interleaving





Interleaving (Example)





Information Capacity Theorem (Shannon Limit)

The information capacity (or channel capacity) C of a continuous channel with bandwidth B Hertz can be perturbed by additive Gaussian white noise of power spectral density $N_0/2$, provided bandwidth B satisfies

$$C = B \log_2 \left(1 + \frac{P}{N_0 B} \right) \quad bits / \sec ond$$

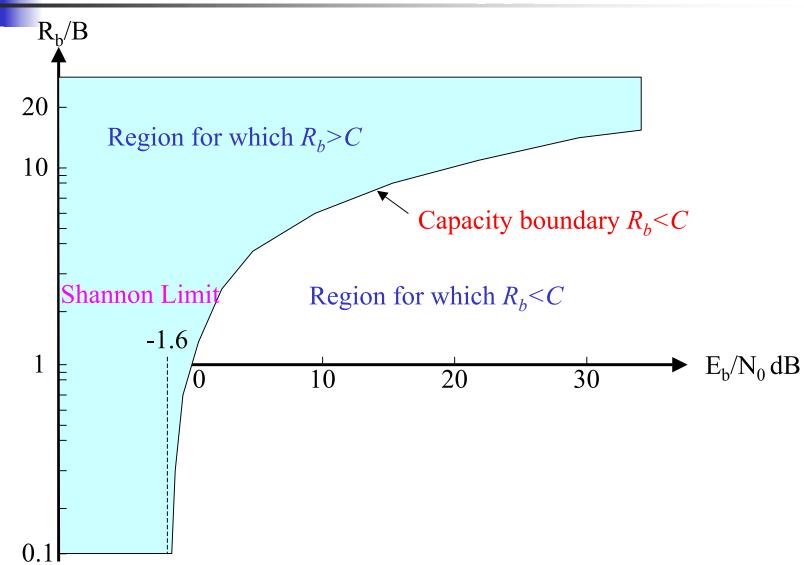
where P is the average transmitted power $P = E_b R_b$ (for an ideal system, $R_b = C$).

 E_b is the transmitted energy per bit,

 R_b is transmission rate.



Shannon Limit



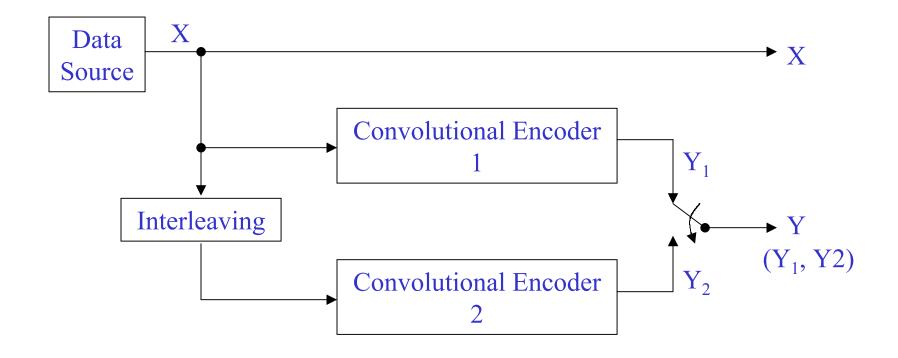


Turbo Codes

- A brief historic of turbo codes :
 - The turbo code concept was first introduced by C. Berrou in 1993. Today, Turbo Codes are considered as the most efficient coding schemes for FEC.
- Scheme with known components (simple convolutional or block codes, interleaver, soft-decision decoder, etc.)
- Performance close to the Shannon Limit $(E_b/N_0 = -1.6 \text{ db})$ if $R_b \rightarrow 0$ at modest complexity!
- Turbo codes have been proposed for low-power applications such as deep-space and satellite communications, as well as for interference limited applications such as third generation cellular, personal communication services, ad hoc and sensor networks.



Turbo Codes: Encoder

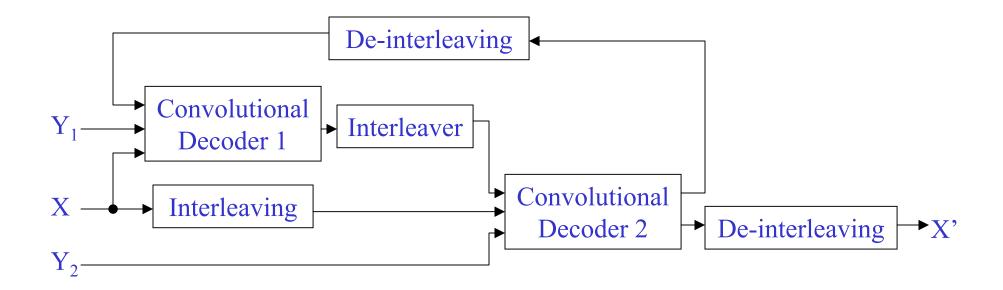


X: Information

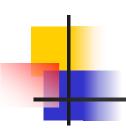
Y_i: Redundancy Information



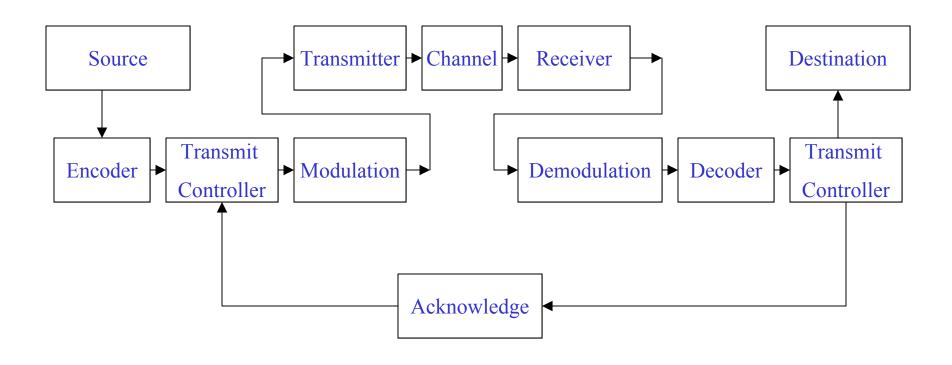
Turbo Codes: Decoder



X': Decoded Information

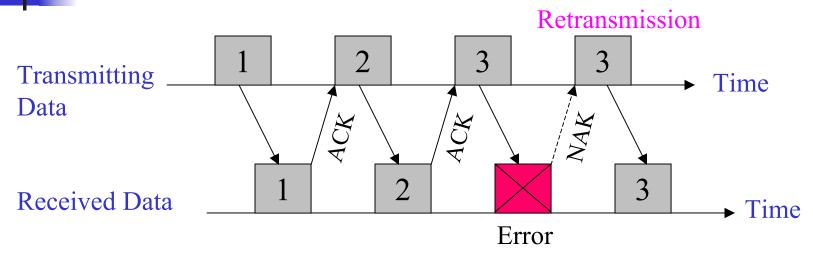


Automatic Repeat Request (ARQ)





Stop-And-Wait ARQ (SAW ARQ)





ACK: Acknowledge

NAK: Negative ACK



Stop-And-Wait ARQ (SAW ARQ)

Throughput:

$$S = (1/T) * (k/n) = [(1 - P_b)^n / (1 + D * R_b/n)] * (k/n)$$

where T is the average transmission time in terms of a block duration

$$T = (1 + D * R_b/n) * P_{ACK} + 2 * (1 + D * R_b/n) * P_{ACK} * (1 - P_{ACK})$$

$$+ 3 * (1 + D * R_b/n) * P_{ACK} * (1 - P_{ACK})^2 +$$

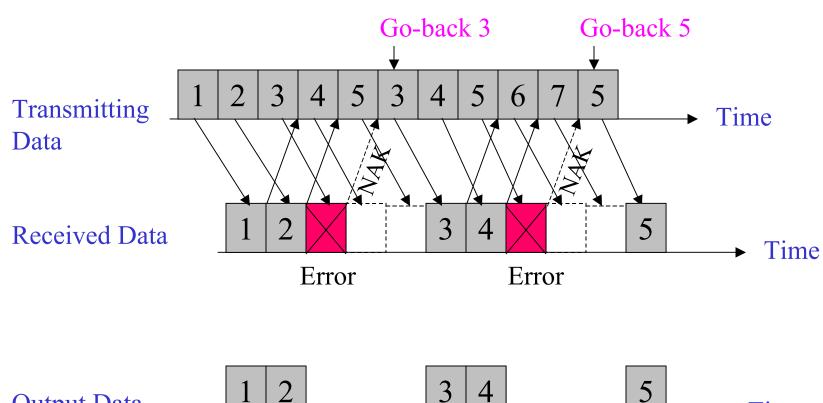
$$= (1 + D * R_b/n) / (1 - P_b)^n$$

where n = number of bits in a block, k = number of information bits in a block, D = round trip delay, $R_b =$ bit rate, $P_b =$ BER of the channel, and $P_{ACK} = (1 - P_b)^n$



Output Data

Go-Back-N ARQ (GBN ARQ)



➤ Time



Go-Back-N ARQ (GBN ARQ)

Throughput

$$S = (1/T) * (k/n)$$

$$= [(1-P_b)^n / ((1-P_b)^n + N * (1-(1-P_b)^n))] * (k/n)$$

where

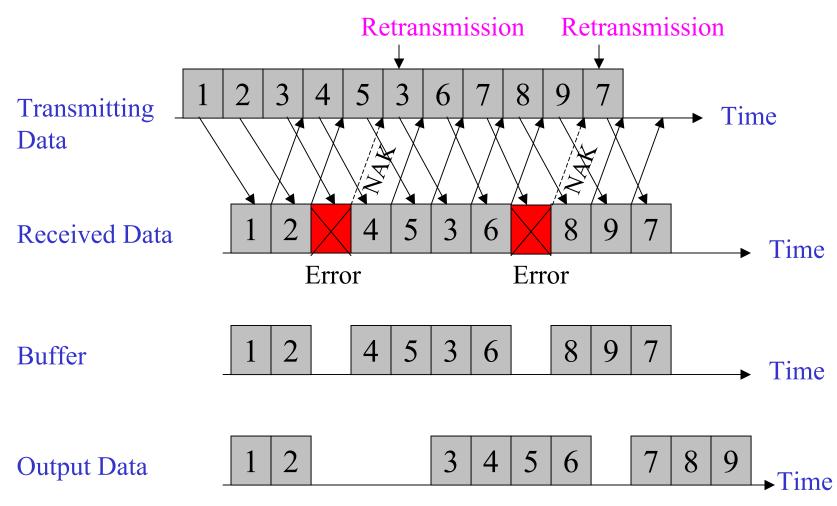
$$T = 1 * P_{ACK} + (N+1) * P_{ACK} * (1-P_{ACK}) + 2 * (N+1) * P_{ACK} *$$

$$(1-P_{ACK})^2 + \dots$$

$$= 1 + (N * [1 - (1-P_b)^n]) / (1-P_b)^n$$



Selective-Repeat ARQ (SR ARQ)





Selective-Repeat ARQ (SR ARQ)

Throughput

$$S = (1/T) * (k/n)$$
$$= (1 - P_b)^n * (k/n)$$

where

$$T = 1 * P_{ACK} + 2 * P_{ACK} * (1-P_{ACK}) + 3 * P_{ACK} * (1-P_{ACK})^{2}$$

$$+ \dots$$

$$= 1/(1-P_{b})^{n}$$