

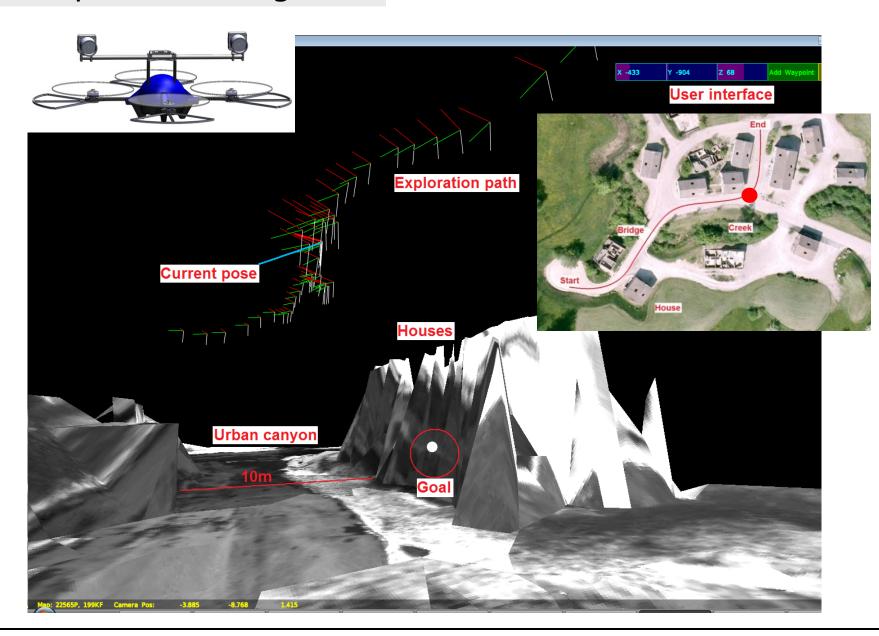
Autonomous Micro Aerial Vehicles: Design, Perception, and Control

Robot Sensors

Sensor Fusion



Perception & Navigation



Robot Sensors: The Need of a Pose Estimate

Efficiency

Limited calculation power on MAVs

Every 10 grams need 1W to lift

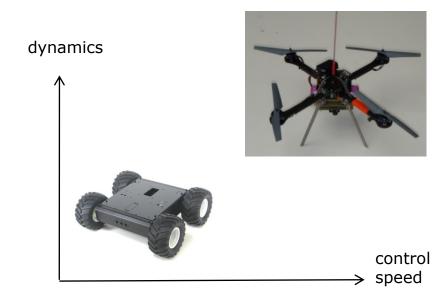
Sensors with high information density



Real-Time State Estimation

There is no "stop" position like on ground vehicles

Agile and unstable platforms such as quadcopters need high-speed control

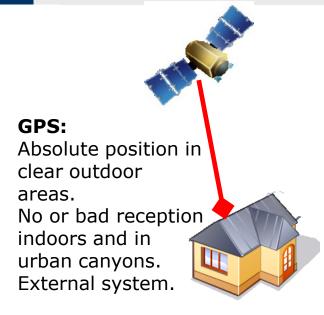


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Sensors and Their Usage





Magnetic field sensors:

2 Absolute angles in open areas.

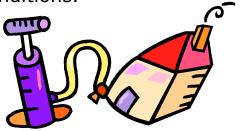
Disturbances from (electric) objects.

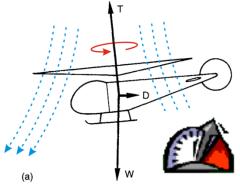
Cameras:

Vast information (angular and intensities/color).
Salient patterns must be visible.
Susceptive to light changes.

Barometric / Pressure sensors:

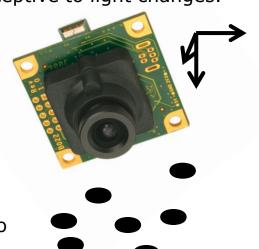
Absolute altitude measurement. Sudden pressure changes in indoor environments. Changing weather conditions.





Airspeed sensors:

Reliable for fix wing UAVs Application critical. Difficult to use on helicopters.



"The Global Positioning System (GPS) is a space-based global navigation satellite system that provides reliable location and time information in all weather and at all times and anywhere on or near the Earth when and where there is an unobstructed line of sight to four or more GPS satellites." [Wikipedia]

Full Operational Capability in April 1995 (24 Satellites) Since March 2008: 31 actively broadcasting satellites (nonuniform arrangement).

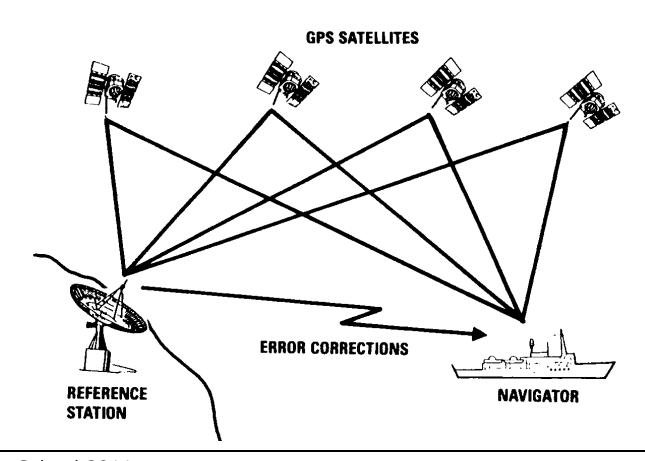
Carrier frequency is about 1.5GHz, clouds are +/- transparent

Accuracy: Difficult to assess. Different factors play a role:

- 1. Signal quality (atmosphere conditions, # of visible satellites)
- 2. Support method: for ex. differential GPS (DGPS)
- 3. Observation time

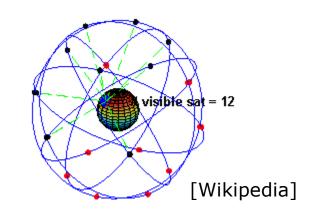
DGPS: differential GPS:

Ground station at known location sends error corrections to the mobile user

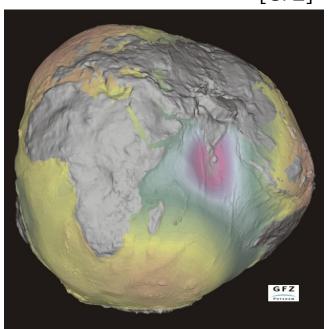


GPS Segments:

- Space Segment:
 - 31 orbiting satellites with atomic clocks
 - About 8 always visible
- Control Segment
 - Master control for satellites
 - Update satellite position (Kalman Filter) and atomic clocks
- User Segment
 - General GPS receiver with antenna and precise clock (crystal, less precise than atomic clock!)
 - Geoid model for height estimation



[GFZ]



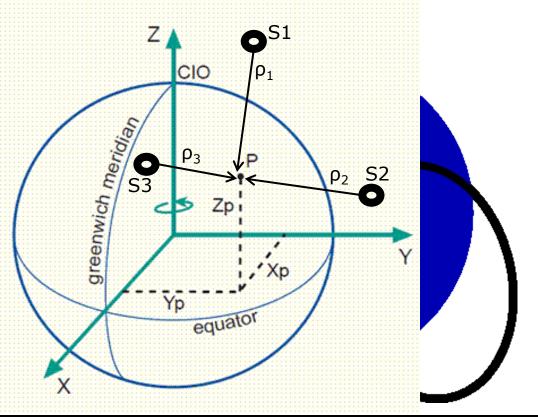
Navigation Equations: Intuition

Basic Principle:

Distance to 3 satellites is known as ρ_i .

Intersect 3 spheres with radii ρ_i and satellite position as center

→ up to two solutions S₁



Navigation Equations: Mathematics 1

Distance to satellite $\rho_i = v t = c \cdot (t_r - t_i)$

c: speed of light

t_r: time signal received

t_i: time signal sent (satellite i)

Navigation equation:

$$\rho_i = \sqrt{(x_i - x)^2 + (y_i - y)^2 + (z_i - z)^2} + c \cdot \Delta t = c \cdot (t_r - t_i)$$

 (x_i, y_i, z_i) : position of satellite i

(x,y,z): receiver position

∆t: receiver time offset to GPS time

→ 4 unknowns require 4 equations (i.e. satellites)

Navigation Equations: Mathematics 2

Solution of the Navigation Equation System (with $b = c \cdot \Delta t$)

$$\rho_{1} = \sqrt{(x_{1} - x)^{2} + (y_{1} - y)^{2} + (z_{1} - z)^{2}} + b = c \cdot (t_{r1} - t_{1})$$

$$\rho_{2} = \sqrt{(x_{2} - x)^{2} + (y_{2} - y)^{2} + (z_{2} - z)^{2}} + b = c \cdot (t_{r2} - t_{2})$$

$$\rho_{3} = \sqrt{(x_{3} - x)^{2} + (y_{3} - y)^{2} + (z_{3} - z)^{2}} + b = c \cdot (t_{r3} - t_{3})$$

$$\rho_{4} = \sqrt{(x_{4} - x)^{2} + (y_{4} - y)^{2} + (z_{4} - z)^{2}} + b = c \cdot (t_{r4} - t_{4})$$

Least squares solution using Bancroft's method

Error Sources

Influence of receiver time offset $b = c \cdot \Delta t$

c = 299792458 m/s = 0.299792458 m/nsec

→ 1ns offset reflects in 30cm distance error (1ms in 300km!)

Atmospheric Distortions and Satellite Errors

Satellite signal is delayed as a function of the atmosphere state Satellite information is inaccurate (position, time sync)

→ Phase measurements on different receivers eliminate errors

Sensor Model

Recall:

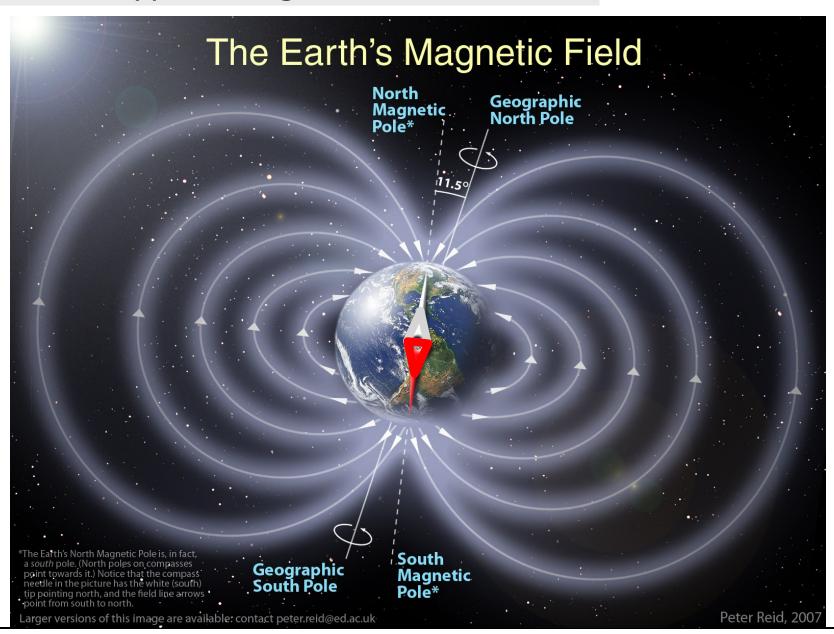
$$\rho_i = \sqrt{(x_i - x)^2 + (y_i - y)^2 + (z_i - z)^2} + c \cdot \Delta t = c \cdot (t_i - t_i)$$

Clock Noise: $t_r^* = t_r + b_{tr} + n_{tr}$, b_{tr} , b_{ti} as random walk, n_{tr} , n_{ti} as white noise $t_i^* = t_i + b_{ti} + n_{ti}$

- Satellite Position Noise: (x_i, y_i, z_i) is a result of the GPS satellite update. Precise model requires error propagation from this algorithm!
- → GPS noise model requires noise propagation from all measurements!

Simplified model for GPS position measurement: $\mathbf{p}^* = \mathbf{p} + \mathbf{b_p} + \mathbf{n_p}$, $\mathbf{b_p}$ as a bounded random walk, $\mathbf{n_p}$ as zmwgn

Sensor Types: Magnetic Field Sensor





Sensor Types: Magnetic Field Sensor

Earth's magnetic field in practice

International Geomagnetic Reference Field

Highly complex model with several coefficients, download as text files.

Models the earth's magnetic field at any location and time between 1900 and 2015.

IGRF models are standardized for a particular year, reflecting the most accurate measurements available at that time, and indicating a small-scale, slow time variation of the Earth's overall magnetic field.

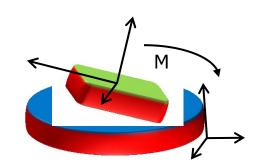
Currently at the 11th generation: IGRF-11

Sensor Types: Magnetic Field Sensor

Sensor Model

Define noise model of measurement vector $\mathbf{m}^* = (\mathbf{m}_{x}^*, \mathbf{m}_{y}^*, \mathbf{m}_{z}^*)$

$$\underline{\vec{m}}^* = \underline{\vec{b}_C} + \underline{S} \cdot \underline{M} \cdot \underline{\vec{m}} + \underline{\vec{b}} + \underline{W}\underline{\vec{n}} + \overline{\vec{o}}$$
 measurement calibration estimation modelling



Where: \mathbf{m}^* = measured magnetic vector

 $\mathbf{b_c}$ = long term constant bias

s = scale factor

M = misalignement correction matrix

b = short term constant bias

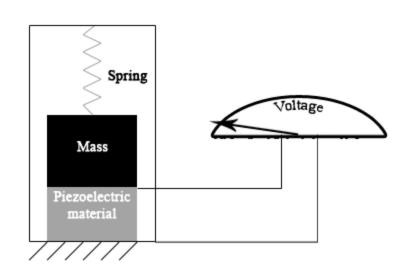
W=crosscorrelation matrix for white noise vector **n**

o = other (environmental) influences

Sensor Types: Accelerometer

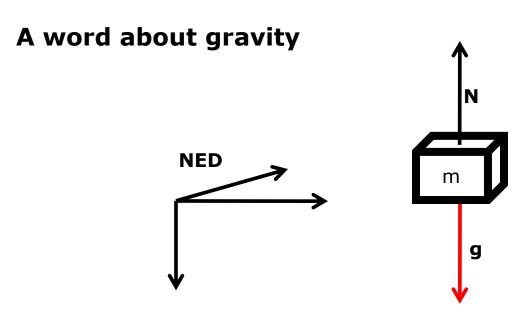
"An accelerometer is a device that measures proper acceleration, the acceleration experienced relative to freefall." [Wikipedia]

Basic principle: mass inertia effect



Most widely used types: MEMS devices (Microelectromechanical systems)

Sensor Types: Accelerometer



Note: $\mathbf{g} = (0,0,9.81)$

But the measured acceleration in hover mode is $\mathbf{a} = (0,0,-9.81)$

Sensor Types: Accelerometer

Sensor Model

Define noise model of measurement vector $\mathbf{a}^* = (\mathbf{a}_{x}^*, \mathbf{a}_{y}^*, \mathbf{a}_{z}^*)$

$$\vec{\underline{a}}^* = \vec{\underline{b}}_C + s \cdot \underline{M} \cdot \vec{\underline{a}} + \vec{\underline{b}} + \underline{W}\vec{\underline{n}} + \vec{o}$$
 measurement calibration estimation modelling



a* = measured acceleration vector

 $\mathbf{b_c}$ = long term constant bias

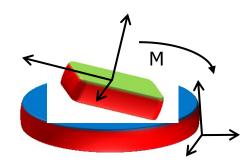
s = scale factor

M = misalignement correction matrix

b = short term constant bias

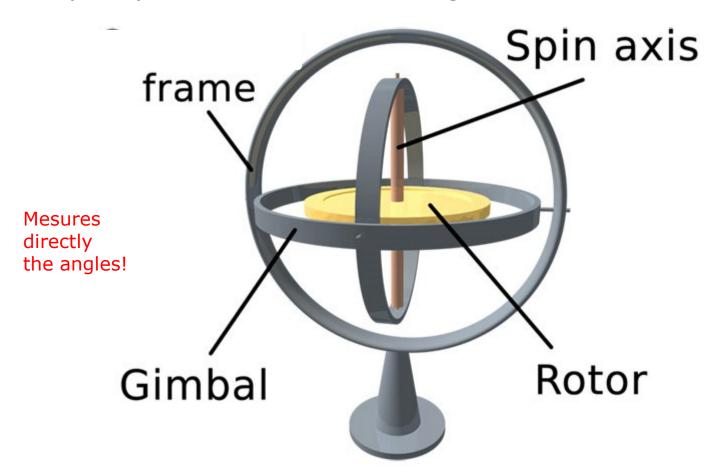
W=crosscorrelation matrix for white noise vector **n**

o = other (environmental) influences



Sensor Types: Gyroscope

" A spinning wheel is a device for measuring or maintaining orientation, based on the principles of conservation of angular momentum." [Wikipedia]



Sensor Types: Gyroscope

Gyroscope

Use the Coriolis effect to directly measure angular velocity:

Proof masses vibrate at frequency ω_r giving the position $p = A\sin(\omega_r t)$ The Coriolis effect introduces the acceleration: $a = -2(v \times \omega)$ with the velocity $v = \dot{p} = A\omega_r \cos(\omega_r t)$

We can measure the introduced acceleration with an accelerometer

Note: Angular velocity is directly measurable. Linear velocity is not! (we have to integrate linear acceleration)

Sensor Types

MEMS: Vibrating elements as Micro Electro-Mechanical System

Fiber Optic: Interference of light to detect mechanical rotation (5km fibre!)

Sensor Types: Gyroscope

Sensor Model

Define noise model of measurement vector $\mathbf{\omega}^* = (\omega_x^*, \omega_y^*, \omega_z^*)$

$$\underline{\vec{\phi}}^* = \underline{\vec{b}_C} + \underline{s} \cdot \underline{M} \cdot \underline{\vec{\phi}} + \underline{\vec{b}} + \underline{W}\underline{n} + \overline{o}$$
 measurement calibration estimation modelling

Where:

$$\omega$$
 * = measured angular velocities

 $\mathbf{b_c}$ = long term constant bias

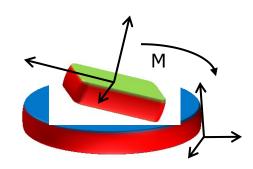
s = scale factor

M = misalignement correction matrix

b = short term constant bias

W=crosscorrelation matrix for white noise vector **n**

o = other (environmental) influences



Sensor Types: Air Pressure Sensor

Air pressure as a function of atmospheric height:

$$p = p_0 \cdot e^{-Mgh/RT}$$

$$h = \frac{(\ln(p) - \ln(p_0)) \cdot RT}{Mg}$$

Where:p = pressure at height h

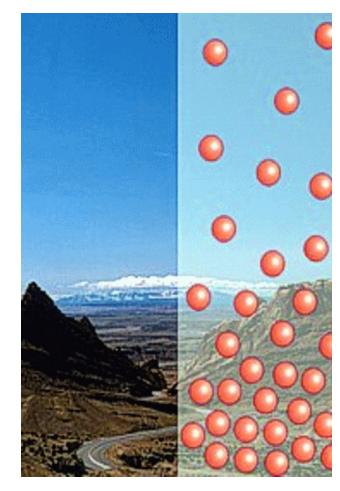
T = temperature at height h in K

R = 8.31432 N·m/(mol·K) universal gas const.

g = gravity

M = 0.0289644 kg/mol Molar mass of Earth's air

 $p_0 = 1.2250 \text{ kg/m}^3$



This model of the low altitude atmosphere is valid up to an altitude of 11000m.

Sensor Types: Air Pressure Sensor

Error Sources

Outdoor use: highly dependent on weather changes!

Pressure and temperature changes during flight time

 \rightarrow initially measured p₁ and T₁ are not reliable anymore

Indoor use: sudden pressure or/and temperature changes and pressure spikes

Closing doors provoke pressure spikes

Open windows or ventilation changes pressure and temperature

Heating/air conditioning changes temperature



sensor model has to account for bias, scaling factor and other noise terms

Sensor Types: Air Pressure Sensor

Sensor Model

Define noise model of measurement p* and T*

$$p^* = b_{\mathit{Cp}} + s_p \cdot p + b_p + n_p + o_p$$

$$T^* = b_{\mathit{CT}} + s_T \cdot T + b_T + n_T + o_T$$
 measurement calibration estimation modelling

Where: p^* and T^* = measured pressure and Temperature

 b_c = long term constant bias

s = scale factor

b = short term constant bias

n = white noise

o = other (environmental) influences

Ideally, the noise model has to be propagated for $\Delta h!$

Approximation is

 $\Delta h^* = \Delta h + b + n$ for a calibrated sensor with random walk bias b and white noise n

Sensor Types: Air Speed Sensor (Pitot tube)

" A pitot tube is a pressure measurement instrument used to measure fluid flow velocity" [Wikipedia]



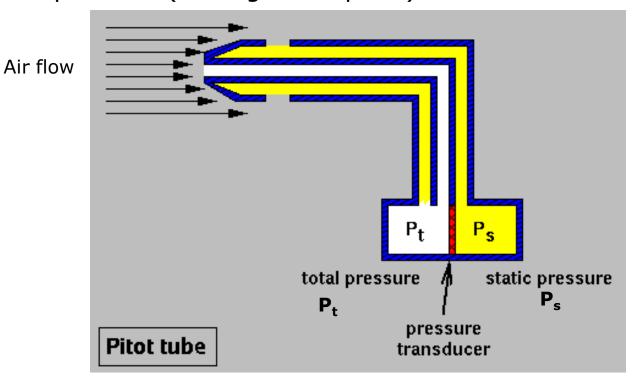




Sensor Types: Air Speed Sensor (Pitot tube)

Functioning

Measured air pressure (at stagnation point) is related to the air speed:



Bernoulli's eqution:

$$p_t = p_s + (0.5 \cdot \rho v^2)$$
 With $\rho = \text{air density}$ $v = \text{air flow velocity}$

Quaternions: Overview and Basics

Attitude Representation

Quaternion as a complex number: $q = q_4 + q_1i + q_2j + q_3k$ with: $i^2=j^2=k^2=-1$ -ij=ji=k -jk=kj=i -ki=ik=jConvention as proposed by JPL* (i.e. not Hamilton)

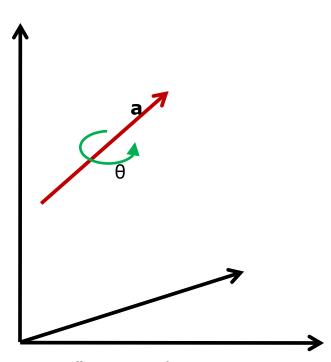
Define:
$$q = \begin{bmatrix} \vec{r} \\ q_4 \end{bmatrix} = [q_1 \quad q_2 \quad q_3 \quad q_4]^T$$

If the following is fulfilled:

$$\vec{r} = \begin{bmatrix} a_x \sin(\theta/2) \\ a_y \sin(\theta/2) \\ a_z \sin(\theta/2) \end{bmatrix} = \vec{a} \sin(\theta/2) , |\vec{a}| = 1$$

$$q_4 = \cos(\theta/2)$$

Then: q is a "quaternion of rotation"



*W.G. Breckenridge, "Quaternions – Proposed Standard Conventions", JPL Tech. Rep.

Quaternions: Overview and Basics

Quaternion Properties

Recall:

$$q = \begin{bmatrix} a_x \sin(\theta/2) \\ a_y \sin(\theta/2) \\ a_z \sin(\theta/2) \\ \cos(\theta/2) \end{bmatrix} = \begin{bmatrix} \vec{a} \sin(\theta/2) \\ \cos(\theta/2) \end{bmatrix}, |\vec{a}| = 1$$

Hence:

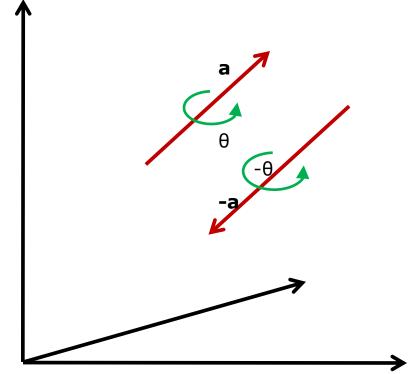
q and -q describe same rotation to final coordinate system position

$$|\vec{q}| = \sqrt{\vec{q}^T \vec{q}} = \sqrt{|\vec{r}|^2 + q_4^2} = 1$$

Other conventions:

$$q = q_1 + q_2 i + q_3 j + q_4 k$$





Quaternions: Calculation Rules

Multiplication:
$$q \otimes p = (q_1 i + q_2 j + q_3 k + q_4)(p_1 i + p_2 j + p_3 k + p_4)$$

In matrix form:
$$q \otimes p = M_q \cdot p = M_p q$$

also: $a \otimes b \otimes c = M_{ab} \cdot c = M_{ac} \cdot b = M_{bc} \cdot a$ coeff. are linear!

Note: $q \otimes p \neq p \otimes q$

Neutral Element q_T:

Hence:
$$q \otimes q_I = q_I \otimes q = q$$

$$q_I = \begin{bmatrix} 0 & 0 & 0 & 1 \end{bmatrix}^T$$

Inverse Element q⁻¹:

Hence:
$$q \otimes q^{-1} = q^{-1} \otimes q = q_I$$

$$q^{-1} = \begin{bmatrix} -\vec{r} \\ q_4 \end{bmatrix} = \begin{bmatrix} -\vec{a}\sin(\theta/2) \\ \cos(\theta/2) \end{bmatrix} = \begin{bmatrix} \vec{a}\sin(-\theta/2) \\ \cos(-\theta/2) \end{bmatrix}$$

complex conjugate

Quaternions: Calculation Rules

Vector Rotation

Imaginary quaternion: $v = v_1 i + v_2 j + v_3 k$, $v_4 = 0$, Note: |v|! = 1

Vector relation in different coordinate systems:

$$^{L}v=_{G}^{L}q\otimes^{G}v\otimes_{G}^{L}q^{-1}=_{G}^{L}C\cdot^{G}v$$

("global vector v expressed in local coordinate system")

Approximations

For very small rotations δq : $q = \delta q \otimes \hat{q}$

$$\delta q = \begin{bmatrix} \delta \vec{r} \\ \delta q_4 \end{bmatrix} = \begin{bmatrix} \vec{a} \sin(\delta \theta / 2) \\ \cos(\delta \theta / 2) \end{bmatrix} \approx \begin{bmatrix} 0.5 \delta \vec{\theta} \\ 1 \end{bmatrix}$$

with $\delta \vec{\theta}$ in the direction of the axis of rotation and the magnitude of the angle of rotation

Quaternions: Conversions

Quaternion to Rotation Matrix

Recall:
$${}^Lv=_G^Lq\otimes^Gv\otimes_G^Lq^{-1}=_G^LC\cdot^Gv$$

$${}^{L}v = \begin{bmatrix} (2q_4^2 - 1) \cdot I_{3x3} - 2q_4 \lfloor \vec{r}_X \rfloor + 2\vec{r}\vec{r}^T \\ 0 \end{bmatrix} \cdot \begin{bmatrix} {}^{G}v \\ 0 \end{bmatrix}$$

And thus:
$${}_{G}^{L}C({}_{G}^{L}q) = (2q_{4}^{2}-1)\cdot I_{3x3} - 2q_{4}\lfloor\vec{r}_{X}\rfloor + 2\vec{r}\vec{r}^{T}$$

Small angle approximation:

From:
$$\delta q \approx \begin{bmatrix} 0.5\delta \vec{\theta} \\ 1 \end{bmatrix}$$

We have:
$$C(\delta q) \approx I_{3x3} - \lfloor \delta \vec{\theta}_X \rfloor$$

$$\begin{bmatrix} \vec{v}_X \end{bmatrix} = \begin{bmatrix} 0 & -v_3 & v_2 \\ v_3 & 0 & -v_1 \\ -v_2 & v_1 & 0 \end{bmatrix}$$

skew symetric matrix

Quaternions: Derivatives & Integration

Quaternion Derivative

Recall:
$$\delta q = \frac{L(t + \Delta t)}{L(t)} q \approx \begin{bmatrix} 0.5 \delta \vec{\theta} \\ 1 \end{bmatrix}$$

Thus:
$$\vec{\omega} = \lim_{\Delta t \to 0} \frac{\delta \theta}{\Delta t}$$

Then:
$$^{L(t)}_{G}\dot{q} = \lim_{\Delta t \to 0} \frac{1}{\Delta t} \left(^{L(t+\Delta t)}_{G}q - ^{L(t)}_{G}q \right) = \frac{1}{2} \Omega(\vec{\omega}) \cdot ^{L(t)}_{G}q$$

With:
$$\Omega(\vec{\omega}) = \begin{pmatrix} 0 & \omega_z & -\omega_y & \omega_x \\ -\omega_z & 0 & \omega_x & \omega_y \\ \omega_y & -\omega_x & 0 & \omega_z \\ -\omega_x & -\omega_y & -\omega_z & 0 \end{pmatrix}$$

And $\vec{\omega} = L(t) \vec{\omega}$ in the local body frame L(t)

Quaternions: Derivatives & Integration

Quaternion Integration

Find a solution for the following first order differential equation:

$$_{G}^{L(t)}\dot{q}=rac{1}{2}\Omega(\vec{\omega})\cdot_{G}^{L(t)}q$$

Zeroth order quaternion integrator (for small angular velocities):

$${}^{L(t+\Delta t)}_{G}q = \left(I_{4x4} + \frac{\Delta t}{2}\Omega(\vec{\omega})\right) \cdot {}^{L(t)}_{G}q$$

First order quaternion integratior is slightly more complex*

^{*}Trawny N., Roumeliotis S.I., "Indirect Kalman Filter for 3D Attitude Estimation", Tech. Rep.

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Sensor coupling classification 1 How is the second sensor used?

Correction

Uses information from one sensor to correct or verify another.

Ex: Integrate the IMU accelerations for a position estimate and compensate for the drift with a DGPS measurement.

(Note: normal GPS is not accurate)

Ex: Estimate the rotation using vision, verify accuracy with gyros

Colligation

Merges different parts of the sensors

Ex: Use gyro information to eliminate rotational effects on optical flow translation estimation

Fusion

Merges sensor information in a (statistically) optimal way

Sensor coupling classification 2 Loose/Tight coupling

Loose coupling

Using separate INS and SFM blocks running at different rates and exchanging information

Tight coupling

Combining the disparate raw data of vision and inertial sensors in a single block

Sensor coupling classification 3 Statistical/Deterministical fusion

Statistical fusion

Fusion using a filter based strategy (Kalman filter, EKF, UKF, Particle filter)

Deterministic fusion

Fusion using a closed-form solution

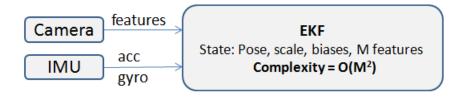
Four variations

Pose estimator (black box) unscaled pose EKF
State: Pose, scale, biases
Complexity = constant

- Loosely coupled
 - EKF-based pose estimation filter with visual pose and IMU measurements (scale as additional state variable) [Weiss, ICRA 2011]
 - Closed-form solution for scale determination using short-term integration of inertial readings (delta observations) and consecutive SLAM poses in relative scale [Kneip, IROS 2011]

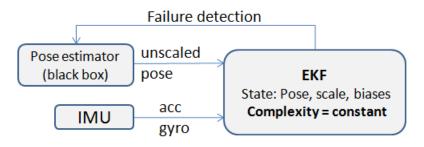
Statistical fusion

Tightly coupled



- EKF-based SLAM with centripetal acceleration motion model (using intertial readings) [Strelow 2004, Jones 2009, Kelly 2009]
- Closed-form solution for direct speed determination using inertial delta observations and single feature correspondences [Kneip et al., ICRA 2011]

Four variations



- Loosely coupled
 - EKF-based pose estimation filter with visual pose and IMU measurements (scale as additional state variable) [Weiss, ICRA 2011]
 - Closed-form solution for scale determination using short-term integration of inertial readings (delta observations) and consecutive SLAM poses in relative scale [Kneip, IROS 2011]

Deterministic fusion

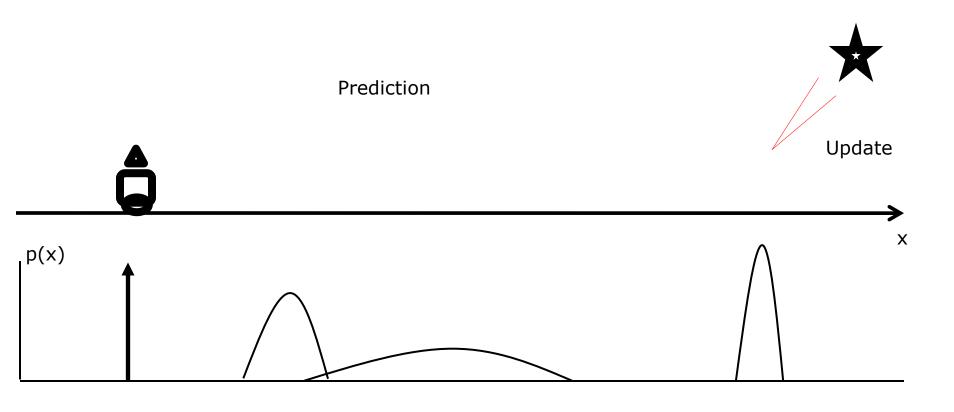
Tightly coupled

- Camera features

 IMU acc State: Pose, scale, biases, M features

 Complexity = O(M²)
- EKF-based SLAM with centripetal acceleration motion model (using intertial readings) [Strelow 2004, Jones 2009, Kelly 2009]
- Closed-form solution for direct speed determination using inertial delta observations and single feature correspondences [Kneip et al., ICRA 2011]

Review Linear Kalman Filter



^{*} Peter Maybeck "Stochastic Models, Estimation, and Control" Vol 1-3

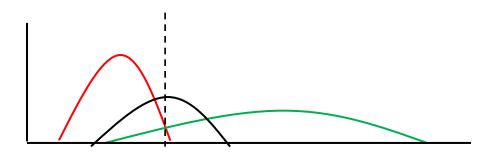
Review Linear Kalman Filter

Prediction step:

$$p(x_{t+1}|u_t) = \int_{x_t} p(x_{t+1}|u_t, x_t) p(x_t) dx_t = p(x_{t+1}|u_t, x_t) \otimes p(x_t)$$

Update step: bayes rule

$$p(x_{t+1}|z_{t+1}) = \frac{p(z_{t+1}|x_{t+1}) \cdot p(x_{t+1})}{p(z_{t+1})} = \eta p(z_{t+1}|x_{t+1}) p(x_{t+1})$$



Review Linear Kalman Filter

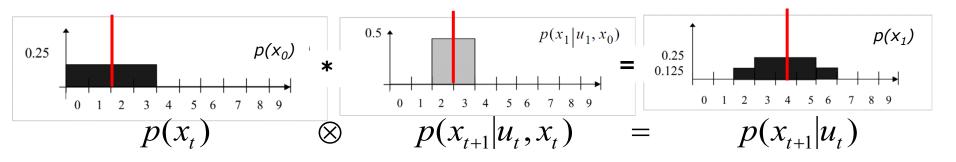
Linear Filter System: $\vec{x}_{k+1} = A\vec{x}_k + B\vec{u}_k + W\vec{n}_x$

$$\vec{z} = H\vec{x} + V\vec{n}_z$$

Prediction step:

propagating the mean and variance of the gaussian distribution p(x)

mean: $\vec{x}_{k+1} = A\vec{x}_k + B\vec{u}_k$



covariance matrix P: use the error propagation law

$$P_{k+1} = AP_k A^T$$

Review Linear Kalman Filter

Linear Filter System:
$$\vec{x}_{k+1} = A\vec{x}_k + B\vec{u}_k + W\vec{n}_x$$

$$\vec{z} = H\vec{x} + V\vec{n}_z$$

Update step: Minimum mean-square error estimator
$$\min(E(\widetilde{z}_k - z_k)^2)$$

Error:
$$y = \widetilde{z} - Hx$$

Error covariance (innovation):
$$S = R + HPH^T$$
 $R = V \operatorname{cov}(n_z)V^T$

Optimal kalman gain:
$$K = P_k H^T S^{-1}$$

State:
$$x_{k+1} = x_k + K\vec{y}$$

Covariance:
$$P_{k+1} = \text{cov}(\widetilde{x}_k - x_k) = (I - KH)P_k(I - KH)^T + KRK^T$$

Peter Maybeck "Stochastic Models, Estimation, and Control" Vol 1-3

Extended Kalman Filter

Non-linear Filter System: $\dot{\chi} =$

$$\dot{x} = f(x, u) + Wn_x$$

$$z = h(x) + Vn_z$$

Use jacobians for prediction and update step:

$$A = J(f(x, u))$$

$$H = J(h(x))$$

Note:

This is a linear approximation step around the current state. It may yield poor results in highly nonlinear systems. More robust filters are in this case particle or unscented kalman filters. (at the cost of calculation power)