

Passive Respiration Detection via mmWave Communication Signal Under Interference

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Abstract—Recent research has highlighted the detection of human respiration rate using commodity WiFi devices. Nevertheless, these devices encounter challenges in accurately discerning human respiration amidst the prevailing human motion interference encountered in daily life. To tackle this predicament, this paper introduces a passive sensing and communication system designed specifically for respiration detection in the presence of robust human motion interference. Operating within the 60.48 GHz millimeter wave (mmWave) band, the proposed system aims to detect human respiration even when confronted with substantial human motion interference within close proximity. Subsequently, a neural network is trained using the collected data by us to enable human respiration detection. The experimental results demonstrate a consistently high accuracy rate over 95% of the human respiration detection under interference, given an adequate sensing duration. Finally, the accuracy of the proposed passive detection system of counting human respiration achieves 90% overall.

Keywords—Integrated Sensing and Communication (ISAC), mmWave, Device-free Respiration Monitoring (DFRM), Deep Learning

I. INTRODUCTION

The ability to accurately and reliably detect the human respiratory patterns has long been a crucial aspect of healthcare monitoring, sleep analysis, and physical activity tracking. A person's respiration provides valuable insights into his overall well-being. For example, a respiratory rate value greater than 27 breaths per minute (bpm) has been identified as a significant predictor of cardiac arrest and is utilized in the prediction of conditions such as pneumonia or lower respiratory tract infections [1–3]. Traditionally, respiration detection has relied on the use of sensors or wearable devices, which often impose constraints on the individual's freedom of movement and may lead to discomfort or inconvenience [4].

Passive sensing, a branch of the ISAC technology, has paved the way for device-free respiration monitoring (DFRM) [5, 6]. This innovative approach eliminates the need for physical contact by capturing and processing signals influenced by subtle movements associated with respiration. Importantly, passive sensing has demonstrated its ability to accurately capture the motion of the chest caused by respiration without significantly occupying the communication bandwidth.

There have been several existing methods to estimate the respiratory rate by using non-contact techniques, with most of

them relying on channel state information (CSI) or received signal strength indicator (RSSI) [7–10]. However, the CSI methods typically utilize only a small part of the signal, resulting in increased delay and compromised accuracy. Moreover, the CSI method has to eliminate the carrier frequency offset (CFO) between the transmitter and receiver, realizing the synchronization between them.

Currently, most DFRM systems employ commercial WiFi [7–10], offering convenience and ease of implementation. However, the detection precision of these systems is limited by the wavelength with the following restriction:

$$f_d = \frac{v \cos(\theta)}{\lambda}, \quad (1)$$

where f_d is Doppler Frequency, θ is the angle between the direction of velocity and the direction of arrival of the signal, λ is the wavelength of the signal.

To overcome the limitations associated with lower sensitivity, the utilization of mmWave offers distinct advantages in DFRM systems. Mmwave has shorter wavelengths, enabling higher precision in detecting subtle respiratory movements. Furthermore, mmWave technology is extensively employed in next-generation communication systems, creating an opportunity for synergistic integration.

In general, there are two types of interference:

- 1) **Static Environmental Interference:** Environmental interference encompasses external factors that potentially compromise the accuracy and reliability of a measurement or detection process, such as the presence of reflected signals from the static environment. Within the domain of respiration detection, interference stemming from the static environment can significantly impact the precision of measurements.
- 2) **Motion Interference:** The interference caused by human motion is induced through random human motion around the detected target in the area during respiration detection. Movements such as walking, running, or even slight body motions can introduce artifacts and distort the respiration signal. Motion interference poses a challenge because it can lead to inaccurate measurements and make it difficult to extract the true respiratory pattern.

To address these challenges, we propose a DFRM system based on passive mmWave radar with strong anti-interference

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ability, acceptable processing speed, and improved accuracy. It will offer a powerful solution for advancing both healthcare monitoring and communication capabilities.

The subsequent sections of this paper are structured as follows. Section II presents an introduction to the passive sensing and communication system based on mmWave. The signal processing algorithm for passive sensing and the methodology for approximating classification accuracy are expounded in Section III. Section IV outlines the system implementation, accompanied by the demonstration and discussion of experimental results. Lastly, the paper is concluded in Section V.

II. OVERVIEW OF THE SYSTEM

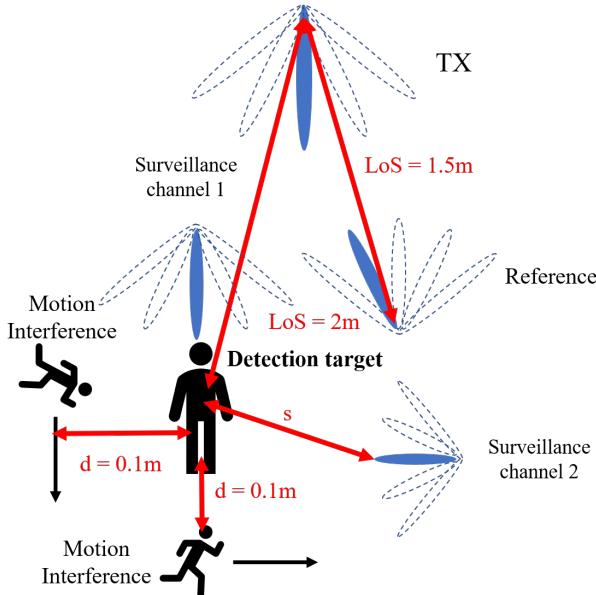


Fig. 1. Model of the Sensing System.

The model of passive sensing and communication system working on the mmWave waveband is elaborated in Fig. 1. The transmitter portion is composed of a Sivers mmWave phased array block with 16 antennas, which transmits the communication signal with a carrier frequency of 60.48 GHz over the downlink. The receiver portion consists of three different channels: The first channel denoted as the reference channel is settled to receive the communication signal. The second and the third channels denoted as surveillance channel are used to receive the reflected signal from the chest respectively. To be specific, the receiver towards the reference channel obtains the communication signal without Doppler frequency and interference. The receivers toward the two surveillance channels receive the echo signal reflected from the moving target in the environment, and the echo signal is processed by the receiver subsequently to obtain the Doppler frequency information of the target [11], which can derive the approximate human respiratory rate by deep learning method. And s in Fig. 1 represents the distance between the target and the closest receiver which is a variable in different experiments.

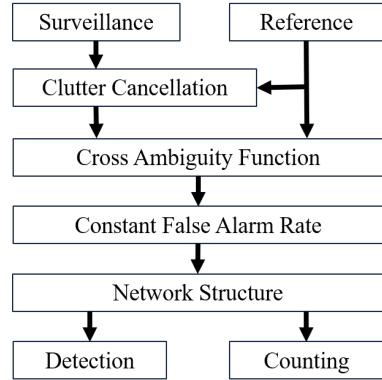


Fig. 2. Signal Processing Diagram.

Subsequently, a clutter cancellation method is applied to mitigate the interference based on static scattering within the signals obtained from both the reference channel and the surveillance channel. The signals from both the reference channel and the surveillance channel after clutter cancellation are then subjected to the Cross Ambiguity Function (CAF) and Constant False Alarm Rate (CFAR) algorithms to extract the Doppler frequency information and reduce the chances of false detection. The Time-Doppler spectrograms obtained from the CAF and CFAR are subsequently fed into a neural network that has been trained using the collected data, enabling the detection of human respiration.

The primary objective of the proposed system is twofold: Firstly, to detect the presence of the human respiration in the surrounding environment under interference, and secondly, to accurately count the respiratory rate.

III. SYSTEM MODEL

A. Signal Model

The transmitter transmits the communication signal $s(t)$ to the receiver, and the received signal via the reference channel, denoted as $y_r(t)$, can be expressed in (2)

$$y_r(t) = \alpha_r s(t - \tau_r) e^{-j2\pi f_\Delta t} + n_r(t), \quad 0 \leq t \leq T, \quad (2)$$

where α_r and τ_r denote the attenuation coefficient and time delay of the reference channel respectively, f_Δ denotes the frequency offset between the transmitter and receiver, $n_r(t)$ denotes the noise and other interference (e.g., sensing target, the surrounding static scattering clusters), T denotes the duration of the transmit signal. Based on the scattering loss, the interference power is often much weaker than the signal power in the reference channel.

The received signal via the i -th ($i = 1, 2$) surveillance channel, denoted as $y_{s,i}(t)$, can be expressed in (3)

$$y_{s,i}(t) = \alpha_{s,i}^{\text{tar}}(t) s(t - \tau_{s,i}^{\text{tar}}(t)) e^{-j2\pi f_\Delta t} e^{-j2\pi f_{s,i}^{\text{tar}}(t)t} + \sum_{k=1}^{L_s} \alpha_{s,i}^k s(t - \tau_{s,i}^k) e^{-j2\pi f_\Delta t} + n_{s,i}(t), \quad 0 \leq t \leq T, \quad (3)$$

where $\alpha_{s,i}^{\text{tar}}(t)$, $\tau_{s,i}^{\text{tar}}(t)$, and $f_{s,i}^{\text{tar}}(t)$ denote the time-varying attenuation coefficient, time delay and Doppler frequency of the

scattered path off the sensing target respectively, f_Δ denotes the frequency offset between the transmitter and receiver, L_s denotes the number of paths from clutter interference based on static scattering, $\alpha_{s,i}^k$ and $\tau_{s,i}^k$ denote the attenuation coefficient and time delay of the k -th path of static clutter interference respectively, and $n_{s,i}(t)$ denotes the noise.

B. Signal Processing of Passive Sensing

The received signals from reference and surveillance channels are sampled at the baseband with a period T_s , which can be expressed by

$$y_r[n] = y_r(nT_s) \text{ and } y_{s,i}[n] = y_{s,i}(nT_s), \quad (4)$$

where $n = 1, 2, \dots, T/T_s$, $i = 1, 2$ which denote the two surveillance channels. The received signals in two surveillance channels undergo the clutter cancellation firstly to mitigate the clutter interference originating from the static environment. Subsequently, it is processed by the Cross Ambiguity Function and Constant False Alarm Rate algorithms to calculate the Doppler Frequency within the vicinity. The resulting information is then forwarded to a network architecture to facilitate both the detection and counting tasks of human respiration.

1) *Adaptive Clutter Cancellation Algorithm:* Note that the signal components with zero Doppler frequency in $y_{s,i}[n]$ can seriously interfere the estimation of the target Doppler frequency $f_{s,i}^{\text{tar}}(t)$ in (3). To solve this problem, the Adaptive Clutter Cancellation Algorithm is used to eliminate the interference of reflected signal from the static environment [12]. The algorithm encompasses the Least Mean Square (LMS), recursive minimum algorithm, and normalized least mean square algorithm. In this work, an adaptive direct path clutter cancellation algorithm based on LMS is used.

The \mathbf{V}_r is the matrix of the delayed reference signal in the reference channel $y_r[n]$ expressed by

$$\mathbf{V}_r = \begin{bmatrix} y_r[1] & y_r[0] & \cdots & y_r[-P+2] \\ y_r[2] & y_r[1] & \cdots & y_r[-P+3] \\ \vdots & \vdots & \vdots & \vdots \\ y_r[N] & y_r[N-1] & \cdots & y_r[N-P+1] \end{bmatrix}, \quad (5)$$

where P is the number of delays, N is the number of sample points in one Coherent Integration Time (CIT).

Suppose $y_{s,i}$ is the received signal vector of the surveillance channel in one CIT time, then $\mathbf{V}_{s,i}$ is the matrix of the i -th surveillance channel represented by

$$\mathbf{V}_{s,i} = [y_{s,i}[1], y_{s,i}[2], \dots, y_{s,i}[N]]^T, \quad (6)$$

according to the LMS method, the weight matrix can be represented by

$$\mathbf{K} = (\mathbf{V}_r^H \mathbf{V}_r)^{-1} \mathbf{V}_r^H \mathbf{V}_{s,i}, \quad (7)$$

the received signal of the surveillance channel in one CIT after eliminating clutter interference is denoted as $y_{s,i}^*$, which can be expressed by

$$y_{s,i}^* = y_{s,i} - \mathbf{V}_r \mathbf{K}. \quad (8)$$

2) *Doppler Frequency Estimation Algorithm:* Cross Ambiguity Function (CAF) is a crucial radar signal processing tool that plays a vital role in representing the matched filter response. It provides valuable information about the time delay presented in the echo signal as well as the energy of the Doppler-shifted echo component. By estimating the correlation between signals collected from the reference channel and surveillance channel, the CAF enables the estimation of target parameters such as time delay and Doppler frequency shift. The expression of the CAF in one coherent integration time (CIT) is shown below:

$$R_i(f_d) = \max_{\tau} \sum_{n=0}^{N-1} y_{s,i}^*[n] y_r^\dagger[n-\tau] e^{-j2\pi f_d n T_s}, \quad (9)$$

where i denotes the i -th surveillance channel, $(\cdot)^\dagger$ denotes the complex conjugate, f_d denotes the Doppler frequency, N denotes the number of sample points in one CIT, and τ denotes the matched time delay.

By examining the CAF, it becomes apparent that the CAF traverses various combinations of time delay and Doppler frequency within their respective feasible ranges. When the values of τ and f_d align with the actual time delay and Doppler frequency shift between the surveillance channel and the reference channel, a distinct local peak marked as $R_i(f_d)$ emerges. Utilizing this information, one can estimate the precise values of the time delay between the reference channel and surveillance channel, and the Doppler frequency of the sensing target. However, in this work, since we only care about the feature extraction of the Doppler frequency, the time delay τ is not considered as a parameter in (9).

C. Methodologies to Mitigate Interference

Due to factors such as scattering from different body parts and noise interference in the environment, the Doppler frequency is not unique. The approach to cancel the possible errors used in this work is the two-dimensional Constant False Alarm Rate algorithm (CFAR) [13]. In the CFAR algorithm, the unit that needs to be compared with the threshold is referred to as the cell under test (CUT). Typically, an equal number of units in the surrounding of the CUT are used to compute the background noise power, and these units are called training units (w). In addition, there are several units between the CUT and the training units that do not participate in the calculation of the background noise power, and these units are referred to as guard units. The threshold $\beta = \alpha P_n$ is determined by the background noise power P_n and the threshold factor α . The background noise power is the mean of the training units and can be written as

$$P_n = \frac{1}{N} \sum_{w \in W} w, \quad (10)$$

where N is the number of training units, and W is the set of training units. The threshold factor α is determined by the pre-defined false alarm rate and can be expressed as

$$\alpha = N(P_{fa}^{-1/N} - 1), \quad (11)$$

where P_{fa} represents the false alarm rate.

The CFAR algorithm dynamically adjusts the threshold for target detection based on the level of noise present in the environment. This adaptive adjustment helps to reduce the chances of false detection, missed detection, and false alarms. Furthermore, the CFAR algorithm effectively handles the scenarios where no Doppler frequency is present, ensuring accurate detection and improved performance overall.

D. Respiration Detection Based on Deep Learning

For the two primary research tasks of respiration detection, namely the relationship between the detection of respiration presence or absence under interference and sensing duration, and the counting of the respiratory rate under interference, deep learning methods are chosen for the analysis. For the consideration of generalizability and time cost, ResNet [14] is adopted in this work.

The basic building block of ResNet is the residual block, consisting of a skip connection and batch normalization. The skip connection directly adds the input to the output, enabling the network to transfer the gradient well to avoid the gradient disappearing and exploding during the training process.

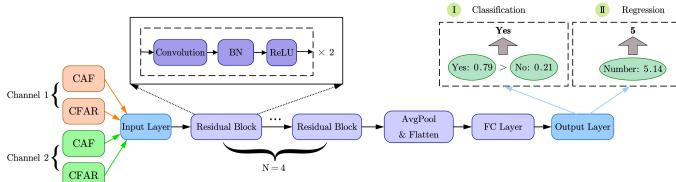


Fig. 3. Network Structure.

The network structure is elaborated in Fig. 3, which uses four residual blocks. The input of the ResNet are Time-Doppler spectrograms obtained from CAF and CFAR results of two surveillance channels as raw CAF results provide more micro information and CFAR provides more macro information. The output of the ResNet is different for respective tasks. For the detection of respiration presence or absence within sensing duration, treat it as a binary classification task and then the output layer has two nodes. On the other hand, for the respiratory rate counting problem, treat it as a regression task and the output layer has only one node. Since we only consider the case where the number of breaths is an integer, the output of the model will be rounded. Then we can evaluate the accuracy of the regression results which contain different discrete counting values as a classification problem.

IV. EXPERIMENTS

A. Implementation

The system diagram, as depicted in Fig. 4, presents an overview of the overall system configuration. In the transmitter section, a NI USRP-2954R is utilized to generate an intermediate frequency (IF) signal centered at 500 MHz. This IF signal is subsequently upconverted to 60.48 GHz and transmitted using a Sivers 60.48 GHz phased array. The transmission signal comprises a training sequence lasting 16 μ s and is

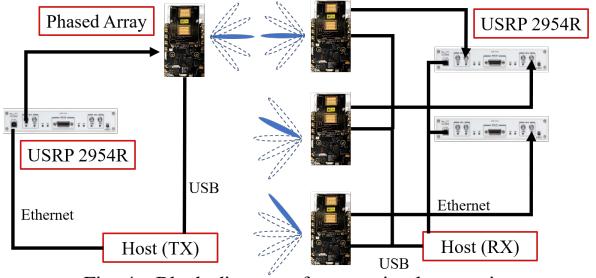


Fig. 4. Block diagram of system implementation.

succeeded by an Orthogonal Frequency Division Multiplexing (OFDM)-modulated data payload with a duration of 200 μ s. On the receiver side, three 60 GHz phased arrays called Ref, Sur1 and Sur2 are connected to three channels of two USRPs, enabling the reception of signals from both the reference channel and two surveillance channels. Note that Sur2 outputs the clock signal to Sur1 and Ref which synchronizes the clocks and eliminates the carrier frequency offset among them and these three phased arrays are controlled by laptops, facilitating collaborative beam switching functionality. After a predefined duration, the receiver host reads the data samples collected by two USRPs through Gigabit Ethernet ports. In this work, the duration of CIT is 0.1 s.

Fig. 5 depicts the experimental layout, which corresponds to the model presented in Fig. 1, wherein the Line-of-Sight (LoS) channel. The distance from the transmitter to the detection target (experimenter 1) is 2 m. And experimenter 2 moving around in the vicinity serves as interference. The distance between the target and the closest receiver changes in different experimental designs.

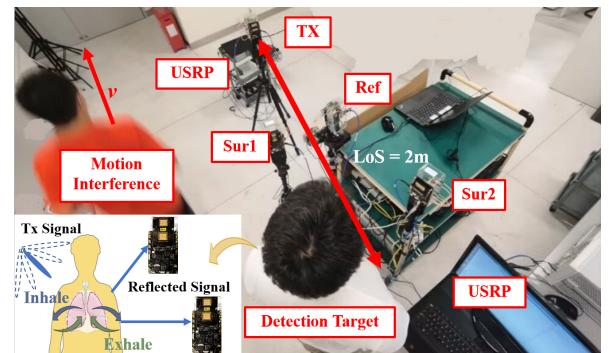


Fig. 5. Experiment Layout.

B. Dataset

The experiment involves sensing two types of human respiration: detection and counting the human respiration under interference. The interference caused by human motion is induced through random walking around the detected target in the surrounding area. Notice that there are no restrictions on the respiratory intensity of the target, i. e. both deep and calm breathing are allowed. In the experimental scenario, the detection target can breathe freely. To ensure the robustness of respiratory detection system, different volunteers (variations in height, gender, body shape and so forth) are invited as

TABLE I
DATASETS SETUP

Detection Dataset							
Positive Samples		Negative Samples		Train:Test		K-Fold Method	
200		200		8:2		5-Fold	
Counting Dataset							
2 Respirations	3Respirations	4 Respirations	5 Respirations	6 Respirations	Train:Test	K-Fold Method	
200	200	200	200	200	8:2	5-Fold	

the detection target. Furthermore, a group of volunteers is instructed to walk around the target with varying directions and speeds.

Initially, we detect the presence of human respiration amidst the interference, considering periods of 2.5, 5, 7, and 10 seconds. Subsequently, we quantify the occurrence of human respiration under the influence of interference, conducting multiple counts ranging from 2 to 6 instances. Each situation of the experiment is sampled 100 times with the passive sensing system in LoS scenarios, resulting in two datasets, illustrated in Table I.

C. CAF and CFAR Results

Fig. 6(a) and Fig. 6(e) illustrate the time-Doppler spectrograms and CFAR results for the detection of human respiration in the presence of motion interference. Conversely, Fig. 6(b) and Fig. 6(f) display the time-Doppler spectrograms and CFAR results for human respiration in the presence of static environment interference. In contrast, Fig. 6(c)(d)(g)(h) solely depict the spectrograms and CFAR results of interference signals. These spectrograms provide clear visual representations of distinct patterns that distinguish human respiration across various time periods and interference, thereby emphasizing noticeable disparities between interference signals and respiration signals.

D. Respiration Detection Results

1) *Detection:* This work investigates the binary classification detection of breathing presence in the environment across various detection time intervals. As elaborated in Fig. 7, the analysis reveals that when the detection time exceeds 7 seconds, accurate identification of potential breathing signals within the detection environment can be achieved with an accuracy of 100%. Also, the experiment verifies the robustness of the proposed approach by changing the distance between the target and the closest receiver. Note that analytical approximation is achieved by a logistic fitting curve, which can better reveal the tendency of real accuracy. The definition of the accuracy is shown in (12):

$$\text{Accuracy} = \frac{TN + TP}{\text{All Predicted Labels}}, \quad (12)$$

where TN and TP denote the number of correctly classified positive samples and negative samples, respectively. Moreover, the classification accuracy increases with respect to the length of detection time, in line with the expectations.

2) *Counting:* In the task of counting the human respiration, five experimental scenarios are performed: 2 respirations, 3 respirations, 4 respirations, 5 respirations and 6 respirations within 10 seconds. In order to achieve this counting task, we

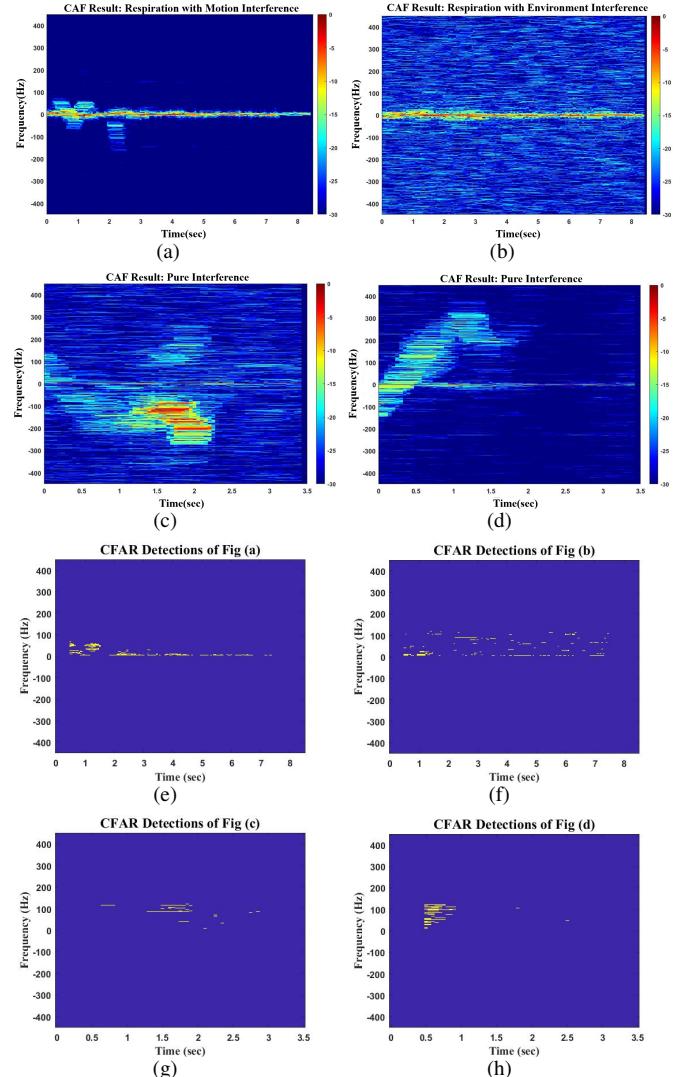


Fig. 6. Fig (a) and (e) represent the results of Cross Ambiguity Function (CAF) and Constant False Alarm Rate (CFAR) algorithms applied to respiration signals in the presence of human motion interference. Fig (b) and (f) display the CAF and CFAR results of respiration signals affected by static environment interference. On the other hand, Fig (c), (d), (g), and (h) exhibit the CAF and CFAR results obtained from environments where no respiration activity is present.

train the network with 800 samples (160 samples per type) and treat the remaining samples as the test set. The results with the closest distance is 0.3 m are illustrated in Fig. 8(a). The experimental results show that the accuracy of counting the human respiration rate is around 90%.

The comparison of other studies is shown in Table II.

TABLE II
COMPARISON OF STUDIES

Name	Working frequency	System	Interference	Accuracy	Application
Zhang et al.[9]	WiFi	2 commercial WiFi	No	-	Human Respiration
MIT CSAIL[15]	Milimeter Wave	2 rows of FMCW radar	No	96%	Human skeleton
Sun et al.[12]	WiFi	Commercial WiFi and Horn Antenna	No	-	Human Motion
Ali et al.[16]	Milimeter Wave	Simulation	No	-	Vehicle Motion
Ours	Milimeter	4 Sivers phased array	Yes	90%	Human Respiration

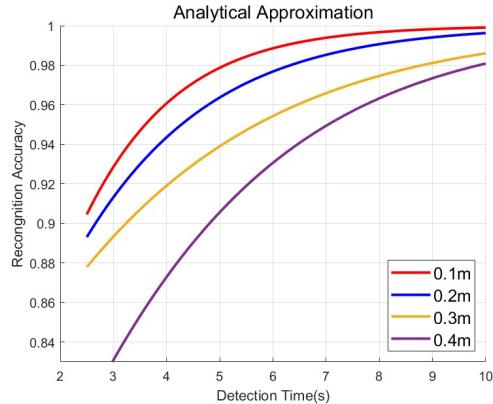


Fig. 7. Classification accuracy versus the sensing time.

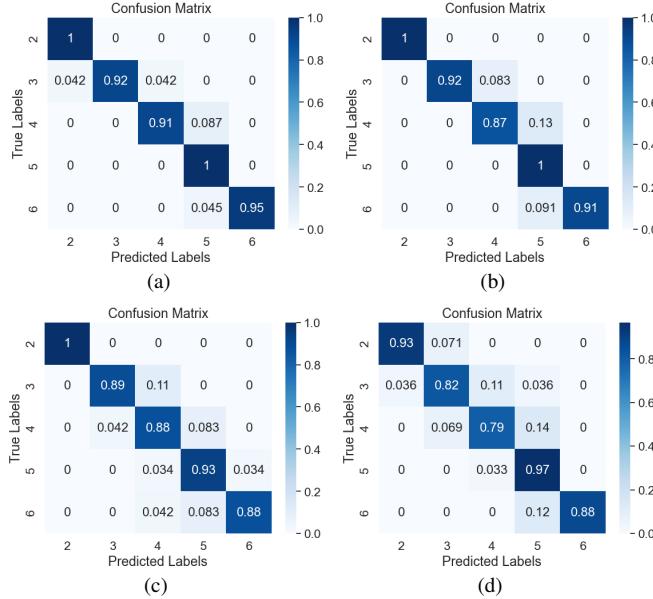


Fig. 8. Confusion matrix for counting results (Although it is a regression task, the confusion matrix is used to evaluate the accuracy of the rounded regression results). (a) Represents the results of the distance between the target and the closest receiver, which is 0.3 m. The distances in (b), (c) and (d) are 0.6 m, 0.9 m and 1.2 m respectively.

V. CONCLUSION

This paper proposes a novel respiration detection technique based on a passive mmWave communication system. The system is demonstrated by executing five distinct respiratory rates in the surveillance channel and collecting a dataset of received signals using the aforementioned system. Subsequently,

a ResNet model is trained using the collected dataset for both respiration detection and respiratory rate counting. The experimental results showcase remarkable respiration detection classification accuracy, over 95%, for detection time exceeding 10 seconds. The accuracy of counting the respiratory rate of human respiration is around 90% overall.

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REFERENCES

- [1] A. Nicolò, C. Massaroni, E. Schena, and M. Sacchetti, "The Importance of Respiratory Rate Monitoring: From Healthcare to Sport and Exercise," *Sensors*, vol. 20, no. 21, p. 6396, Nov. 2020.
- [2] J. Rumiński, "Analysis of the parameters of respiration patterns extracted from thermal image sequences," *Biocybernetics and Biomedical Engineering*, vol. 36, no. 4, pp. 731–741, Jan. 2016.
- [3] C. Chourpaliadis and A. Bhardwaj, "Physiology, Respiratory Rate," in *StatPearls*. StatPearls Publishing, 2023.
- [4] C. Massaroni, A. Nicolò, D. Lo Presti, M. Sacchetti, S. Silvestri, and E. Schena, "Contact-Based Methods for Measuring Respiratory Rate," *Sensors*, vol. 19, no. 4, p. 908, Jan. 2019.
- [5] E. Vanegas, R. Igual, and I. Plaza, "Sensing Systems for Respiration Monitoring: A Technical Systematic Review," *Sensors (Basel, Switzerland)*, vol. 20, no. 18, p. 5446, Sep. 2020.
- [6] J. Wang *et al.*, "Integrated Sensing and Communication: Enabling Techniques, Applications, Tools and Data Sets, Standardization, and Future Directions," *IEEE Internet of Things Journal*, vol. 9, no. 23, pp. 23 416–23 440, Dec. 2022.
- [7] T. Nakamura, M. Bouazizi, K. Yamamoto, and T. Ohtsuki, "Wi-Fi-Based Fall Detection Using Spectrogram Image of Channel State Information," *IEEE Internet of Things Journal*, vol. 9, no. 18, pp. 17 220–17 234, Sep. 2022.
- [8] X. Wang, C. Yang, and S. Mao, "Resilient Respiration Rate Monitoring With Realtime Bimodal CSI Data," *IEEE Sensors Journal*, vol. 20, no. 17, pp. 10 187–10 198, Sep. 2020.
- [9] Y. Zeng, D. Wu, R. Gao, T. Gu, and D. Zhang, "FullBreathe: Full Human Respiration Detection Exploiting Complementarity of CSI Phase and Amplitude of WiFi Signals," *Proceedings of the ACM on Interactive, Mobile, Wearable and Ubiquitous Technologies*, vol. 2, no. 3, pp. 148:1–148:19, Sep. 2018.
- [10] Y. Zeng *et al.*, "MultiSense: Enabling Multi-person Respiration Sensing with Commodity WiFi," *Proceedings of the ACM on Interactive, Mobile, Wearable and Ubiquitous Technologies*, vol. 4, no. 3, pp. 102:1–102:29, Sep. 2020.
- [11] C. Wang *et al.*, "Multi - Touch in the Air: Device-Free Finger Tracking and Gesture Recognition via COTS RFID," in *IEEE INFOCOM 2018 - IEEE Conference on Computer Communications*, Apr. 2018, pp. 1691–1699.
- [12] H. Sun, L. G. Chia, and S. G. Razul, "Through-Wall Human Sensing With WiFi Passive Radar," *IEEE Transactions on Aerospace and Electronic Systems*, vol. 57, no. 4, pp. 2135–2148, Aug. 2021.
- [13] M. Kronauge and H. Rohling, "Fast Two-Dimensional CFAR Procedure," *IEEE Transactions on Aerospace and Electronic Systems*, vol. 49, no. 3, pp. 1817–1823, Jul. 2013.
- [14] K. He, X. Zhang, S. Ren, and J. Sun, "Deep residual learning for image recognition," in *Proceedings of the IEEE conference on computer vision and pattern recognition*, 2016, pp. 770–778.
- [15] M. Zhao *et al.*, "Through-Wall Human Pose Estimation Using Radio Signals," in *2018 IEEE/CVF Conference on Computer Vision and Pattern Recognition*. Salt Lake City, UT: IEEE, Jun. 2018, pp. 7356–7365.
- [16] Z. Ali, A. Duel-Hallen, and H. Hallen, "Early Warning of mmWave Signal Blockage and AoA Transition Using sub-6 GHz Observations," *IEEE Communications Letters*, vol. 24, no. 1, pp. 207–211, Jan. 2020.