Fourier transforms of commonly occurring signals

Colophon

An annotatable worksheet for this presentation is available as Worksheet 13.

- You can view the notes for this presentation as a webpage (HTML). This page is downloadable as a PDF file.

If you have been reading both Karris and Boulet you may have noticed a difference in the notation used in the definition of Fourier Transform:

• Karris uses $F(\omega)$

- **Note on Notation**

• Boulet uses $F(i\omega)$ I checked other sources and Hsu (Schaum's Signals and Systems) and Morrell (The Fourier Analysis Video Series on YouTube)

According to Wikipedia Fourier Transform: Other Notations both are used only by electronic engineers anyway and either would be acceptible.

There is some advantage in using Boulet's notation $F(j\omega)$ in that it helps to reinforce the idea that Fourier Transform is a

special case of the Laplace Transform and it was the notation that I used in the last section. In these notes, I've used the other convention on the basis that its the more likely to be seen in your support materials.

However, I am happy to change back if you find the addition of j useful. You should be aware that Fourier Transforms are in general complex so whatever the notation used to represent the transform,

we are still dealing with real and imaginary parts or magnitudes and phases when we use the actual transforms in analysis.

Agenda

 Tables of Transform Pairs Examples of Selected Transforms

frequency $F(\omega)$:

The Fourier Transform

Reminder of the Definitions Last time we derived the Fourier Transform by evaluating what would happen when a periodic signal was made periodic. Let us

Relationship between Laplace and Fourier

• Fourier Transforms of Common Signals

restate the definitions.

The Inverse Fourier Transform In the signals and systems context, the *Inverse Fourier Transform* is used to convert a function of frequency $F(\omega)$ to a function of time f(t):

 $\mathcal{F}^{-1}\left\{F(\omega)\right\} = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(\omega)e^{j\omega t} d\omega = f(t).$

In the signals and systems context, the Fourier Transform is used to convert a function of time f(t) to a function of radian

 $\mathcal{F}\left\{f(t)\right\} = \int_{-\infty}^{\infty} f(t)e^{-j\omega t} dt = F(\omega).$

Duality of the transform Note the similarity of the Fourier and its Inverse.

Remarks

a > 0

1 Constant energy at all frequencies.

also known as sign function

 $e^{j\omega t_0}$

 $2\pi\delta(\omega-\omega_0)$

<u>Transfom—WolframMathworld</u> for more complete references. $F(\omega)$ f(t)Name

Dirac delta

Time sample

Unit step

Single pole

Note, the factor 2π is introduced because we are changing units from radians/second to seconds.

This has important consequences in filter design and later when we consider sampled data systems.

3. Phase shift 4. Signum

2

5.

8.

 $\frac{1}{j\omega} + \pi\delta(\omega)$ $\pi \left[\delta(\omega - \omega_0) + \delta(\omega + \omega_0) \right]$ 6. Cosine $\cos \omega_0 t$ 7. $\sin \omega_0 t - j\pi \left[\delta(\omega - \omega_0) - \delta(\omega + \omega_0)\right]$ Sine

 $e^{-at}u_0(t)$

 $u_0(t)$

 $\delta(t)$

 $e^{j\omega t_0}$

sgn(x)

 $\delta(t-t_0)$

	0.	Single pole	$c u_0(i)$	$j\omega + a$	
	9.	Double pole	$te^{-at}u_0(t)$	$\frac{1}{(j\omega+a)^2}$	a > 0
	10.	Complex pole (cosine component)	$e^{-at}\cos\omega_0 t\ u_0(t)$	$\frac{j\omega + a}{((j\omega + a)^2 + \omega^2)}$	a > 0
	11.	Complex pole (sine component)	$e^{-at} \sin \omega_0 t \ u_0(t)$	$\frac{\omega}{((j\omega+a)^2+\omega^2)}$	a > 0
	12.	Gating function ($aka \ rect(T)$)	$A\left[u_0(t+T)-u_0(t-T)\right]$	$2AT\frac{\sin \omega T}{\omega T}$	
Some Selected Fourier Transforms					
In most of the work we will do in this course, and in practice, the <i>signals</i> that we use with the Fourier transform will be a real continuous aperiodic functions of time that are zero when $t = 0$.					
The Fourier transforms of such signals will be complex continous functions of frequency which have real and imaginary parts and will exist at both positive and negative values of ω .					
It is often most convenient to deal with the transformed "spectrum" by considering the magnitude and phase and we will therefore often plot $F(\omega)$ on two separate graphs as magnitude $ F(\omega) $ and phase $\angle F(\omega)$ (where phase is measured in radians) plotted against frequency $\omega \in [-\infty, \infty]$ (in radians/second).					
We most often represent the system by its so-called frequency response and we will be interested on what effect the system					

The Dirac Delta

Have these ideas in mind as we go through the examples in the rest of this section.

f(t)

Proof: uses sampling and sifting properties of $\delta(t)$.

has on the signal f(t).

transforming the result.

 $\delta(t)$

 $\delta(t) \Leftrightarrow 1$

 $F(\omega)$

 $F(\omega)$

 $2\pi\delta(\omega)$

As for the Laplace transform, this is more conveniently determined by exploiting the time convolution property. That is by

performing a Fourier transform of the signal, multiplying it by the system's frequency response and then inverse Fourier

syms t; fourier(dirac(t)) ans =

 $\delta(t-t_0) \Leftrightarrow e^{-j\omega t_0}$

 $1 \Leftrightarrow 2\pi\delta(\omega)$

f(t)

ans =

Sinewave

Signum (Sign)

The transform is:

fourier(A,omega)

2*pi*dirac(omega)

Related by frequency shifting property:

Note: f(t) is real and even. $F(\omega)$ is also real and even.

Matlab:

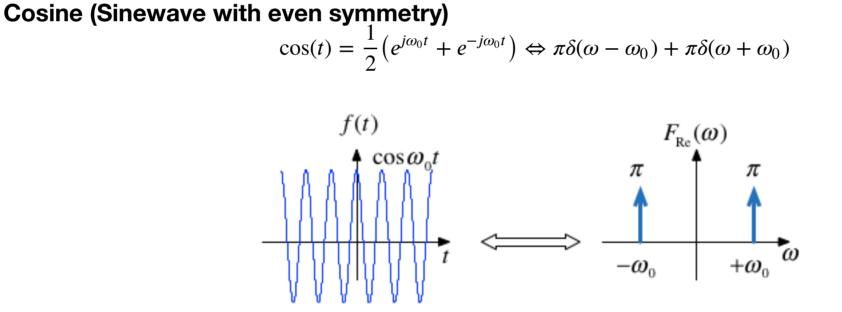
1

Related:

DC

In [1]:

Matlab: syms t omega; In [2]: A = sym(1);



 $\sin(t) = \frac{1}{j2} \left(e^{j\omega_0 t} - e^{-j\omega_0 t} \right) \Leftrightarrow -j\pi \delta(\omega - \omega_0) + j\pi \delta(\omega + \omega_0)$

 $\operatorname{sgn} x = \begin{cases} -1 & x < 0 \\ 0 & x = 0 \end{cases}$

 $\mathcal{F}\left\{u_0(t)\right\} = \pi\delta(\omega) + \frac{1}{j\omega}$

 $\operatorname{sgn} t = 2u_0(t) - 1$

 $\operatorname{sgn} x = 2u_0(t) - 1$

 $u_0(t) = \frac{1}{2} \big[1 + \operatorname{sgn} x \big]$

 $u_0(t) \Leftrightarrow \pi \delta(\omega) + \frac{1}{i\omega}$

 $F_{\rm Im}(\omega)$

 ω

 $e^{j\omega_0 t} \Leftrightarrow 2\pi\delta(\omega-\omega_0)$

 $\operatorname{sgn} x = u_0(t) - u_0(-t) = \frac{2}{j\omega}$ f(t)

Note: f(t) is real and odd. $F(\omega)$ is imaginary and odd.

The signum function is a function whose value is equal to

This function is often used to model a *voltage comparitor* in circuits. **Example 4: Unit Step** Use the signum function to show that

Clue

Define

Does that help?

Proof

so

QED

Graph of unit step

Example 5

Example 6

Example 7

substituting s by $j\omega$.

Solution to example 8

Solution to example 9

rectangular pulse

periodic time function

triangular pulse

for.

Boulet gives the graph of this function.

Given that

Compute

Example 8: Single Pole Filter

Boulet gives the graph of this function.

Example 9: Complex Pole Pair cos term

Use the results derived so far to show that

Hint: linearity plus frequency shift property.

Use the results derived so far to show that

Hint: Euler's formula plus solution to example 2.

See <u>worked solution</u> for the corrected proof.

f(t)

From previous results $1 \Leftrightarrow 2\pi\delta(\omega)$ and $\operatorname{sgn} x = 2/(j\omega)$ so by linearity

 $F_{\rm Im}(\omega)=1/(j\omega)$. The real part is even, and theimaginary part is odd.

 $u_0(t) \Leftrightarrow \pi \delta(\omega) + \frac{1}{i\omega}$ f(t)

Unit step is neither even nor odd so the Fourier transform is complex with real part $F_{\rm Re}(\omega)=\pi\delta(\omega)$ and imaginary part

 $e^{j\omega_0 t}u_0(t) \Leftrightarrow \pi\delta(\omega-\omega_0) + \frac{1}{j(\omega-\omega_0)}$

 $\sin \omega_0 t \ u_0(t) \Leftrightarrow \frac{\pi}{j2} [\delta(\omega - \omega_0) - \delta(\omega + \omega_0)] + \frac{\omega_0}{\omega_0^2 - \omega^2}$

Use the result of Example 6 to determine the Fourier transform of $\cos \omega_0 t \ u_0(t)$. Solution to example 7 $\cos \omega_0 t \ u_0(t) \Leftrightarrow \frac{\pi}{2} [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)] + \frac{j\omega}{\omega_0^2 - \omega^2}$

Important note: the equivalent example in Karris (Section 8.4.9 Eq. 8.75 pp 8-23-8-24) is wrong!

Derivation of the Fourier Transform from the Laplace Transform

If a signal is a function of time f(t) which is zero for $t \leq 0$, we can obtain the Fourier transform from the Lpalace transform by

 $\mathcal{L}\left\{e^{-at}u_0(t)\right\} = \frac{1}{s+a}$

 $\mathcal{F}\left\{e^{-at}u_0(t)\right\}$

 $\frac{j\omega + a}{(j\omega + a)^2 + \omega_0^2}$

We shall conclude this session by computing as many of the the Fourier transform of some common signals as we have time

Given that $\mathcal{L}\left\{e^{-at}\cos\omega_0 t\ u_0(t)\right\} = \frac{s+a}{(s+a)^2 + \omega_0^2}$ Compute $\mathcal{F}\left\{e^{-at}\cos\omega_0t\ u_0(t)\right\}$

 unit impulse train (model of regular sampling) I will not provide notes for these, but you will find more details in Chapter 8 of Karris and Chapter 5 of Boulet and

Fourier Transforms of Common Signals

both use the $F(\omega)$ notation.

• The source code for this page is <u>content/fourier_transform/2/ft2.ipynb</u>.

Summary Tables of Transform Pairs • Examples of Selected Transforms Relationship between Laplace and Fourier

Particular highlights which we will not have time to cover: • The Duality of the Fourier transform (pp 191–192).

 Fourier Transforms of Common Signals **Next Section** The Fourier Transform for Systems and Circuit Analysis

Suggestions for Further Reading Boulet has several interesting amplifications of the material presented by Karris. You would be well advised to read these. • Time multiplication and its relation to amplitude modulation (pp 182-183). • Fourier transform of the complex exponential signal $e^{(\alpha+j\beta)t}$ with graphs (pp 184–187).

I have created some worked examples (see Blackboard and the OneNote notebook) to help with revision.

• Use of inverse Fourier series to determine f(t) from a given $F(j\omega)$ and the "ideal" low-pass filter (pp 188–191).