Worksheet 12

To accompany Chapter 5.1 Defining the Fourier Transform ¶

Colophon

This worksheet can be downloaded as a PDF file. We will step through this worksheet in class.

An annotatable copy of the notes for this presentation will be distributed before the second class meeting as Worksheet 12 in the Week 6: Classroom Activities section of the Canvas site. I will also distribute a copy to your personal Worksheets section of the OneNote Class Notebook so that you can add your own notes using OneNote.

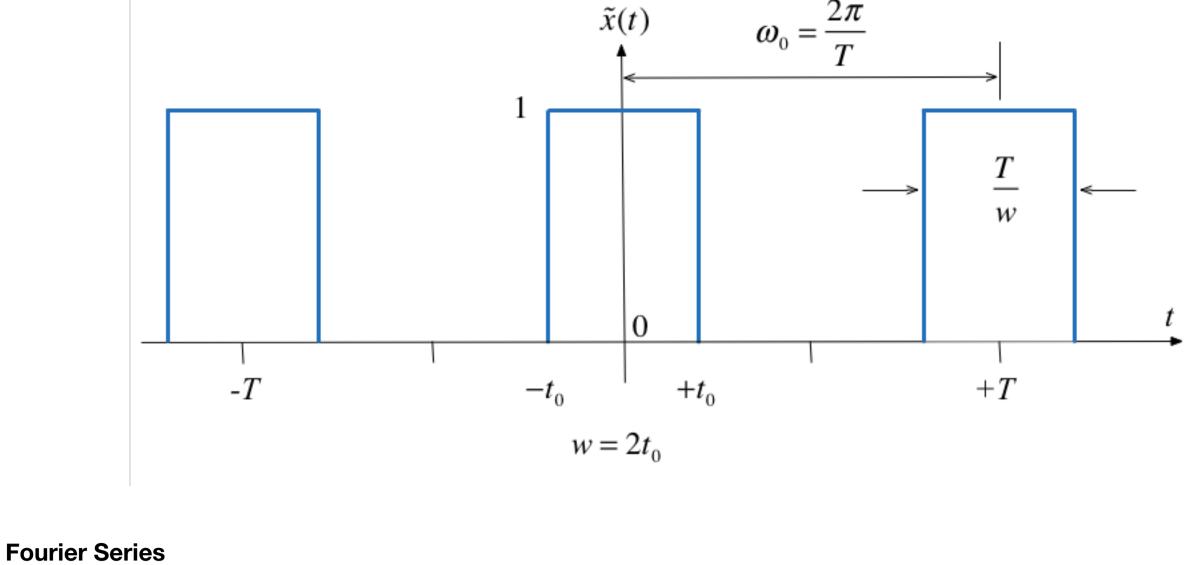
You are expected to have at least watched the video presentation of Chapter 5.1 of the notes before coming to class. If you haven't watch it afterwards! After class, the lecture recording and the annotated version of the worksheets will be made available through Canvas.

Fourier Transform as the Limit of a Fourier Series We start by considering the pulse train that we used in the last lecture and demonstrate that the discrete line spectra for the Fourier Series becomes a

continuous spectrum as the signal becomes aperiodic.

Let $\tilde{x}(t)$ be the Fourier series of the rectangular pulse train shown below:

This analysis is from Boulet pp 142-144 and 176-180.



to compute the line spectra.

From the Time Point of View

In the previous section we used

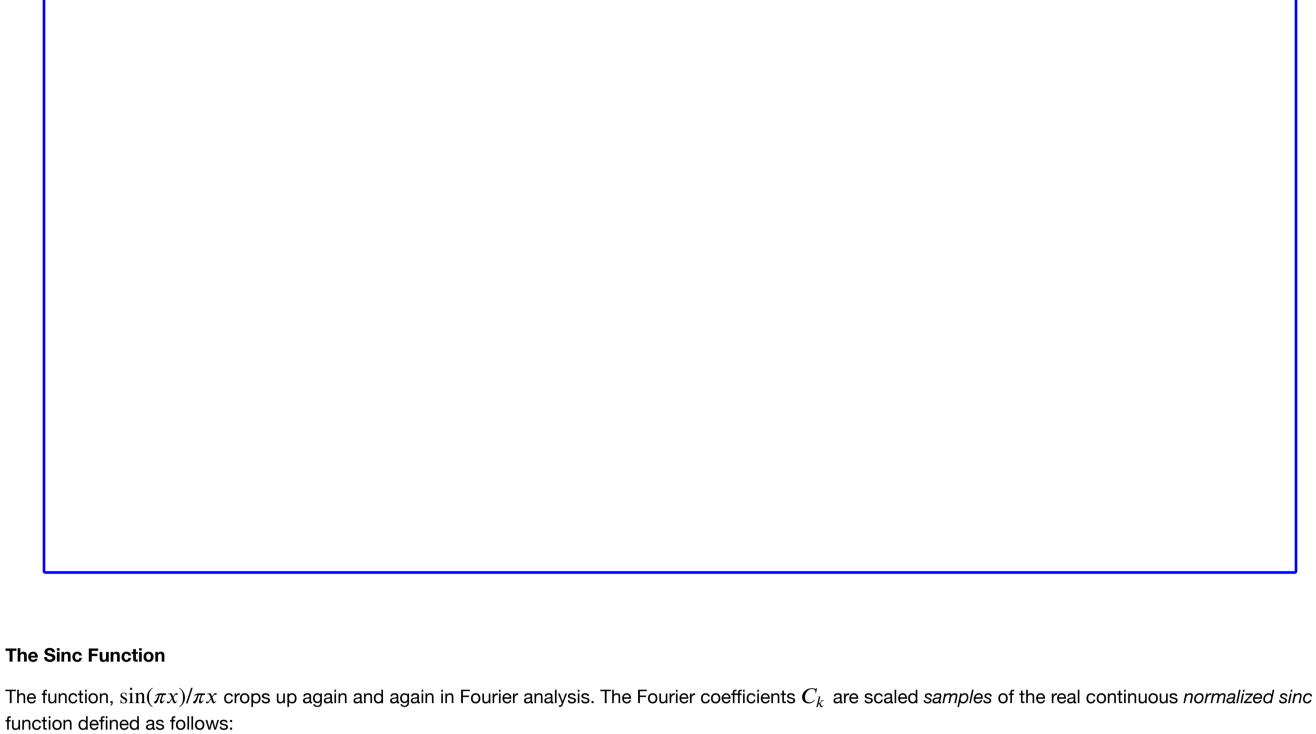
 $C_k = \frac{1}{2\pi} \int_{-\pi/w}^{\pi/w} A e^{-jk(\Omega_0 t)} d(\Omega_0 t) = \frac{A}{2\pi} \int_{-\pi/w}^{\pi/w} e^{-jk(\Omega_0 t)} d(\Omega_0 t)$

If we instead take a time point-of-view and let A=1

Let's complete the analysis in the whiteboard.

$$d \, \text{let} \, A = 1$$

 $C_k = \frac{1}{T} \int_{-t_0}^{t_0} e^{-jk\Omega_0 t} dt.$



Plot the sinc function Plots: $sinc(u) = \frac{\sin \pi u}{\pi u}, \ u \in \mathbb{R}$

 $\operatorname{sinc} u := \frac{\sin \pi u}{\pi u}, \ u \in \mathbb{R}.$

imatlab_export_fig('print-svg') % Static svg figures. cd ../matlab format compact

The function is equal to 1 at $u=0^1$ and has zero crossings at $u=\pm n,\ n=1,2,3,\ldots$ as shown below.

In []: x = linspace(-5, 5, 1000);

In []: clear all

plot(x,sin(pi.*x)./(pi.*x))grid

ylabel('sinc(u)')

xlabel('u')

Duty cycle

increase T:

In []: open duty_cycle

• 50% ($\eta = 1/2$)

title('Graph of sinc function')

• The duty cycle is often given as a percentage.

• We define the duty cycle $\eta = 2t_0/T$ of the rectangular pulse train as the fraction of the time the signal is "on" (equal to 1) over one period.

 $C_k = \frac{2t_0}{T} \frac{\sin\left(\frac{\pi k 2t_0}{T}\right)}{\frac{\pi k 2t_0}{T}} = \frac{2t_0}{T} \operatorname{sinc}\left(\frac{k 2t_0}{T}\right)$ $C_k = \eta \operatorname{sinc}(k\eta)$

The spectral coefficients expressed using the normalized sinc function and the duty cycle can be written as

$$TC_k = T\eta \operatorname{sinc}(k\eta) = 2t_0\operatorname{sinc}\left(k\frac{2t_0}{T}\right)$$
 Then the normalized coefficents TC_k of the rectangular wave is a sinc envelope with constant amplitude at the origin equal to $2t_0$, and a zero crossing at fixed

Let us normalize the spectral coefficients of $\tilde{x}(t)$ by mutiplying them by T, and assume t_0 is fixed so that the duty cycle $\eta = 2t_0/T$ will decrease as we

Demo

frequency π/t_0 rad/s, both independent of T.

Normalize the spectral coefficients

Run duty_cycle with values of:

• 25% ($\eta = ?$)

• 12.5% ($\eta = ?$) • 5% ($\eta = ?$)

- **Comments** • As the fundamental period increases, we get more spectral lines packed into the lobes of the sinc envelope. • These normalized spectral coefficients turn out to be samples of the continuous sinc function on the spectrum of $\tilde{x}(t)$
- The zero-crossing points of sinc envelope are independent of the period T. They only depend on t_0 .

Intuition leading to the Fourier Transform

• The envelope is this function.

Inverse Fourier Transform:

• An aperiodic signal that has been made periodic by "repeating" its graph every T seconds will have a line spectrum that becomes more and more dense as the fundamental period is made longer and longer. • The line spectrum has the same continuous envelope. • As T goes to infinity, the line spectrum will become a continuous function of ω .

Doing the Maths See the <u>notes</u>.

 $X(j\omega) = \int_{-\infty}^{\infty} x(t)e^{-j\omega t} dt := \mathcal{F}\left\{x(t)\right\}.$

• The two spectra are plotted against the frequency variable $k\omega_0$ with units of rad/s rather than index of harmonic component

 $x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega := \mathcal{F}^{-1} \{X(j\omega)\}$ Similarly, given the expression we have already seen for an arbitrary x(t): **Fourier Transform:**

• The two equations on the previous slide are called the *Fourier transform pair*.

You will find a number of these in the accompanying notes.

• The first zeros of each side of the main lobe are at frequencies $\omega = \pm \pi/t_0$ rad/s

Properties of the Fourier Transform Again, we will provide any properties that you might need in the examination.

Name

Linearity

Frequency

differentiation

shifting

Time

Time

Time

integration

Conjugation

No.

1.

5.

8.

Fourier Transform Pair

Table of Properites of the Fourier Transform As was the case of the Laplace Transform, properties of Fourier transforms are usually summarized in Tables of Fourier Transform properties. For example this one: Properties of the Fourier Transform (Wikpedia) and Table 8.8 in Karris (page 8-17). More detail and some commentry is given in the printable version of these notes.

 $a_1 f_1(t) + a_2 f_2(t) + \dots + a_n f_n(t)$

Symmetry Time and frequency scaling

 $\frac{d^n}{dt^n} f(t)$ $(j\omega)$ Frequency differentiation $(-jt)^n f(t)$ $(j\omega)$ $F(j\omega)$

 $F(j\omega)$

 $a_1F_1(j\omega)$ $+ a_2 F_2$

 $+ a_n F_n$ $(j\omega)$

 $(j\omega)$

 $F(j\omega$

 $-j\omega_0$

 $(j\omega)^n F$

jω

 $+\pi F(0$

 $)\delta(\omega)$

 $F^*(-j\omega)$

 $F_1(j\omega$

 $f(\alpha t)$ \$\$\frac{1}{

F(t)

Remarks

}F\left(j\frac{\omega}

{\alpha}\right)\$\$

time compression is

frequency expansion

and vice versa

Definition

 $^2\,d\$

F(j\omega)

RMS

follows from this

Fourier transform is a linear

A time shift corresponds to a phase shift in frequency domain

Multiplying a signal by a complex

exponential results in a frequency

f(t)

 $2\pi f(-j\omega)$

 $e^{j\omega_0 t}f(t)$

 $\int_{-\infty}^{t} f(\tau)d\tau$

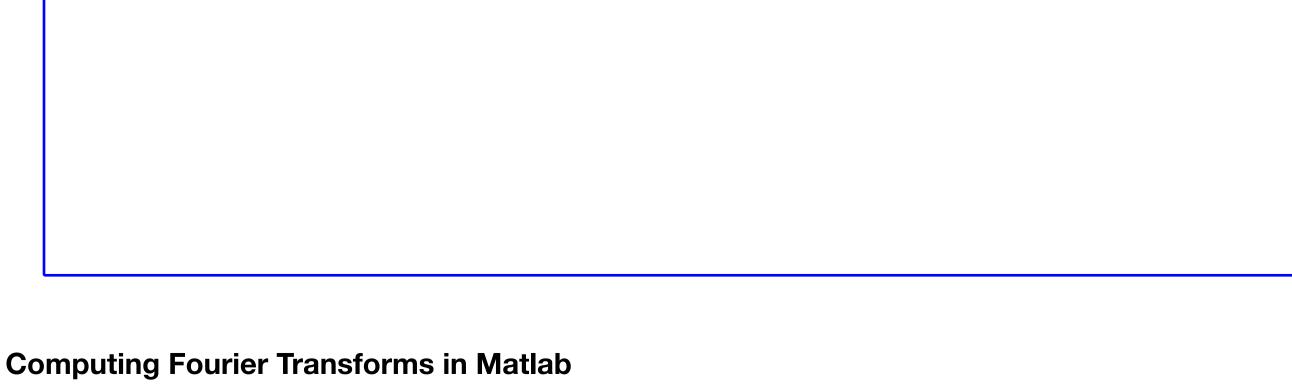
 $f^*(t)$

 $f_1(t) * f_2(t)$ 10. Compare with Laplace Transform convolution $)F_2(j\omega)$ This has application to amplitude Frequency $f_1(t)f_2(t)$ modulation as shown in Boulet pp 11. convolution 182 - 183. * $F_2(j\omega)$ $\int_{-\infty}^{\infty} f(t) dt = F(0)$ Area under Way to calculate DC (or average) 12. f(t)value of a signal $f(0) = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(j\omega) \, d\omega$ Area under 13. $F(j\omega)$ Energy- $\label{eq:comega1} $$ \sup_{2\pi_1 \neq 1}:=\frac{1}{2\pi_2}:=\frac{1}{2\pi_2}.$ F(j\omega) Density 14. $^2\,d\$ Spectrum $^2\,dt=\frac{1}{2\pi}\int_{-\infty}^2\int_{-\infty}^{\infty}\int_{-\infty}^{\infty}dt$ Parseval's 15. f(t) \$\displaystyle{\int_{-\infty}^{\infty}} \infty}^{\infty} theorem See also: Wikibooks: Engineering Tables/Fourier Transform Properties and Fourier Transform—WolframMathworld for more complete references. **Examples** 1. Amplitude Modulation 2. Impulse response 3. Energy computation

Example 2: Impulse response A system has impulse response $f(t) = e^{-t}u_0(t)$. Compute the frequency sprectrum of this system.

Example 1: Amplitude Modulation

Hint use Euler's identity and the frequency shift property



For now, here's an example:

Example Use Matlab to confirm the Fourier transform pair:

Fw = fourier(ft,omega)

Check by computing the inverse using ifourier In []: ft = ifourier(Fw)

Example 3: Energy computation An aperiodic real signal f(t) has Fourier transform $F(j\omega)$. Compute the energy contained the signal between 5kHz and 10kHz.

MATLAB has the built-in fourier and ifourier functions that can be used to compute the Fourier transform and its inverse. We will explore some of thes in the next lab.

In []: syms t v omega x;

 $ft = \exp(-t^2/2);$

 $e^{-\frac{1}{2}t^2} \Leftrightarrow \sqrt{2\pi}e^{-\frac{1}{2}\omega^2}$

In []: pretty(Fw)

Compute the result of multiplying a signal f(t) by a carrier waveform $\cos \omega_c t$.