

# Communication Systems

## EE-351

Lectures 22 to 23

# Frequency Demodulation:

- Two devices for frequency demodulation:
  - Frequency discriminator
    - Relies on **slope detection** followed by envelope detection
  - Phase-locked loop
    - Performs frequency demodulation in a somewhat indirect manner

# Frequency discriminator:

- It is a demodulator that consists of a differentiator followed by an envelope detector.

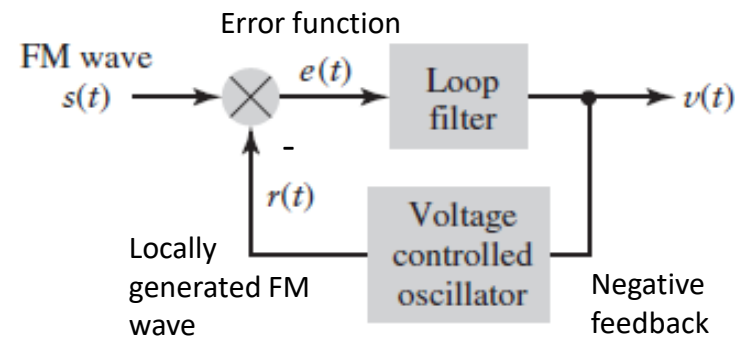


# Phase-Locked Loop (PLL):

- The *phase-locked loop* is a feedback system whose operation is closely linked to frequency modulation.
- Purpose is to extract message signal.
- Applications: It is commonly used for
  - carrier synchronization, and
  - indirect frequency demodulation

# Phase-Locked Loop (PLL):

- Basically, the phase-locked loop consists of three major components:
  - **Voltage-controlled oscillator (VCO)**, which performs frequency modulation on its own control signal.
  - **Multiplier**, which multiplies an incoming FM wave by the output of the voltage-controlled oscillator.
  - **Loop filter** of a low-pass kind, the function of which is to remove the high-frequency components contained in the multiplier's output signal and thereby shape the overall frequency response of the system.



**FIGURE 4.14** Block diagram of the phase-locked loop.

# Phase-Locked Loop (PLL):

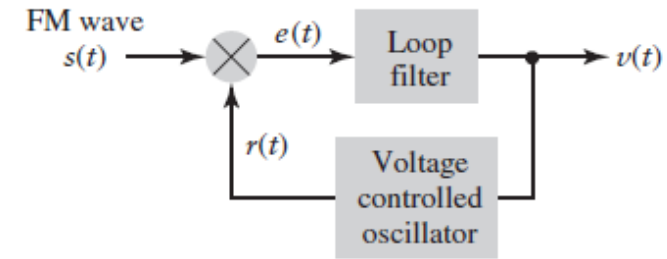


FIGURE 4.14 Block diagram of the phase-locked loop.

Purpose is to adjust:

$$\varphi_2(t) = \varphi_1(t)$$

Hence, setting stage for frequency demodulation:

If phase error,  $\varphi_e(t) = 0$ ,

PLL is said to be in **phase-lock**.

- To demonstrate the **operation of the phase-locked loop** as a frequency demodulator, we assume that the VCO has been adjusted so that when the control signal (i.e., input) is zero, two conditions are satisfied:
  1. The frequency of the **VCO is set precisely at the unmodulated carrier frequency** of the incoming FM wave.
  2. The VCO output has a **90-degree phase-shift** with respect to the unmodulated carrier wave.

These three components are connected together to form a **closed-loop feedback system**.

# Phase-Locked Loop (PLL):

- Free-running frequency range ———→ (Signal cannot be demodulated)
- Capture range ———→ (Near-phase-lock when phase error  $\varphi_e(t)$  is small)
- Lock-in range (ideal case)

# Phase-Locked Loop (PLL):

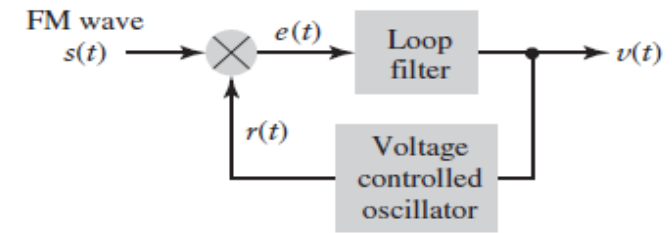


FIGURE 4.14 Block diagram of the phase-locked loop.

Capture range: (a range where difference reduces)

$$f_r < f_c$$

Let,

$$r(t) \rightarrow f_r$$

$$s(t) \rightarrow f_c$$

$$f_c - f_r \text{ reduces}$$

$v(t)$  allows VCO to produce that frequency  $f_r$  which reduces this difference ( $f_c - f_r$ ).

After sometime,  $f_r$  converts into  $f_c$  (feedback system)

$$f_c - f_r = 0$$

$$v(t) = 0$$



# Phase-Locked Loop (PLL):

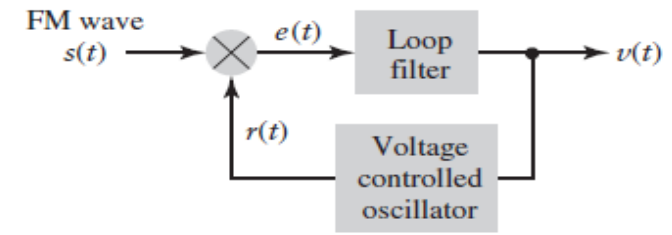


FIGURE 4.14 Block diagram of the phase-locked loop.

$$s_{FM}(t) = A_c \cos \left( 2\pi f_c t + 2\pi k_f \int_0^t m(t) dt \right)$$

Let,

$$\begin{aligned} \varphi_1(t) &= 2\pi k_f \int_0^t m(t) dt \\ s_{FM}(t) &= A_c \cos(2\pi f_c t + \varphi_1(t)) \\ r(t) &= A_v \sin(2\pi f_c t) \end{aligned}$$

@ Perfect lock-in,  $v(t) = 0$

when  $v(t) \neq 0$ , capture range

$$r(t) = A_v \sin(2\pi f_c t) + 2\pi k_v \int_0^t v(t) dt$$

# Phase-Locked Loop (PLL):

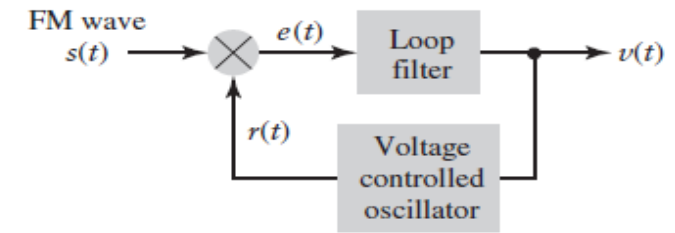


FIGURE 4.14 Block diagram of the phase-locked loop.

$$r(t) = A_v \sin(2\pi f_c t) + 2\pi k_v \int_0^t v(t) dt$$

$k_v$  is freq. sensitivity factor of VCO,  $A_v$  is the amplitude

$$\varphi_2(t) = 2\pi k_v \int_0^t v(t) dt$$

So,

$$\begin{aligned} e(t) &= s(t) \times [-r(t)] \\ &= A_c \cos(2\pi f_c t + \varphi_1(t)) \times (-A_v \sin(2\pi f_c t) + \varphi_2(t)) \\ &\quad 2\cos A \sin B = \sin(A + B) - \sin(A - B) \\ e(t) &= \frac{-A_c A_v}{2} [\sin(4\pi f_c t + \varphi_1(t) + \varphi_2(t)) - \sin(\varphi_1(t) - \varphi_2(t))] \end{aligned}$$

# Phase-Locked Loop (PLL):

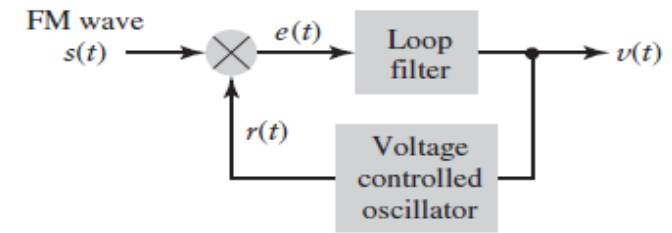


FIGURE 4.14 Block diagram of the phase-locked loop.

$$\begin{aligned} e(t) &= \frac{-A_c A_v}{2} [\sin(4\pi f_c t + \varphi_1(t) + \varphi_2(t)) - \sin(\varphi_1(t) - \varphi_2(t))] \\ &= \frac{A_c A_v}{2} \sin(\varphi_1(t) - \varphi_2(t)) \\ e(t) &= \frac{A_c A_v}{2} \sin \varphi_e(t) \end{aligned}$$

where,

$$\begin{aligned} \varphi_e(t) &= \varphi_1(t) - \varphi_2(t) \\ \varphi_e(t) &\ll 1 \\ \therefore \sin \varphi_e(t) &\approx \varphi_e(t) \end{aligned}$$

# Phase-Locked Loop (PLL):

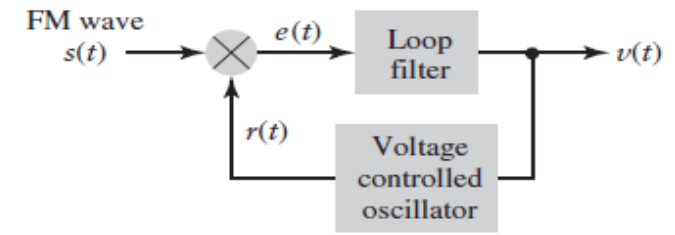


FIGURE 4.14 Block diagram of the phase-locked loop.

$$e(t) = \frac{A_c A_v}{2} \varphi_e(t)$$

Lock-in range > capture range

$$v(t) = e(t) * H(t)$$

$$v(\omega) = e(\omega) \times H(\omega)$$

$$v(t) = \frac{A_c A_v}{2} \varphi_e(t) * H(t)$$

$$v(\omega) = \frac{A_c A_v}{2} \varphi_e(\omega) \times H(\omega)$$

# Phase-Locked Loop (PLL):

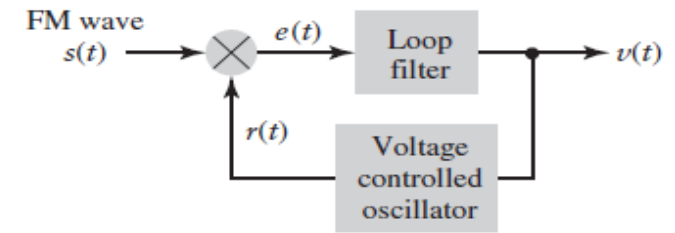


FIGURE 4.14 Block diagram of the phase-locked loop.

$$\begin{aligned}
 \varphi_e(t) &= \varphi_1(t) - \varphi_2(t) \\
 &= k_f \int_0^t m(t) dt - k_v \int_0^t v(t) dt \\
 \varphi_e(t) &= \varphi_1(t) - k_v \int_0^t \frac{A_c A_v}{2} \varphi_e(t) * H(t) dt \\
 \frac{d}{dt} \varphi_e(t) &= \frac{d}{dt} \varphi_1(t) - k_v \frac{A_c A_v}{2} \frac{d}{dt} \int_0^t \varphi_e(t) * H(t) dt \\
 &= \frac{d}{dt} \varphi_1(t) - k_v \frac{A_c A_v}{2} \varphi_e(t) * H(t)
 \end{aligned}$$

# Phase-Locked Loop (PLL):

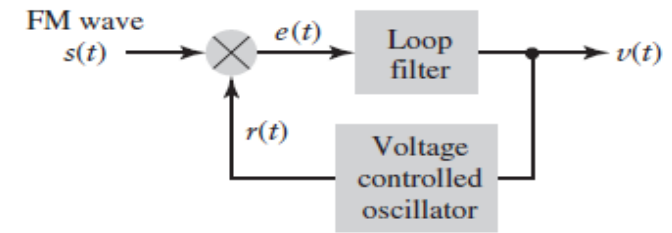


FIGURE 4.14 Block diagram of the phase-locked loop.

$$\begin{aligned}\frac{d}{dt}\varphi_e(t) &= \frac{d}{dt}\varphi_1(t) - k_v \frac{A_c A_v}{2} \varphi_e(t) * H(t) \\ j\omega\varphi_e(\omega) &= j\omega\varphi_1(\omega) - k_v \frac{A_c A_v}{2} \varphi_e(\omega) \times H(\omega) \\ j\omega\varphi_e(\omega) + k_v \frac{A_c A_v}{2} \varphi_e(\omega) \times H(\omega) &= j\omega\varphi_1(\omega) \\ \varphi_e(\omega) \left[ j\omega + k_v \frac{A_c A_v}{2} H(\omega) \right] &= j\omega\varphi_1(\omega) \\ \varphi_e(\omega) &= \frac{j\omega\varphi_1(\omega)}{j\omega \left[ 1 + \frac{k_v A_c A_v}{2j\omega} H(\omega) \right]}\end{aligned}$$

# Phase-Locked Loop (PLL):

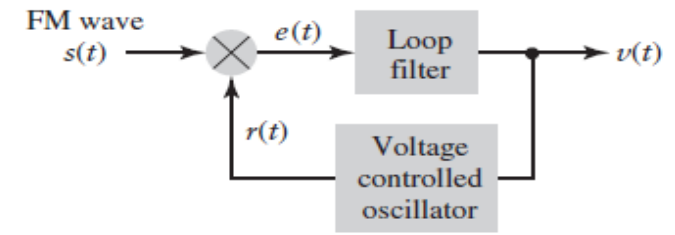


FIGURE 4.14 Block diagram of the phase-locked loop.

$$\varphi_e(\omega) = \frac{\varphi_1(\omega)}{1 + \frac{k_v A_c A_v}{2j\omega} H(\omega)}$$

$$H(\omega) \uparrow \Rightarrow \varphi_e(\omega) \downarrow$$

$\Downarrow$

$$\varphi_1(t) \approx \varphi_2(t)$$
$$H(\omega) = \infty \Rightarrow \varphi_e(\omega) = 0$$

$\Downarrow$

$$\varphi_1(t) = \varphi_2(t)$$

# Phase-Locked Loop (PLL):

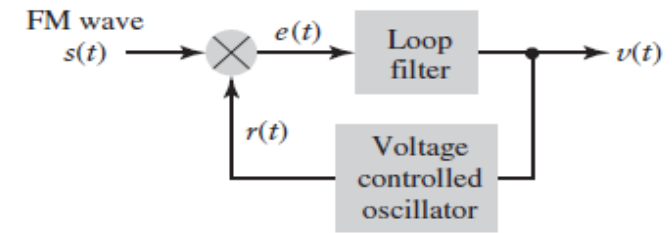


FIGURE 4.14 Block diagram of the phase-locked loop.

$$\begin{aligned}
 \varphi_e(\omega) &= \frac{\varphi_1(\omega)}{\frac{k_v A_c A_v}{2j\omega} H(\omega)} \\
 v(\omega) &= e(\omega) \times H(\omega) \\
 &= \frac{A_c A_v}{2} \varphi_e(\omega) H(\omega) \\
 &= \frac{A_c A_v}{2} \frac{\varphi_1(\omega)}{\frac{k_v A_c A_v}{2j\omega} H(\omega)} H(\omega) \\
 v(\omega) &= \frac{j\omega \varphi_1(\omega)}{k_v}
 \end{aligned}$$



# Phase-Locked Loop (PLL):

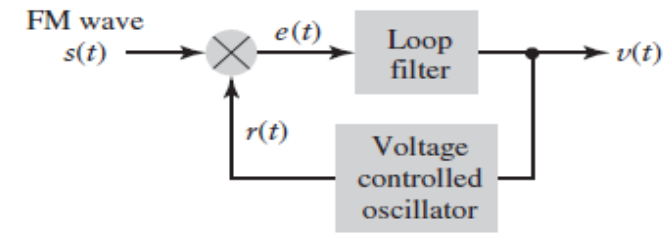


FIGURE 4.14 Block diagram of the phase-locked loop.

- Taking IFT of  $v(\omega) = \frac{j\omega\varphi_1(\omega)}{k_v}$

$$v(t) = \frac{1}{k_v} \frac{d}{dt} \varphi_1(t)$$

$$\frac{d}{dt} \varphi_1(t) = \frac{d}{dt} k_f \int_0^t m(t) dt = k_f m(t)$$

$$v(t) = \frac{k_f}{k_v} m(t)$$

$$v(t) \propto m(t)$$

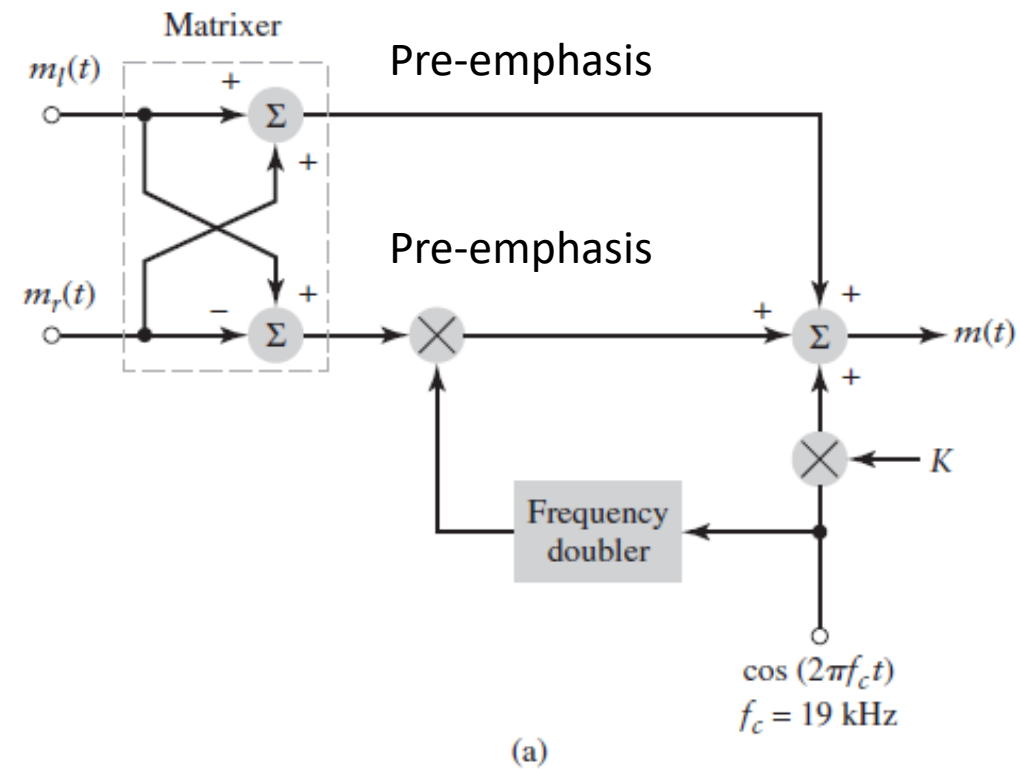
# Phase-Locked Loop (PLL):

- PLL is advanced in comparison to all receivers.
- Does not depend on change in amplitude
  - Noise does not affect this system
- Highly linear
  - so distortion-less

# Stereophonic FM broadcasting: FM Stereo Multiplexing:

- Spatial and temporal distribution both
  - Left and right ear listening a different sound with the help of stereophonic FM broadcasting
- Range of FM signal: 88 MHz to 108 MHz
- $\Delta f = 75\text{kHz}$  (for 1 signal)
  - For both USB and LSB, 150kHz
- Distance between successive carriers  $\approx 200\text{kHz}$  (+ guard band)
- Intermediate freq. 10.7 MHz

# FM Stereo Multiplexing:



# FM Stereo Multiplexing:

