

Question-1

Code

```
clear; close all;
tc = 5;
s = tf('s');
Gp = tf(1.15,[50 15 1]);
Gm = tf(1,[50 15 1]);
% PID controller
Gc = tf([50 15 1],[tc 0]);
% Disturbance tf
Gd = tf(1,[5 1]);
lambdavec = 0.001:0.001:0.5;
r1 = ones(length(lambdavec),1);
r2 = r1;
for k = 1:length(lambdavec)
    % Feedforward controller
    Gff = -Gd*1/(lambdavec(k)*s+1)/Gp;
    % sys is Y/Do
    sys = (Gff*Gp+Gd)/(1+Gp*Gc);
    S = stepinfo(sys);
    % get settling time as close as possible to 15
    r2(k) = S.SettlingTime;
    r1(k) = abs(S.SettlingTime-15);
end
[val,loc] = min(r1);
lambda = lambdavec(loc);
```

Question-2

Tables

Table 12.1 IMC Controller Settings for Parallel-Form PID Controller (Chien and Fruehauf, 1990)

Case	Model	$K_c K$	τ_I	τ_D
A	$\frac{K}{\tau s + 1}$	$\frac{\tau}{\tau_c}$	τ	—
B	$\frac{K}{(\tau_1 s + 1)(\tau_2 s + 1)}$	$\frac{\tau_1 + \tau_2}{\tau_c}$	$\tau_1 + \tau_2$	$\frac{\tau_1 \tau_2}{\tau_1 + \tau_2}$

Table 12.4 Controller Design Relations Based on the ITAE Performance Index and a First-Order-plus-Time-Delay Model (Lipták, 2006)*[†]

Type of Input	Type of Controller	Mode	A	B
Disturbance	PI	P	0.859	-0.977
		I	0.674	-0.680
Disturbance	PID	P	1.357	-0.947
		I	0.842	-0.738
		D	0.381	0.995
Set point	PI	P	0.586	-0.916
		I	1.03 [†]	-0.165 [†]
Set point	PID	P	0.965	-0.85
		I	0.796 [†]	-0.1465 [†]
		D	0.308	0.929

*Design relation: $Y = A(\theta/\tau)^B$ where $Y = KK_c$ for the proportional mode, τ/τ_I for the integral mode, and τ_D/τ for the derivative mode.

[†]For set-point changes, the design relation for the integral mode is $\tau/\tau_I = A + B(\theta/\tau)$.

Code

```
clear;close all;
s = tf('s');
%% Given Data
Kv = 0.9; Kip = 0.75;
t = (0:1:11)';
T =
([12,12.5,13.4,14,14.8,15.4,16.1,16.4,16.8,16.9,17,16.9]'
-12)/2;
plot(t,T);
% Can't see any inverse response, so mostly no zero
assume first order plus
% time delay.
%% Model Estimation
[X,RESNORM,RESIDUAL,EXITFLAG] = lsqcurvefit(@resp,[5 2
1],t,T);
K = X(1)*Kv*Kip;
tau1 = X(2);
tau2 = X(3);
Gp = tf(K,conv([tau1 1],[tau2 1]));
%% Part a) IMC
tauc = max(tau1 ,tau2)/2;
Kc = (tau1 +tau2)/(K*tauc);
tauI = tau1 + tau2;
tauD = (tau1*tau2)/(tau1 + tau2);
Gc_imc = Kc*(1+1/(tauI*s)+tauD*s);
%% Part b) ITAE (setpoint)
% FOPTD approximation
D = tau2/2;
```

```

tau = tau1 + tau2/2;
% Use tables
AP = 0.965;
BP = -0.85;
Kc_b = AP*(D/tau)^BP/K;
AI = 0.796;
BI = -0.1465;
tauI_b = tau/(AI + BI*(D/tau));
AD = 0.308;
BD = 0.929;
tauD_b = AD*(D/tau)^BD*tau;
Gc_b = Kc_b*(1+1/(tauI_b*s)+tauD_b*s);
%% Part c) ITAE (disturbance)
AP = 1.357;
BP = -0.947;
Kc_c = AP*(D/tau)^BP/K;
AI = 0.842;
BI = -0.738;
tauI_c = tauI/(AI*(D/tau)^BI);
AD = 0.381;
BD = 0.995;
tauD_c = AD*(D/tau)^BD*tau;
Gc_c = Kc_c*(1+1/(tauI_c*s)+tauD_c*s);
%% Function to give step response for lsqcurvefit
function Y = resp(params,tvec)
    K = params(1);
    tau = params(2);
    tau2 = params(3);
    Gp = tf(K,conv([tau 1],[tau2 1]));
    Y = step(Gp,tvec);
end

```

Question-4

Part a) Feedforward controller

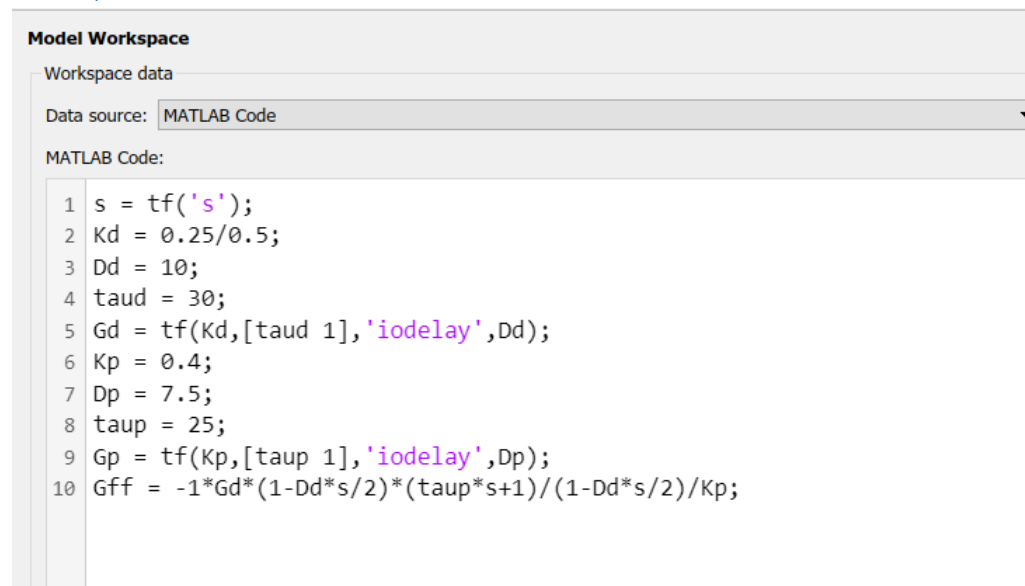


Figure 1: SIMULINK DIAGRAM of the system with just a feed-forward controller

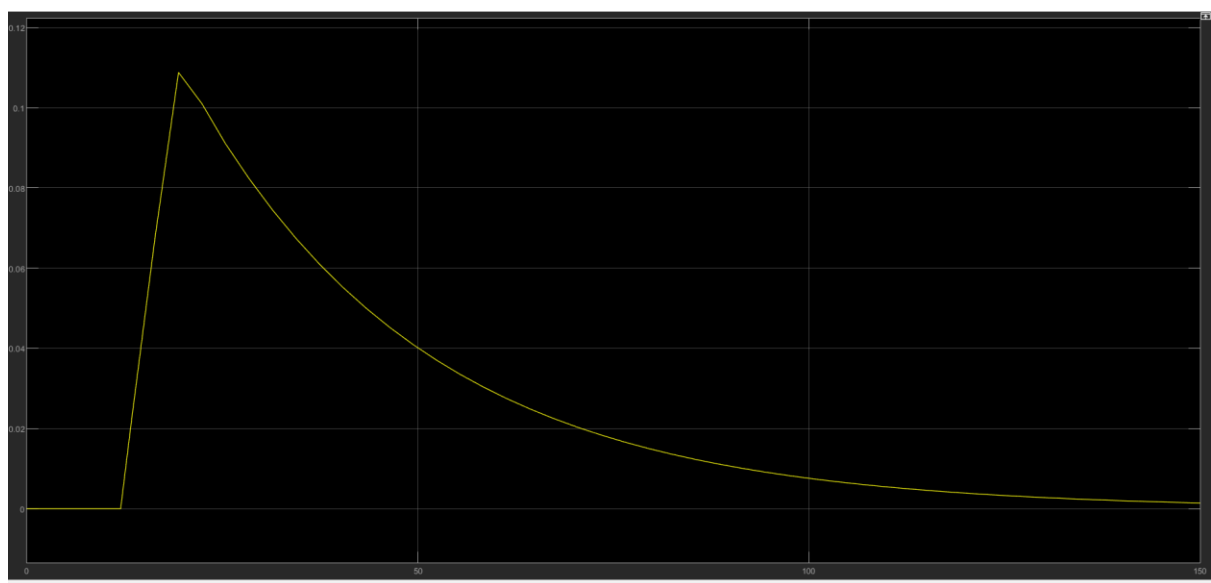


Figure 2: Disturbance rejection performance

Part b) Tuned PID Controller

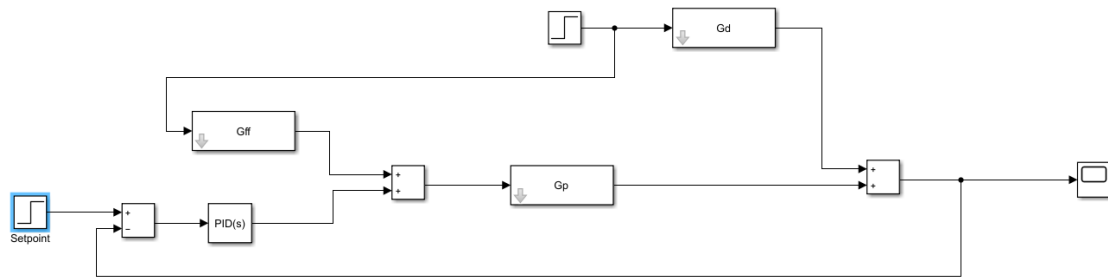


Figure 3: Feedforward in combination with a PID controller

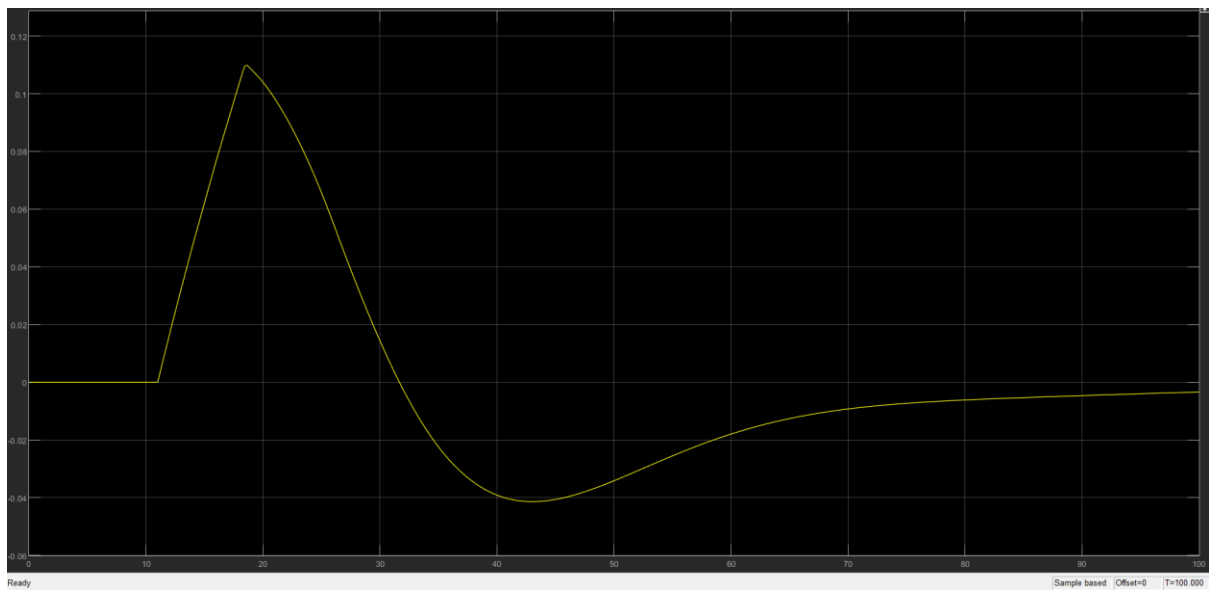


Figure 4: Response for combined efforts of feedforward and PID controller

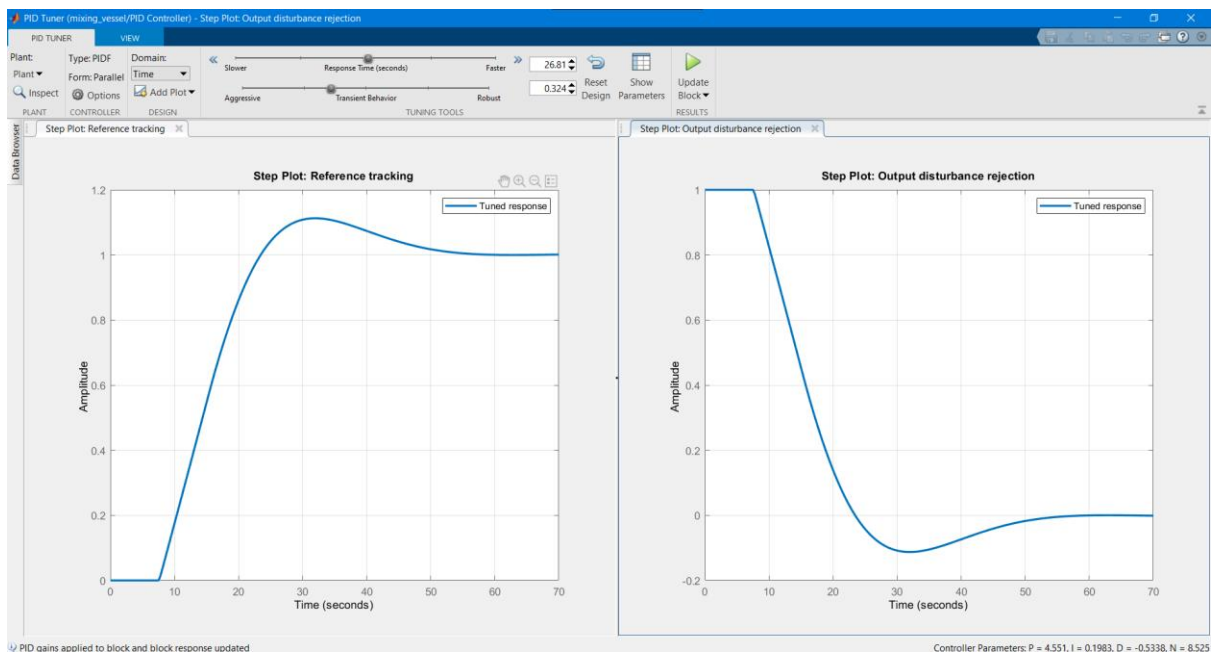


Figure 5: Tuning of the PID Controller (It linearizes the closed loop, and then we manually tune it on the basis of the responses/settling time requirements)

```
1 - model = 'mixing_vessel';
2 - load_system(model);
3 - out = sim(model);
4 - y = out.simout.data;
5 - t = out.tout;
6 - iae = trapz(t,abs(y));
```

IAE with FF controller alone was obtained to be 3.5326

IAE with FF + PID controller was obtained to be 2.3819

As expected, we see an improvement when we use a PID controller in addition.

Part c) MPC

- Firstly, I will obtain the step response models of the process and disturbance using the corresponding transfer functions we have.
- Given the time delay and time constants, I will choose a sampling interval of about 2.5 minutes.
- In this way I can obtain step response model length as about $5 \times 25 / 2.5 = 50$.
- Given this n , and multiple delays involved I would want to have larger prediction horizon, $p = 25$. Roughly half of what we have for n . (Note that obviously if we go for $p > n$, the system might exhibit instability)
- It is always safe to have the control horizon to be smaller than the predictive horizon. We can probably have $m = 10-15$. And tune as per the response we get.
- Higher m is more aggressive but we need more computational power (because we need to optimize more variables).
- Input constraints can be decided based on the expected disturbance inputs that might occur. Let's say if 0.5 is the maximum expected disturbance (this causes a change of 0.25 in output) we can constrain the valve to have absolute value of input moves within $0.25 / 0.4 = 0.625$ psig.