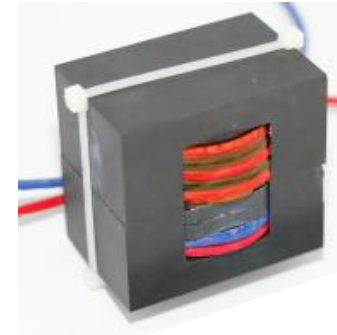
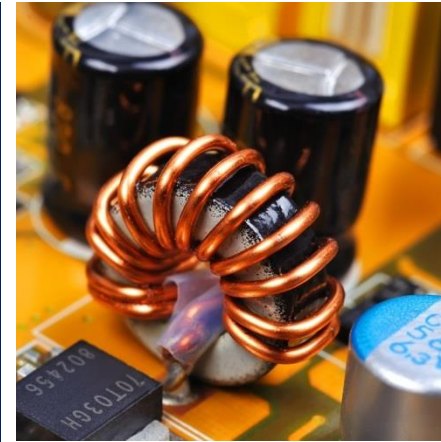




Power Electronics and Electrical Drives
Prof. Dr.-Ing. Joachim Böcker



PADERBORN
UNIVERSITY



Overview: Core Loss in Ferrites

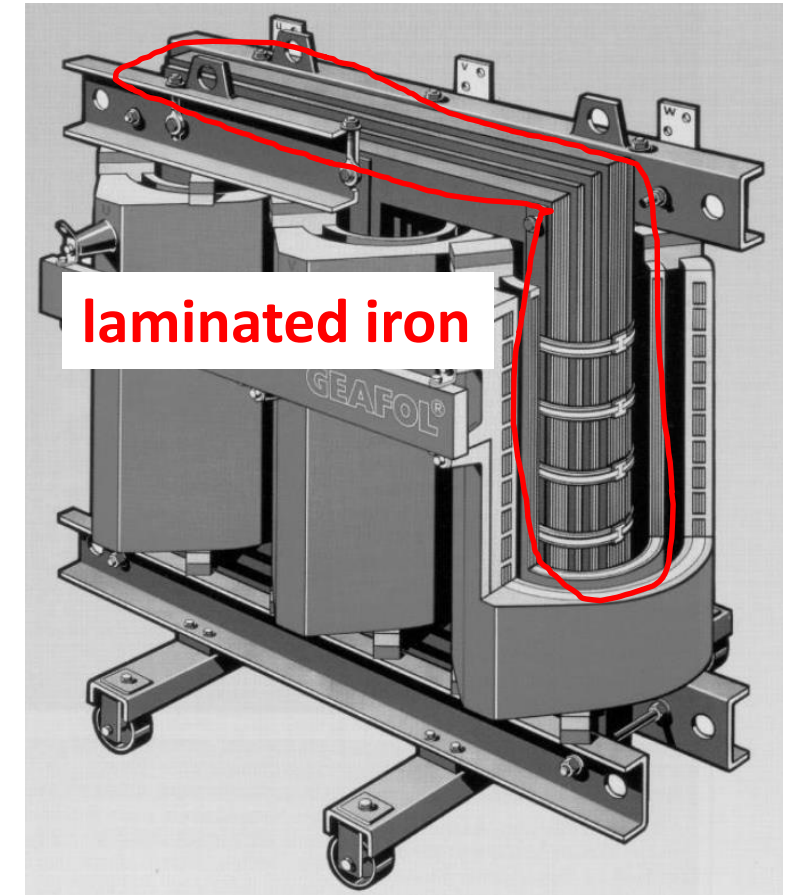
Till Piepenbrock

- Where do we use ferrite cores?
- Temperature problem
- Some physics behind
- Loss modelling approaches
- BH curve modelling approaches
- Measurement methods

TRANSFORMERS...



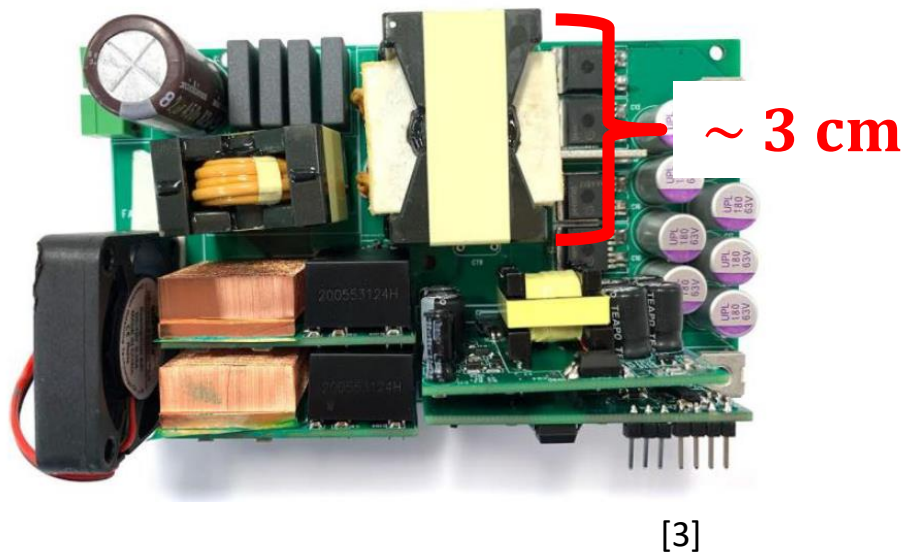
[1]



[2]

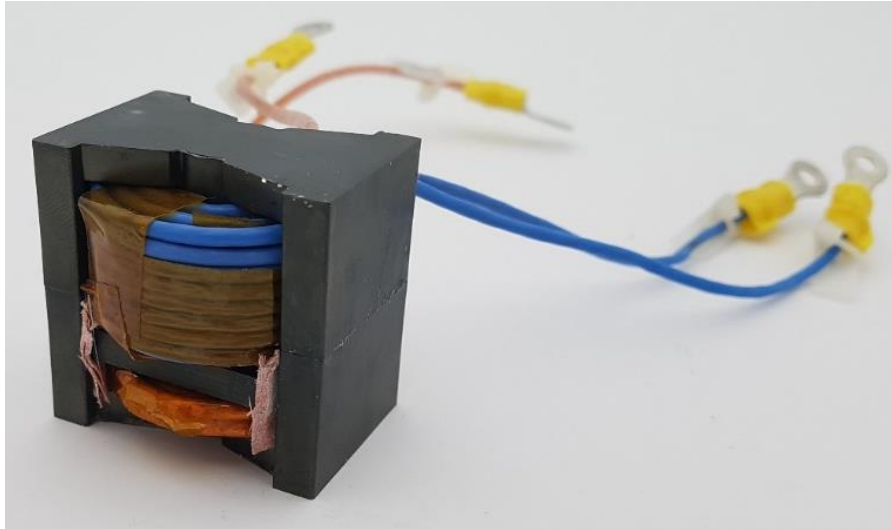
~ 100 kHz (frequency)

~ 1 kW (power)



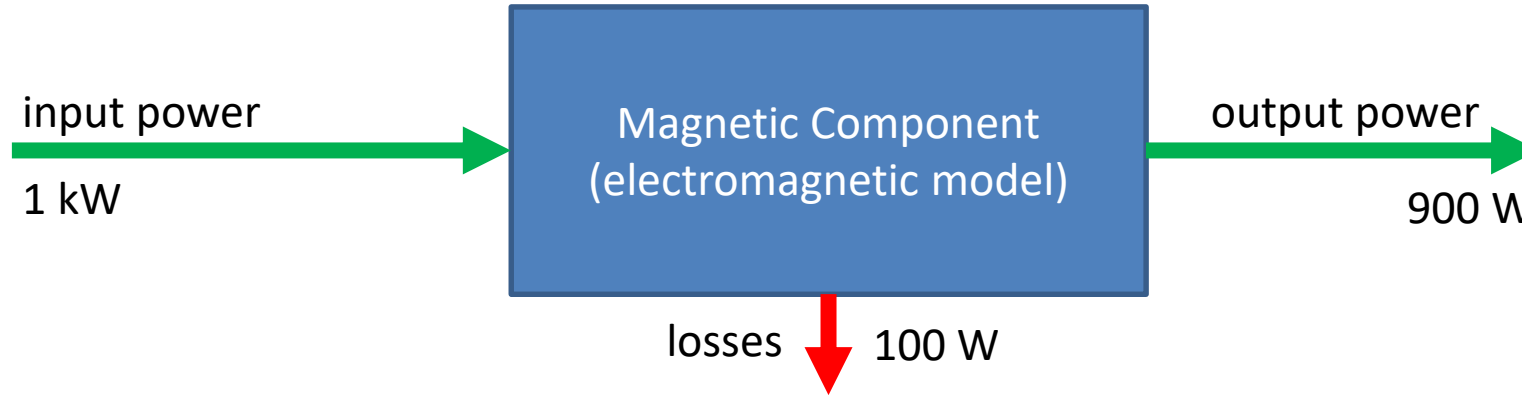
ferrite powder





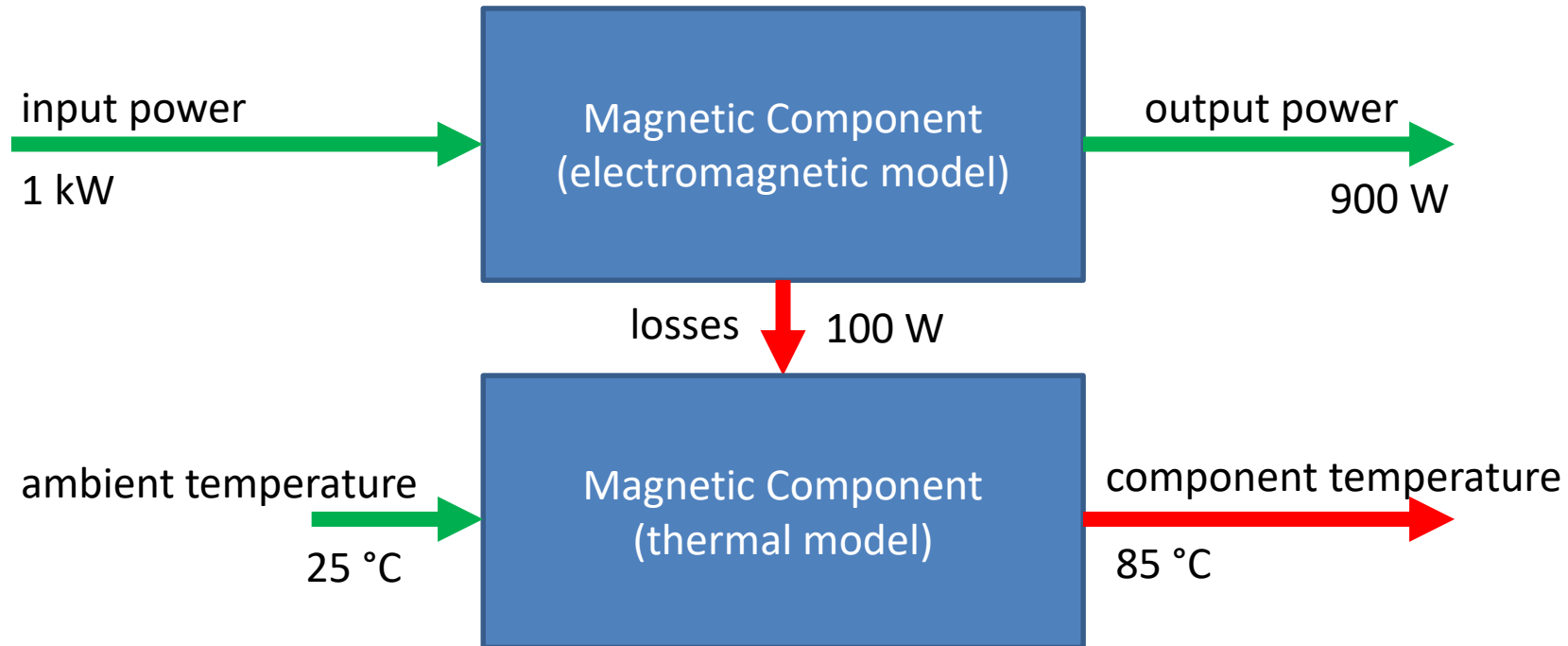
becomes hot in operation
-> damage at appr. 150 °C

Why does the transformer become hot?



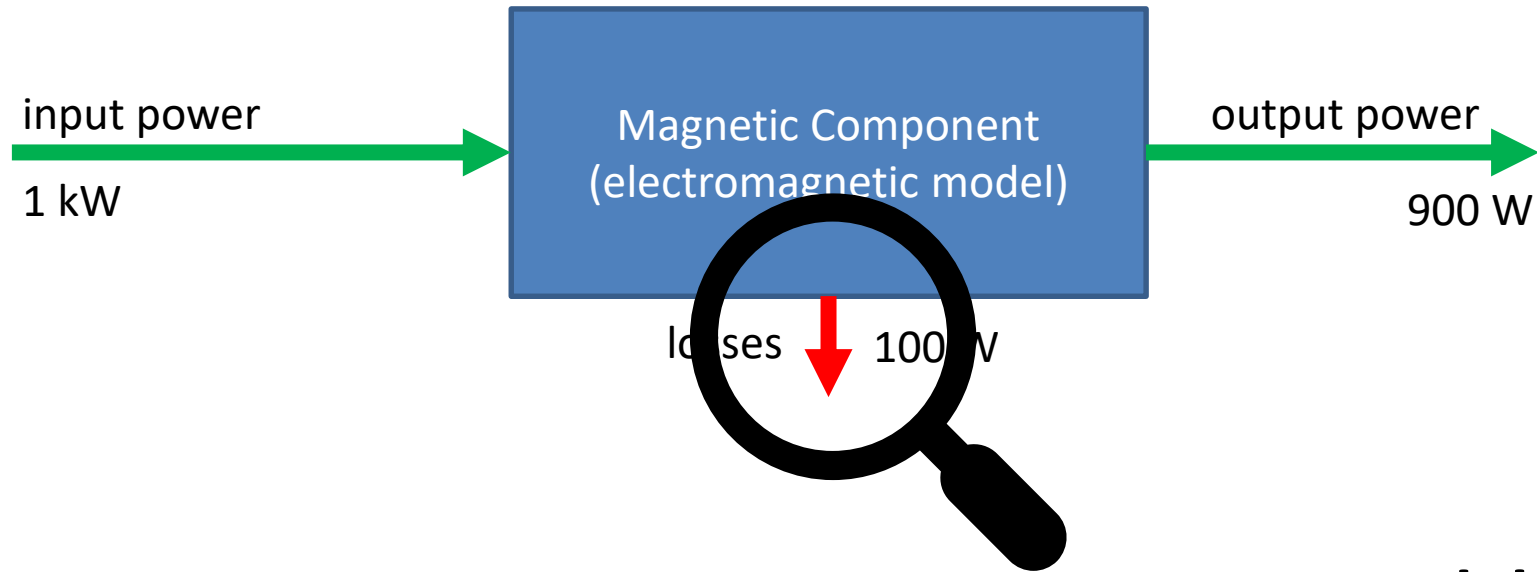
$$\text{Efficiency: } \eta = \frac{\text{output power}}{\text{input power}} = \frac{\text{input power} - \text{losses}}{\text{input power}} = \frac{900 \text{ W}}{1000 \text{ W}} = 90 \%$$

Why does the transformer become hot?



$$\text{Efficiency: } \eta = \frac{\text{output power}}{\text{input power}} = \frac{\text{input power} - \text{losses}}{\text{input power}} = \frac{900 \text{ W}}{1000 \text{ W}} = 90 \%$$

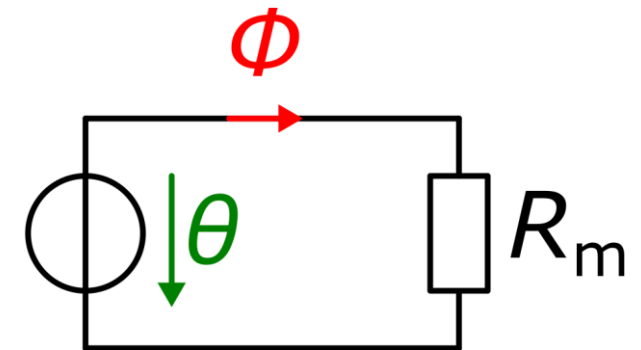
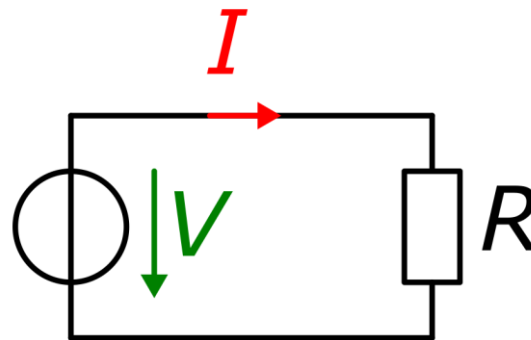
Why does the transformer become hot?

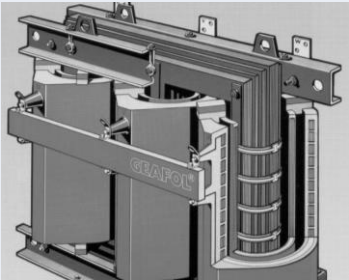



How to model the losses?

Electrical vs. magnetic circuits

Property	Electrical circuit	Magnetic circuit
Specific conductance	κ	μ
Resistance	$R = \frac{l}{\kappa A}$	$R_m = \frac{l}{\mu A}$
Ohm's law	$U = R I$	$\theta = R_m \phi$



Property	Electrical circuit	Magnetic circuit
Specific conductance	κ	μ
Resistance	$R = \frac{l}{\kappa A}$	$R_m = \frac{l}{\mu A}$
Ohm's law	$U = R I$	$\theta = R_m \phi$
Iron sheet 	High electric conductance κ	Very high magnetic conductance μ
Ferrite powder 	Low electric conductance κ	High magnetic conductance μ

Mechanism 1: Eddy Currents

Property	Electrical circuit
Specific conductance	κ
Resistance	$R = \frac{l}{\kappa A}$
Ohm's law	$U = R I$
Iron sheet	High electric conductance κ
Ferrite powder	Low electric conductance κ

← crucial for eddy currents:
high electrical resistance → low eddy current losses

→ High eddy currents

→ Low eddy currents

(recap)

Mechanism 2: Magnetic Hysteresis

Property	Magnetic circuit
Specific conductance	μ
Resistance	$R_m = \frac{l}{\mu A}$
Ohm's law	$\theta = R_m \phi$
Ferrite powder	High magnetic conductance μ

← “Imperfection of μ causes losses”

...too complicated to explain in one slide...

Speaking:

„magnetic conductance“ μ is called **permeability**

Mechanism 1: Eddy Currents

- Almost linear material behaviour
- Precise modelling in FEM simulation
- In our case: **can probably be neglected** completely

Mechanism 2: Magnetic Hysteresis

- Highly nonlinear material behaviour
- Several approaches for modelling the losses
- In our case: is the **key point of investigation**

Mechanism 2: Magnetic Hysteresis

Property	Magnetic circuit
Specific conductance	μ
Resistance	$R_m = \frac{l}{\mu A}$
Ohm's law	$\theta = R_m \phi$
Ferrite powder	High magnetic conductance μ

Switch from
circuit
to material



$$b = \mu h$$

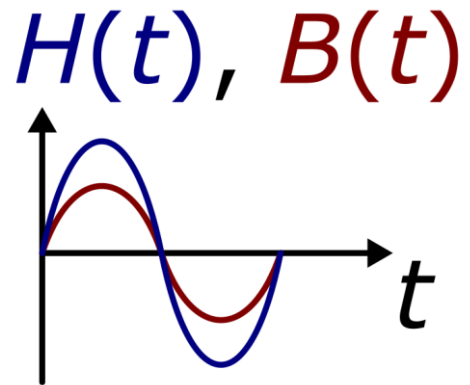
Material equation

Speaking:

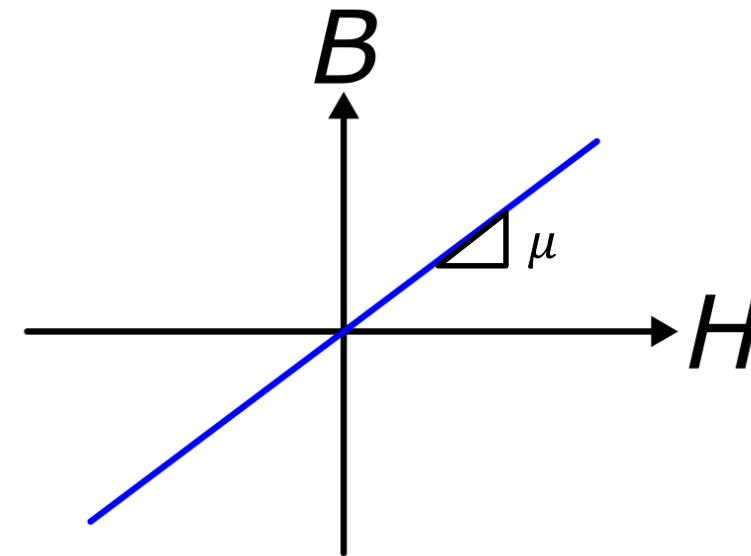
„magnetic conductance“ μ is called **permeability**

$$B(t) = \mu H(t)$$

Material equation

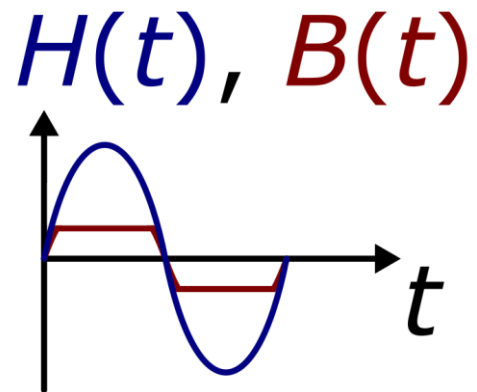


In a “perfect”/linear world:
Independent of the shape of B and H

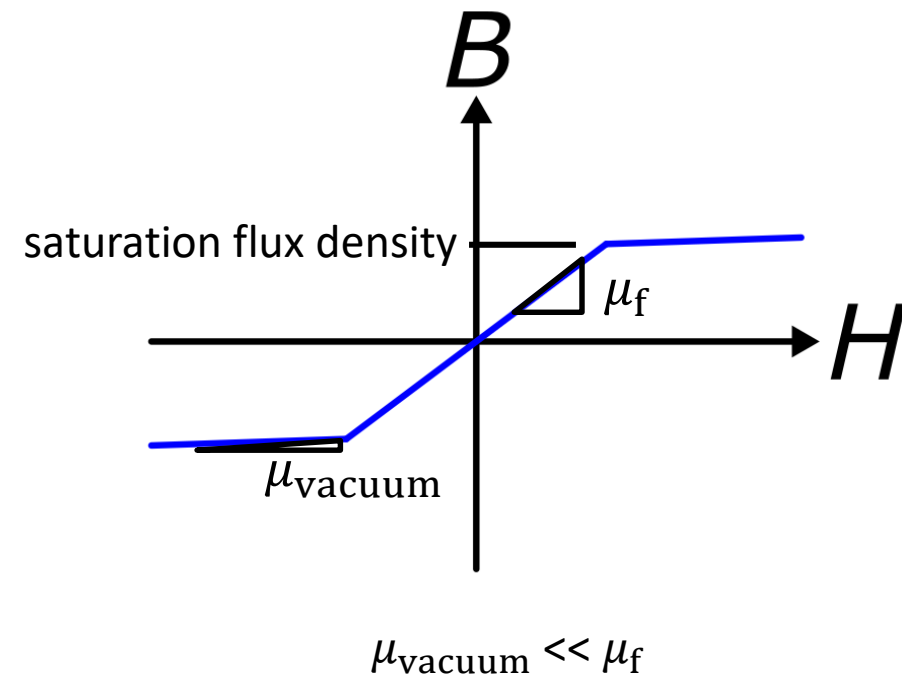


$$B(t) = \mu(H) H(t)$$

Material equation

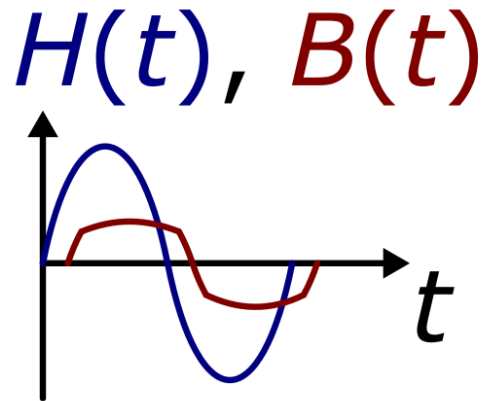


More realistic:

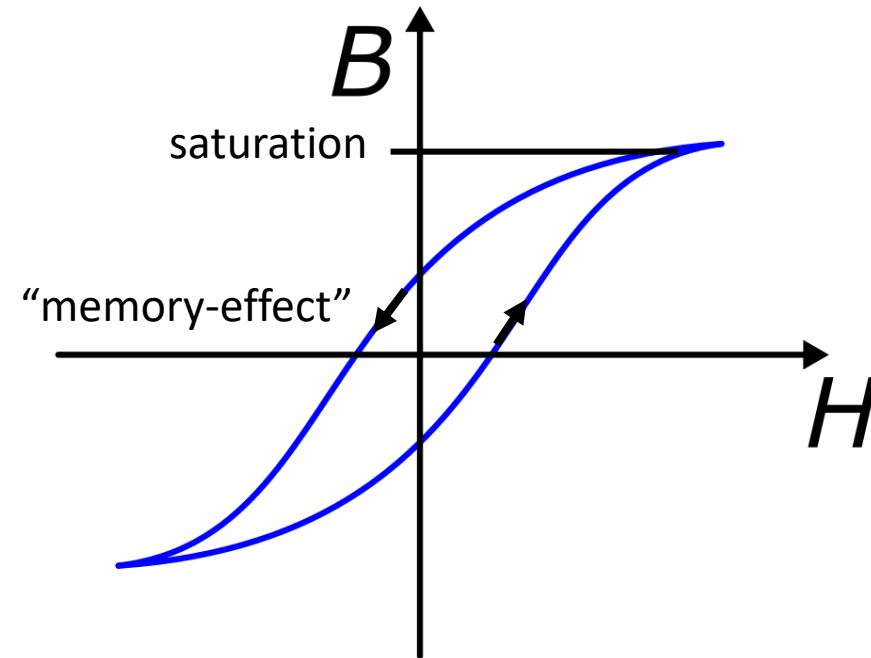


$$B(t) = \mu(H, T, f, \dots) H(t)$$

Material equation

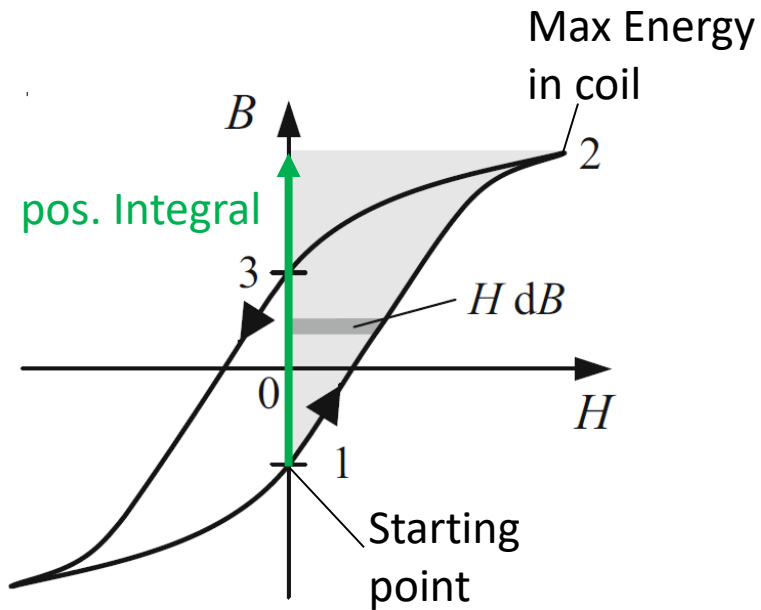


Reality:



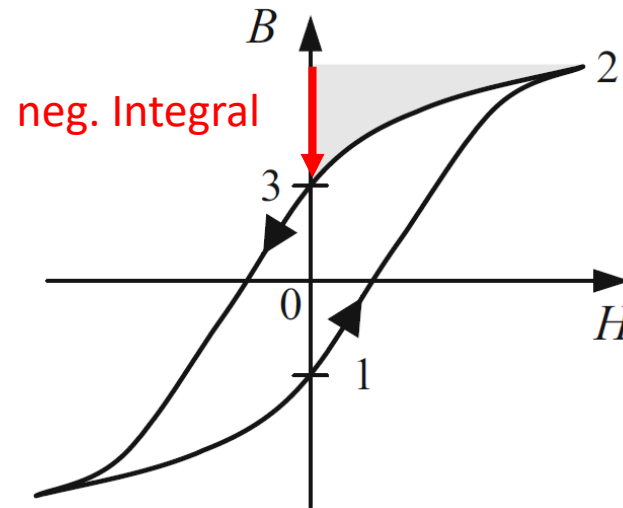
Losses due to Magnetic Hysteresis

Loading the coil

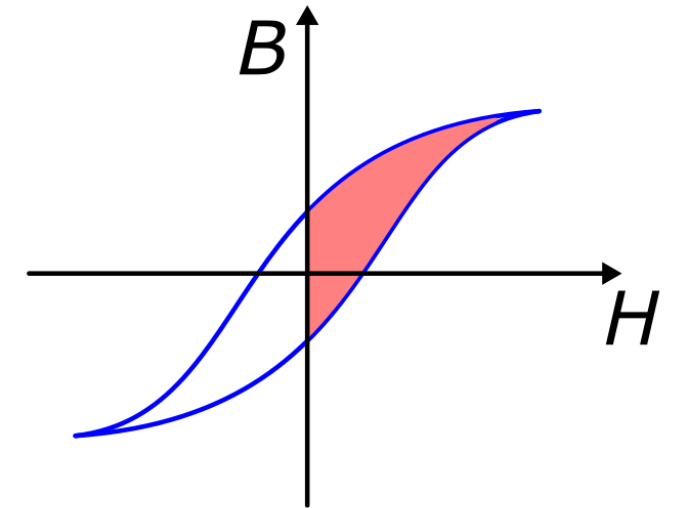


$$[H \cdot B] = \frac{A}{m} \cdot \frac{Vs}{m^2} = \frac{J}{m^3}$$

Unloading the coil



Dissipated energy w_v



$$w_v = \int H(B) \, dB$$

Mechanism 2: Magnetic Hysteresis

$$B(t) = \mu(H, T, f, \dots) H(t)$$

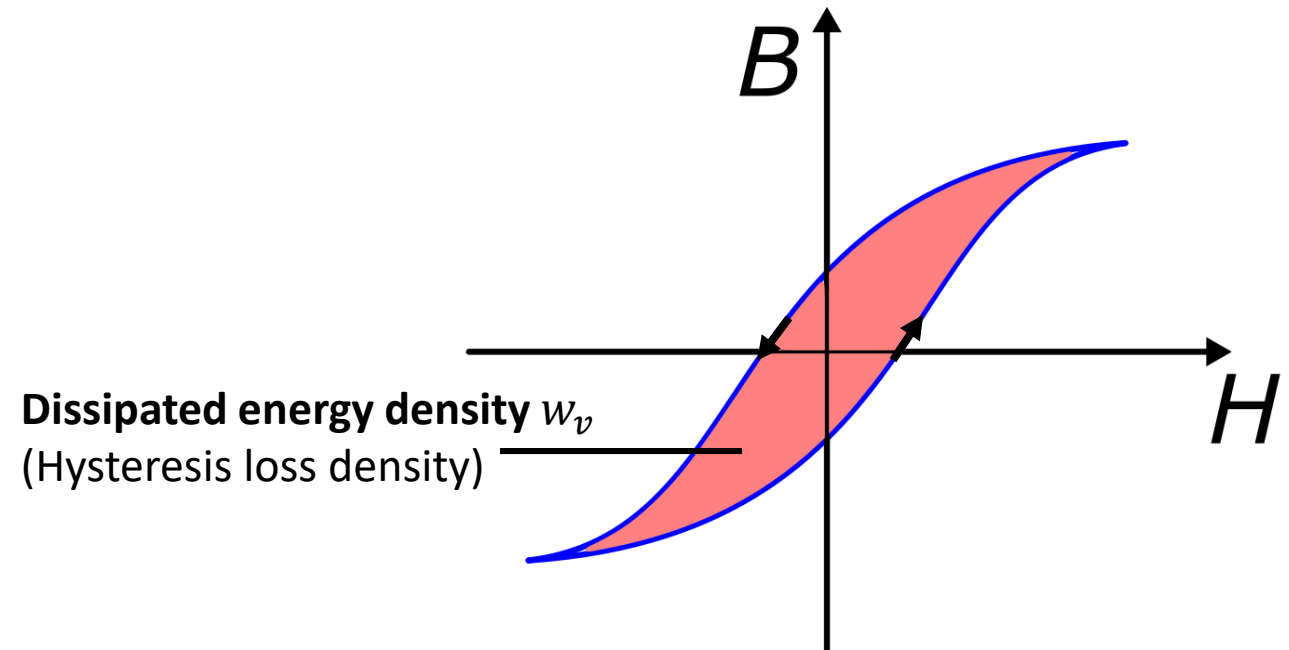
Material equation

$$p_v = \frac{w_v}{T} = f \int H(B) dB$$

power loss density

$$[p_v] = \frac{W}{m^3}$$

This curve is run through repetitively with period $T = \frac{1}{f}$



For a known curve: an exact calculation – Problem: impossible to measure each possible curve

Concept: Complex Numbers

Complex numbers:

$$\underline{x} = a + jb = \hat{x} (\cos(\varphi) + j \sin(\varphi)) = \hat{x} e^{j\varphi}$$

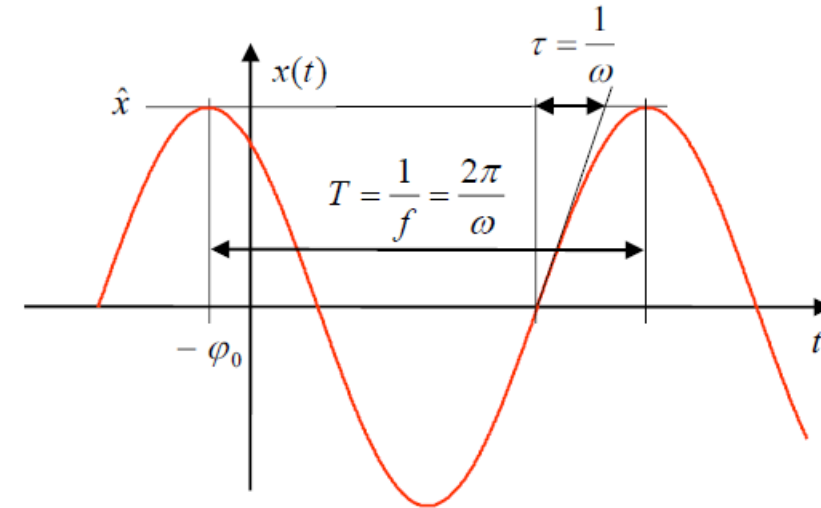
Real part:

$$\operatorname{Re}\{\underline{x}\} = a = \hat{x} \cos(\varphi)$$

Imaginary part:

$$\operatorname{Im}\{\underline{x}\} = b = \hat{x} \sin(\varphi)$$

-> used to model sinusoidal signals



Time domain:

$$x(t) = \hat{x} \cos(\omega t + \varphi_0) = \operatorname{Re}\{\underline{x}(t)\}$$

Use complex numbers:

$$\underline{x}(t) = \hat{x} e^{j(\omega t + \varphi_0)} = \underline{\hat{x}} e^{j\varphi_0} e^{j\omega t} = \underline{\hat{x}} e^{j\varphi_0} \underset{\substack{\uparrow \\ \text{often}}}{e^{j\omega t}}$$

Usually only the complex amplitude \underline{x} or \hat{x} is denoted!

(below saturation, sinusoidal excitation)

$$\underline{B} = \underline{\mu}(H, T, f) \underline{H}$$

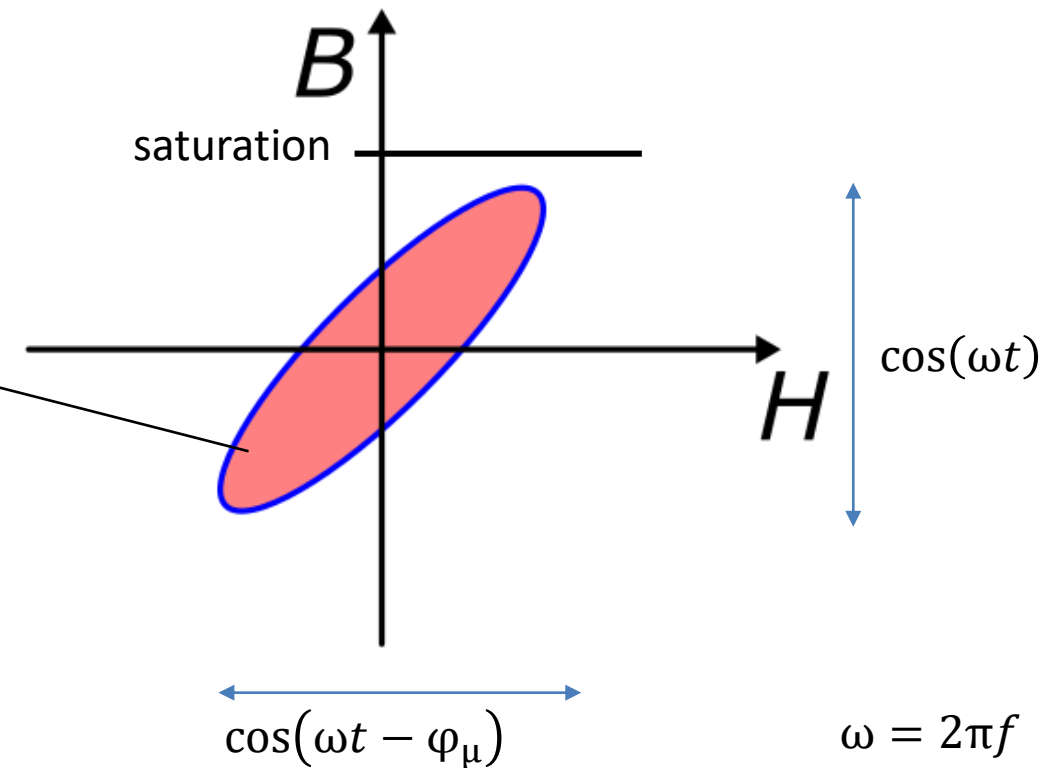
Material equation

$$p_v = \frac{1}{2} \cdot \omega \cdot \text{Im} \{ \underline{\mu} \} \cdot \hat{H}^2$$

Loss density

Benefits:	Drawbacks:
physical model	Only below saturation
simple	Only sinusoidal
fast	

elliptic approximation



Loss mechanisms: Steinmetz equation

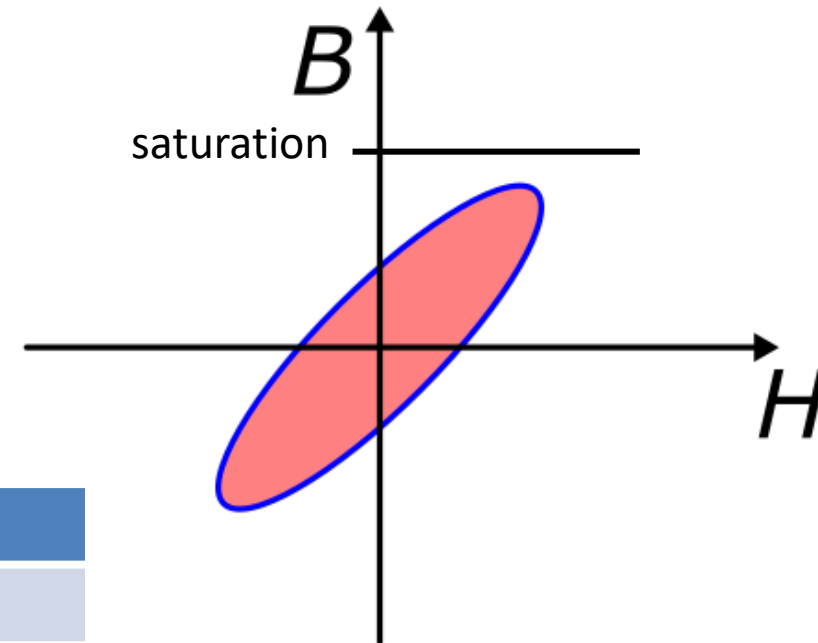


(below saturation, sinusoidal excitation)

$$p_v = k f^\alpha B^\beta$$

Steinmetz equation (SE) from 1892

$$\alpha = 1 \dots 1.5$$
$$\beta = 2 \dots 2.5$$



Benefits:	Drawbacks:
simple	Empirical / non physical
fast	Only below saturation
	Only sinusoidal

Loss mechanisms: improved Generalized Steinmetz Equation

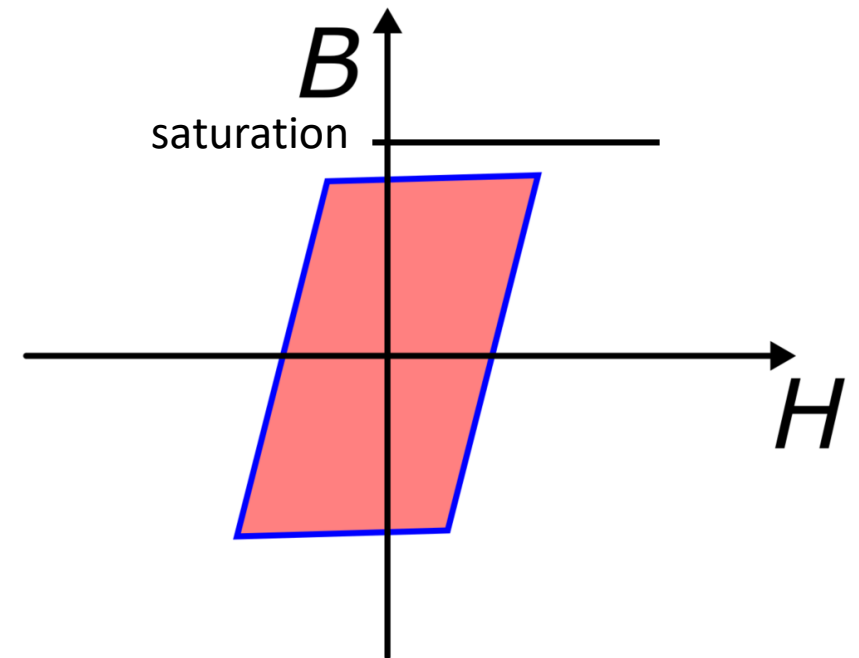
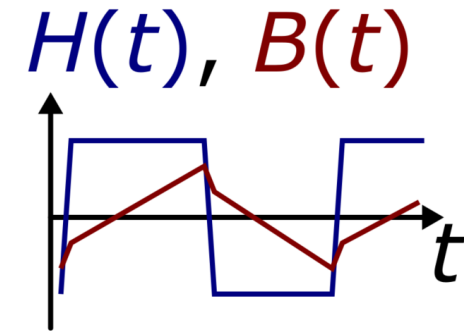


(below saturation)

$$p_v = \frac{1}{T} \int_0^T k_i \left| \frac{dB}{dt} \right|^\alpha (\Delta B^{\beta-\alpha}) dt$$

improved Generalized Steinmetz Equation (iGSE)

$$\text{with } k_i = \frac{k}{(2\pi)^{\alpha-1} \int_0^{2\pi} |\cos\theta|^\alpha 2^{\beta-\alpha} d\theta}$$



Benefits:	Drawbacks:
Allows arbitrary signals (AC)	Empirical / non physical
Simple (only sinusoidal loss measurements needed)	Becomes inaccurate (based on sinusoidal measurement data)
fast	Only valid below saturation

Loss mechanisms: improved Generalized Steinmetz Equation

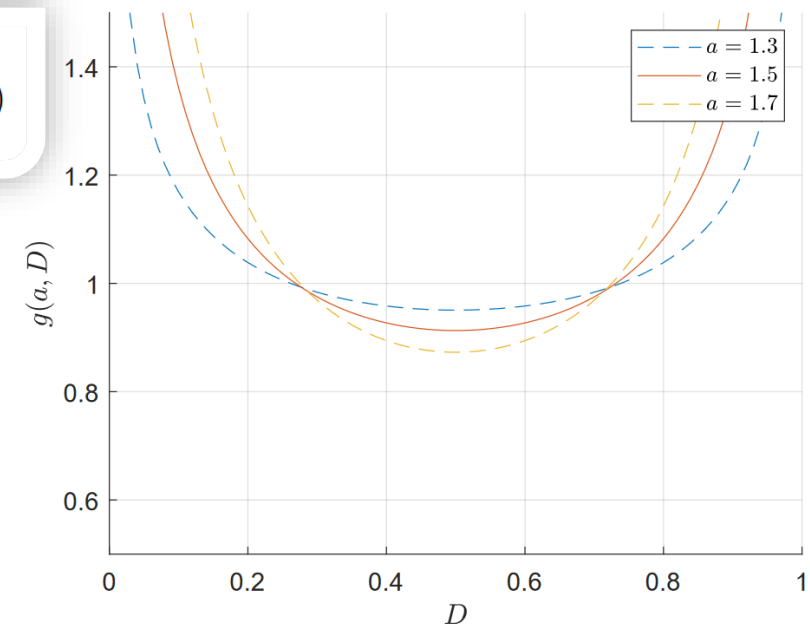
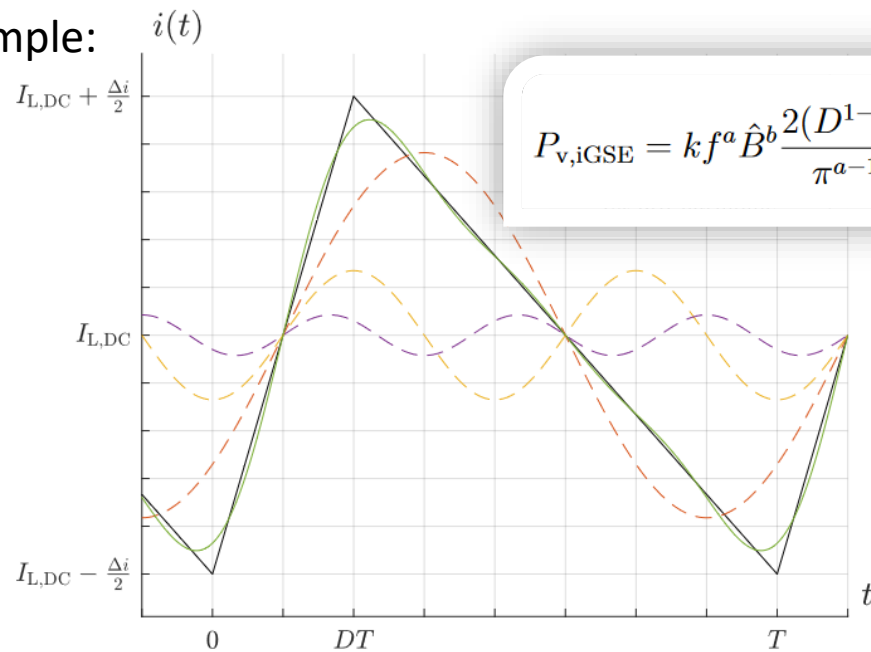
(below saturation)

$$p_v = \frac{1}{T} \int_0^T k_i \left| \frac{dB}{dt} \right|^\alpha (\Delta B^{\beta-\alpha}) dt$$

$$\text{with } k_i = \frac{k}{(2\pi)^{\alpha-1} \int_0^{2\pi} |\cos \theta|^\alpha d\theta 2^{\beta-\alpha}}$$

improved Generalized Steinmetz Equation (iGSE)

Example:



Most approaches directly try to model the losses without exact modelling of the BH curve.

➤ **Steinmetz Equation (SE), 1890s**

3 parameters

$$P_V = k \cdot f^\alpha \cdot \hat{B}^\beta$$

➤ **Improved Generalized Steinmetz Equation (iGSE), 2000s**

$$P_V = \frac{1}{T} \int_0^T k_i \cdot \left| \frac{dB}{dt} \right|^\alpha \cdot (\Delta B)^\beta dt$$

3 parameters

➤ **Improved-improved GSE (i²GSE), 2010s**

$$P_V = \frac{1}{T} \int_0^T k_i \cdot \left| \frac{dB}{dt} \right|^\alpha \cdot (\Delta B)^\beta dt + \sum_{l=1}^n Q_{rl} \cdot P_{rl}$$

8 parameters

(from MagNet webinar)

Alternatively, it can be tried to model the full BH curve and calculate the losses by integrating the loop area.

Preisach model (1935)

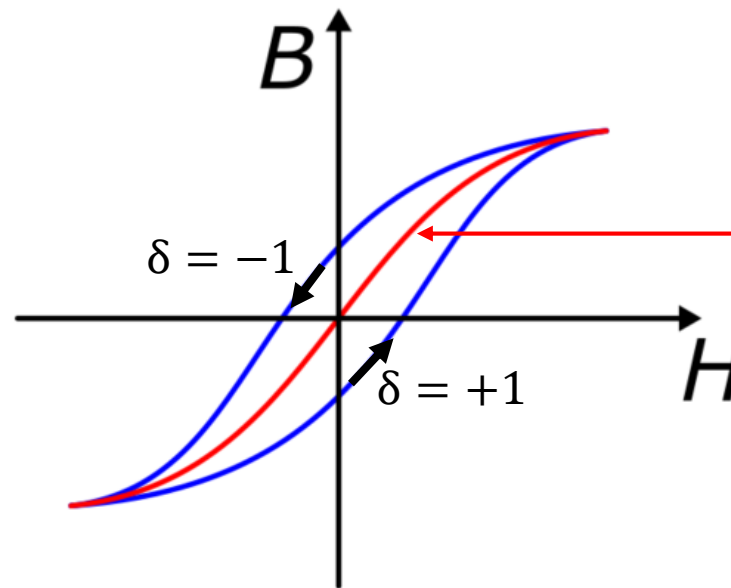
Chua (1970)

Jiles-Atherton model (1980)

Coleman-Hogdon (1986)

Usually magnetization is used:

$$B = \mu_{\text{vacuum}} M$$



An hysteretic
magnetization curve:

$$M_{\text{an}}^{\text{iso}} = M_s \left(\coth \left(\frac{H_e}{a} \right) - \frac{a}{H_e} \right)$$

$$\frac{\partial M}{\partial H} = \frac{1}{1+c} \cdot \frac{M_{\text{an}} - M}{\nu_0 \delta k - \alpha(M_{\text{an}} - M)} + \frac{c}{(1+c)} \frac{\partial M_{\text{an}}}{\partial H}$$

α, c, k must be estimated

Alternatively, it can be tried to model the full BH curve and calculate the losses by integrating the loop area.

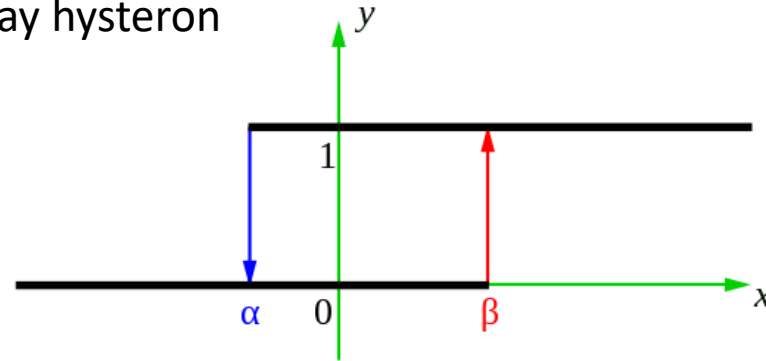
Preisach model (1935)

Chua (1970)

Jiles-Atherton model (1980)

Coleman-Hogdon (1986)

relay hysteron



$$y(x) = \begin{cases} 1 & \text{if } x \geq \beta \\ 0 & \text{if } x \leq \alpha \\ k & \text{if } \alpha < x < \beta \end{cases}$$

Loss mechanisms: Overview BH representations

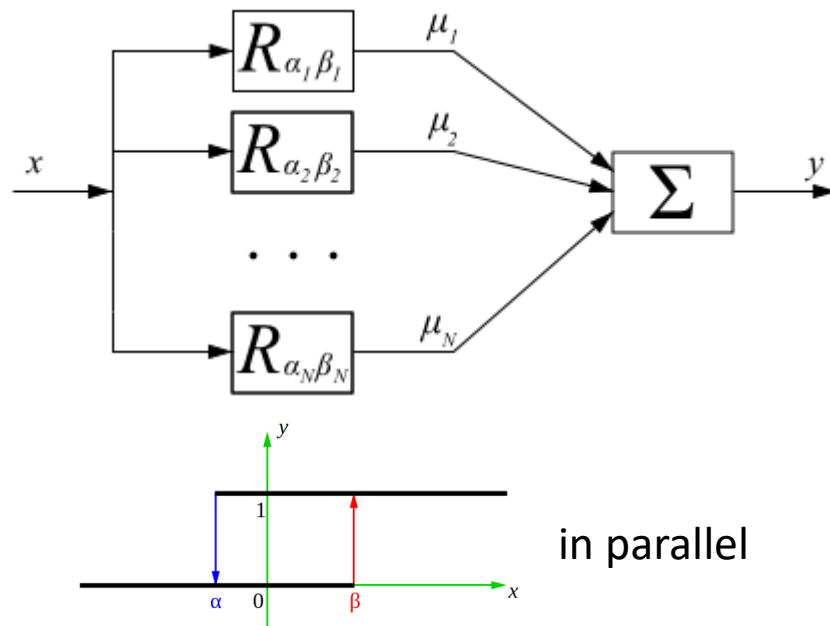
Alternatively, it can be tried to model the full BH curve and calculate the losses by integrating the loop area.

Preisach model (1935)

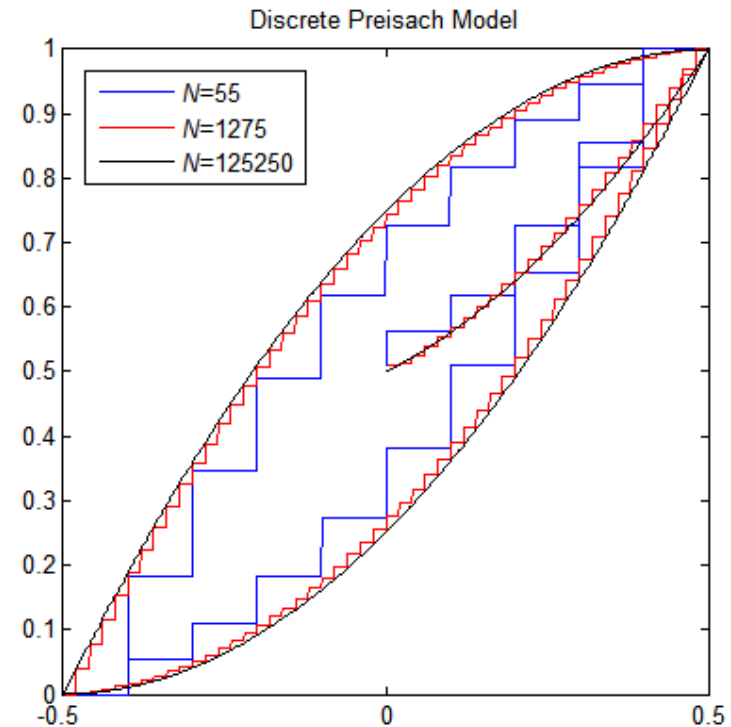
Chua (1970)

Jiles-Atherton model (1980)

Coleman-Hogdon (1986)



in parallel



$2N$ parameters have to be determined!

Alternatively, it can be tried to model the full BH curve and calculate the losses by integrating the loop area.

Preisach model (1935)

Chua (1970)

Jiles-Atherton model (1980)

Coleman-Hogdon (1986)

	Error	Complexity of parameter estimation procedure	Complexity of required measurements
Jiles - Atherton	Medium	High	Low
Preisach	Low	Low	High
Chua	Medium	Medium	Low
Coleman - Hogdon	Medium	Medium	Low
Bouc - Wen	Medium	High	Low

Alternatively, it can be tried to model the full BH curve and calculate the losses by integrating the loop area.

Preisach model (1935)

Chua (1970)

Jiles-Atherton model (1980)

Coleman-Hogdon (1986)

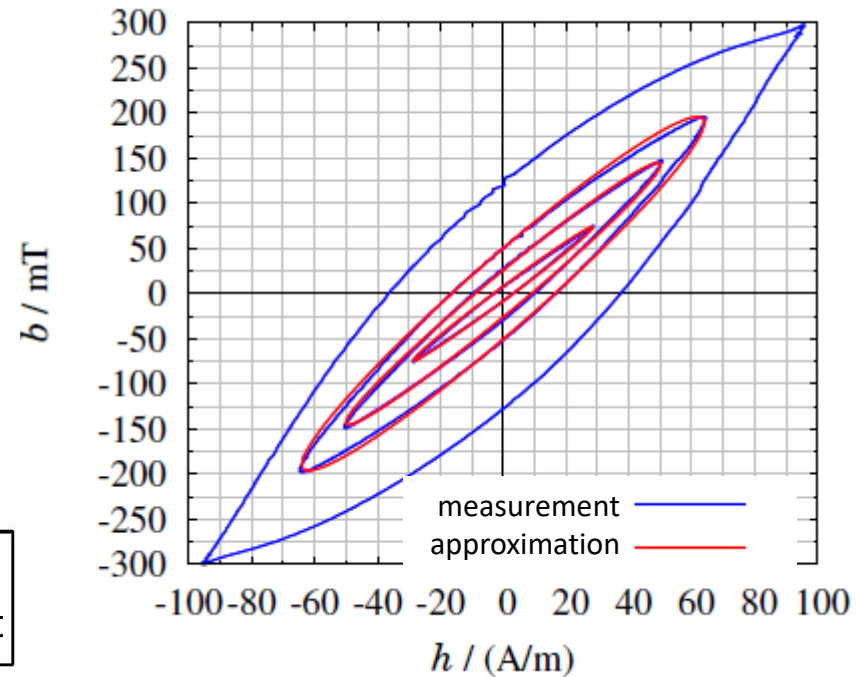
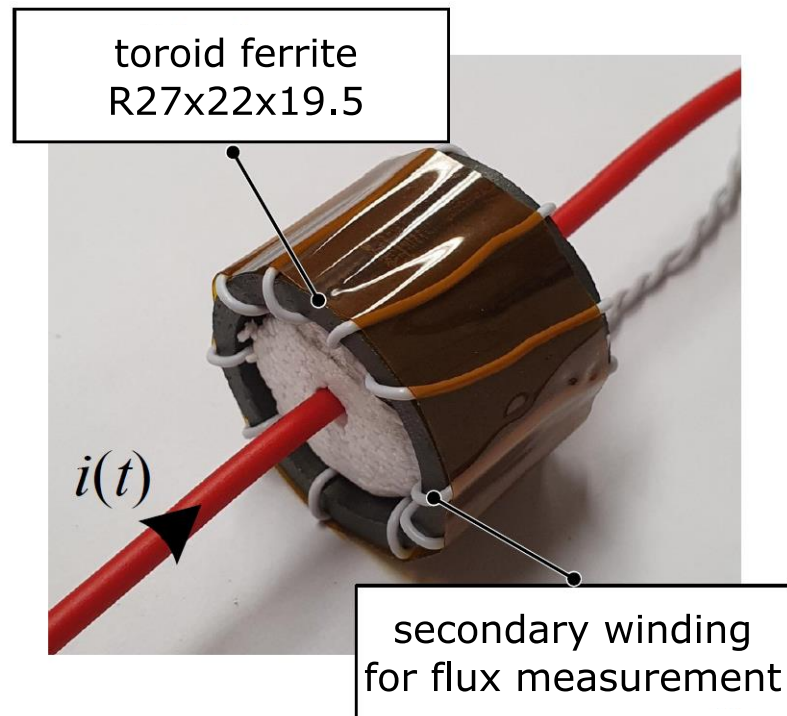
Why are they (almost) never used in Power Electronics?

“The modelling of

- Temperature
- Flux density shape
- Frequency
- (DC field premagnetisation) not covered in Challenge

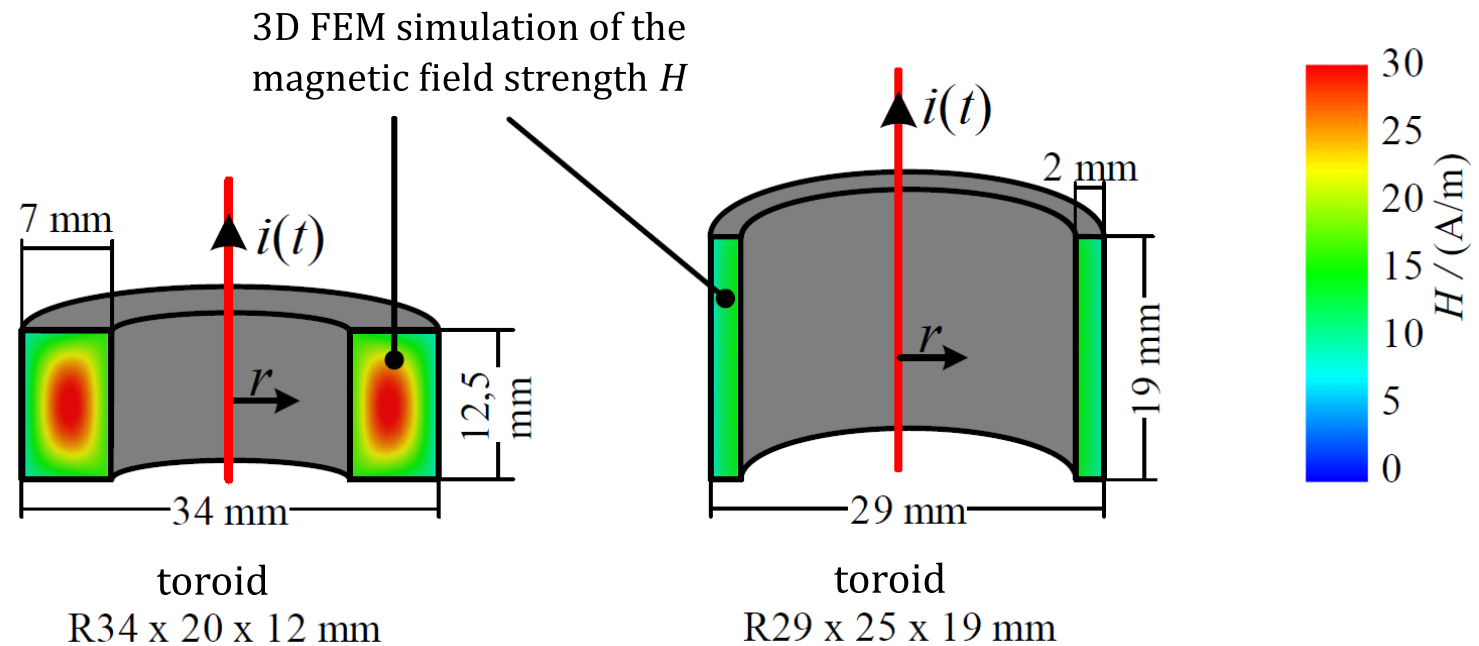
requires extensive data sets for the various materials materials, but these are not made available by the manufacturers and have to be have to be collected through time-intensive measurements.”

(Albach, translated)



Assumption: homogeneous flux fields

- High frequency measurements require very thin toroid cores



$$f = 2 \text{ MHz}$$

$$\underline{\mu_r} = 1750e^{-j10^\circ}$$

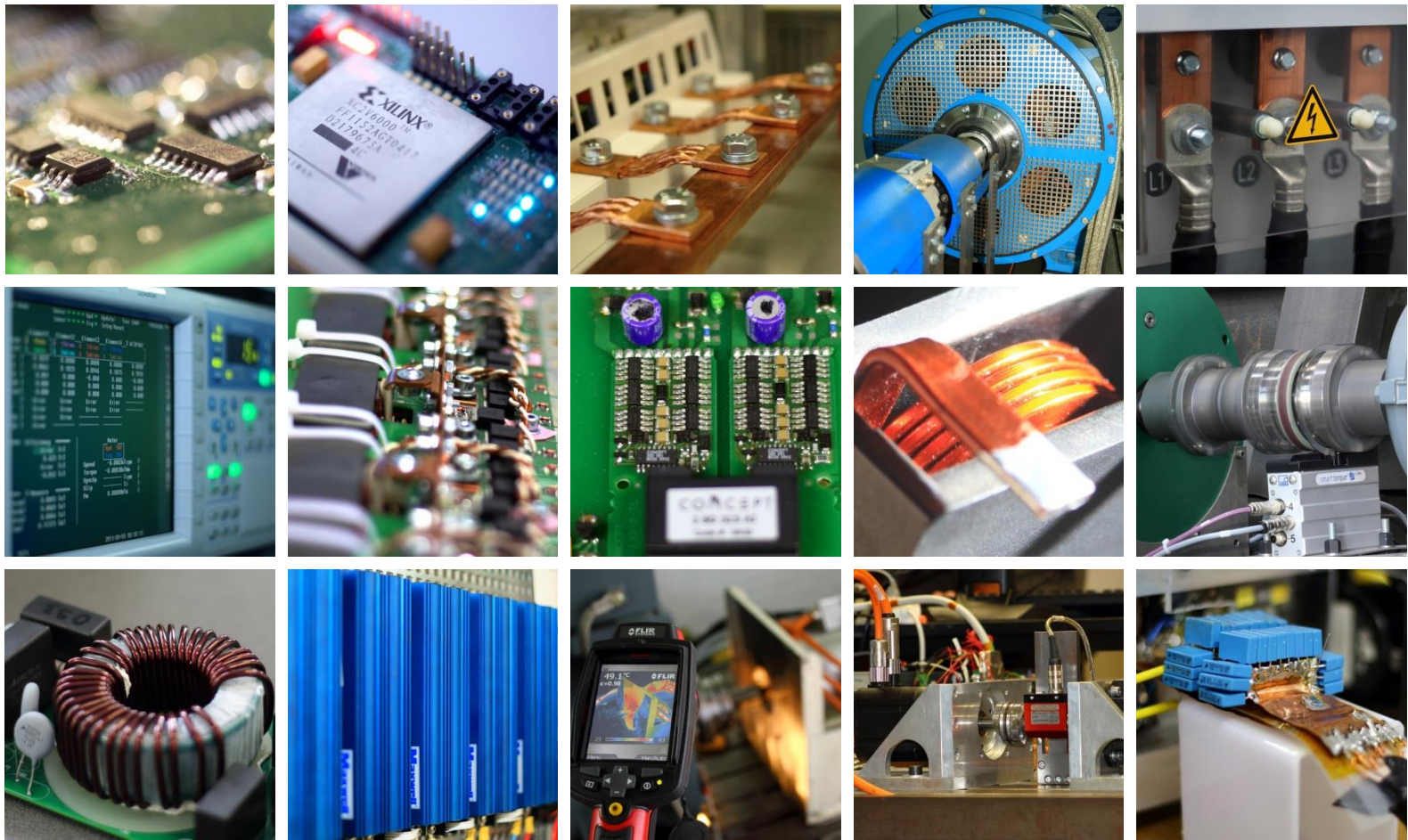
$$\underline{\varepsilon_r} = 60000e^{-j10^\circ}$$

- Goal:

$$p_v = \text{function}(\text{waveform}, \text{frequency}, \text{temperature})$$

- Here: only hysteresis
 - Usually calculated without exact modelling the BH curve (not enough data)
 - Big dataset is given:
 - 1. model BH loop
 - 2. intergrate for losses p_v
- } feasible?

Power Electronics and Electrical Drives Paderborn University



- [1] <https://www.reinhausen.com/de/service-details/transformatoren-serviceleistungen/modernisierung-und-austausch>
- [2] https://www.ew.tu-darmstadt.de/media/ew/rd/ew_praktika/lv_praktika/anleitungen/M2_Drehstromtransformator_v2.18.pdf
- [3] GaN Systems - Reference Design “*GaN-Based 3KW Full Bridge LLC Resonant Converter*”