

Wireless Communication Systems

HW3

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For simplicity, I set $T = 1$ for all cases in my simulation.

1. Filtered Gaussian Noise Method :

I use first-order low-pass digital filter to generate the channel output as following

$$\begin{pmatrix} g_{I,k+1} \\ g_{Q,k+1} \end{pmatrix} = \zeta \begin{pmatrix} g_{I,k} \\ g_{Q,k} \end{pmatrix} + (1 - \zeta) \begin{pmatrix} w_{1,k} \\ w_{2,k} \end{pmatrix} \quad (1)$$

where $g_{I,k}, g_{Q,k}$ are in $-$ phase and quadrature component at time kT

$$w_{1,k}, w_{2,k} \stackrel{iid}{\sim} \mathcal{N}(0, \sigma^2)$$

$$\zeta = 2 - \cos\left(\frac{\pi f_m T}{2}\right) - \sqrt{\left(2 - \cos\left(\frac{\pi f_m T}{2}\right)\right)^2 - 1}$$

$$\sigma^2 = \left(\frac{1 + \zeta}{1 - \zeta}\right) \frac{\Omega_p}{2}, \quad \frac{\Omega_p}{2} \text{ is the PSD of input noise source}$$

and I use the equation at the lecture note CH2-p93 to simulate the autocorrelation $\phi_{g_I g_I}$

$$\phi_{g_I g_I}(n) = \phi_{g_Q g_Q}(n) = \frac{1 - \zeta}{1 + \zeta} \sigma^2 \zeta^{|n|}$$

Furthermore, because I generate the output by (1) with initial value $g_{I,0}, g_{Q,0}$ (here I set 0 for both), the value of envelope level around 0 is significantly affected by the initial value, so I use the last 300 points of g instead

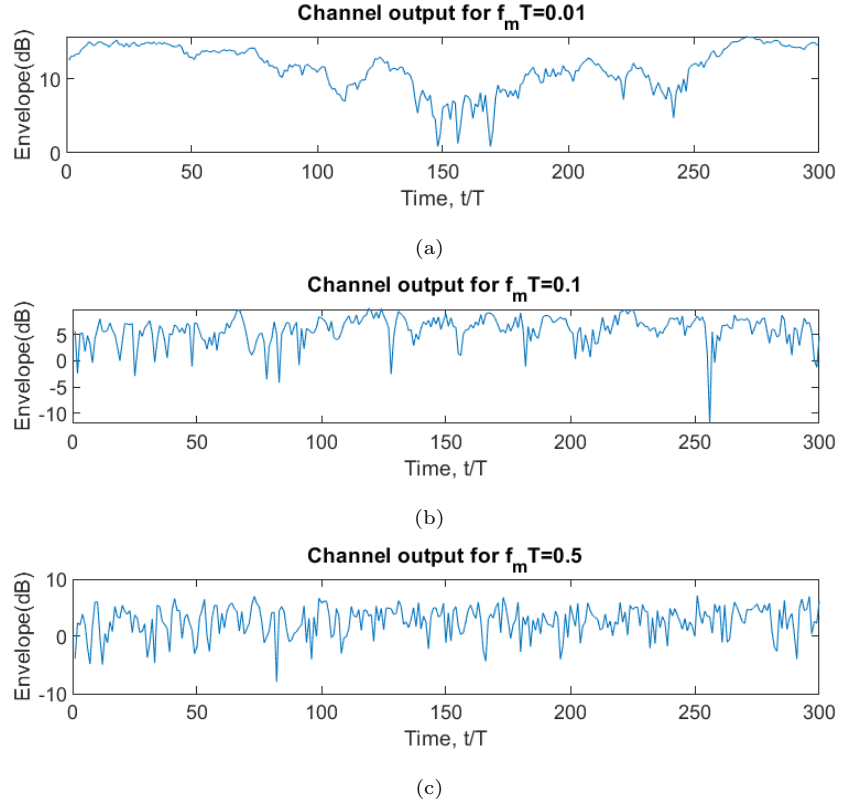
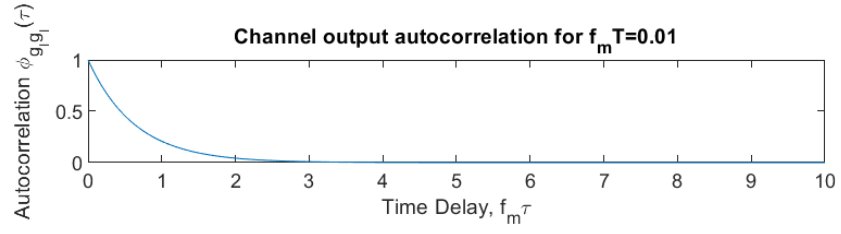


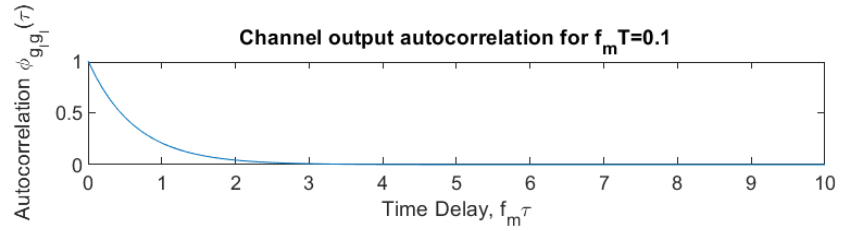
Figure 1: Envelope level(dB) of channel output (a) $f_m T = 0.01$
(b) $f_m T = 0.1$ (c) $f_m T = 0.5$

$$f_m T \uparrow \Rightarrow \zeta \downarrow \Rightarrow (1 - \zeta) \uparrow$$

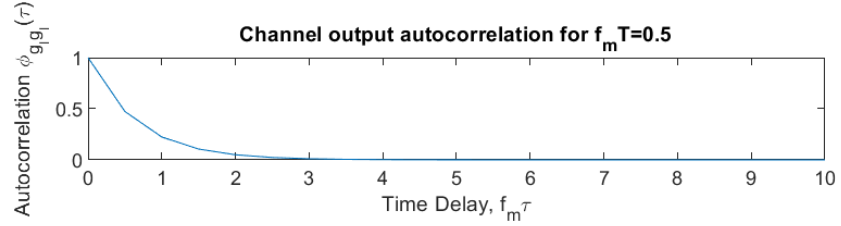
Thus, as $f_m T$ become large, the noise affects output more seriously



(a)



(b)



(c)

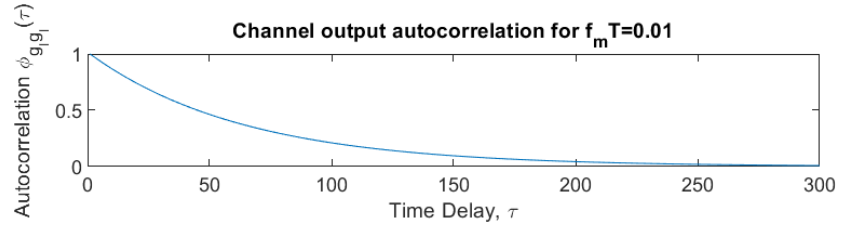
Figure 2: Autocorrelation with normalization of in-phase component of channel output (a) $f_m T = 0.01$ (b) $f_m T = 0.1$ (c) $f_m T = 0.5$

Since the channel output is $g(t) = g_I(t) + jg_Q(t)$, the autocorrelation of channel output is given as

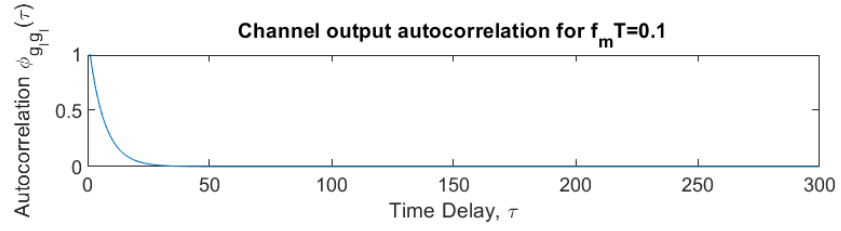
$$\phi_{gg}(\tau) = 2\phi_{g_I g_I}(\tau) = 2\phi_{g_Q g_Q}(\tau)$$

Thus, I present the autocorrelation $\phi_{g_I g_I}(\tau)$ instead.

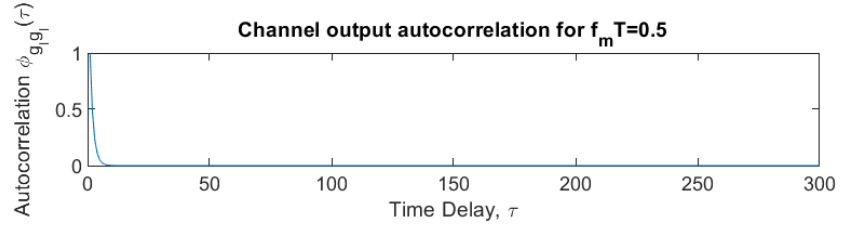
And due to normalization, Figure 2-(a), 2-(b), 2-(c) are look like the same. I will show non-normalized results later



(a)



(b)



(c)

Figure 3: Autocorrelation without normalization of in-phase component of channel output (a) $f_m T = 0.01$ (b) $f_m T = 0.1$ (c) $f_m T = 0.5$

Without normalization, it is easy to see that as f_m become large, $\phi_{g_I g_I}(\tau)$ decays more rapidly

2. Sum of Sinusoids Method :

I generate the channel output as following

$$g(t) = g_I(t) + jg_Q(t)$$

$$= 2\sqrt{2} \left\{ \left[\sum_{n=1}^M \cos(\beta_n) \cos(2\pi f_n t) \right] + j \left[\sum_{n=1}^M \sin(\beta_n) \cos(2\pi f_n t) \right] \right\}$$

where I ignored the term of $\cos(\alpha) \cos(2\pi f_m t)$ of $g_I(t)$ and the term of $\sin(\alpha) \cos(2\pi f_m t)$ of $g_Q(t)$ is equal to 0 as choose $\alpha = 0$

Note that $\beta_n = \frac{\pi n}{M}$, $f_n = f_m \cos(\theta_n)$, where $\theta_n = \frac{2\pi n}{N}$

And I simulate the autocorrelation by using the MATLAB function `autocorr`

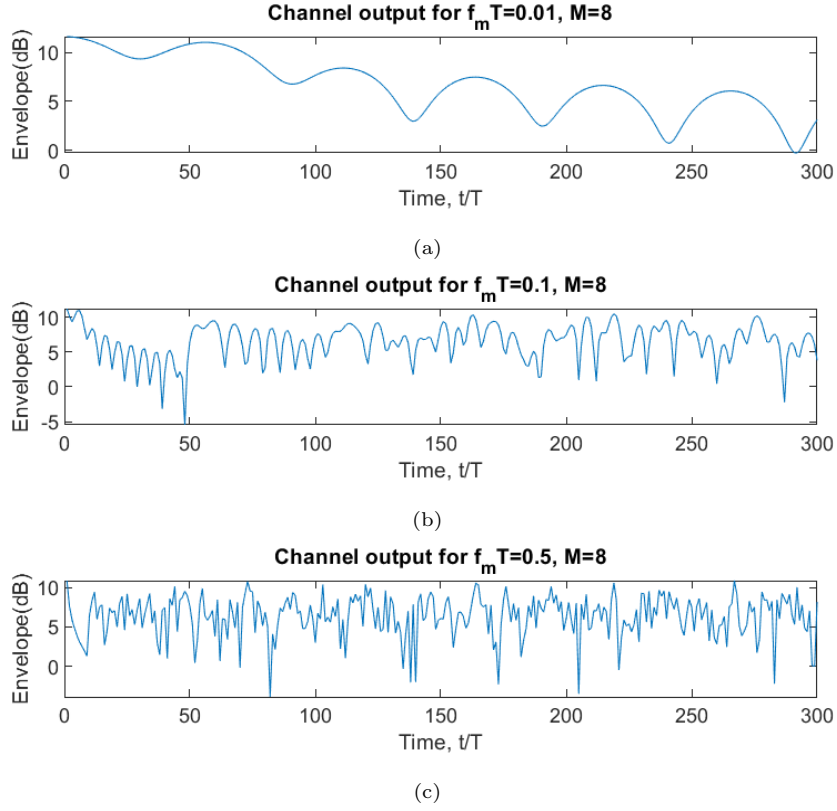


Figure 4: Envelope level(dB) of channel output for $M = 8$ (a) $f_m T = 0.01$ (b) $f_m T = 0.1$ (c) $f_m T = 0.5$

Since the initial phase(the phase at $t = 0$) is 0 for all sinusoid components of g_I, g_Q , it occurs constructive interferences around $t = 0$. Thus, the envelope level get relatively large value around $t = 0$. Otherwise, as f_m become larger, the output of channel is more noise-like

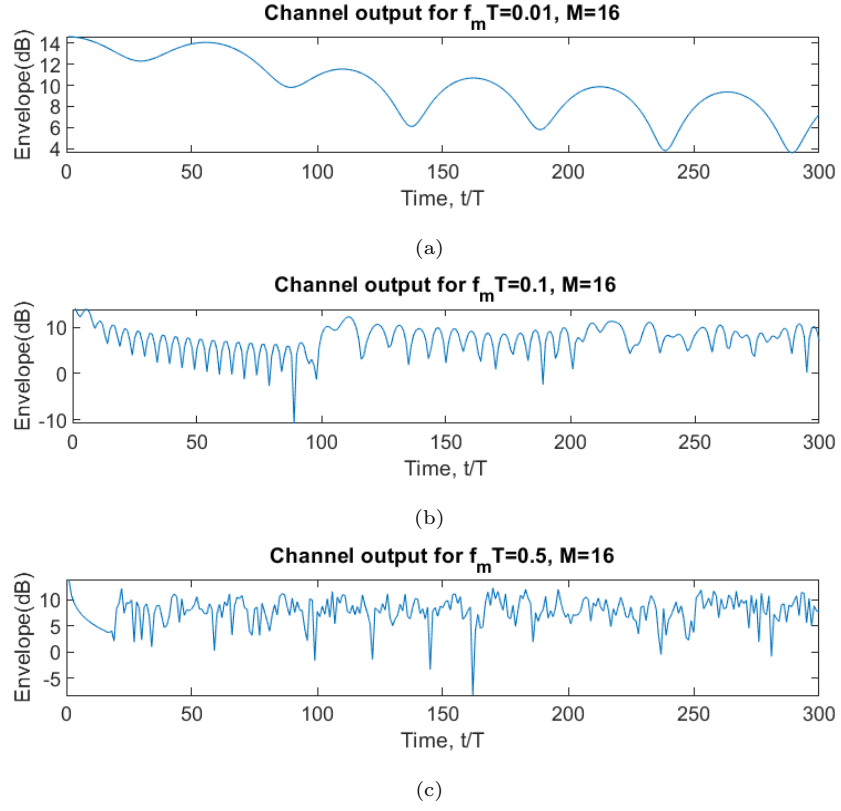
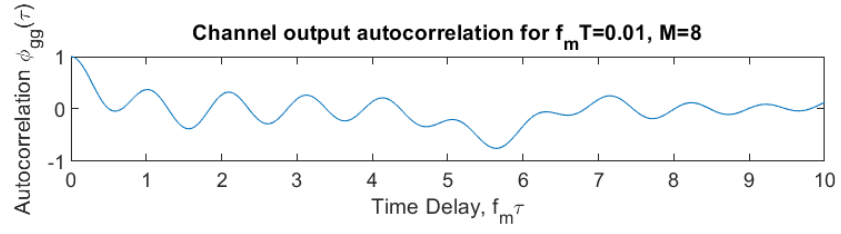
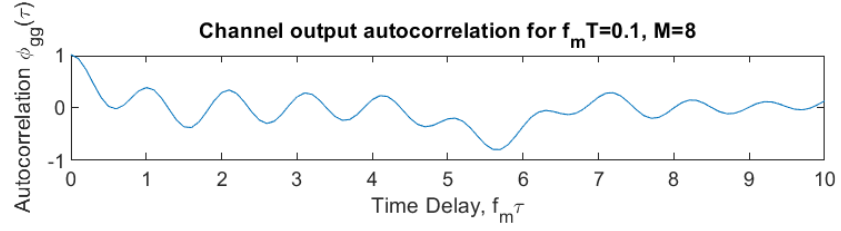


Figure 5: Envelope level(dB) of channel output for $M = 16$ (a) $f_m T = 0.01$ (b) $f_m T = 0.1$ (c) $f_m T = 0.5$

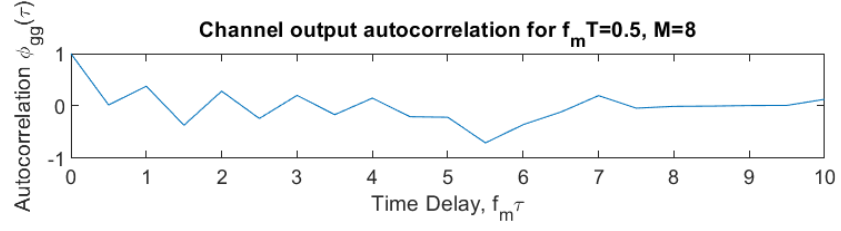
Compare the case of $M = 8$ with $M = 16$, we can observe that the duration of constructive interference around $t = 0$ of $M = 16$ is longer than $M = 8$



(a)



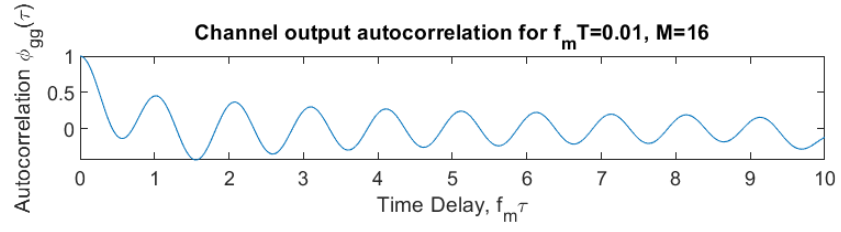
(b)



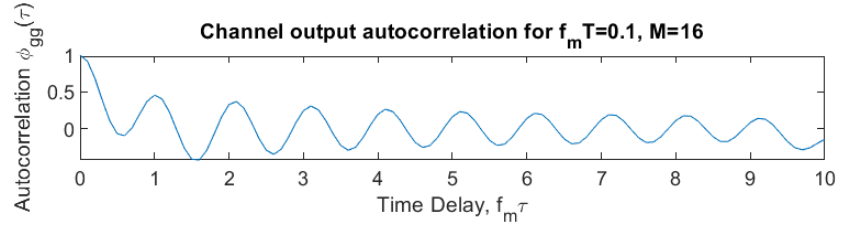
(c)

Figure 6: Autocorrelation with normalization of channel output for $M = 8$
(a) $f_m T = 0.01$ (b) $f_m T = 0.1$ (c) $f_m T = 0.5$

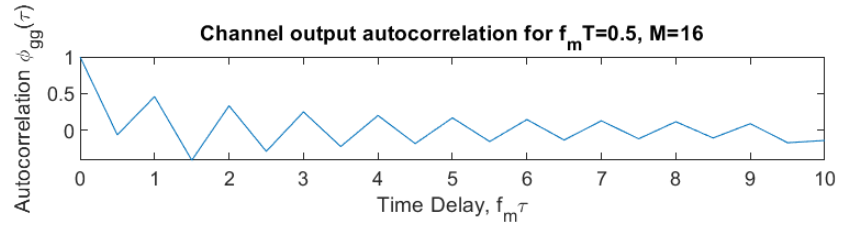
We can observe that the value of autocorrelation is much inaccurate after $f_m \tau = 6$ no matter what the value of $f_m T$ is



(a)



(b)



(c)

Figure 7: Autocorrelation with normalization of channel output for $M = 16$
(a) $f_m T = 0.01$ (b) $f_m T = 0.1$ (c) $f_m T = 0.5$

As we increase M to 16, the value of autocorrelation is more accurate

3. Discussion :

Compare the envelope level and autocorrelation generated by Filtered Gaussian Noise Method with Sum of Sinusoids Method, we can see that the envelope level generated by Filtered Gaussian Noise Method is more noise-like, and maintain the randomness. For Sum of Sinusoids Method, even though we can choose different start time to get the property that is similar to randomness, it is still a deterministic periodic signal indeed. However, as M is sufficient large, the value of autocorrelation generated by Sum of Sinusoids Method is much more like the real value of autocorrelation than Filtered Gaussian Noise Method.