Experiments in Spectroscopy

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Part I Fundamentals

Part II Two Level Systems

Part III Nonlinear Spectroscopy

Part IV Plasmonics

Chapter 17

Plasmonic Nano-Rods

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Tasks

- Model the angle-dependent extinction spectrum of an array of plasmonic particles, as measured by Simon Durst (Bayreuth) and shown below. Discuss the observed phenomena.
- · Assume that the embedding medium contains a dye with a narrow Lorentzian absorption line. Model and discuss the dispersion relation.

How this is measured

Waveguides

Let us start by investigating the optical modes of a thin wire. We assume cylindrical symmetry and coordinates z and ρ . The wire has a radius a and is made of a material (first dielectric, later metal) of dielectric function ϵ_{in} . It is embedded in a dielectric matrix characterized by $\epsilon_{out}.$ We solve Maxwell's equation with these boundary conditions.¹

The electric field is described inside and outside the wire by cylindrical Bessel J and Hankel $H^{(1)}$ functions of the first kind, respectively. The lowest order mode is²

$$\begin{split} \mathbf{E}_{in} = & E_0 \left(J_0(\kappa_1 r) \, \hat{\mathbf{z}} + \frac{i k_z}{\kappa_1} J_1(\kappa_1 r) \, \hat{\mathbf{r}} \right) e^{i(k_z z - \omega t)} \\ \mathbf{E}_{out} = & E_0 \left(H_0^{(1)}(\kappa_2 r) \, \hat{\mathbf{z}} - \frac{i k_z}{\kappa_2} H_1^{(1)}(\kappa_2 r) \, \hat{\mathbf{r}} \right) e^{i(k_z z - \omega t)} \end{split}$$

where k_z is the component of the wave vector (length in vacuum $k_0 = 2\pi/\lambda$) along the wire. The components perpendicular to the wire are defined as

$$\kappa_i = k_0 \sqrt{\epsilon_i - (k_z/k_0)^2} \tag{17.1}$$

with $\epsilon_i=(\epsilon_{in},\epsilon_{out}).$ The boundary condition at r=a leads to a condition for k_z (see also H/N eq. 12.53)

$$\frac{\epsilon_{in}}{\kappa_{1}a} \frac{J_{1}(\kappa_{1}a)}{J_{0}(\kappa_{1}a)} - \frac{\epsilon_{out}}{\kappa_{2}a} \frac{H_{1}^{(1)}(\kappa_{2}a)}{H_{0}^{(1)}(\kappa_{2}a)} = 0 \tag{17.2}$$

¹ Details see Saleh /Teich chapter fiber optics

² Takahara

Guiding of waves requires that the electric field is localized near the wire, i.e., the radial component of the wave vector outside the wire κ_2 has to be imaginary, or $k_z/k_0 > \sqrt{\epsilon_{out}}$. At the same time, the wave should propagate inside the wire, i.e., k_z should be (almost³) real, or $\sqrt{\epsilon_{in}} > k_z/k_0$.

For dielectric waveguides, the core has to have a higher index of refraction than the embedding medium. When the radius of the wire is decrease, the decay of the field outside the wire becomes slower and slower. By Fourier transformation from κ to ρ one finds a characteristic lower limit for radius R of the field distribution, independent of the wire diameter,

$$R > \frac{\lambda}{2\sqrt{\epsilon_{in}}} \tag{17.3}$$

Dielectric waveguides are this limited in their size of the mode field to approximately the wavelength in the core medium.

Plasmonic waveguides in contrast can become very small. When the dielectric function ϵ_{in} is negative, the mode field remains bound close to the wire even for small a wire radius. The downside is that losses increase. The wave vector in propagation direction becomes complex. The real part describes the effective index of refraction of the mode, the imaginary part the losses due to absorption in metal. These losses increase drastically when the wire becomes smaller, as shown in Fig XXX

The large component of the wave vector in propagation direction k_z in a plasmonic waveguide corresponds to a short effective wavelength $\lambda_{in}=2\pi/k_z.$ The thinner the waveguide, the shorter the effective wavelength. Novotny shows that a linear relation to the vacuum wavelength exists XXX cite

$$\lambda_{in} = n_1 + n_2 \frac{\lambda}{\lambda_p} \tag{17.4}$$

where $\lambda_p=2\pi c/\omega_p$ is the plasma wavelength of the Drude metal and the n_i are constants depending on geometry and refractive indices. ⁴

Side note: Leakage radiation

The component of the wave vector in propagation direction k_z is larger than the maximum possible length of a wave vector in the embedding medium $k_0\sqrt{\epsilon_{out}}$. This means that momentum conservation does not allow photons to leave the waveguide. Plane waves with such a value of k_z are evanescent (have an imaginary k_\perp) in the embedding medium. In this way, a mode can propagate without loses to the environment. However, this also hinders observation of such a propagation. One only could detect the emission at the end of the waveguide or at defects.

One way around is leakage radiation. When the waveguide is placed on or near a medium with a higher index of refraction $k_z < k_0 \sqrt{\epsilon_{substrate}}$, then this substrate supports suitable free-space modes. The distance between waveguide and substrate defines the coupling (as seen by the observer) or the losses (from the point of view of the waveguide). In such an experiment one can see bright emission along the whole waveguide.

 $^{
m 3}$ When the dielectric functions are complex-valued, also k_z becomes complex-valued

Figure 17.1: Al wire alpha and beta similar to Fig. 12.20 in H/N

⁴ see eq. 14 in XXX

Fabry-Perot modes in nano-rods

We now cut out a piece of a thin plasmonic waveguide and call this object a nano-rod. The propagation of the plasmon mode along the waveguide is the same as in the preceding section, but now the wave is reflected back at the ends of the waveguide. The free space around the nano-rod does not support modes of sufficient high wave vector, so that the light can not just propagate out. The two ends thus form two mirrors of a Fabry-Perot cavity and the short piece of waveguide in between is similar to the medium in the cavity. We expect to find periodic resonances when varying the length of the rod. This is indeed what is observed. However, the apparent length of the rod is larger by some offset length L_o

$$L_{res} = n \frac{\lambda_{in}}{2} + L_o \tag{17.5}$$

The exact value of the offset L_o depends on details of the waveguide end, for example how this is rounded or cut flat, and also from where to where on measures the rod length L_{res} .

The physical origin of this apparent additional length L_o is that a propagating plasmon is a quasi particle combined of free electrons and an electromagnetic field. The electrons have to stay inside the metal to within less than a lattice constant. The optical field however extends all around the nanorod and thus also extends over the ends of the rod. This gives an additional length.

In some publications this additional length L_o is called a reflection phase $\Delta \phi$, as one could also describe the process in terms of phases as

$$\frac{2L_{res}}{\lambda_{in}} 2\pi = n 2\pi + 2\Delta\phi \tag{17.6}$$

This is in analogy to the phase acquired by an optical beam undergoing total internal reflection. In this case, $k_{z,1}$ is real-valued, but $k_{z,2}$ is complex, as it describes an evanescent wave. The reflection coefficient

$$r_{12}^s = \frac{k_{z,1} - k_{z,2}}{k_{z,1} + k_{z,2}} \tag{17.7}$$

becomes then also complex, so that the field acquires a phase shift

$$\Delta \phi = \arg(r_{12}^s) \tag{17.8}$$

Excitation of resonances in nano-rods

We shine a plane wave on a nano-rod, i.e., a short piece of a thing plasmonic wire, and want to calculate the response of the particle in terms of scattered field and absorption. We follow here the analytical model by Jens Dorfmüller et al. XXX cite, model the effect of the wire itself and assume an ad-hoc parameter for the effect of the wire ends. At the wire surface, the boundary condition for electric fields relates the z components if incident, scattered and internal fields

$$E_z^{inc}(r=a) + E_z^{scat}(r=a) = E_z^{inside}(r=a) \quad . \eqno(17.9)$$

The plane wave is incident with an angle θ to the symmetry axis of the wire

$$E_z^{inc}(r=a) = E_0 \sin(\theta) e^{i k_{\parallel} z}$$
 with $\cos(\theta) = \frac{k_{\parallel}}{k_0}$ (17.10)

where we omitted the time dependence $e^{-i\omega t}$ and have out the phase factor of $e^{ik_r r}$ into E_0 . k_\parallel is the component of the plane wave's wave vector along the wire direction.

The scattered field is described by the current inside the wire. An oscillating current is source of radiation, as for example in an antenna. We assume that the current is spatially homogeneous and running only along z direction (not radially). One finds

$$E_z^{scat}(r=a) = \frac{i\,\tilde{Z}}{4\pi\epsilon_0\omega}\,\left(\frac{\partial^2}{\partial z^2} + k^2\right)I_z(z) \tag{17.11}$$

where \tilde{Z} is the (antenna) impedance of the wire, which is unknown at this point.

Current and field inside the wire are connected by the conductivity σ or the imaginary part of the dielectric function of the wire ϵ_{in}

$$I_z(z) = \sigma E_z^{inside}(z) = \epsilon_0 \omega \Im(\epsilon_{in}) E_z^{inside}(z)$$
 (17.12)

Putting everything in eq XXX we get a differential equation for $I_z(z)$

$$E_0 \sin(\theta) e^{i k_z z} + \frac{i \tilde{Z}}{4 \pi \epsilon_0 \omega} \left(\frac{\partial^2}{\partial z^2} + k^2 \right) I_z(z) = \frac{1}{\epsilon_0 \omega \Im(\epsilon_{in})} I_z(z) \quad \text{(17.13)}$$

We make the ansatz that the current either flows with the wave vector k_{\parallel} of the external field projected on the wire direction, or with the wave vector k_p of the plasmon mode calculated by eq XXX above (where it was called k_z):

$$I_z(z) = I_{\parallel} e^{ik_{\parallel}z} + I_{+p} e^{ik_pz} + I_{-p} e^{-ik_pz} \tag{17.14} \label{eq:17.14}$$

Additionally we know that the current has t vanish at the end of the wire, as the electrons can not flow out, i.e.,

$$I_z\left(z=\pm\tilde{L}/2\right)=0\tag{17.15}$$

where \tilde{L} is the effective length of the nano-rod, i.e., longer than the geometrical size to include the phase effect of the reflection at the wire end. All together one then finds XXX cite dirfmüller

$$\tilde{Z} = +i \frac{4\pi}{\Im(\epsilon_{in})} \frac{1}{k_p^2 - k^2}$$
 (17.16)

$$I_{\parallel}=-i\,rac{\omega\epsilon_0\, ilde{Z}}{4\pi}\,rac{1}{k_p^2-k_{\parallel}^2}\,E_0\,\sin heta$$
 (17.17)

$$I_{\pm p} = -I_{\parallel} \frac{\sin([k_p \pm k_{\parallel}] \tilde{L}/2)}{\sin(k_p \tilde{L})}$$
 (17.18)

The antenna impedance \tilde{Z} of the wire (eq. XXX) has a resonance that goes with the momentum mismatch between free-space modes and plasmonic wire modes. For the excitation of the current in the wire (Eq. xxx) only the momentum mismatch between the plasmon mode and the wave vector projected on the wire axis k_{\parallel} matters. As function of nano-rod length we find back the Fabry-Perot resonances at

$$k_p \, \tilde{L} = n \, \pi \tag{17.19}$$

where the index n enumerates the resonances.

Dipole model of the modes in a nano-rod

Before we continue with the work of Durfmüller, let us have a look at these resonances. We simplify the discussion by assuming that the incident plane wave propagates perpendicular to the wire axis, i.e. $\theta\,=\,90^\circ$ and $k_\parallel\,=\,0.$ The amplitudes of the plasmon currents $I_{\pm p}$ become identical and come into resonance for

$$k_p \tilde{L} = n\pi \quad \text{for odd } n$$

By perpendicular incidence we can only excite odd resonances of the rod! In this case, the current distribution in the wire also simplify, as the term with I_{\parallel} is spatially constant, as $k_{\parallel}=0.$ We thus have

$$I_z(z) = I_{\parallel} + 2I_{\pm p}\cos(k_p z) \approx 2I_{\pm p}\cos(k_p z)$$
 (17.21)

where the approximation certainly hold on resonance. We thus have current maxima at positions z along the nano-rod where $|\cos(k_p\,z)|$ is maximal, or at

$$z_{
u}=\pm\,
u\,rac{ ilde{L}}{n}$$
 where $u=0,1,\ldots,rac{n-1}{2}$

As the current oscillates with $e^{i\omega t}$, we can replace $I_z(z)$ by n dipoles placed at the positions z_{ν} , when neighboring dipoles oscillate 180° out of phase.

This dipole model then also explains why we can not excite even Fabry-Perot orders. Here one half of the dipoles would require a field pointing into +z direction, the other one a field pointing in -z direction, which can not be fulfilled simultaneously. For odd-order modes, one of the two directions always has one dipole more that the other direction, so that some net excitation efficiency remains. But one also sees that the excitation efficiency drops with increasing order, as most dipoles cancel with each other. We will find this back in the data by Dorfmüller et al.

Field distributions

Now that we know the current in the center of the nano-rod, we can go back to fields. We get the z-compinent of the field in the center of the wire from Eq. XXX. At this position the radial component has to vanish due to symmetry. As we have seen in the beginning of this chapter, the field is described by Bessel functions inside the wire. For small values of $\kappa_1 r$ we can approximate

$$J_0(z) \approx 1 - \frac{z^2}{4} + \cdots$$
 and $J_1(z) \approx \frac{z}{2} - \frac{z^3}{4} + \cdots$ (17.23)

so that the z component is roughly constant inside the wire and the radial component increases linearly with r. Both components are related by $\nabla \cdot \mathbf{E} =$ 0. Just inside the surface of the wire (r = a) this leads to

$$E_r(r=a) \approx -\frac{a}{2} \frac{\partial}{\partial z} E_z$$
 (17.24)

Together with eq. XXX we thus now have the full vectorial field inside the nano-rod. The partial derivate is at the end applied to eq. XXX and just produces pre-factors of ik_{\parallel} and $\pm ik_{p}$.

The fields just outside the nano-rod are obtained from the continuity conditions

$$E_{\parallel}^{in}=E_{\parallel}^{out}$$
 and $\epsilon_{in}E_{\perp}^{in}=\epsilon_{out}E_{\perp}^{out}$ (17.25)

Dorfmüller et al. (XXX) measure the r component of the electric field a few nanometers outside the nano-rod by near-field microscopy. The field above the wire is well described by the field just outside the wire. In the experiment, one also finds field amplitude above and outside the wire ends. This is not included here. It would need near-field propagation of the fields on the wire surface, which is beyond the scope of this chapter.

To compare more systematically the model with the experiment, Jens Dorfmüller et al. vary the rod length and measure near-fields. They compare then the maximum near-field amplitude under a certain illumination angle with the model. To this end, the exact calculation of the spatial dependence of the field can be skipped. If the wire is just long enough, at some point all three contribution in Eq. will add up, so that the maximum field is approximately

$$|E_r|_{max} \approx |k_{\parallel} I_{\parallel}| + |k_{+p} I_{+p}| + |k_{-p} I_{-p}|$$
 (17.26)

where the k_i are a consequence of the partial derivative in eq XX. This simplified model describes the dependence of the Farbry-Perot resonances on the effective rod-length \tilde{L} .

Far-field emission pattern

Far-field emission pattern according to Dorfmüller, thesis, appendix, eq. B.16

$$\frac{dP}{d\Omega} = \frac{I_0^2}{8\pi^2 c\epsilon_0} \frac{k^2}{k_p^2} \sin^2 \theta \, \frac{f^2 \left(n \, \pi k_{\parallel} / (2k_p)\right)}{\left(1 - k_{\parallel}^2 / k_p^2\right)^2} \tag{17.27}$$

where $k_{\parallel}=k\,\cos{ heta}$ and

$$f(x) = \begin{cases} \sin(x) & \text{for } n \text{ even} \\ \cos(x) & \text{for } n \text{ odd} \end{cases}$$

with the order n of the Fabry-Perot resonance.

Near-field excitation of modes

Nonlinear effects with modes

