Fourier transforms of commonly occurring signals

Colophon

An annotatable worksheet for this presentation is available as Worksheet 13. • The source code for this page is fourier_transform/2/ft2.ipynb.

- You can view the notes for this presentation as a webpage (HTML).
- This page is downloadable as a PDF file.

If you have been reading both Karris and Boulet you may have noticed a difference in the notation used in the definition of

Note on Notation

• Karris uses $F(\omega)$

Fourier Transform:

• Boulet uses $F(j\omega)$ I checked other sources and Hsu (Schaum's Signals and Systems) {cite} schaum and Morrell (The Fourier Analysis Video

Series on YouTube) both use the $F(\omega)$ notation. According to Wikipedia Fourier Transform: Other Notations both are used only by electronic engineers anyway and either would be acceptible.

There is some advantage in using Boulet's notation $F(j\omega)$ in that it helps to reinforce the idea that Fourier Transform is a special case of the Laplace Transform and it was the notation that I used in the last section.

In these notes, I've used the other convention on the basis that its the more likely to be seen in your support materials. However, I am happy to change back if you find the addition of j useful.

You should be aware that Fourier Transforms are in general complex so whatever the notation used to represent the transform, we are still dealing with real and imaginary parts or magnitudes and phases when we use the actual transforms in analysis.

Agenda Tables of Transform Pairs

 Relationship between Laplace and Fourier • Fourier Transforms of Common Signals

frequency $F(\omega)$:

The Fourier Transform

Reminder of the Definitions Last time we derived the Fourier Transform by evaluating what would happen when a periodic signal was made periodic. Let us

Examples of Selected Transforms

restate the definitions.

The Inverse Fourier Transform In the signals and systems context, the *Inverse Fourier Transform* is used to convert a function of frequency $F(\omega)$ to a function

 $\mathcal{F}\left\{f(t)\right\} = \int_{-\infty}^{\infty} f(t)e^{-j\omega t} dt = F(\omega).$

In the signals and systems context, the Fourier Transform is used to convert a function of time f(t) to a function of radian

of time f(t):

Duality of the transform

1.

2.

3.

5.

transforming the result.

The Dirac Delta

Matlab:

syms t;

ans =

1

DC

fourier(dirac(t))

fourier(A,omega)

2*pi*dirac(omega)

Related by frequency shifting property:

Cosine (Sinewave with even symmetry)

Note: f(t) is real and odd. $F(\omega)$ is imaginary and odd.

The signum function is a function whose value is equal to

This function is often used to model a *voltage comparitor* in circuits.

Signum (Sign)

The transform is:

Example 4: Unit Step

Clue

Define

Proof

so

Use the signum function to show that

ans =

In [2]:

Proof: uses sampling and sifting properties of $\delta(t)$.

In [1]: imatlab_export_fig('print-svg') % Static svg figures.

Note the similarity of the Fourier and its Inverse.

Table of Common Fourier Transform Pairs

Dirac delta

Time sample

Phase shift

Signum

Unit step

 $\mathcal{F}^{-1}\left\{F(\omega)\right\} = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(\omega)e^{j\omega t} d\omega = f(t).$ Note, the factor 2π is introduced because we are changing units from radians/second to seconds.

This table is adapted from Table 8.9 of Karris. See also: Wikibooks: Engineering Tables/Fourier Transform Table and Fourier

 $\delta(t)$

 $\delta(t-t_0)$

 $e^{j\omega t_0}$

sgn(x)

 $u_0(t)$

Remarks

1 Constant energy at all frequencies.

also known as sign function

 $e^{-j\omega t_0}$

 $2\pi\delta(\omega-\omega_0)$

 $\frac{1}{j\omega} + \pi \delta(\omega)$

<u>Transforr</u>—WolframMathworld for more complete references. $F(\omega)$ f(t)Name

$\pi \left[\delta(\omega - \omega_0) + \delta(\omega + \omega_0)\right]$ 6. Cosine $\cos \omega_0 t$ $\sin \omega_0 t - j\pi \left[\delta(\omega - \omega_0) - \delta(\omega + \omega_0)\right]$ 7. Sine $e^{-at}u_0(t)$ 8. Single pole a > 0 $te^{-at}u_0(t)$ 9. Double pole a > 0 $\frac{j\omega + a}{((j\omega + a)^2 + \omega^2)}$ 10. Complex pole (cosine component) $e^{-at} \cos \omega_0 t \ u_0(t)$ a > 0Complex pole (sine component) $e^{-at} \sin \omega_0 t \ u_0(t)$ a > 011.

Relationship between
$$f(t)$$
 and $F(\omega)$
 In most of the work we will do in this course, and in practice, the $signals$ that we use with the Fourier transform will be a real continuous aperiodic functions of time that are zero when $t=0$. The Fourier transforms of such signals will be complex continuous functions of frequency which have real and imaginary parts and will exist at both positive and negative values of ω . It is often most convenient to deal with the transformed "spectrum" by considering the magnitude and phase and we will therefore often plot $F(\omega)$ on two separate graphs as $magnitude |F(\omega)|$ and $phase \angle F(\omega)$ (where phase is measured in radians) plotted against frequency $\omega \in [-\infty, \infty]$ (in radians/second). We most often represent the $system$ by its so-called $frequency$ $response$ and we will be interested on what effect the system has on the signal $f(t)$.

 $\delta(t) \Leftrightarrow 1$

 $F(\omega)$

 $F(\omega)$

 $2\pi\delta(\omega)$

As for the Laplace transform, this is more conveniently determined by exploiting the time convolution property. That is by performing a Fourier transform of the signal, multiplying it by the system's frequency response and then inverse Fourier

Have these ideas in mind as we go through the examples in the rest of this section.

Related: $\delta(t-t_0) \Leftrightarrow e^{-j\omega t_0}$

f(t)

 $1 \Leftrightarrow 2\pi\delta(\omega)$

 $e^{j\omega_0 t} \Leftrightarrow 2\pi\delta(\omega - \omega_0)$

 $\cos(t) = \frac{1}{2} \left(e^{j\omega_0 t} + e^{-j\omega_0 t} \right) \Leftrightarrow \pi \delta(\omega - \omega_0) + \pi \delta(\omega + \omega_0)$

Note: f(t) is real and even. $F(\omega)$ is also real and even. **Sinewave** $\sin(t) = \frac{1}{j2} \left(e^{j\omega_0 t} - e^{-j\omega_0 t} \right) \Leftrightarrow -j\pi \delta(\omega - \omega_0) + j\pi \delta(\omega + \omega_0)$

f(t)

$$\operatorname{sgn} x = \begin{cases} -1 & x < 0 \\ 0 & x = 0 \\ +1 & x > 0 \end{cases}$$

$$\operatorname{sgn} x = u_0(t) - u_0(-t) = \frac{2}{j\omega}$$

$$f(t)$$

$$F_{\operatorname{Im}}(\omega)$$

 $\mathcal{F}\left\{u_0(t)\right\} = \pi\delta(\omega) + \frac{1}{j\omega}$

 $\operatorname{sgn} t = 2u_0(t) - 1$

f(t)

 $\operatorname{sgn} x = 2u_0(t) - 1$

 $u_0(t) = \frac{1}{2} \left[1 + \operatorname{sgn} x \right]$

 $u_0(t) \Leftrightarrow \pi \delta(\omega) + \frac{1}{i\omega}$

 $u_0(t) \Leftrightarrow \pi \delta(\omega) + \frac{1}{j\omega}$

Unit step is neither even nor odd so the Fourier transform is complex with real part $F_{\rm Re}(\omega)=\pi\delta(\omega)$ and imaginary part

 $F(\omega)$

 ω

Does that help?

From previous results $1 \Leftrightarrow 2\pi\delta(\omega)$ and $\operatorname{sgn} x = 2/(j\omega)$ so by linearity

f(t)

QED Graph of unit step

 $F_{\rm Im}(\omega)=1/(j\omega)$. The real part is even, and theimaginary part is odd. **Example 5** Use the results derived so far to show that $e^{j\omega_0 t}u_0(t) \Leftrightarrow \pi\delta(\omega-\omega_0) + \frac{1}{j(\omega-\omega_0)}$ Hint: linearity plus frequency shift property. **Example 6** Use the results derived so far to show that $\sin \omega_0 t \ u_0(t) \Leftrightarrow \frac{\pi}{j2} [\delta(\omega - \omega_0) - \delta(\omega + \omega_0)] + \frac{\omega_0}{\omega_0^2 - \omega^2}$

Important note: the equivalent example in Karris (Section 8.4.9 Eq. 8.75 pp 8-23—8-24) is wrong!

 $\cos \omega_0 t \ u_0(t) \Leftrightarrow \frac{\pi}{2} [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)] + \frac{j\omega}{\omega_0^2 - \omega^2}$

If a signal is a function of time f(t) which is zero for $t \leq 0$, we can obtain the Fourier transform from the Laplace transform by

 $\mathcal{F}\left\{e^{-at}u_0(t)\right\}$

 $\frac{1}{j\omega+1}$

Derivation of the Fourier Transform from the Laplace Transform

Use the result of Example 6 to determine the Fourier transform of $\cos \omega_0 t \ u_0(t)$.

Hint: Euler's formula plus solution to example 2.

See worked solution for the corrected proof.

Example 7

Solution to example 7

substituting s by $j\omega$.

Solution to example 8

Boulet gives the graph of this function.

Compute

Given that

Example 8: Single Pole Filter Given that $\mathcal{L}\left\{e^{-at}u_0(t)\right\} = \frac{1}{s+a}$

 $\mathcal{F}\left\{e^{-at}\cos\omega_0t\ u_0(t)\right\}$ Solution to example 9 $j\omega + a$ $\frac{1}{(j\omega+a)^2+\omega_0^2}$

Fourier Transforms of Common Signals We shall conclude this session by computing as many of the the Fourier transform of some common signals as we have time for. rectangular pulse triangular pulse periodic time function unit impulse train (model of regular sampling)

Suggestions for Further Reading

Summary Tables of Transform Pairs

 Fourier Transforms of Common Signals References

 $\mathcal{L}\left\{e^{-at}\cos\omega_0 t\ u_0(t)\right\} = \frac{s+a}{(s+a)^2 + \omega_0^2}$ Compute Boulet gives the graph of this function.

Example 9: Complex Pole Pair cos term

• Use of inverse Fourier series to determine f(t) from a given $F(j\omega)$ and the "ideal" low-pass filter (pp 188–191). The Duality of the Fourier transform (pp 191 – 192).

Boulet has several interesting amplifications of the material presented by {cite} karris. You would be well advised to read these. Particular highlights which we will not have time to cover: • Time multiplication and its relation to amplitude modulation (pp 182-183). • Fourier transform of the complex exponential signal $e^{(\alpha+j\beta)t}$ with graphs (pp 184—187).

 Examples of Selected Transforms Relationship between Laplace and Fourier

See Bibliography. Next Section

I will not provide notes for these, but you will find more details in Chapter 8 of Karris and Chapter 5 of Boulet and I have created some worked examples (see Blackboard and the OneNote notebook) to help with revision.

• The Fourier Transform for Systems and Circuit Analysis