

## Linear Systems

ECE3150

Test III

November 18, 2019

Name:

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Signature:

Robert Campbell

Instructions:

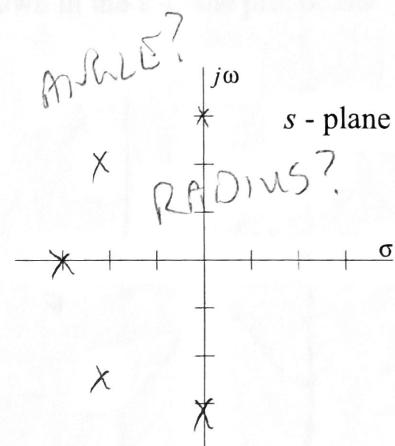
- 1) This exam is closed book, closed notes, and closed neighbor. You may have one  $8.5'' \times 11''$  note sheet with notes written on one side only. **Turn in your notesheet with your test.**
- 2) There are 8 pages to this exam including this cover sheet. You have 50 minutes to work the exam. Start when the instructor tells you to start.
- 3) Work the problems on the exam in the space provided. If you need additional space, *use the back side of the previous page.*
- 4) If you believe a problem cannot be solved, for full credit state exactly *why* it cannot be solved.
- 5) If you believe a problem has ambiguous notation, ask the instructor for clarification.

Question #	Max Points	Points
1	25	<u>12</u>
2	25	<u>20</u>
3	25	<u>10</u>
4	25	<u>24</u>
Total:	100	(66)

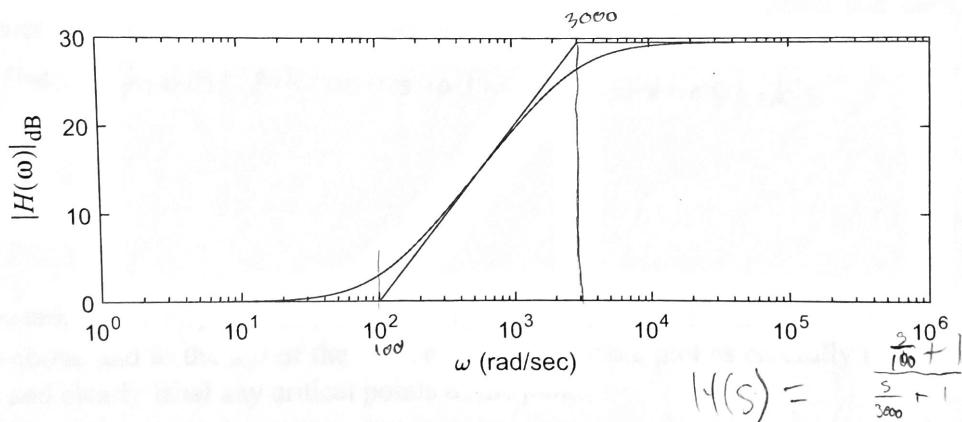
(12)

1) (25 pts)

- a) A 5th-order Butterworth *lowpass* filter has cutoff frequency  $\omega_c = 2000 \text{ rad/sec}$ . Draw the pole-zero constellation for this filter on the  $s$ -plane shown. Clearly label the axis tic marks so the plot scale is defined. Also, clearly label the position of each point in the pole-zero constellation so it is clearly defined.



- b) The magnitude  $|H(\omega)|_{\text{dB}}$  and bode plot asymptotes for a filter are shown below. If  $|H(0)|_{\text{dB}} = 0$ , write an expression for the filter transfer function  $H(s)$ .



$$H(s) = \frac{\frac{s}{100} + 1}{\frac{s}{2000} + 1}$$

- c) The transfer function for a system is given by  $H(s) = \frac{4(s-2)(s+2)}{s(s+1)}$ . Is this system (circle one)  
 (1) stable, (2) marginally stable, or (3) unstable? Clearly explain your answer.

The rightmost pole lies on the ~~jω~~ axis.

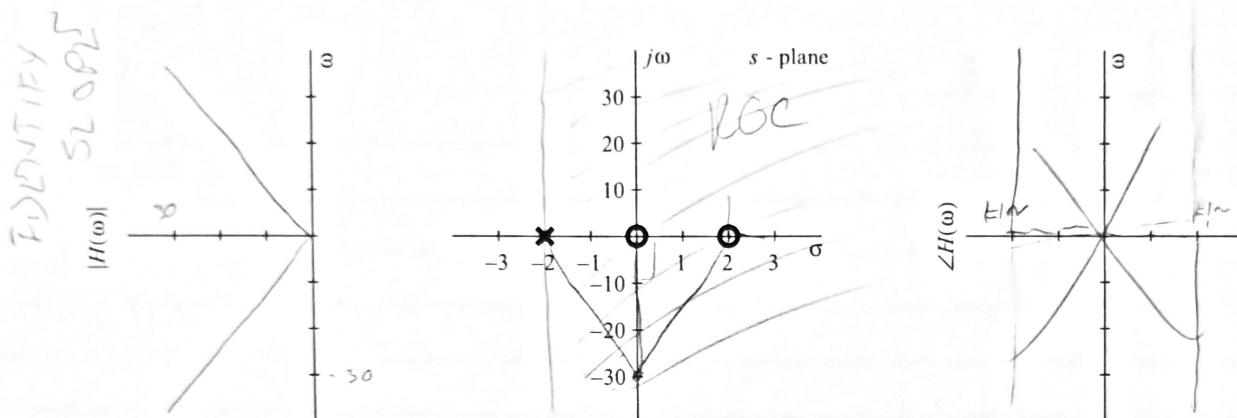
AT  $s=0$   
 + NO OTHER  
~~jω~~ ON  
 AXIS

- d) The transfer function for a system is given by  $H(s) = \frac{4(s-2)(s+2)}{s(s+1)}$ . Find  $h(0^+)$  and  $h(\infty)$ . If either do not exist, then simply state that and explain why.

$$H(s)$$



- 2) (25 pts) The pole-zero constellation for *transfer function*  $H(s)$  is shown in the  $s$ -plane plot below.



- a) Draw the *Region of Convergence* for this transfer function directly on the  $s$ -plane plot.

- b) Is this system *stable*? Clearly explain why or why not.

Yes, the rightmost pole occurs to the left of the  $\sigma$ -axis.

- c) Draw the *magnitude response*  $|H(\omega)|$  over the frequency range  $[-30, 30]$  rad/sec directly on the diagram above and to the *left* of the  $s$ -plane plot. Draw the plot as carefully and accurately as you can and clearly label any critical points of the plot.

$$\frac{w \sqrt{\omega^2 + 2\zeta^2}}{\sqrt{\omega^2 + 2\zeta^2}}$$

- d) Draw the *phase response*  $\angle H(\omega)$  over the frequency range  $[-30, 30]$  rad/sec directly on the diagram above and to the *right* of the  $s$ -plane plot. Draw the plot as carefully and accurately as you can and clearly label any critical points of the plot.

- e) Write a *complete expression* for the transfer function  $H(s)$  assuming that  $|H(10)| = 20$ .

$$H(s) = \frac{k s(s-2)}{s+2}$$

$$H(10) = \left| \frac{k(10)(10-2)}{j10+2} \right| = \frac{k \cdot 80}{12} = \frac{k \cdot 20}{3} = 20$$

$$k = 3$$

$$H(s) = \frac{3(s)(s-2)}{s+2}$$

10

10

$$\frac{-b \pm \sqrt{b^2 - 4ac}}{2a}$$

$$\frac{1 \pm \sqrt{1-184}}{4} = \text{Zeros}$$

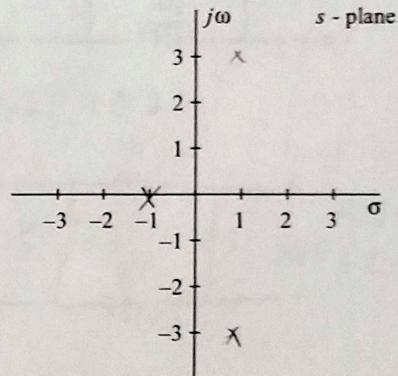
- 3) (25 pts) The transfer function for a Linear, Time-Invariant system is shown at the right.

$$H(s) = \frac{2s^2 - s + 23}{(s^2 - 2s + 10)(s + 1)}$$

- a) Plot the *pole-zero constellation* on the  $s$ -plane plot shown at the right

- b) Is this system *stable*? Clearly explain why or why not.

Unstable, poles occur to the right of the  $\sigma$ -axis.



- c) Find the *impulse response*  $h(t)$ .

- \* Carefully organize your solution
- \* Circle your final answer.
- \* For full credit, show all your work.

$$\frac{2 \pm \sqrt{4-4(10)}}{2} = 1 \pm j\frac{\sqrt{36}}{2}$$

$$s = 1 \pm j3$$

$$H(s) = \frac{2s^2 - s + 23}{(s^2 - 2s + 10)(s + 1)} = \frac{2s^2 - s + 23}{((s-2)^2 + 6)(s+1)} = \frac{2s^2 - s + 23}{\cancel{(s-2)^2 + 6}}(s+1)$$

$$\begin{aligned} \frac{2s^2 - s + 23}{s+1} &= \frac{A}{s+1} + \frac{Bs + C}{s^2 - 2s + 10} \\ &= As^2 - 2As + 10A + Bs^2 + Cs + B + C \end{aligned}$$

$$2s^2 = As^2 + Bs^2 \quad -s = -2As + Cs + Bs \quad 23 = 10A + C$$

$$H(s) = 2 \frac{1}{s+1}$$

$$\begin{aligned} A &= \left. \frac{2s^2 - s + 23}{s^2 - 2s + 10} \right|_{s+1=0} = \frac{2+1+23}{1+2+10} \\ &= \frac{26}{13} = 2 \end{aligned}$$

$$\begin{aligned} Bs + C &= \left. \frac{2s^2 - s + 23}{s+1} \right|_{s=-1} \\ &= \frac{2(-1)^2 - (-1) + 23}{-1+1} = 26 \\ &= 10j3 \end{aligned}$$

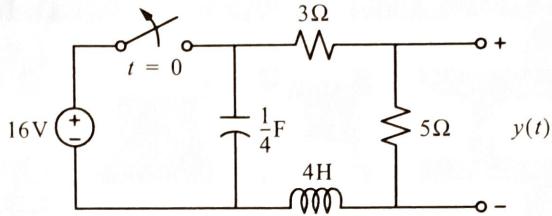
$$\begin{aligned} 2(1+j3)^2 - 10j3 + 23 \\ 1+j3 + 1 \end{aligned}$$

$$h(t) = e^{-t} \frac{t}{2} +$$

~~$$= e^{-t} +$$~~

24

- 4) (25 pts) The circuit shown at the right contains a switch that changes state at  $t = 0$ . Solve for the output voltage  $y(t)$  over the entire time range from  $[-\infty, \infty]$ .



- \* Carefully organize your solution
- \* Circle the steady-state part of your final answer.
- \* Underline the transient part of your final answer.
- \* For full credit, show all your work.

$$i(t) = \underline{Y} e^{-t} + u(t)$$

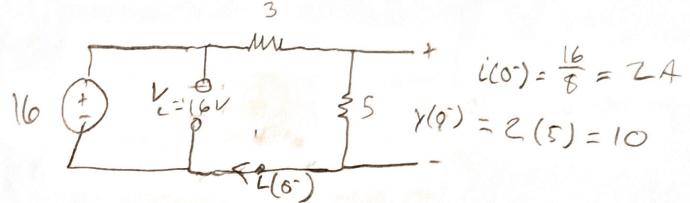
$$+ 2e^{-t} u(t)$$

$$- 2e^{-t} \frac{t}{F} u(t)$$

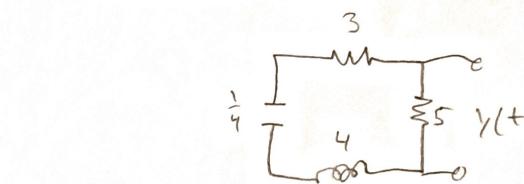
$$i(t) = \underline{e^{-t}} + u(t) + \underline{2e^{-t} u(t)}$$

**TRANSIENT**

Case I:  $t < 0$



Case II:  $t \geq 0$



$$\frac{16}{s} = \frac{V(0)}{s}$$

$$V_L = L \frac{di}{dt}$$

$$V_L = L(I(s)) - i(0)$$

$$= L I(s) - L i(0)$$

$$S4 \quad L i(0) = 2 \cdot 4$$

$$KVL: 4 \frac{1}{5} I(s) + 3I(s) + 5I(s) - 2 \cdot 4 + 4s I(s)$$

$$+ \frac{16}{s} = 0$$

$$4s I(s) + 8I(s) + \frac{4}{5} I(s) + \frac{16}{s} - 8 = 0$$

$$I(s) \left[ 4s + 8 + \frac{4}{5} \right] = \frac{16}{s} - 8$$

$$I(s) = \frac{\frac{16}{s} + 8}{4s + 8 + \frac{4}{5}} \quad \left( \frac{s}{s} \right)$$

$$I(s) = \frac{\frac{4}{s} + 8s}{4s^2 + 8s + 1} = \frac{\frac{4}{s} + 2s}{s^2 + 2s + 1}$$

$$= \frac{4}{(s+1)^2} + \frac{2s}{(s+1)^2} = \underline{4 \frac{1}{(s+1)^2}} + \frac{2(s+1) + 2}{(s+1)^2} \frac{2}{(s+1)^2}$$