

① Digital Butterworth low pass Filter. Using IIT at $T=1$.

$$A_p = -1.932 \text{ dB.}$$

$$\omega_p = 0.2\pi \text{ rad} = 0.6285.$$

$$A_s = -133 - 13.9794 \text{ dB. } \omega_s = 0.6\pi \text{ rad} = \underline{\underline{1.8857.}}$$

Solution: For IIT.

$$\Omega_p = \frac{\omega_p}{T} = 0.6285 \text{ rad/s.}$$

$$\Omega_s = \frac{\omega_s}{T} = \underline{\underline{1.8857 \text{ rad/s.}}}$$

Step 1: Order $-N$.

$$N \geq \frac{\log_{10} \left[\frac{\omega^{0.1 A_s} - 1}{10^{0.1 A_p} - 1} \right]}{2 \log_{10} \frac{\Omega_s}{\Omega_p}} = \frac{\log_{10} \left[\frac{10^{0.1 \times 13.9794} - 1}{10^{0.1 \times 1.932} - 1} \right]}{2 \log_{10} \left[\frac{1.8857}{0.6285} \right]}$$

$$N \geq 1.7099 \quad \boxed{\therefore N=2.}$$

Step 2: Calculating cutoff Ω_c .

$$\Omega_{cp} = \frac{\Omega_p}{\left(10^{0.1 A_p} - 1 \right)^{1/2N}} = \frac{0.6285}{\left(10^{0.1 \times 1.932} - 1 \right)^{1/4}} = \underline{\underline{0.726.}}$$

$$\Omega_{cs} = \frac{\Omega_s}{\left(10^{0.1 A_s} - 1 \right)^{1/2N}} = \frac{1.8857}{\left(10^{0.1 \times 13.9794} - 1 \right)^{1/4}} = \underline{\underline{0.852.}}$$

$$\therefore \Omega_c = \frac{\Omega_{cp} + \Omega_{cs}}{2} = \underline{\underline{0.7889 \text{ rad/s.}}}$$

①

Step 3: Designing Normalised Filter.

$$P_R = \pm e^{j \frac{(2R+1)\pi}{2N}}$$

$$P_0 = \pm (0.707 + 0.707j)$$

$$P_1 = \pm (-0.707 - 0.707j)$$

$$H(s) = \frac{1}{(s + 0.707 - 0.707j)(s + 0.707 + 0.707j)}$$

$$= \frac{1}{(s + a + jb)(s + a - jb)} = \frac{1}{s^2 + 2as + a^2 + b^2}$$

$$= \frac{1}{s^2 + 1.414s + 1}$$

Step 4: De normalising:

$$s \rightarrow \frac{s}{\omega_c} = \frac{s}{0.789}$$

$$\therefore H(s) = \frac{1}{s^2 + 1.414s + 1} \quad s \rightarrow \frac{s}{0.789}$$

$$H(s) = \frac{0.623}{s^2 + 1.11s + 0.623}$$

Step 5: $s \rightarrow z$ transformation.

$$\frac{b}{(s+a)^2 + b^2} = \frac{1.78 \times 0.561}{(s + 0.555)^2 + (0.561)^2} \rightarrow \frac{e^{-at} \sin bt z^{-1}}{1 - 2e^{-at} \cos bt z^{-1} + e^{-2at} z^{-2}}$$

$$H(z) = \frac{e^{-0.555s} \sin(0.561) z^{-1}}{1 - 2e^{-0.555s} \cos(0.561) z^{-1} + e^{-2(0.555s)} z^{-2}}$$

$$H(z) = \frac{0.305 z^{-1}}{1 - 0.9721 z^{-1} + 0.3295 z^{-2}}$$

② Design a Chebyshev Filter.

$$0.75 \leq H(\omega) \leq 1$$

$$0 \leq \omega \leq 0.25\pi$$

$$|H(\omega)| \leq 0.23$$

$$0.63\pi \leq \omega \leq \pi$$

$$\text{BLT: } T=2.$$

Sol:

$$A_p = 0.75$$

$$\omega_p = 0.25\pi$$

$$\omega_p = \frac{2}{2} \tan \frac{\omega_p}{2} = 0.414 \text{ rad/s}$$

$$A_s = 0.23$$

$$\omega_s = 0.63\pi$$

$$\omega_s = \frac{2}{2} \tan \frac{\omega_s}{2} = 1.522 \text{ rad/s}$$

$$A_p = -20 \log(0.75) = 2.49 \text{ dB}$$

$$\omega_p = 0.414 \text{ rad/s}$$

$$A_s = -20 \log(0.23) = 12.78 \text{ dB}$$

$$\omega_s = 1.522 \text{ rad/s}$$

Normalizing:

$$\omega_p' = \frac{\omega_p}{\omega_p} = 1$$

$$\omega_s' = \frac{\omega_s}{\omega_p} = 3.678$$

Finding ϵ .

$$\epsilon = \sqrt{10^{0.1 A_p} - 1} = 0.8798$$

Order N

$$N \geq \frac{A_s - 20 \log_{10}(\epsilon)}{6 + 20 \log_{10}(\omega_s')}$$

$$\geq \frac{12.78}{6 + 20 \log_{10}(3.678)} = 2.657 = 3$$

$$\underline{N=3}$$

$$H(s) = \frac{K}{(s-s_1)(s-s_2)}$$

$$s_1 = -0.3583 + j 0.7927$$

$$s_2 = -0.3583 - j 0.7927$$

$$H(s) = \frac{K}{s^2 + 0.7166s + 0.7567}$$

Finding K .

$$K = \frac{b_0}{\sqrt{1-\varepsilon^2}} \quad \left. \vphantom{\frac{b_0}{\sqrt{1-\varepsilon^2}}} \right\} \therefore \text{Nus. num.}$$

$$b_0 = 0.7567$$

$$\therefore K = 0.86$$

Frequency transformation -

LP - LP

$$s \rightarrow \frac{s}{0.414} = \frac{s}{0.414}$$

$$H(s) = \frac{0.86 (0.414)^2}{s^2 + 0.7166 \times 0.414 s + 0.7567 \times (0.414)^2}$$

$$s^2 + 0.2966s + 0.129$$

$$H(s) = \frac{0.0959}{s^2 + 0.2966s + 0.129}$$

Finding $H(z)$.

$$S \rightarrow \frac{z}{T} \left[\frac{1-z^{-1}}{1+z^{-1}} \right] \quad T=2.$$

$$S \rightarrow \left[\frac{1-z^{-1}}{1+z^{-1}} \right]$$

$$H(z) = \frac{0.0959}{\quad}$$

$$\frac{(1-z^{-1})^2}{(1+z^{-1})^2} + 0.3 \left[\frac{1-z^{-1}}{1+z^{-1}} \right] + 0.129.$$

$$\frac{0.0959 [1+2z^{-1}+z^{-2}]}{\quad}$$

$$1.429 + (-1.742)z^{-1} + 0.829z^{-2}$$