# **Answers to Frequently Asked Questions**

A PDF version of this FAQ is FAQ.pdf.

### **Acknowledgements**

I thank the following students who helped me to recognise the holes in my course notes and were the inspiration for these answers.

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#### **Harmonic frequencies**

Fundamental frequency -- A periodic signal  $f(t) = f(t + nT), n \in \mathbb{Z}$  has period T s and a fundamental frequency  $f_0 = 1/T$ Hz. When used in Fourier series and Fourier transforms, frequencies are expressed as  $\omega$  in radians/second. The **fundamental frequency** is  $\omega = \Omega_0 = 2\pi f_0$  or, equivalently,  $\Omega_0 = 2\pi/T$  rad/s.

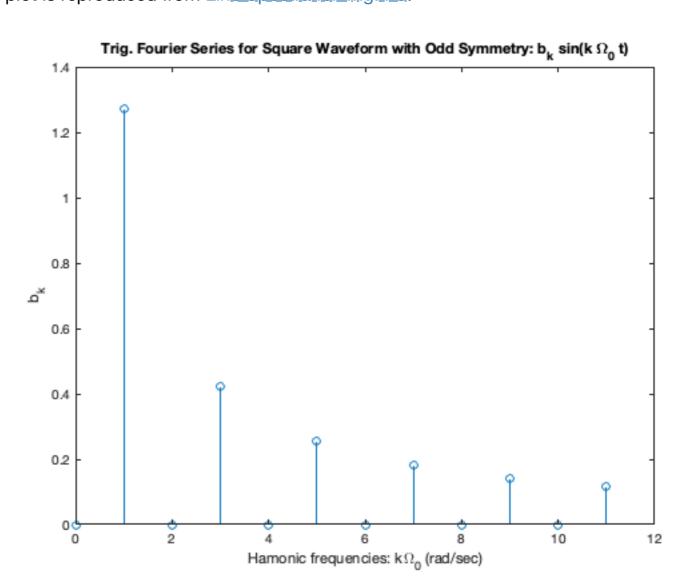
Harmonic frquencies (or Harmonics) are simply integer multiples of the fundamental frequency  $\Omega_0$ . So the zero-th harmonic is  $\Omega_0=0$  rad/s or DC. The first harmonic is  $1.\Omega_0=\Omega_0$ , the second is  $2\Omega_0$ , the third  $3\Omega_0$  etc.

In general, we can express the k-th harmonic as  $k\Omega_0$ ,  $k \in \mathbb{Z}$ .

#### Line spectra

In trig. Fourier series, the coefficients  $a_k$  and  $b_k$  are the amplitudes of the  $\cos(k\Omega_0 t)$  and  $\sin(k\Omega_0 t)$  terms respectively. We usually show these terms as lines (or spectra) with height  $a_k$  and/or  $b_k$  plotted against the harmonic frequency index k.

An example of such a plot is reproduced from Line Spectra for Trig. FS.

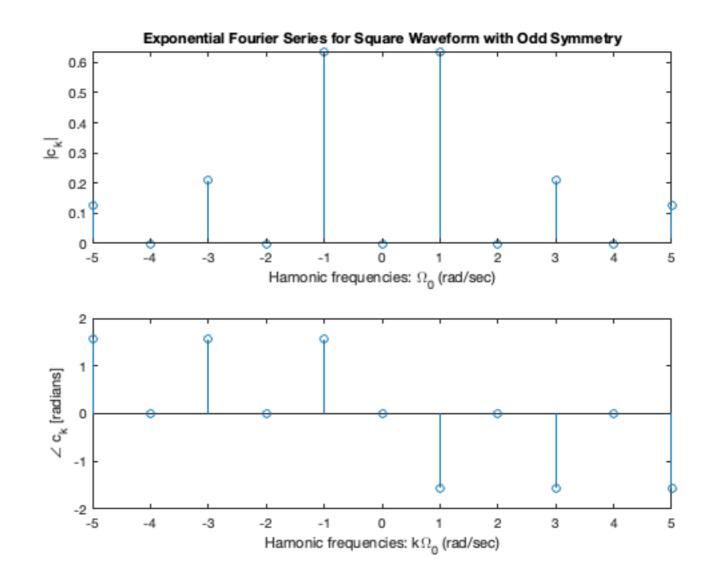


For exponential Fourier Series, the harmonic terms are defined as

$$C_k \exp(jk\Omega_0 t) k = -\infty, \dots, -2, -1, 0, 1, 2, \dots, \infty.$$

And  $|C_k|$  and  $\angle C_k$  are plotted as the line spectra against k. Note that  $C_{-k}$  and  $C_k$  are complex conjugates  $\forall k > 0$ . Hence, the spectrum will be symmetric around k = 0.

An example of such a plot is reproduced from Line Spectra for Exp. FS.



## Filter attenuation

In the notes Steady-State Response of an LTI System to a Periodic Signal we state (without proof) that the output of an LTI system to a periodic function with period T represented by a Fourier series is given by:

$$y(t) = \sum_{k=-\infty}^{\infty} C_k H(jk\Omega_0) e^{jk\Omega_0 t}.$$

As a consequence

Thus y(t) is a Fourier series itself with coefficients  $D_k$ :

$$D_k = C_k H(jk\Omega_0).$$

What is missing from this analogies is a discussion of what  $H(jk\Omega_0)$  looks like.

As an example, consider the simple first-order Butterworth low-pass (LP) filter with cut-off frequency  $\omega_c$ :

$$H(s) = \frac{\omega_c}{s + \omega_c}$$

For this filter

$$H(j\omega) = \frac{\omega_c}{j\omega + \omega_c}.$$

Let us say that we wish to compute the attenuation and phase of this filter at  $\omega=\Omega_0$  .

harmonic terms are changed (attenuated in magnitude and shifted in phase) by the filter.

To compute the *magnitude*:

$$|H(j\Omega_0)| = \left| \frac{\omega_c}{j\Omega_0 + \omega_c} \right|$$
$$= \frac{\omega_c}{\sqrt{\Omega_0^2 + \omega_c^2}}$$

We note that is  $|H(j\Omega_0)| < 1$  so the filter will attenuate the incoming harmonic frequency. This will be true for all harmonics, so in general, for a LP filter:  $D_k = C_k |H(jk\Omega_0)| < C_k$ .

The phase will be given by

Phases are additive so

$$\phi = \angle H(j\omega) = \tan^{-1} \left( \frac{\Im(H(j\omega))}{\Re(H(j(\omega)))} \right)$$

where

$$H(jk\Omega_0) = \frac{\omega_c^2}{(k\Omega_0)^2 + \omega_c^2} - j \frac{k\Omega_0 \omega_c}{(k\Omega_0)^2 + \omega_c^2}$$
$$\phi_k = \tan^{-1} \left( -\frac{K\Omega_0 \omega_c}{\omega_c^2} \right)$$
$$= \tan^{-1} \left( -\frac{k\Omega_0}{\omega_c} \right)$$

$$= \tan^{-1} \left( -\frac{k\Omega_0}{\omega_c} \right)$$

By doing such analysis, we can examine the effect of a filter on a periodic signal, just by considering how the coefficients of the

 $\angle D_k = \angle C_k + \phi_k$ .