

Exercise 8

TTK4130 Modeling and Simulation

Problem 1 (Implementation of friction models)

In this problem we will implement different *static* and *dynamic* friction models. To test the developed models, we will use a simple *testbench* as illustrated in Figure 1. The testbench is based on a model based on Newton's law of a box of mass $m = 1\text{kg}$ sliding on a table due to application of an external force F_a (assumed to be a ramp function). A friction force F_f opposes motion. We will use the parameters

$F_c = 1$	Coulomb friction
$F_s = 1.2$	Stiction (static friction)
$F_v = 0.05$	Viscous friction
$v_s = 0.1$	Characteristic Stribeck velocity

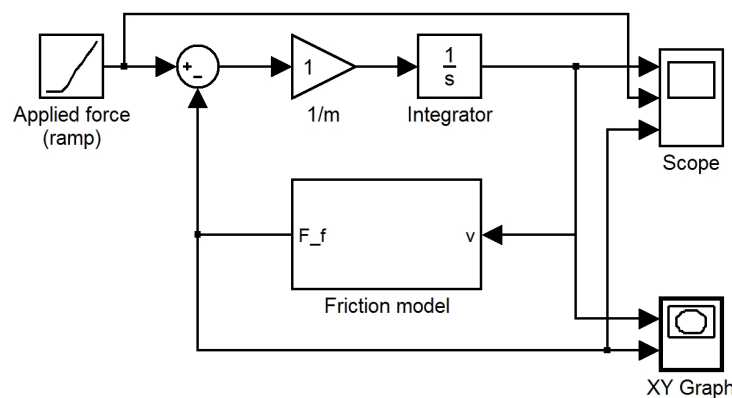


Figure 1: Testbench for friction models

(a) We start with static friction models, more specifically *Coulomb's model*, given by

$$F_f = F_c \text{sign}(v), \quad v \neq 0 \quad (1)$$

where v is velocity. Implement this model in Simulink using a sign-block (do not use the built-in Coulomb friction block). Simulate the model over 10s, with ramp slope of 0.5 and 2.0. Use first a variable step solver. Explain what happens (hint: does the sign block include zero-crossing detection?). Choose the fixed-step Euler solver with sample time 0.01 instead – how does the model simulate now, for both values of the ramp?

The above (Coulomb) model has the disadvantage that it is not defined at $v = 0$. The Karnopp friction model (for Coulomb friction) remedies this by defining

$$F_f = \begin{cases} \text{sat}(F_a, F_c), & v = 0 \\ F_c \text{sign}(v), & v \neq 0 \end{cases}$$

where the sat-function is as defined in the book (the Simulink saturation function can be used for this saturation function by “hard-coding” the upper and lower limit at the second argument (F_c)). For this model to work properly, we must either use variable-step methods with event-detection to detect exactly when $v = 0$, or we have to use some kind of dead-zone around zero-velocity to treat the velocity as zero when it is small. Here, we will implement both of these approaches.

- (b) Implement Karnopp's friction model in the setup in Figure 3, that is, fill in the two If Action Subsystems. Note that the if-block (by default) generate events when the if-clause changes. The Merge-block can be found under Signal Routing, and the If Action Subsystems under Ports & Subsystems.

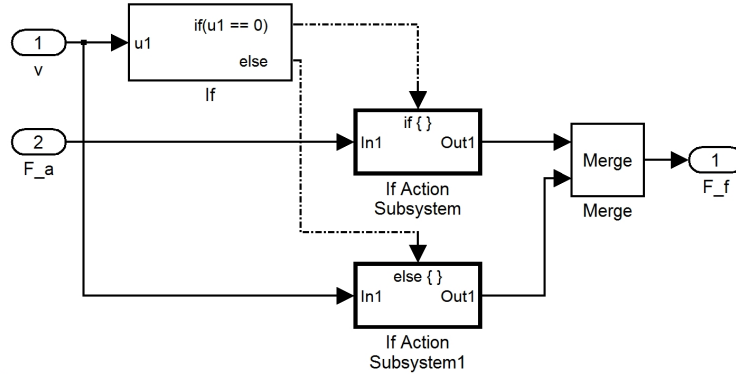


Figure 3: Setup for implementing Karnopp's friction model with event detection

Simulate using a variable step method (for example ode45) and comment. How does event-detection help? Note: It might be necessary to turn off event-detection for the sign-function inside the If Action Subsystem (why?). On the other hand, event-detection for the saturation function inside the other If Action Subsystem block should be turned on (why?).

- (c) Implement Karnopp's model without relying on event-detection, by using dead-zone:

$$F_f = \begin{cases} \text{sat}(F_a, F_c), & |v| \leq \delta \\ F_c \text{sign}(v), & |v| > \delta \end{cases}$$

Choose $\delta = 0.5$. Simulate using a fixed-step solver, and comment. Test also using an initial velocity of $v(0) = -1$.

- (d) Extend the model in (b) with sticking, Stribeck-effect (see 5.2.6 in book) and linear viscous friction,

$$F_f = \begin{cases} \text{sat}(F_a, F_s), & v = 0 \\ \left[F_c + (F_s - F_c)e^{-(v/v_s)^2} \right] \text{sign}(v) + F_v v, & v \neq 0 \end{cases}$$

(note slight error in book). Compare with Coulomb friction. Use ramp slope 0.5.

- (e) Finally, we shall implement the LuGre dynamic friction model,

$$\begin{aligned} \dot{z} &= (v - \sigma_0 \frac{|v|}{g(v)} z), \\ F_f &= \sigma_0 z + \sigma_1 \dot{z} + \sigma_2 v, \end{aligned}$$

where z may represent a small displacement in the stick-zone and σ_0 "spring-stiffness" of asperities. Set $\sigma_0 = 500$, $\sigma_1 = 0$ and $\sigma_2 = F_v$, and implement $g(v)$ to get the same steady-state solution as in (d). Simulate and compare with solution in (d). Play around with the parameters σ_0 and σ_1 , and comment.

Problem 2 (Transmission line)

A lossless transmission line is described by

$$\frac{\partial u}{\partial t} = -cz_0 \frac{\partial y}{\partial x} \quad (2a)$$

$$\frac{\partial y}{\partial t} = -\frac{c}{z_0} \frac{\partial u}{\partial x} \quad (2b)$$

where c is wave velocity (speed of sound) and z_0 is line (characteristic) impedance. The wave variables are

$$a = u + z_0 y,$$

$$b = u - z_0 y.$$

- (a) We first assume that the equations describe a lossless hydraulic transmission line filled with water. What physical variables are u and y in this case?

The transmission line is a pipe with circular cross section with radius $R = 10$ cm. Water has speed of sound

$$c = \sqrt{\frac{\beta}{\rho}} = 1500 \text{ m/s} \quad (3)$$

and density $\rho = 980 \text{ kg/m}^3$. What is the bulk modulus of water, and what is (characteristic) line impedance for the transmission line? What are the units?

- (b) We now assume the equations describe an electrical transmission line in the form of a high voltage line in air. The wave velocity is the speed of light $c = 3 \cdot 10^8 \text{ m/s}$ and the characteristic impedance is $z_0 = 300 \Omega$. What are the line inductance L and capacitance C per unit length? What physical variables are u and y in this case?

Problem 3 (Passivity of transmission line)

A lossless transmission line is described by (2). We use index 1 to denote input-side and index 2 to denote output-side. That is, u_1 is input and y_2 is the output/measurement. On the output side we have a load impedance $Z_2(s)$ such that

$$U_2(s) = Z_2(s)Y_2(s). \quad (4)$$

The corresponding wave variables fulfil

$$B_2(s) = H_2(s)A_2(s) \quad (5)$$

where

$$H_2(s) = \frac{G_2(s) - 1}{G_2(s) + 1} \quad (6)$$

and

$$G_2(s) := \frac{Z_2(s)}{z_0}. \quad (7)$$

See e.g. Chapter 1.6.8 in book.

A linear system with input u and output y is passive if the transfer function $Y(s)/U(s) = G(s)$ is positive real.

- (a) Show that if $G_2(s)$ is positive real, then $H_2(s)$ is bounded real (hint: see book Chapter 2.4.11).

- (b) Define $H_1(s)$ by

$$B_1(s) = H_1(s)A_1(s). \quad (8)$$

Show that if $H_2(s)$ is bounded real, then $H_1(s)$ is also bounded real. Use that

$$H_1(s) = \frac{B_1(s)}{A_1(s)} = e^{-2Ts} H_2(s).$$

(c) We have the load impedance $Z_1(s)$ such that

$$Z_1(s) = \frac{Y_1(s)}{U_1(s)}. \quad (9)$$

Further, we define the transfer function

$$G_1(s) = Z_1(s)z_0. \quad (10)$$

Show that

$$G_1(s) = \frac{1 - H_1(s)}{1 + H_1(s)} \quad (11)$$

and that if $H_1(s)$ is bounded real, then $G_1(s)$ is positive real. (Hint 1: Check how Eq. 6 is found.

Hint 2: $2 \operatorname{Re} [G_1(s)] = G_1(s) + G_1^*(s)$).

(d) Show that if $Z_2(s)$ is positive real, then the system with input u_1 and output y_1 is passive.