

# Worksheet 17

## To accompany Chapter 6.4 Models of Discrete-Time Systems

### Colophon

This worksheet can be downloaded as a [PDF file \(https://cpjobling.github.io/eg-247-textbook/worksheets/worksheet17.pdf\)](https://cpjobling.github.io/eg-247-textbook/worksheets/worksheet17.pdf). We will step through this worksheet in class.

An annotatable copy of the notes for this presentation will be distributed before the second class meeting as **Worksheet 9** in the **Week 4: Classroom Activities** section of the Canvas site. I will also distribute a copy to your personal **Worksheets** section of the **OneNote Class Notebook** so that you can add your own notes using OneNote.

You are expected to have at least watched the video presentation of [Chapter 6.4 \(https://cpjobling.github.io/eg-247-textbook/dt\\_systems/4/dt\\_models\)](https://cpjobling.github.io/eg-247-textbook/dt_systems/4/dt_models) of the [notes \(https://cpjobling.github.io/eg-247-textbook\)](https://cpjobling.github.io/eg-247-textbook) before coming to class. If you haven't watch it afterwards!

After class, the lecture recording and the annotated version of the worksheets will be made available through Canvas.

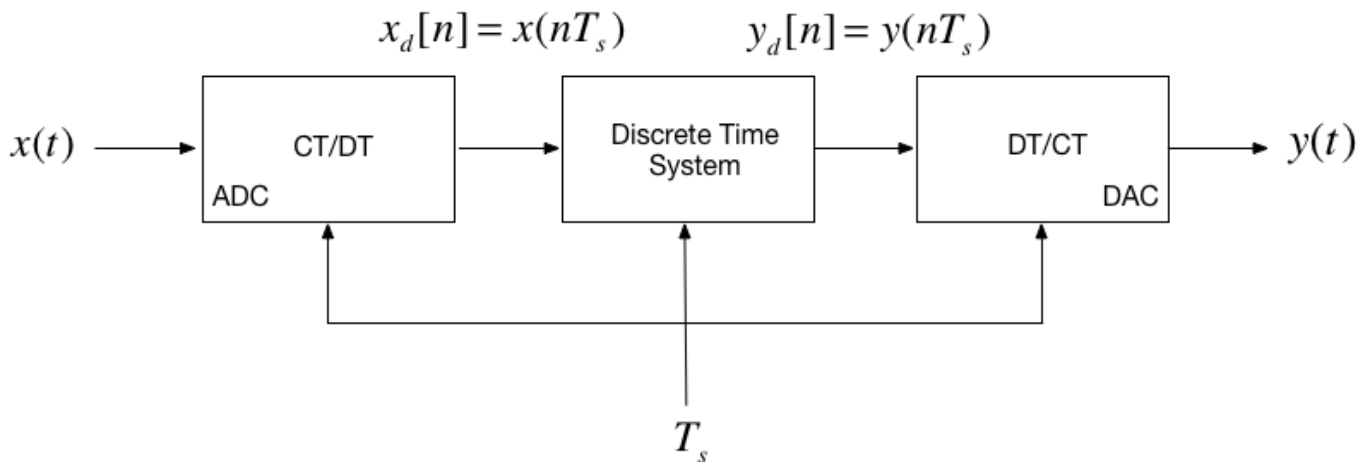
### Agenda

- Discrete Time Systems (Notes)
- Transfer Functions in the Z-Domain (Notes)
- Modelling digital systems in MATLAB/Simulink

- Continuous System Equivalents
- In-class demonstration: Digital Butterworth Filter

## Discrete Time Systems

In the lecture that introduced the z-transform we talked about the representation of a discrete-time (DT) system by the model shown below:



In this session, we want to explore the contents of the central block.

### Example 5

Karris Example 9.10:

The difference equation describing the input-output relationship of a DT system with zero initial conditions, is:

$$y[n] - 0.5y[n - 1] + 0.125y[n - 2] = x[n] + x[n - 1]$$

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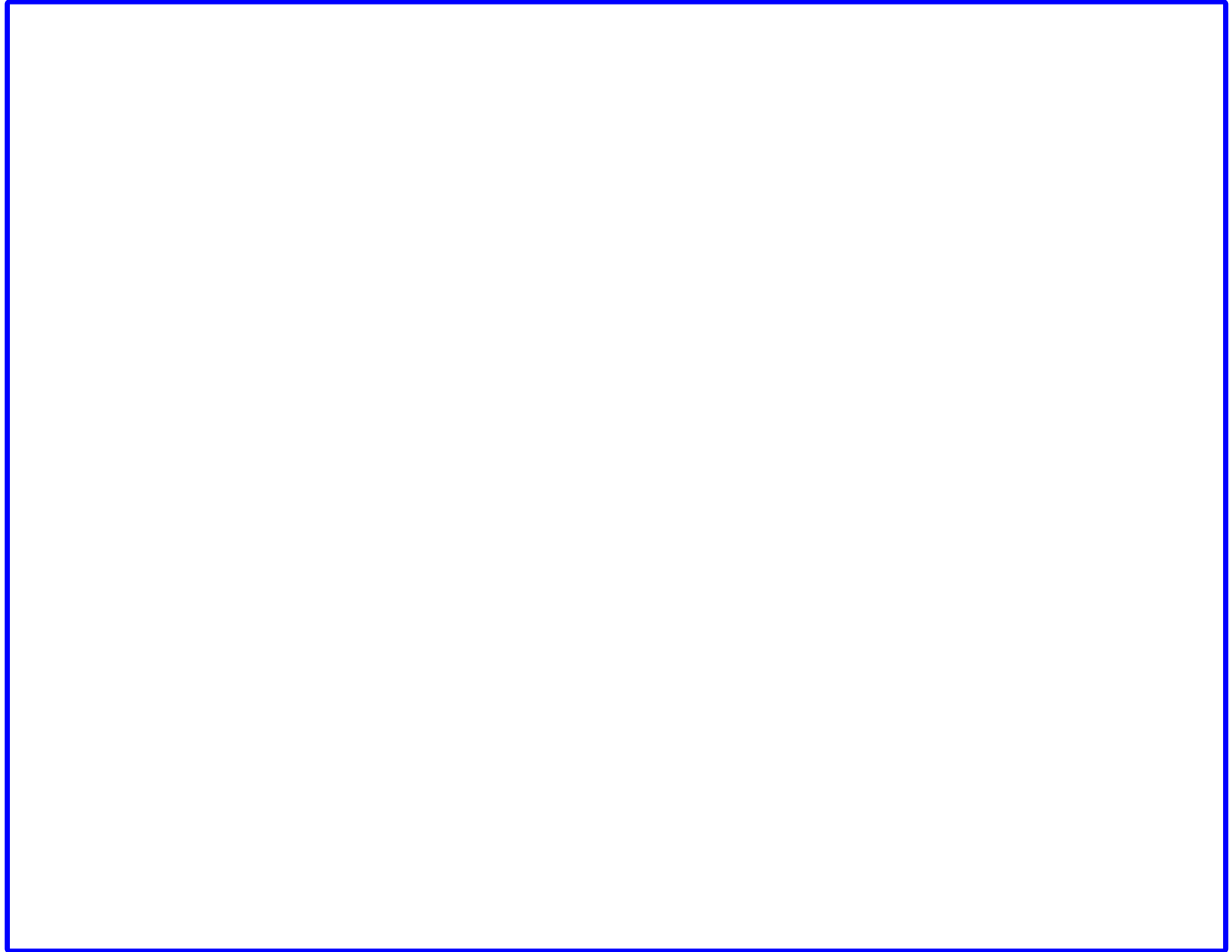
Compute:

1. The transfer function  $H(z)$
2. The DT impulse response  $h[n]$
3. The response  $y[n]$  when the input  $x[n]$  is the DT unit step  $u_0[n]$

## 5.1. The transfer function

$$H(z) = \frac{Y(z)}{U(z)} = \dots ?$$

$$H(z) = \frac{Y(z)}{U(z)} = \dots ?$$



## 5.2. The DT impulse response

Start with:

$$\frac{H(z)}{z} = \frac{z - 1}{z^2 + 0.5z + 0.125}$$

$$\frac{H(z)}{z} = \frac{z - 1}{z^2 + 0.5z + 0.125}$$



## MATLAB Solution

In [ ]:

```
clear all
cd matlab
pwd
format compact
```

See [dtm\\_ex1\\_2.mlx \(matlab/dtm\\_ex1\\_2.mlx\)](#). (Also available as [dtm\\_ex1\\_2.m \(matlab/dtm\\_ex1\\_2.m\)](#).)

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## Transfer function

Numerator  $z + 1$

In [ ]:

```
Nz = [0 1 1];
```

Denominator  $z^2 - 0.5z + 0.125$

In [ ]:

```
Dz = [1 -0.5 0.125];
```

## Poles and residues

In [ ]:

```
[r,p,k] = residue(Nz,Dz)
```

## Impulse Response

In [ ]:

```
Hx = tf(Nz,Dz,1)
hn = impulse(Hx, 15);
```

## Plot the response

In [ ]:

```
stem([0:15], hn)
grid
title('Example 5 - Part 2')
xlabel('n')
ylabel('Impulse response h[n]')
```

## Response as stepwise continuous y(t)

In [ ]:

```
impulse(Hz,15)
grid
title('Example 5 - Part 2 - As Analogue Signal')
xlabel('nTs [s]')
ylabel('Impulse response h(t)')
```

### 5.3. The DT step response

$$Y(z) = H(z)X(z)$$

$$Y(z) = H(z)X(z)$$

$$u_0[n] \Leftrightarrow \frac{z}{z-1}$$

$$u_0[n] \Leftrightarrow \frac{z}{z-1}$$

$$\begin{aligned} Y(z) = H(z)U_0(z) &= \frac{z^2+z}{z^2+0.5z+0.125} \cdot \frac{z}{z-1} \\ &= \frac{z(z^2+z)}{(z^2+0.5z+0.125)(z-1)} \end{aligned}$$

$$\begin{aligned} Y(z) = H(z)U_0(z) &= \frac{z^2+z}{z^2+0.5z+0.125} \cdot \frac{z}{z-1} \\ &= \frac{z(z^2+z)}{(z^2+0.5z+0.125)(z-1)} \end{aligned}$$

$$\frac{Y(z)}{z} = \frac{z^2 + z}{(z^2 + 0.5z + 0.125)(z - 1)}$$

$$\frac{Y(z)}{z} = \frac{z^2 + z}{(z^2 + 0.5z + 0.125)(z - 1)}$$

Solved by inverse Z-transform.



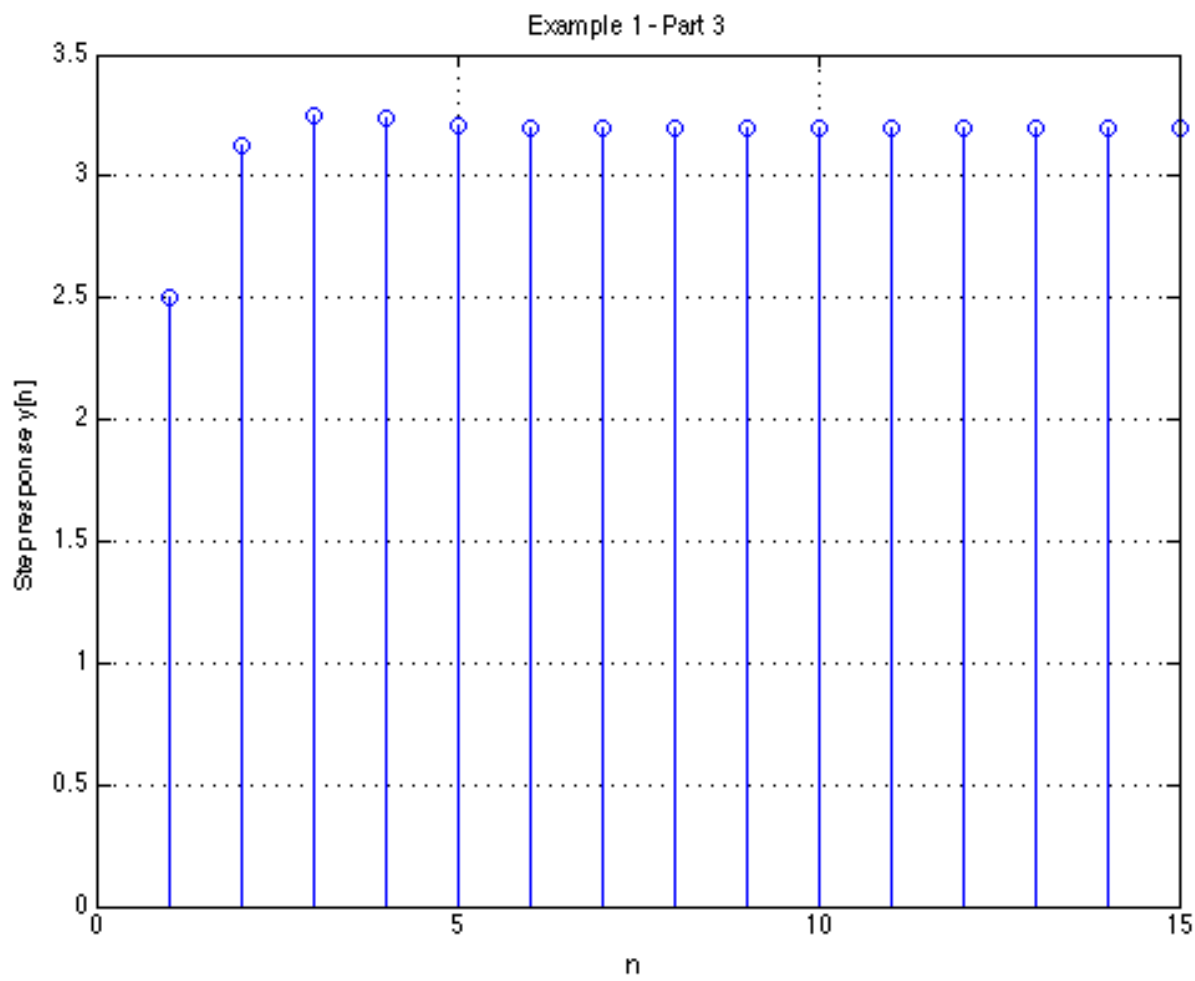
## MATLAB Solution

See [dtm\\_ex1\\_3.mlx \(matlab/dtm\\_ex1\\_3.mlx\)](#). (Also available as [dtm\\_ex1\\_3.m \(matlab/dtm\\_ex1\\_3.m\)](#).)

In [ ]:

```
open dtm_ex1_3
```

# Results



## Modelling DT systems in MATLAB and Simulink

We will consider some examples in class



# MATLAB

Code extracted from [dtm\\_ex1\\_3.m](#) ([matlab/dtm\\_ex1\\_3.m](#)):

In [ ]:

```
Ts = 1;  
z = tf('z', Ts);
```

In [ ]:

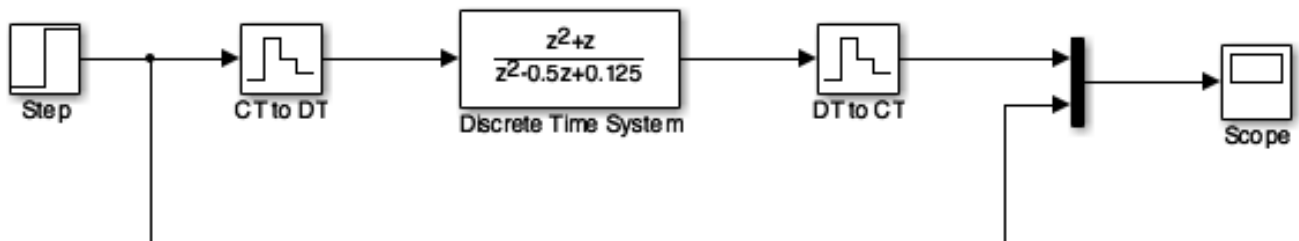
```
Hz = (z^2 + z)/(z^2 - 0.5 * z + 0.125)
```

In [ ]:

```
step(Hz)  
grid  
title('Example 1 - Part 3 - As Analogue Signal')  
xlabel('nTs [s]')  
ylabel('Step response y(t)')  
axis([0,15,0,3.5])
```

## Simulink Model

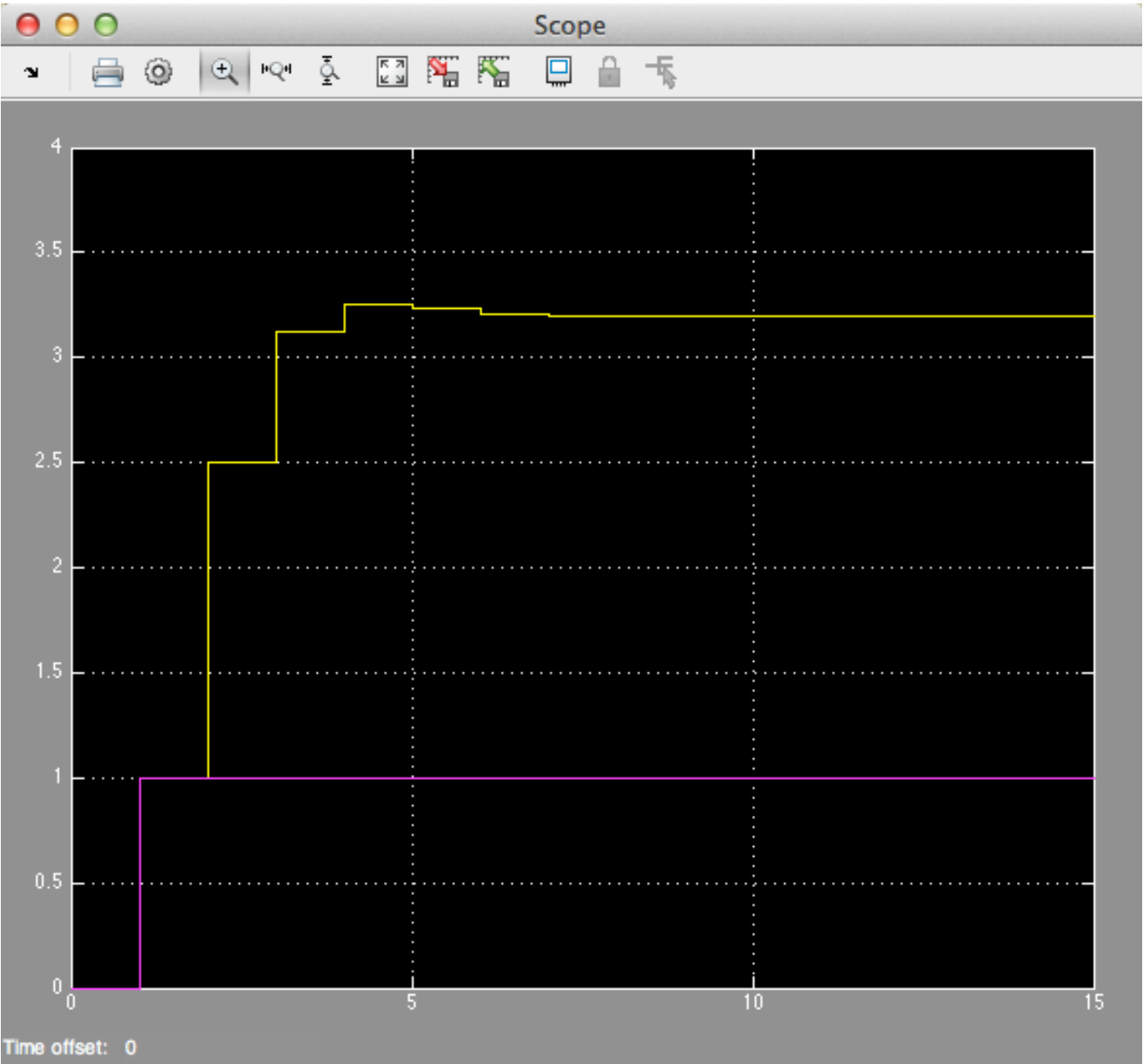
See [dtm.slx](#) ([matlab/dtm.slx](#)):



In [ ]:

```
dtm
```

Results



Converting Continuous Time Systems to Discrete Time Systems

# Continuous System Equivalents

- There is no digital system that uniquely represents a continuous system
- This is because as we are sampling, we only have knowledge of signals being processed at the sampling instants, and need to *reconstruct* the inter-sample behaviour.
- In practice, only a small number of transformations are used.
- The derivation of these is beyond the scope of this module, but here we'll demonstrate the ones that MATLAB provides in a function called `c2d`

## MATLAB c2d function

Let's see what the help function says:

In [ ]:

```
help c2d
```

In [ ]:

```
doc c2d
```

## Example 6

- Design a 2nd-order butterworth low-pass anti-aliasing filter with transfer function  $H(s)H(s)$  for use in sampling music.
- The cut-off frequency  $\omega_c = 20\omega_c = 20$  kHz and the filter should have an attenuation of at least  $-80-80$  dB in the stop band.
- Choose a suitable sampling frequency for the audio signal and give the transfer function  $H(z)H(z)$  and an algorithm to implement  $h[n]h[n]$

## Solution

See [digi\\_butter.mlx \(matlab/digi\\_butter.mlx\)](#).

First determine the cut-off frequency  $\omega_c$

$$\omega_c = 2\pi f_c = 2 \times \pi \times 20 \times 10^3 \text{ rad/s}$$

$$\omega_c = 2\pi f_c = 2 \times \pi \times 20 \times 10^3 \text{ rad/s}$$

In [ ]:

```
wc = 2*pi*20e3
```

$$\omega_c = 125.66 \times 10^3 \text{ rad/s}$$

$$\omega_c = 125.66 \times 10^3 \text{ rad/s}$$

From the lecture on filters, we know the 2nd-order butterworth filter has transfer function:

$$H(s) = \frac{Y(s)}{U(s)} = \frac{\omega_c^2}{s^2 + \omega_c \sqrt{2} s + \omega_c^2}$$

$$H(s) = \frac{Y(s)}{U(s)} = \frac{\omega_c^2}{s^2 + \omega_c \sqrt{2} s + \omega_c^2}$$

Substituting for  $\omega_c = 125.6637 \times 10^3$   $\omega_c = 125.6637 \times 10^3$  this is ...?

In [ ]:

```
Hs = tf(wc^2,[1 wc*sqrt(2), wc^2])
```

$$H(s) = \frac{15.79 \times 10^9}{s^2 + 177.7 \times 10^3 s + 15.79 \times 10^9}$$

$$H(s) = \frac{15.79 \times 10^9}{s^2 + 177.7 \times 10^3 s + 15.79 \times 10^9}$$

## Bode plot

MATLAB:

In [ ]:

```
bode(Hs,{1e4,1e8})  
grid
```

## Sampling Frequency

From the bode diagram, the frequency at which  $|H(j\omega)|$  is  $-80$  dB is approx  $12.6 \times 10^6$  rad/s.

To avoid aliasing, we should choose a sampling frequency twice this = ?

$$\omega_s = 2 \times 12.6 \times 10^6 \text{ rad/s.}$$

In [ ]:

```
ws = 2* 12.6e6
```

So

$$\omega_s = 25.2 \times 10^6 \text{ rad/s.}$$

Sampling frequency ( $f_s$ ) in Hz = ?

$$f_s = \omega_s / (2\pi) \text{ Mhz}$$

$$f_s = \omega_s / (2\pi) \text{ Mhz}$$

In [ ]:

```
fs = ws / (2*pi)
```

$$f_s = 40.11 \text{ Mhz}$$

$$f_s = 40.11 \text{ Mhz}$$

Sampling time  $T_s = ?$

$$T_s = 1/f_s \text{ s}$$

In [ ]:

```
Ts = 1/fs
```

$$T_s = 1/f_s \approx 0.25 \mu s$$

$$T_s = 1/f_s \approx 0.25 \mu s$$

## Digital Butterworth

zero-order-hold equivalent

In [ ]:

```
Hz = c2d(Hs, Ts)
```

## Step response

In [ ]:

```
step(Hz)
```

## Algorithm

From previous result:

$$H(z) = \frac{Y(z)}{U(z)} = \frac{486.6 \times 10^{-6} z + 476.5 \times 10^{-6}}{z^2 - 1.956z + 0.9567}$$

$$H(z) = \frac{Y(z)}{U(z)} = \frac{486.6 \times 10^{-6} z + 476.5 \times 10^{-6}}{z^2 - 1.956z + 0.9567}$$

Dividing top and bottom by  $z^2$  ...

$$H(z) = \frac{Y(z)}{U(z)} = \frac{486.6 \times 10^{-6} z^{-1} + 476.5 \times 10^{-6} z^{-2}}{1 - 1.956z^{-1} + 0.9567z^{-2}}$$

$$H(z) = \frac{Y(z)}{U(z)} = \frac{486.6 \times 10^{-6} z^{-1} + 476.5 \times 10^{-6} z^{-2}}{1 - 1.956z^{-1} + 0.9567z^{-2}}$$

expanding out ...

$$Y(z) - 1.956z^{-1}Y(z) + 0.9567z^{-2}Y(z) = 486.6 \times 10^{-6}z^{-1}U(z) + 476.5 \times 10^{-6}z^{-2}U(z)$$

$$Y(z) - 1.956z^{-1}Y(z) + 0.9567z^{-2}Y(z) = 486.6 \times 10^{-6}z^{-1}U(z) + 476.5 \times 10^{-6}z^{-2}U(z)$$

Inverse z-transform gives ...

$$y[n] - 1.956y[n-1] + 0.9567y[n-2] = 486.6 \times 10^{-6}u[n-1] + 476.5 \times 10^{-6}u[n-2]$$

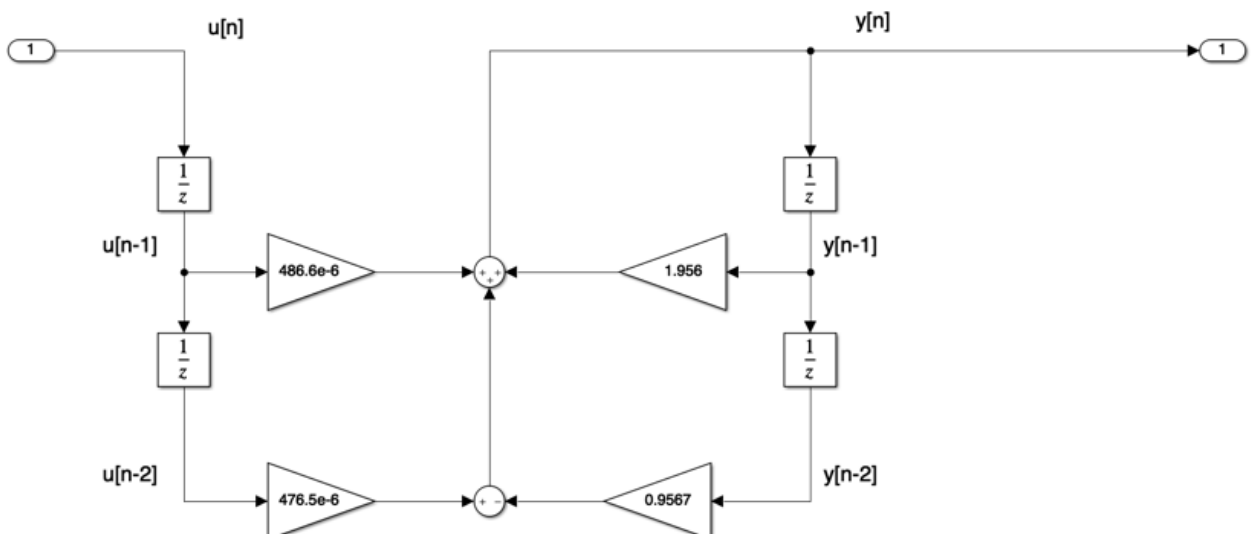
$$y[n] - 1.956y[n-1] + 0.9567y[n-2] = 486.6 \times 10^{-6}u[n-1] + 476.5 \times 10^{-6}u[n-2]$$

in algorithmic form (compute  $y[n]$  from past values of  $u$  and  $y$ ) ...

$$y[n] = 1.956y[n-1] - 0.9567y[n-2] + 486.6 \times 10^{-6}u[n-1] + 476.5 \times 10^{-6}u[n-2]$$

$$y[n] = 1.956y[n-1] - 0.9567y[n-2] + 486.6 \times 10^{-6}u[n-1] + 476.5 \times 10^{-6}u[n-2]$$

## Block Diagram of the digital BW filter



## As Simulink Model

[digifilter.slx \(matlab/digifilter.slx\)](#)

In [ ]:

```
open digifilter
```

## Convert to code

To implement:

$$y[n] = 1.956y[n-1] - 0.9567y[n-2] + 486.6 \times 10^{-6}u[n-1] + 476.5 \times 10^{-6}u[n-2]$$

$$y[n] = 1.956y[n-1] - 0.9567y[n-2] + 486.6 \times 10^{-6}u[n-1] + 476.5 \times 10^{-6}u[n-2]$$

```
/* Initialize */
Ts = 2.4933e-07; /* more probably some fraction of clock speed */
ynm1 = 0; ynm2 = 0; unml = 0; unml2 = 0;
while (true) {
    un = read_adc;
    yn = 1.956*ynm1 - 0.9567*ynm2 + 486.6e-6*unml + 476.5e-6*unml2;
    write_dac(yn);
    /* store past values */
    ynm2 = ynm1; ynm1 = yn;
    unml2 = unml; unml = un;
    wait(Ts);
}
```

## Comments

PC soundcards can sample audio at 44.1 kHz so this implies that the anti-aliasing filter is much sharper than this one as  $f_s/2 = 22.05f_s/2 = 22.05$  kHz.

You might wish to find out what order butterworth filter would be needed to have  $f_c = 20f_c = 20$  kHz and  $f_{\text{stop}}f_{\text{stop}}$  of 22.05 kHz.