

# Complex Baseband Representation

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# Complex Baseband Representation

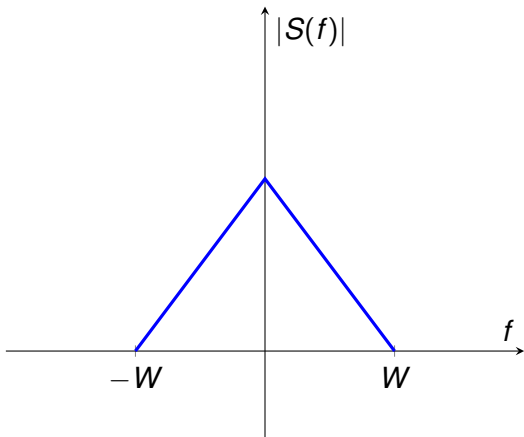
- Contains all the information of a real-valued passband signal
- Requires a smaller sampling rate for discrete-time representation
- Enables modular transceiver design
  - Signal processing algorithms are implemented in the baseband
  - Carrier frequency can be chosen independently

# Baseband Signals

A signal  $s(t)$  is said to be *baseband* if

$$S(f) \approx 0, \quad |f| > W$$

for some  $W > 0$

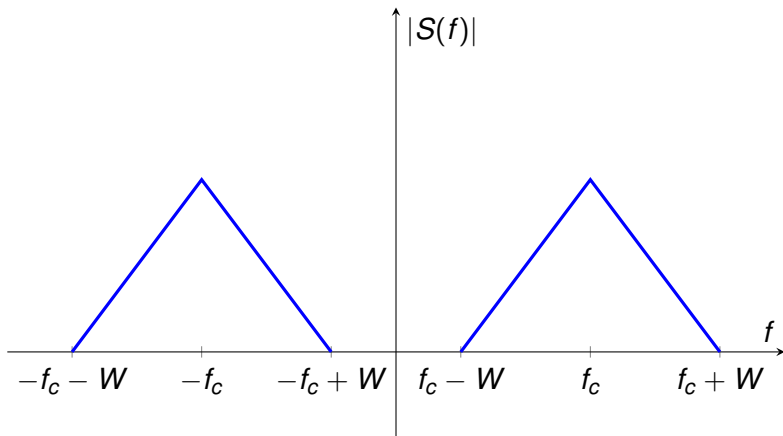


# Passband Signals

A signal  $s(t)$  is said to be *passband* if

$$S(f) \approx 0, \quad |f \pm f_c| > W,$$

where  $f_c > W > 0$



# Sampling Theorem

## Theorem

*If a signal  $s(t)$  is bandlimited to  $B$ ,*

$$S(f) = 0, \quad |f| > B$$

*then a sufficient condition for exact reconstructability is a uniform sampling rate  $f_s$  where*

$$f_s > 2B.$$

Baseband Signals  $B = W$

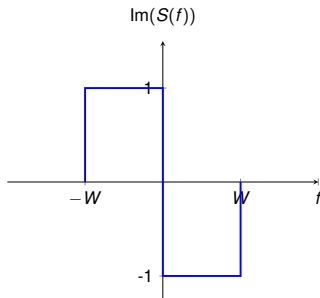
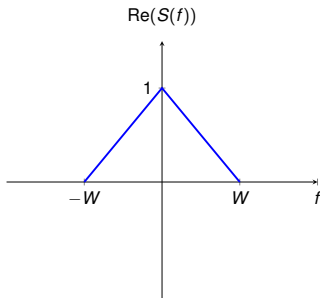
Passband Signals  $B = f_c + W$

Can we reduce the sampling rate for passband signals?

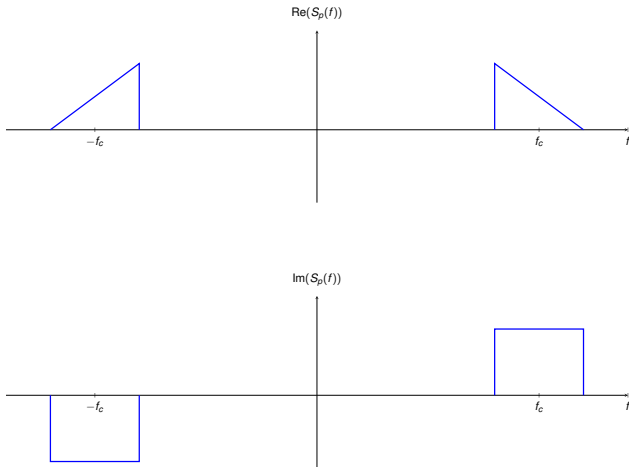
Yes. By using the complex baseband representation.

# Fourier Transform for Real Signals

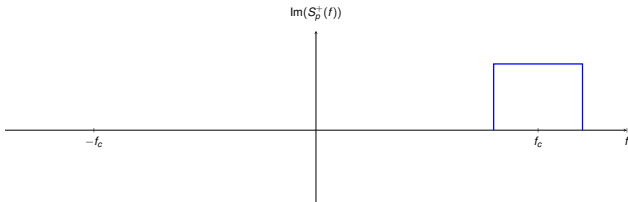
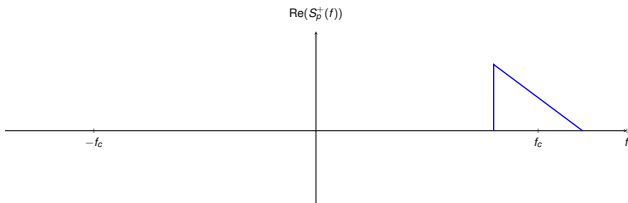
$$\begin{aligned}\operatorname{Im}[s(t)] = 0 &\Rightarrow S(f) = S^*(-f) \\ &\Rightarrow \operatorname{Re}(S(f)) = \operatorname{Re}(S(-f)), \\ &\quad \operatorname{Im}(S(f)) = -\operatorname{Im}(S(-f))\end{aligned}$$



# Fourier Transform of a Real Passband Signal



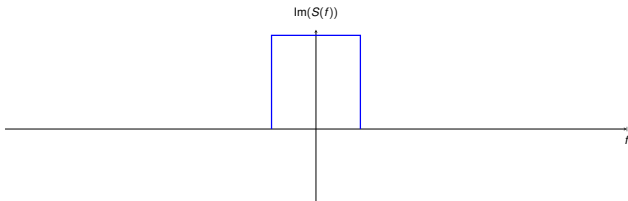
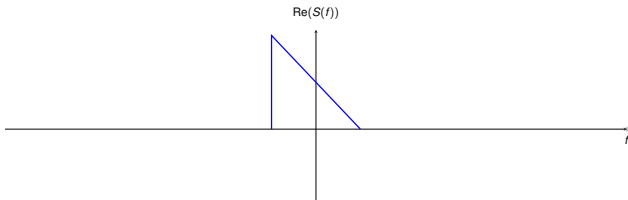
# Positive Spectrum of a Real Passband Signal



$$S_p^+(f) = S_p(f)u(f)$$



# Complex Envelope of a Real Passband Signal



$$S(f) = \sqrt{2}S_p^+(f + f_c) = \sqrt{2}S_p(f + f_c)u(f + f_c)$$

# Complex Envelope in Time Domain

## Frequency Domain Representation

$$S(f) = \sqrt{2}S_p^+(f + f_c) = \sqrt{2}S_p(f + f_c)u(f + f_c)$$

## Time Domain Representation of Positive Spectrum

$$\begin{aligned}S_p^+(f) &= S_p(f)u(f) \\s_p^+(t) &= s_p(t) \star \mathcal{F}^{-1}[u(f)]\end{aligned}$$

## Time Domain Representation of Frequency Domain Unit Step

$$\begin{aligned}u(f) &\leftrightarrow \frac{1}{j2\pi f} + \frac{1}{2}\delta(f) \\ \frac{1}{j2\pi t} + \frac{1}{2}\delta(t) &\leftrightarrow u(-f) \\ \frac{j}{2\pi t} + \frac{1}{2}\delta(t) &\leftrightarrow u(f)\end{aligned}$$

# Complex Envelope in Time Domain

## Time Domain Representation of Positive Spectrum

$$\begin{aligned}s_p^+(t) &= s_p(t) \star \left[ \frac{1}{2} \delta(t) + \frac{j}{2\pi t} \right] \\ &= \frac{1}{2} [s_p(t) + j\hat{s}_p(t)]\end{aligned}$$

## Time Domain Representation of Complex Envelope

$$\begin{aligned}\sqrt{2}S_p(f)u(f) &\leftrightarrow \frac{1}{\sqrt{2}} [s_p(t) + j\hat{s}_p(t)] \\ \sqrt{2}S_p(f + f_c)u(f + f_c) &\leftrightarrow \frac{1}{\sqrt{2}} [s_p(t) + j\hat{s}_p(t)] e^{-j2\pi f_c t} \\ S(f) &\leftrightarrow \frac{1}{\sqrt{2}} [s_p(t) + j\hat{s}_p(t)] e^{-j2\pi f_c t} \\ s(t) &= \frac{1}{\sqrt{2}} [s_p(t) + j\hat{s}_p(t)] e^{-j2\pi f_c t}\end{aligned}$$

# Passband Signal in terms of Complex Envelope

## Complex Envelope

$$s(t) = s_c(t) + js_s(t)$$

$s_c(t)$  In-phase component

$s_s(t)$  Quadrature component

## Time Domain Relationship

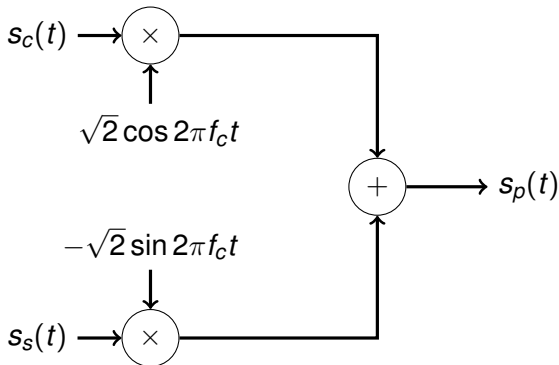
$$\begin{aligned}s_p(t) &= \operatorname{Re} \left[ \sqrt{2}s(t)e^{j2\pi f_c t} \right] \\&= \operatorname{Re} \left[ \sqrt{2}\{s_c(t) + js_s(t)\}e^{j2\pi f_c t} \right] \\&= \sqrt{2}s_c(t) \cos 2\pi f_c t - \sqrt{2}s_s(t) \sin 2\pi f_c t\end{aligned}$$

## Frequency Domain Relationship

$$S_p(f) = \frac{S(f - f_c) + S^*(-f - f_c)}{\sqrt{2}}$$

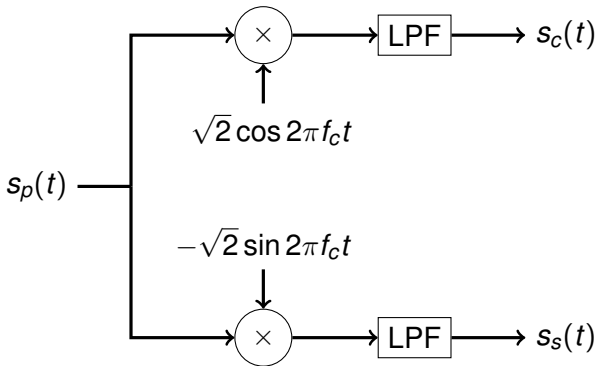
# Upconversion

$$s_p(t) = \sqrt{2}s_c(t) \cos 2\pi f_c t - \sqrt{2}s_s(t) \sin 2\pi f_c t$$



## Downconversion

$$\begin{aligned}\sqrt{2}s_p(t) \cos 2\pi f_c t &= 2s_c(t) \cos^2 2\pi f_c t - 2s_s(t) \sin 2\pi f_c t \cos 2\pi f_c t \\ &= s_c(t) + s_c(t) \cos 4\pi f_c t - s_s(t) \sin 4\pi f_c t\end{aligned}$$



# References

- Section 2.2, *Fundamentals of Digital Communication*, Upamanyu Madhow, 2008