

Worksheet 12

Contents

- To accompany Unit 6.1 The Discrete Fourier Transform
- Z-Transform
- The Discrete-time Fourier Transform
- A summary of the important features of sampling and the DFT

To accompany Unit 6.1 The Discrete Fourier Transform

We will step through this worksheet in class.

You are expected to have at least watched the video presentation of [Unit 6.1: The Discrete Fourier Transform](#) of the [notes](#) before coming to class.

If you haven't watch it afterwards!

List of Abbreviations

- CT – Continuous Time
- DT – Discrete Time
- DF - Discrete frequency
- DFT – Discrete (Time) Fourier Transform
- FFT – Fast Fourier Transform

Notation

In the following we shall denote a DT signal as $x[n]$ and its discrete frequency function as $X[m]$.

[Skip to main content](#)

Z-Transform

Recall that

$$F(z) = \mathcal{Z}f[n] = \sum_{n=0}^{\infty} f[n]z^{-n}.$$

The value of this function on the unit circle in the Z-plane will be

$$F(\exp(j\omega T)) = \sum_{n=0}^{\infty} f[n] \exp(-jn\omega T).$$

This is an infinite sum. So to compute it, we need to truncate it.

The Discrete-time Fourier Transform

Let's assume that instead of an infinite number of points, we have N points, equally distributed around the unit circle, then the truncated version will be:

$$X[m] = \sum_{n=0}^{N-1} x[n] \exp\left(-j2\pi \frac{mn}{N}\right)$$

where

$$\omega = \left(\frac{2\pi}{N}\right)nm$$

and $m = 0, 1, 2, \dots, N-1$ and $n = 0, 1, 2, \dots, N-1$.

We refer to the equation

$$X[m] = \sum_{n=0}^{N-1} x[n] \exp\left(-j2\pi \frac{mn}{N}\right)$$

as the N-point Discrete-time Fourier Transform (DFT) of $x[n]$.

The inverse DFT is defined as

[Skip to main content](#)

$$x[n] = \frac{1}{N} \sum_{m=0}^{N-1} X[m] \exp \left(j2\pi \frac{mn}{N} \right)$$

for $n = 0, 1, 2, \dots, N - 1$.

Note the symmetry of the DFT and the Inverse DFT!

In general, the DFT is complex, and thus it can be expressed as

$$X[m] = \Re \{X[m]\} + \Im \{X[m]\}$$

for $m = 0, 1, 2, \dots, N - 1$.

Since

$$\exp \left(-j2\pi \frac{mn}{N} \right) = \cos \left(2\pi \frac{mn}{N} \right) - j \sin \left(2\pi \frac{mn}{N} \right)$$

the DFT can be expressed as

$$X[m] = \sum_{n=0}^{N-1} x[n] \exp \left(-j2\pi \frac{mn}{N} \right) = \sum_{n=0}^{N-1} x[n] \cos \left(2\pi \frac{mn}{N} \right) - j \sum_{n=0}^{N-1} x[n] \sin \left(2\pi \frac{mn}{N} \right)$$

For $n = 0$ this reduces to

$$X[m] = x[0].$$

Then the real part of $X[m]$ is

$$\Re \{X[m]\} = x[0] + \sum_{n=1}^{N-1} x[n] \cos \left(2\pi \frac{mn}{N} \right) \quad \text{for } m = 0, 1, 2, \dots, N - 1$$

and the imaginary part is

$$\Im \{X[m]\} = - \sum_{n=1}^{N-1} x[n] \sin \left(2\pi \frac{mn}{N} \right) \quad \text{for } m = 0, 1, 2, \dots, N - 1.$$

Note that the summations are from 1 to $N - 1$ because $n = 0$ is covered in the real term, and as $x[0]$ is real, it is zero in the corresponding imaginary term.

[Skip to main content](#)

In Class Example 1

A discrete time signal is defined by the sequence

$x[0] = 1, x[1] = 2, x[2] = 2, x[3] = 1$, and $x[n] = 0$ for all other values of n .

Compute the frequency components $X[m]$.

Solution 1

- Compute the N point DFT for $\Re\{X[m]\}$.

- Compute the four point DFT for $\Im\{X[m]\}$.

- Add these together to find $X[m]$.

In Class Example 2

Use the inverse DFT to compute the discrete-time sequence $x[n]$ from $X[m]$.

Solution 2

- Write down the expression $x[n]$ in terms of $X[m]$.

- Compute $x[0]$ from this result.

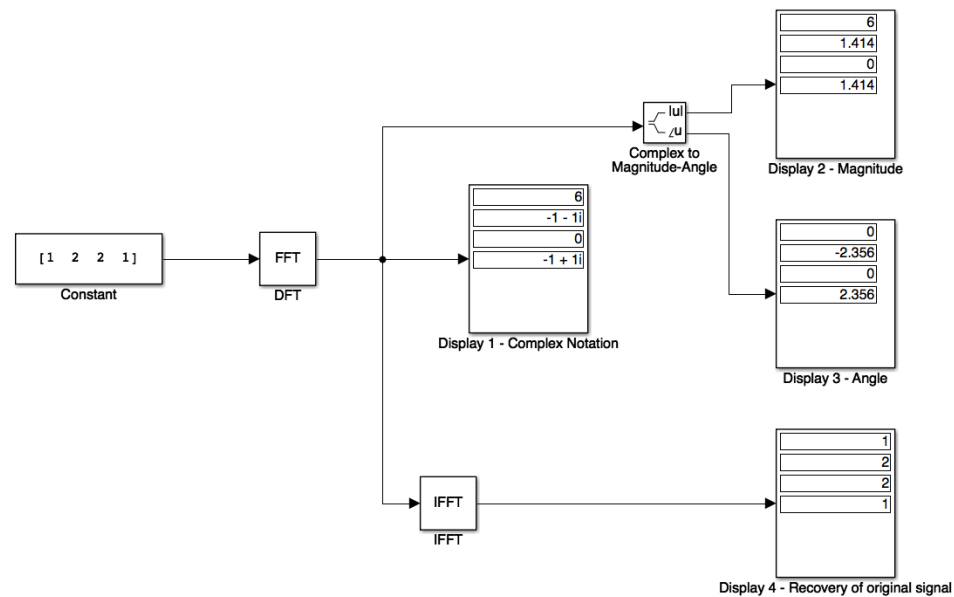
- Repeat for $x[1]$, $x[2]$ and $x[3]$.

Simulink model of the DFT

```
pwd
```

See [dft_ex10_1.slx](#)

```
dft_ex10_1
```



Try inputting your student number.

MATLAB model of the DFT

Karris Example 10.1. To successfully run this script you will need to download the functions [dft.m](#) and [idft.m](#) and make them available on your `MATLABPATH`.

```
xn = [2 0 0 8 9 7 5 0];
```

```
open dft
```

```
Xm = dft(xn,8)
```

```
open idft
```

[Skip to main content](#)

```
xn = idft(Xm,8)
```

A useful compact notation

The term

$$\exp\left(\frac{-j2\pi}{N}\right)$$

is a rotating vector where the range $0 \leq \theta \leq 2\pi$ is divided into N equal segments where N is usually taken to be a power of 2.

It is convenient to represent this as W_N , that is

$$W_N = \exp\left(\frac{-j2\pi}{N}\right)$$

and consequently,

$$W_N^{-1} = \exp\left(\frac{j2\pi}{N}\right).$$

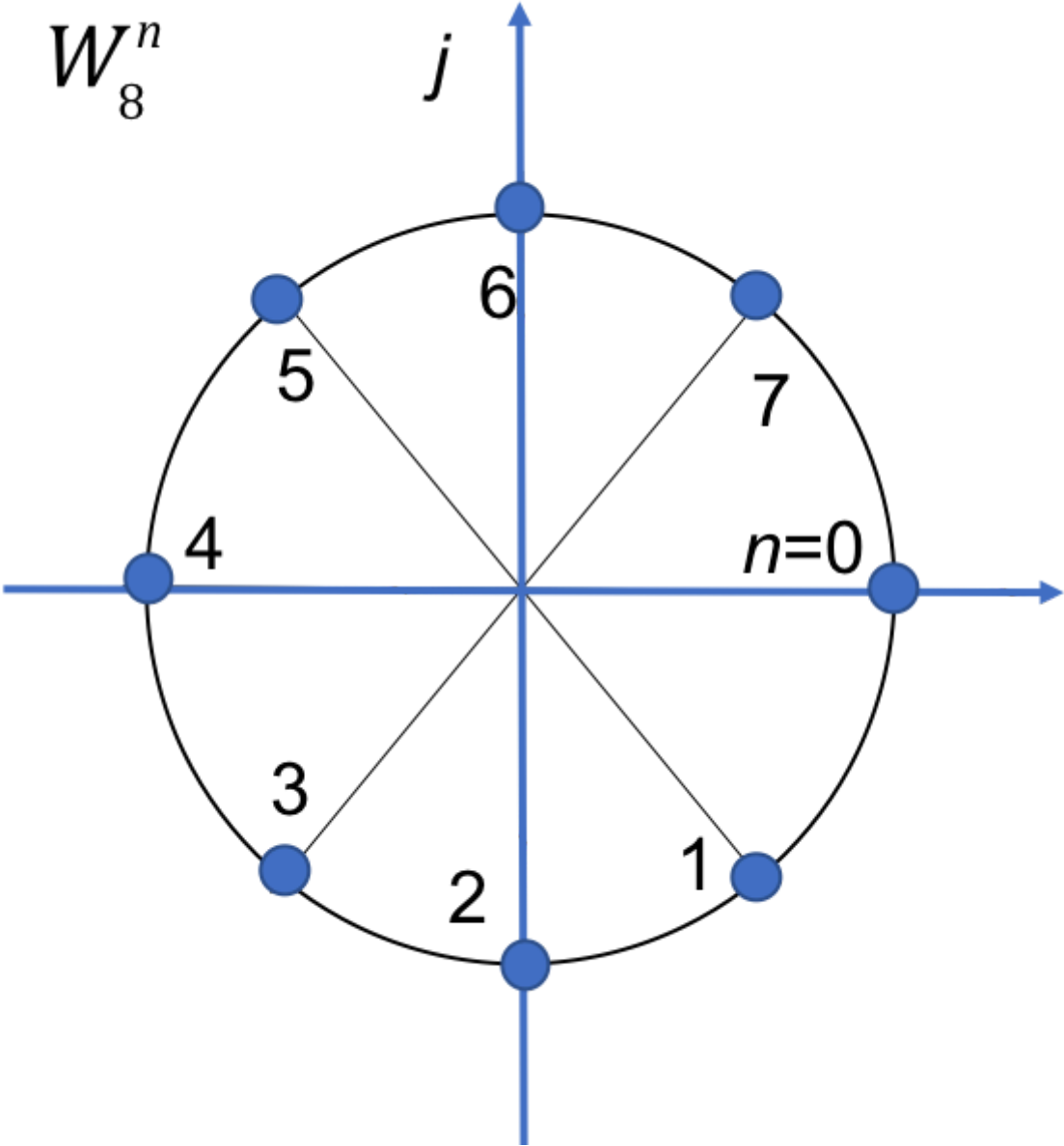
In Class Example 3

Compute the complex numbers represented by the rotating vector W_8

Solution 3

- Rewrite W_8 in exponential form

- Visualize on unit circle



- Complete this table

n	θ	Real	Imaginary	W_8^n
0	0	1	0	1

Using this notation, the DFT and inverse DFT pairs are represented as:

$$X[m] = \sum_{n=0}^{N-1} x[n] W_N^{nm}$$

and

$$x[n] = \frac{1}{N} \sum_{m=0}^{N-1} X[m] W_N^{-nm}$$

MATLAB implementation of DFT

Using the W notation, it is very easy to write a function to implement the DFT.

We will demonstrate this in class.

For example, consider [dft.m](#):

```
function [ Xm ] = dft( xn, N )
% Computes Discrete Fourier Transform
% -----
% [Xm] = dft(xn, N)
% Xm = DFT coeff. array over 0 <= m <= N-1
% xn = N-point finite-duration sequence
% N = length of DFT
%
n = [0:1:N-1];           % row vector for n
m = [0:1:N-1];           % row vector for m
WN = exp(-j*2*pi/N);     % Wn factor
nm = n'*m;               % creates an N by N matrix of nm values
```

[Skip to main content](#)

```

WNnm = WN .^ nm;           % DFT matrix
Xm = xn * WNnm;           % row vector of DFT coefficients

```

Similarly for the inverse DFT [idft.m](#):

```

function [ xn ] = idft( Xm, N )
% Computes Inverse Discrete Fourier Transform
% -----
% [xn] = idft(Xm, N)
% xn = N-point sequence over 0 <= n <= N-1
% Xm = DFT coeff. array over 0 <= m <= N-1
% N = length of DFT
%
n = [0:1:N-1];           % row vector for n
m = [0:1:N-1];           % row vector for m
WN = exp(-j*2*pi/N);      % Wn factor
nm = n'*m;               % creates an N by N matrix of nm values
WNnm = WN .^ (-nm);      % DFT matrix
xn = (Xm * WNnm)/N;      % row vector for IDFT values

```

Notes

In the remainder of these notes, the correspondence between $x[n]$ and $X[m]$ will be written

$$x[n] \Leftrightarrow X[m]$$

In Example 2, we found that, although the DT sequence $x[n]$ was real, the discrete frequency (DF) sequence was complex. However, in most applications we are interested in the magnitude and phase of the DF, that is

$$|X[m]|$$

and

$$\angle X[m].$$

In Class Example 4

Use MATLAB to compute the magnitude of the frequency components of the following DT

[Skip to main content](#)

n	0	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15
$x[n]$	1.0	2	2	2.5	1.5	0.5	-0.5	-1.5	-2.5	-0.5	0.25	1.25	2	1.5	1	0.5

We will compute this in class and make some comments afterwards.

```
xn = [ 1, 2, 3, 2.5, 1.5, 0.5, ...
      -0.5, -1.5, -2.5, -0.5, ...
      0.25, 1.25, 2, 1.5, 1, 0.5];
```

```
stem([0:15],xn),xlabel('n'),ylabel('x[n]'),title('Discrete Time Sequence')
```

```
Xm = dft(xn,16);
```

```
stem([0:15],abs(Xm)),xlabel('m'),ylabel('|X[m]|'),title('Discrete Frequency')
```

Points to note:

- $X[0] = 12$ is the DC component of the DT sequence.
- After the $|X[8]| = 1.4872$ term, the magnitude of the frequency values for the range $9 \leq m \leq 15$ are the mirror image of the values for the range $0 \leq m \leq 7$.
- This is not a coincidence, in fact if $x[n]$ is an N -point *real discrete-time function*, only $N/2$ of the *frequency components* of $|X[m]|$ are unique.

A summary of the important features of sampling and the DFT

- N is the number of samples in frequency.
- f_s sampling frequency, samples per second.
- T_t period of a periodic DT function.
- t_s interval between the N samples in time period T_t .
- f_f period of a periodic DF function.
- F_s interval between the N samples in frequency period T_f .

The relationships between these quantities are:

[Skip to main content](#)

$$t_t = \frac{T_t}{N}$$

$$f_s = \frac{1}{t_t}$$

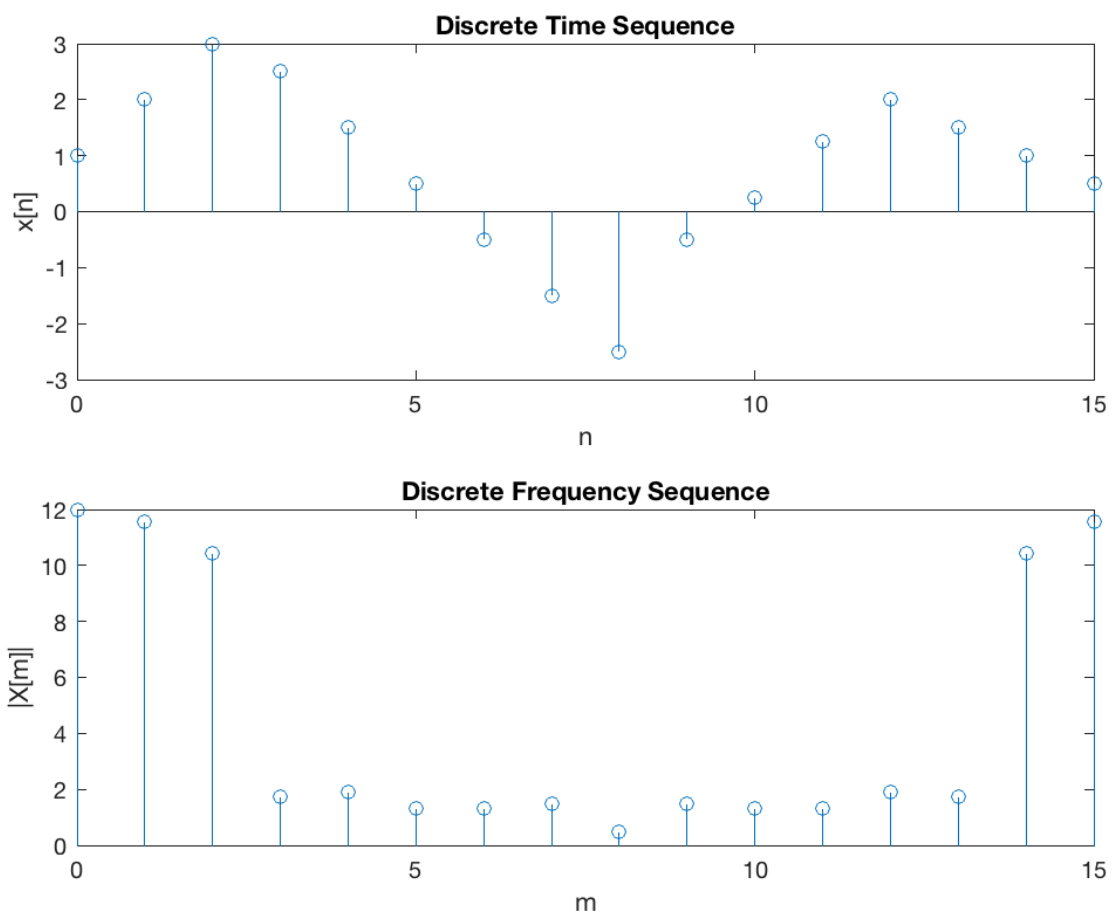
$$t_f = \frac{T_f}{N}$$

$$t_t = \frac{1}{T_f}$$

$$f_f = \frac{1}{T_t}$$

We will add these quantities to the results of Example 4 in class.

Example 4 (continued)



[Skip to main content](#)

To reproduce this plot use [repeat.m](#).

In Class Example 5

The period of a periodic DT function is 0.125 ms and it is sampled at 1024 equally spaced points. It is assumed that with this number of samples, the sampling theorem is satisfied and thus there will be no aliasing.

1. Compute the interval t_t between samples for the periodic signal
2. Compute the period T_f of the frequency spectrum in kHz
3. Compute the interval t_f between frequency components in kHz
4. Compute the sampling frequency f_s .
5. Compute the Nyquist frequency f_n .

Solution

To be done in class.

- Compute the interval t_t between samples for the periodic signal

- Compute the period of the frequency spectrum T_f in kHz

- Compute the sampling frequency f_s .

- Compute the Nyquist frequency f_n .