

# K. J. SOMAIYA COLLEGE OF ENGINEERING

(AFFILIATED TO THE UNIVERSITY OF MUMBAI)

Candidate Roll No. \_\_\_\_\_ (In figures)

Name : \_\_\_\_\_

Date : \_\_\_\_\_

Examination : \_\_\_\_\_

Subject : \_\_\_\_\_

Test Exam.

## Basics of Digital Communication

Junior Supervisor's full  
Signature with Date

200

Branch/Semester \_\_\_\_\_

\_\_\_\_\_

Question No.	1	2	3	4	5	6	7	8	9	10	11	12	Total
Marks Obtained													

PCM

## Digital Representation of Analog Signal:

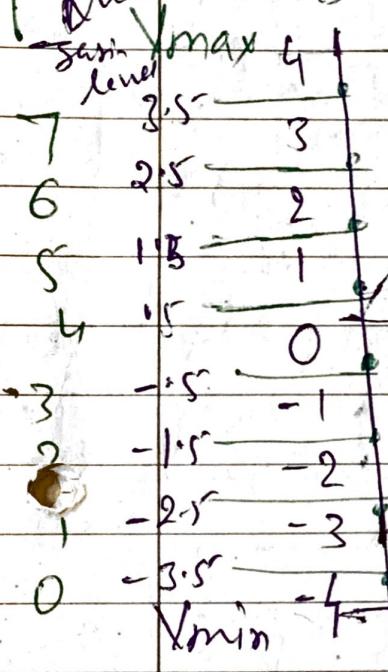
[ Sampling ]

[ Quantization ]

[ Encoding ]

Code No

Ques mVt Volts



m(t)

Analog (A)

Step Size

$$S = \frac{V_H - V_L}{M}$$

M = No. of levels

a Sample Value = 1.3

Step size

$$S = \frac{V_H - V_L}{M}$$

b Nearest Quantization

level → 1.5

c Code No.

S = 1.5

Binary representation = 1010

## Electrical Representation of Binary Digits



101



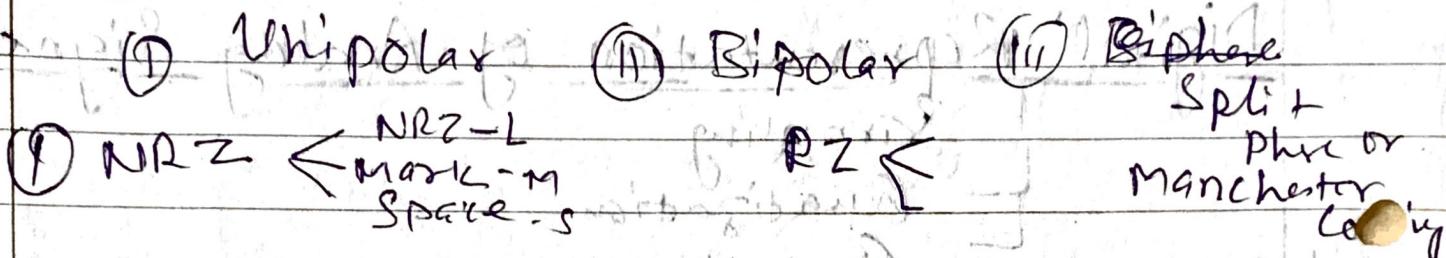
Pulse Representation

Representation by  
Voltage level

## Line Coding

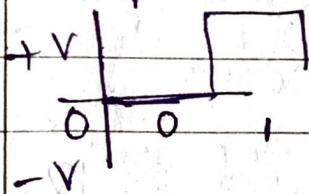
### Issues in Digital Transmission

- (i) Power & B.W. required for transmission
- (ii) ability to extract timing information
- (iii) Error monitoring ability
- (iv) presence of low freq. or DC component which is unsuitable for ac coupled Ckt

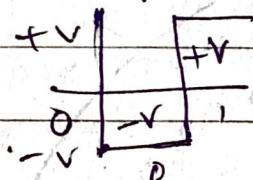


### NRZ

#### ① Unipolar

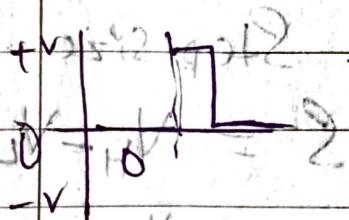


#### Bipolar

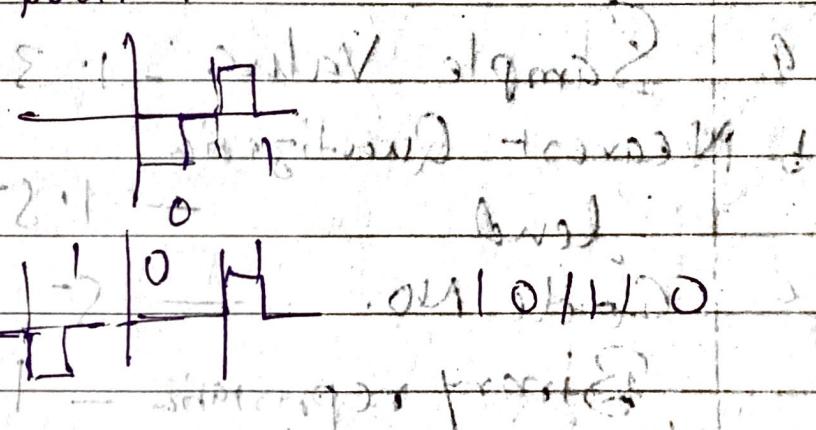


### ② RZ

#### Unipolar



#### Bipolar RZ



### ③

#### Ac-Split Phase (Manchester Coding)



Quantization error is the difference between the Quantized Signal and the Original Signal from which it was derived different from One Another in a random manner. This difference or error may be viewed as a noise due to the Quantization process and is called quantization error.

Mean Square quantization error

$$e = m - m_k$$

Where  $e = \text{diff. between the}$

Original & Quantized Signal Voltage

$$\overline{e^2} = \int_{m_k - S/2}^{m_k + S/2} f(m) (m - m_k)^2 dm$$

(Mean square quantization error)

$$+ \int_{m_k + S/2}^{m_k + 3S/2} f(m) (m - m_k)^2 dm + \dots$$

Where

$$e = m(t) - m_k$$

Error

&  $f(m)dm = \text{Probability that}$   
 $m(t)$  lies in the Voltage  
 $\text{range } m - dm/2 \text{ to } m + dm/2$

Substitute  $S/2 \leq m - m_k$

$$\overline{e^2} = (f^{(1)} + f^{(2)} + \dots) \cdot \int_{-S/2}^{S/2} x^2 dx \left[ \frac{x^3}{3} \right]_{-S/2}^{S/2} = (f^{(1)} + f^{(2)} + \dots) \frac{S^3}{12}$$

$$(f^{(1)}(s) + f^{(2)}(s) + \dots) \frac{s^2}{T^2}$$

$$\overline{e^2} = \frac{s^2}{T^2}$$

$f^{(1)}(s)$  is the prob. that the  
 Signal voltage  $m(t)$  will be

in first quantization range,  $f^{(2)}g$  is the probability that m is in the second quantization range. Hence the sum of terms in the parenthesis has a value of unity.

$\overline{e^2} = \text{mean square quantization error}$  is

$$\boxed{\overline{e^2} = \frac{3}{12}}$$

No. of Quantization levels:

$$Q = 2^N$$

Where  $N = \text{No. of bits/word}$

each Sample Convert to N-bit code-word

No. of bits/sec.

$$\boxed{\text{Signaling rate}} = \text{No. of Samples/sec.} \times \text{No. of bits/sample}$$

$$\text{Sampling rate} = f_s \times N$$

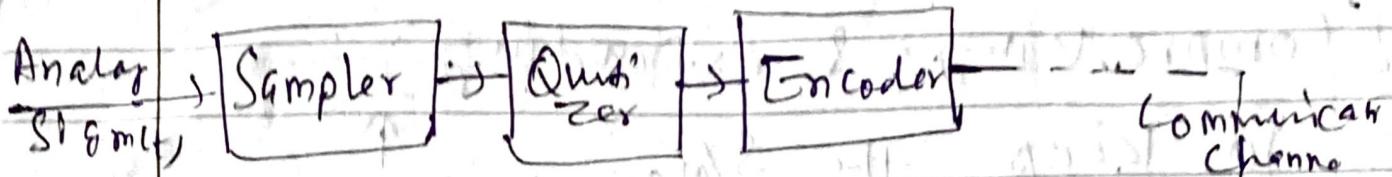
$$\boxed{2f_s N}$$

Transmitted B.W. of PCM

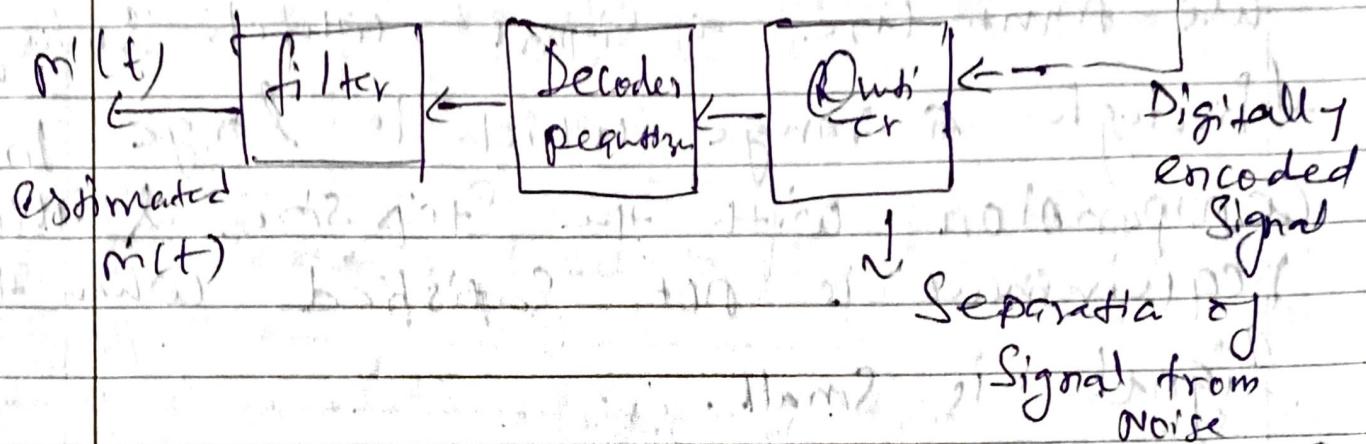
$$\boxed{B_{BW} = \frac{1}{2} N f_s}$$

# PCM

Encoder



Decoder



## Companding

in Quantization Voltage range from  $V_L$  to  $V_H$

① If the signal  $m(t)$  should make excursion beyond the bounds  $V_H$  and  $V_L$ , within these bounds, the instantaneous quantization error never exceeds  $\pm \frac{1}{2}$ . While outside these bounds the error is larger.

② Similarly consider a case in which  $m(t)$  has a Re. P to P Voltage which is less than  $S$  and never crosses one of the transition level. In such a case  $m_q(t)$  will be a fixed dc voltage & will bear no relationship to  $m(t)$ . Reconstruction is NOT Possible.

$$Noise = \frac{S^2}{T^2}$$

||

If no. of quantization level is  $M$

$$\text{then } M_S = 2^N \quad V = \frac{M_S - S}{2}$$

11  
of

The dynamic range can be materially improved by a process called Companding (Compressing & Expanding)

(iii) To keep the S/N (Quantization) we must use a signal which swings through a range which is large in comparison with the step size. If this requirement is not satisfied when the signal is small.

(iv) Before applying the signal to the quantizer we pass it through a N/W which has an Input O/p characteristic as shown

At low amplitude the slope is large than at large amplitude

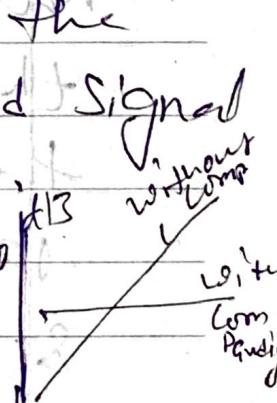
A signal transmitted through such a N/W will have the extremes of its waveform compressed.

The compression producer signal distortion

To undo the distortion, at the receiver we pass the received signal through an expander network.

(Inverse to compression)

(B/I)<sub>O</sub>



Output  $V_O$  vs Input  $V_I$  (compression)

$V_I(\min)$

Expansion

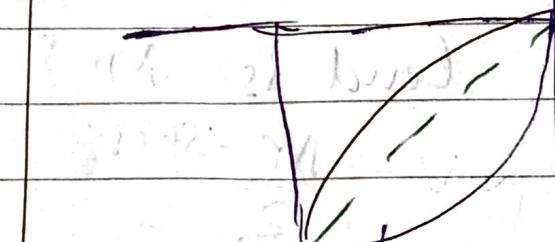
$V_O(\max)$

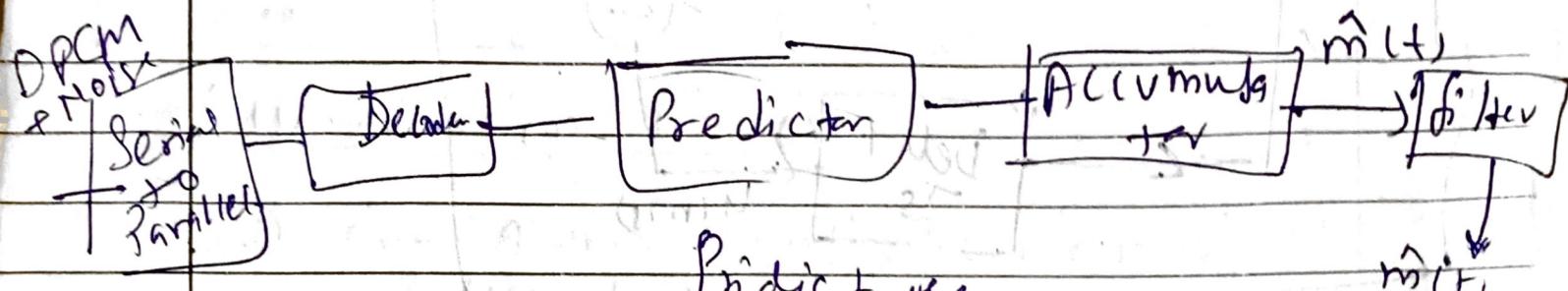
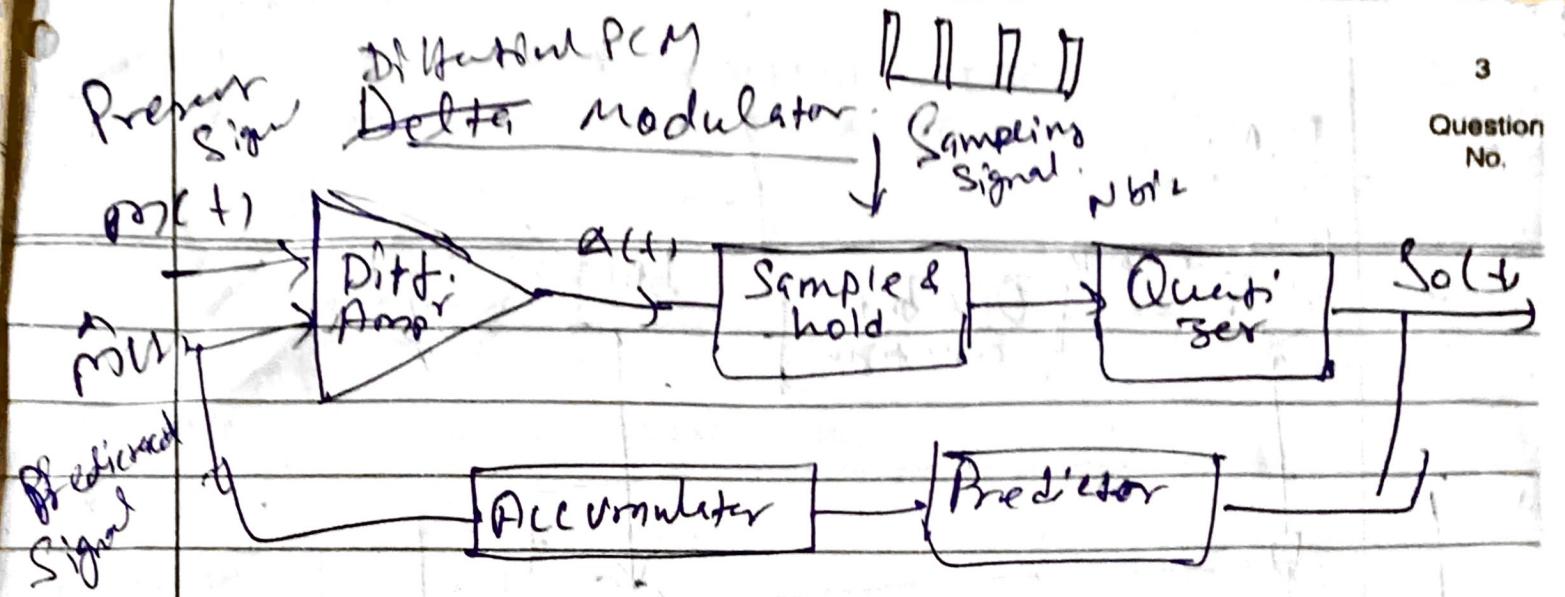
Dynamic Range of transmission

NO Compression

Recinv Expansion

$V_I(\min) \rightarrow$  Input  $V_I$





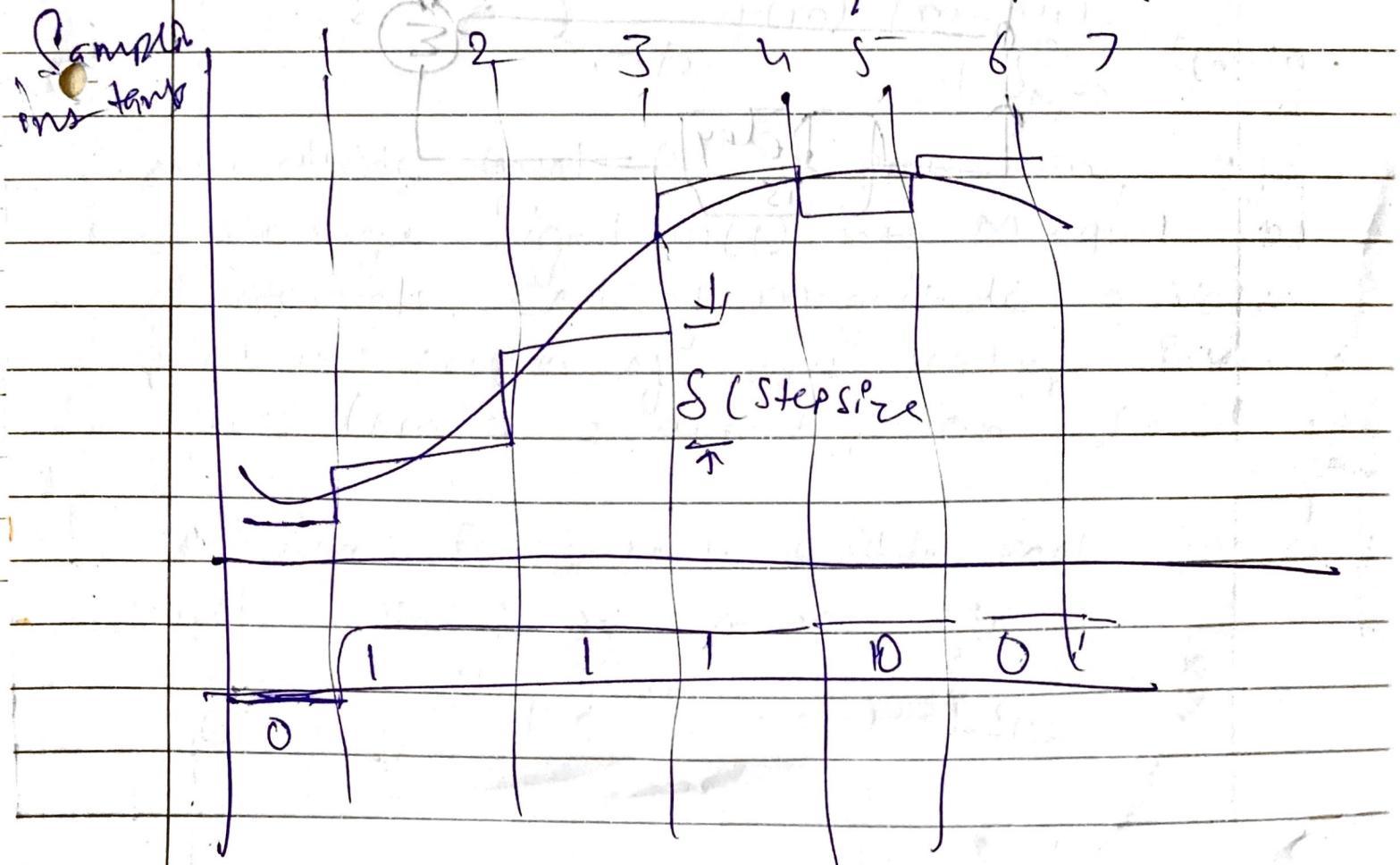
This helps in reducing  $\rightarrow$  redundancy

If  $f_s \gg$  Nyquist rate

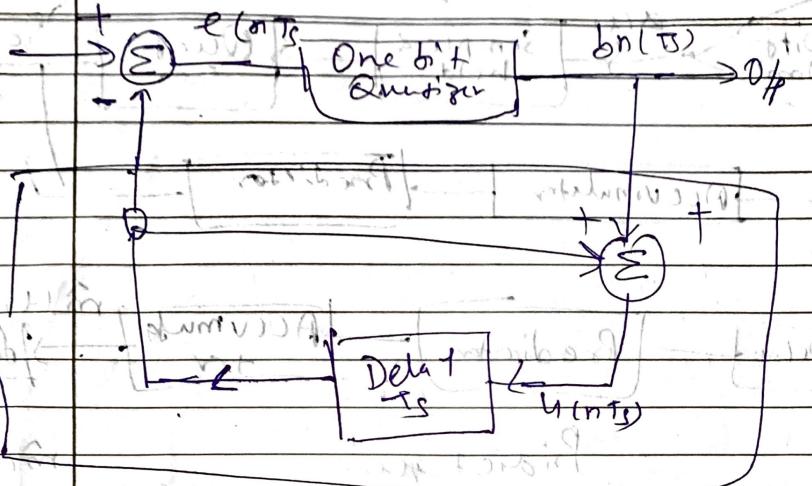
then there is Correlation

$N$  No. of bits transmitted per sample

Delta Modu One bit / Sample

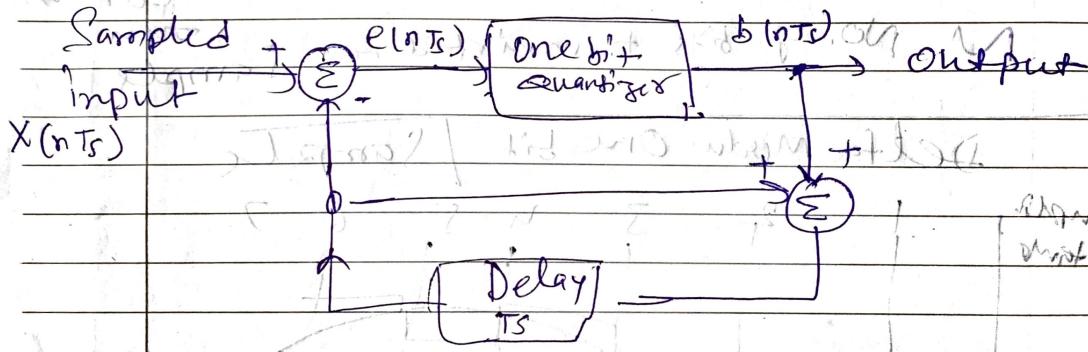


D.M.



Accumulator

D.M. Receiver



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Branch/Semester: \_\_\_\_\_

Junior Supervisor's full  
Signature with Date

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## Quantization Error.

Mean-Square quantization error  $\overline{e^2}$

$e = \text{diff. betn the Original and Quantized Signal Voltages.}$

Quantizer O/p =  $m_k$ ,

$$e = m(t) - m_k \quad m(t) = \text{Message Signal.}$$

Let  $f(m) dm$  be the probability that  $m(t)$  lies in the Voltage range  $m - dm/2$  to  $m + dm/2$

then the mean-square quantization error is

$$\overline{e^2} = \int_{m_1 - S/2}^{m_2 + S/2} f(m) (m - m_1)^2 dm + \int_{m_2 - S/2}^{m_3 + S/2} f(m) (m - m_2)^2 dm + \dots$$

Let's divide total Peak-to-Peak range of the Message Signal  $m(t)$  into  $M$  equal voltage intervals each of magnitude  $S$  Volts at the center of each voltage interval we locate a quantization level  $m_1, m_2, \dots, m_M$ .

Let's  $f(m)$  is constant within each quantization level. then  $\Delta x = m - m_k$

$$\overline{e^2} = (f^{(1)} + f^{(2)} + \dots) \int_{-S/2}^{S/2} x^2 dx$$

$$\frac{S^3}{12}$$

$$f^{(1)}(S) \text{ is the probability that signal voltage will be in } 1^{\text{st}} \text{ bin.}$$

So sum of total probabilities

$$\left[ \overline{\sigma^2} = \overline{T^2} \right].$$

Companding-

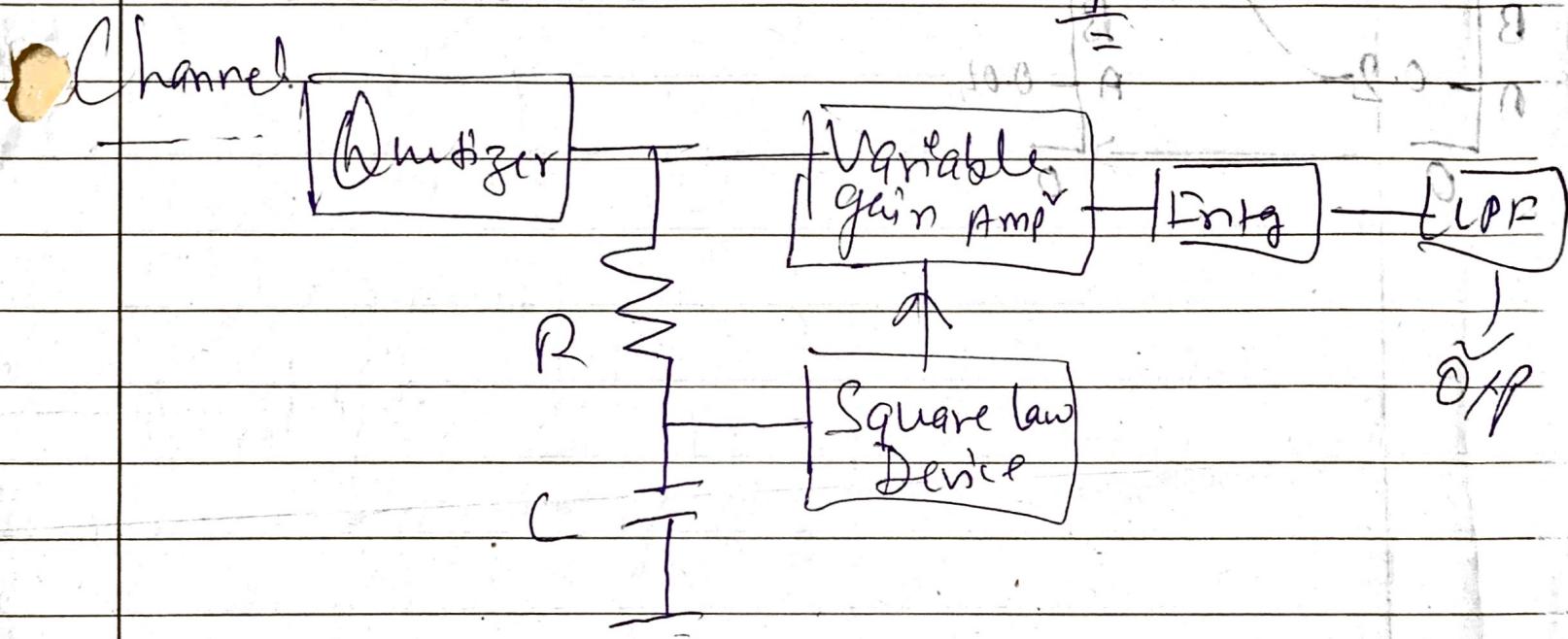
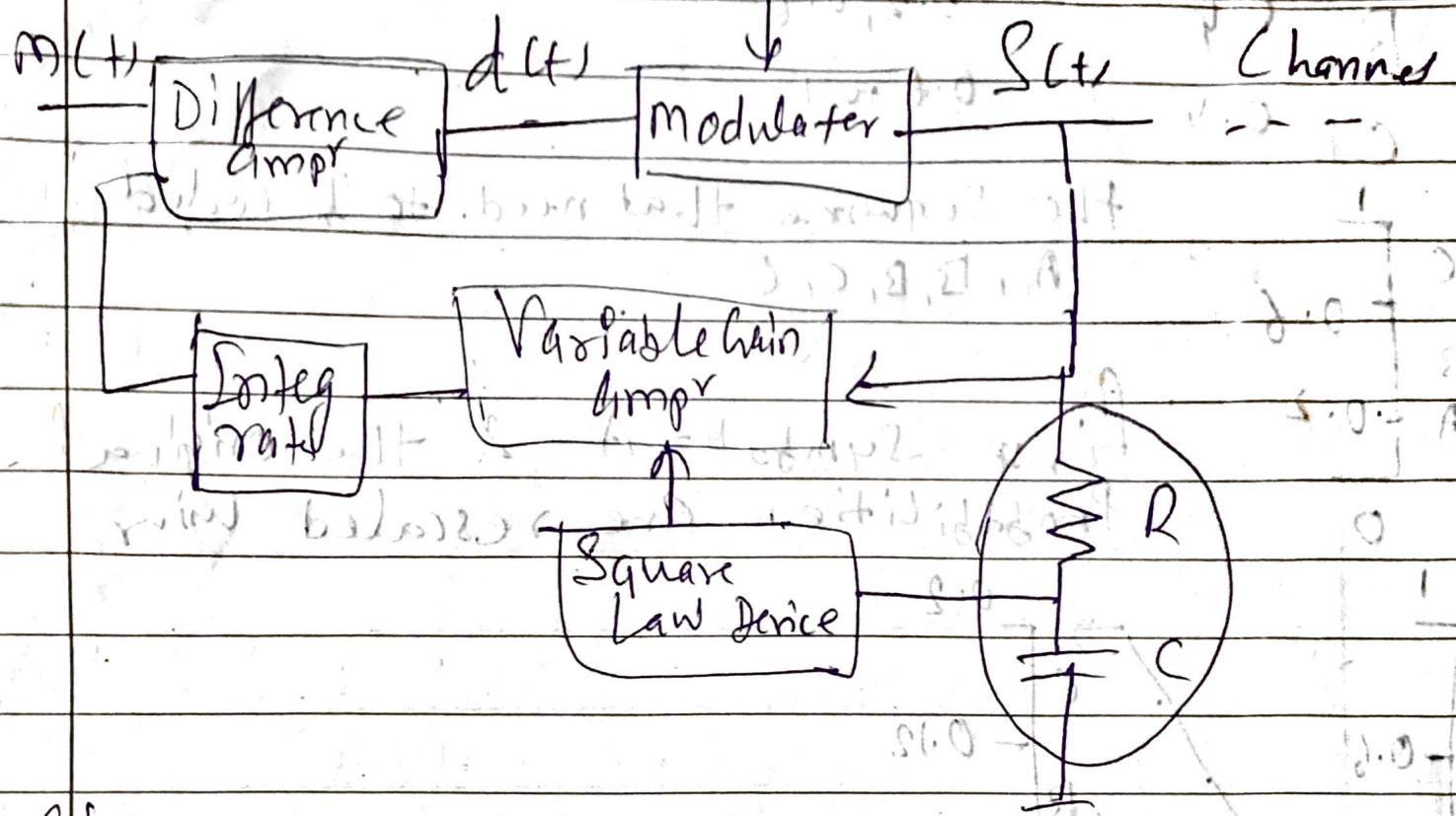
Signal

Quantization Noise ~~versus~~ noise

→ noise reduction, sample - nRMS

Actual noise level - 9

# ADPCM



## Delta modulation

Sampling Sign.

