# SystemVerilog's priority & unique - A Solution to Verilog's "full\_case" & "parallel\_case" Evil Twins!

Clifford E. Cummings

Sunburst Design, Inc.

## **ABSTRACT**

At Boston SNUG 1999, I introduced the evil twins of Verilog synthesis, "full\_case" and "parallel\_case.[2]" In the 1999 Boston SNUG paper I pointed out that the full\_case and parallel\_case directives are always most dangerous ... when they work!

At that time, the best guideline to follow was to avoid using full\_case and parallel\_case in all Verilog designs, with one noteworthy exception: add parallel\_case for a special Verilog onehot case statement encoding style.

The SystemVerilog standard[12] has introduced two case-statement and if-statement modifiers, using the new SystemVerilog keywords "priority" and "unique."

The new keywords, priority and unique are part of the SystemVerilog language, not just comment-style directives, which means that simulation, synthesis and formal tools can all recognize and consistently implement proper cross-tool functionality and testing for RTL code written with these new constructs.

This paper summarizes the problems associated with the use of the <code>full\_case</code> and <code>parallel\_case</code> directives and details how the new SystemVerilog <code>priority</code> and <code>unique</code> keywords solve these problems while adding valuable RTL synthesis capabilities to the new and powerful SystemVerilog language.

# 1.0 The legend of full\_case parallel\_case

Prior to 1999, I found that engineers routinely added full\_case parallel\_case to all RTL case statements. Indeed, full\_case parallel\_case had been two of the most over-used and abused synthesis directives ever employed by Verilog synthesis design engineers. The reasons cited most often by engineers for using full\_case parallel\_case were:

- full\_case parallel\_case makes my designs smaller and faster.
- full case removes latches from my designs.
- parallel\_case removes large, slow priority encoders from my designs.

The above reasons were (and still are) either inaccurate or dangerous.

- Sometimes these directives don't affect a design at all.
- Sometimes these switches make a design larger and slower.
- Sometimes these directives change the functionality of a design.

And finally:

The full\_case and parallel\_case directives are always most dangerous when they work!

For these reasons, I call full\_case and parallel\_case, "the evil twins of Verilog synthesis!"

This paper re-visits the full\_case and parallel\_case problem, and details how the new SystemVerilog enhancements, unique and priority, give the desired full\_case parallel\_case benefits without the dangers these evil twins potentially introduce to a design.

# 1.1 Synopsys tools, versions and useful command aliases

For this paper, Verilog and SystemVerilog experiments were conducted using the VCS Verilog simulator, version 7.2, and Design Compiler (DC) with Design Vision GUI, version V-2004.06-SP2. At the time of this publication, Design Compiler required a special license to enable recognition of SystemVerilog features. Engineers interested in using Design Compiler SystemVerilog features should contact their local sales or field personnel. I also used the following command aliases to run my Verilog-2001 & SystemVerilog simulations

VCS Verilog-2001 (+v2k) aliases to: compile & run; compile, run & dump

```
alias vcsr="vcs -R +v2k"
alias vcsdump="vcs -PP -R +v2k +define+VPD"
```

VCS SystemVerilog (-sverilog) aliases to: compile; compile & run; compile, run & dump

```
alias svcs="vcs -sverilog +define+SV"

alias svcsr="vcs -R -sverilog +define+SV"

alias svcsdump="vcs -PP -R -sverilog +define+SV +define+VPD"
```

NOTE: the -sverilog switch is new with VCS7.2. Earlier versions of VCS used the switch +sysvcs to enable SystemVerilog simulation. The -sverilog switch is an improvement because now both VCS and Design Compiler use a similar sverilog switch name, as shown below.

Within Design Vision, the tcl command that is used to read SystemVerilog files is:

read\_sverilog <filename>

# 1.2 Understanding Verilog case statements

To fully understand all the topics discussed in this paper, the reader must have a good understanding of Verilog case statements and the definitions of the constructs that make up the case statement. If you need a better working knowledge of Verilog case statement constructs, you should review Section 15.0 before proceeding with the rest of this paper.

# 2.0 Verilog & SystemVerilog case statement modifiers

Before going into detail about all of the case statement modifiers, we should look at the big picture as it relates to full\_case, parallel\_case, priority and unique.

First ask yourself the questions: Why are the **full\_case** parallel\_case synthesis directives so dangerous? And is there a way to avoid the danger while retaining their positive capabilities?

# 2.1 The full\_case parallel\_case dangers

The full\_case and parallel\_case directives are dangerous because they tell the synthesis tool something different about the design than what is told to the simulator.

To the Verilog simulator, full\_case and parallel\_case are buried inside of Verilog comments and are completely ignored. To the synthesis tool, full\_case and parallel\_case are command-directives that instruct the synthesis tools to potentially take certain actions or perform certain optimizations that are unknown to the simulator.

This explains the claim that the full\_case and parallel\_case directives are always most dangerous when they work!

The functionality and pitfalls of the full\_case and parallel\_case directives are explained in Section 3.0 and Section 5.0 of this paper.

# 2.2 The SystemVerilog priority and unique modifiers

The only way to avoid the dangers related to adding full\_case and parallel\_case is to give the simulator the same understanding of these directives as the synthesis tool. This has been accomplished by adding the priority and unique case-statement and if-statement modifiers.

The new priority and unique keywords are recognized by all tools, including the SystemVerilog simulators, so now all the tools have the same information about the design. The capabilities of the priority and unique enhancements are described in Section 4.0 and Section 6.0 of this paper.

## 2.3 SystemVerilog priority and unique assertions

Not only do priority and unique modify case-statements and if-statements, they are also both assertions. The engineer asserts that specified case-statements and if-statements contain all possible values for the expressions being tested. If the run-time simulation finds an untested expression, the assertion fails and an error message is immediately reported, helping to identify potential bugs in the design.

The engineer may also optionally *assert* that specified <code>case</code>-statements and <code>if</code>-statements are unique for the expressions being tested and that no tested expression should match more than one of the <code>case</code>-items or <code>if</code>-branches. If the run-time simulation finds any tested expression that could match more than one tested <code>case</code>-item or <code>if</code>-branch, the assertion fails and an error message is immediately reported, again helping to identify potential bugs in the design.

As can be seen from this short discussion, because these key words can not only be used to optimize designs, but can also identify assumption errors on the part of the design engineer, these new modifiers prove to be valuable design assertions that can be used to help rapidly debug a design. These new modifiers are both command directives and design assertions at the same time. Very nice!

# 2.4 Should priority and unique always be used?

And finally, after viewing the enhanced capabilities of the **priority** and **unique** keywords discussed in the remainder of this paper, one might ask if *all* SystemVerilog **case**-statements and **if**-statements should be coded with the **priority** and **unique** modifiers? The answer is no!

A common and efficient RTL coding style example is included in Section 12.1 of this paper that shows a coding style and functionality that break when using the **priority** and **unique** modifiers.

# 3.0 What is a full case statement?

A full case statement is a case statement in which all possible case-expression binary patterns can be matched to a case item or to a case default. If a case statement does not include a case default and if it is possible to find a binary case expression that does not match any of the defined case items, the case statement is not full.

The examples in this section includes the case statement report that is generated when DC reads each Verilog example. For a description of the meaning of the full\_case reports shown in these examples, go to Section 3.6.

#### 3.1 HDL full case statement

From an HDL simulation perspective, a **full case** statement is a **case** statement in which every possible binary, non-binary and mixture of binary and non-binary patterns is included as a **case** item in the **case** statement. Verilog non-binary values are, and VHDL non-binary values include, **z** and **x** and are called metalogical characters by both the IEEE VHDL-1999 RTL Synthesis Standard[9] and the IEEE Verilog-2002 RTL Synthesis Standard[8].

#### 3.2 Synthesis full case statement

From a synthesis tool perspective, a *full* case statement is a case statement in which every possible binary pattern is included as a case item in the case statement.

Verilog does not require case statements to be either synthesis or HDL simulation full, but Verilog case statements can be made full by adding a case default. VHDL requires case statements to be HDL simulation full, which generally requires an others clause.

Example 1 shows a case statement, with case default, for a 3-to-1 multiplexer. The case default causes the case statement to be full. During Verilog simulation, when binary pattern 2'b11 is present on the select lines, the y-output will be driven to an unknown, but synthesis tools will treat the y-output as a "don't care" for the same select-line combination, potentially causing a mismatch to occur between simulation and synthesis. To insure that the pre-synthesis and post-synthesis simulations match, the case default could assign the y-output to either a predetermined constant value, or to one of the other multiplexer input values; however, the x-assignment in this example is being used as an assertion and is valuable during simulation to identify unexpected values on the select inputs. The x-assignment will also be treated as a don't-care for synthesis, which may allow the synthesis tool to further optimize the synthesized design.

5

Example 1 - A case default, full case statement

```
Statistics for case statements in always block at line 7 in file

'.../mux3c.v'

Line | full/ parallel |

auto/auto |

auto/auto |
```

Figure 1 - Case statement report for a case statement with a case default

#### 3.3 Non-full case statements

Example 2 shows a case statement for a 3-to-1 multiplexer that is not full. The case statement does not define what happens to the y-output when binary pattern 2'b11 is placed on the select (sel) lines. In this example, the Verilog simulation will hold the last assigned y-output value and synthesis will infer a latch on the y-output as shown in the latch inference report of Figure 2.

Example 2 - Non-full case statement

Figure 2 - Latch inference report for non-full case statement

#### 3.4 Synopsys full\_case

Synthesis tools recognize two directives when added to the end of a Verilog case header. The directives are // synopsys full\_case parallel\_case. The directives can either be used

together or an engineer can elect to use only one of the directives for a particular case statement. The Synopsys parallel\_case directive is described in section 5.3.

When // synopsys full\_case is added to a case statement header, there is no change in the Verilog simulation for the case statement, since "// synopsys ..." is interpreted to be nothing more than a Verilog comment; however, Synopsys DC parses all Verilog comments that start with "// synopsys ..." and interprets the full\_case directive to mean that if a case statement is not full, then the outputs are "don't care's" for all unspecified case item combinations. If the case statement includes a case default, the full\_case directive will be ignored because a default automatically causes the case statement to become full.

Example 3 shows a case statement for a 3-to-1 multiplexer that is not full but the case header includes a full\_case directive. During Verilog simulation, when binary pattern 2'bl1 is present on the select lines, the y-output will behave as if it were latched, the same as in Example 2, but the synthesis will treat the y-output as a "don't care" for the same select-line combination, causing a functional mismatch to occur between simulation and synthesis.

Example 3 - Non-full case statement with **full\_case** directive

Figure 3 - Case statement report for a non-full case statement with **full\_case** directive

#### 3.5 A full\_case statement with a case default

What happens to a case statement that includes both the full\_case directive and a case default?

Simply stated, any case statement with a case default will disable the effect of a full\_case directive. This would be an example of a full\_case that does not work, the full\_case directive is completely ignored, it is just extra code in the file and so it does not cause any design problems (it was just wasted typing and a potential point of confusion for some engineers who read the code and tried to determine what the directive would do).

# 3.6 Synopsys case statement reports - full\_case

For each case statement that is read by Synopsys DC, a case statement report is generated that indicates one of the following conditions with respect to the full-nature of a each case statement:

• full / auto (Figure 4) - Synopsys Design Compiler (DC) tools have determined that the case statement as coded is full.

Figure 4 - full / auto - Case statement is full

• full / no (Figure 5) - The case statement was not recognized to be full by Synopsys DC.

Figure 5 - full / no - Case statement not full

• full / user (Figure 6) - A Synopsys full\_case directive was added to the case statement header by the user.

```
Statistics for case statements in always block at line ...

Line | full/ parallel |

X | user/auto |
```

Figure 6 - full / user - "// synopsys full\_case" added to the case header

# 4.0 SystemVerilog priority case

SystemVerilog adds the new case statement modifier called priority. This raises the question, aren't case statements already *priority* statements? The answer is yes, case statements are already *priority* statements.

So what does priority mean? When used in the context of a case statement, the priority modifier means that the case statement is a priority statement and that all possible legal cases have been listed. If during simulation the case expression ever becomes a value that cannot be matched to any of the case items, a runtime error shall be reported.

The biggest difference between a full\_case directive and a priority modified case statement is that the priority keyword is part of the SystemVerilog syntax that will be interpreted the same by simulators, synthesis tools and formal verification tools. In essence, the priority case statement is a "safe" full\_case case statement.

If the priority case statement includes a case default statement, then the effect of the priority keyword is disabled because the case statement is full through the use of the case default statement, and no runtime checking will ever find a case expression that cannot match one of the defined case items or default statement (this is the same as full\_case with a case default - see Section 3.5),..

Just like the full\_case directive, the priority case statement does not guarantee the removal of unwanted latches. Any case statement that makes assignments to more than one output in each case item statement can still generate latches if one or more output assignments are missing from other case item statements.

# 4.1 An editorial comment about the priority keyword

Editorial comment: I do not like the keyword priority.

As noted above, case-statements and if-else-statements are already priority statements and there will be confusion surrounding the priority keyword. What this modifier does is to allow the engineer to assert that all possible cases or if-conditions have been defined so a synthesis tool is free to optimize the case-logic or if-logic assuming that all other testable conditions are don't cares. Of course in synthesis, don't cares frequently lead to favorable logic optimization. Since an engineer has asserted that all testable conditions have been included, simulation tools are required to report an error if an unexpected condition is tested. The engineer asserted that all conditions were tested but the assertion failed!

A better keyword would have been either full or all\_possible, but the former is a very commonly used identifier in legacy Verilog designs and the latter is somewhat verbose and clumsy. Since SystemVerilog design and verification engineers have already used the priority keyword in multiple designs and verification suites, we are now stuck with it. Remember,

**priority case** behaves just like **full\_case**, except the simulator now knows about it and is required to react if the assertion is proven false during simulation.

# 5.0 What is a parallel case statement?

A parallel case statement is a case statement in which it is only possible to match any case expression to one and only one case item. If it is possible to find a case expression that would match more than one case item, the matching case items are called "overlapping" case items and the case statement is not parallel.

The examples in this section includes the case statement report that is generated when DC reads each Verilog example. For a description of the meaning of the parallel\_case reports shown in these examples, go to Section 5.6.

## **5.1** Non-parallel case statements

Example 4 shows a casez statement that is not parallel because if the 3-bit irq bus has any one of the patterns 3'b011, 3'b101, 3'b110 or 3'b111, more than one case item could potentially match the irq value. This will simulate like a priority encoder where irq[2] has priority over irq[1], which has priority over irq[0]. This example will also infer a priority encoder when synthesized, as shown in Figure 8.

Example 4 - Non-parallel case statement

Figure 7 - Case statement report for Example 4

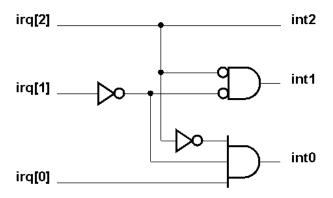


Figure 8 - Correctly implemented interrupt control logic

#### **5.2** Parallel case statements

Example 5 is a modified version of Example 4 such that each of the case items is now unique and therefore parallel. Even though the case items are parallel, this example also happens to infer the same priority encoder logic of Figure 8 when synthesized.

Example 5 - Parallel case statement

Figure 9 - Case statement report for Example 5

# 5.3 Synopsys parallel\_case

Example 6 is the same as Example 4 except that a // synopsys parallel\_case directive has been added to the case header. This example will simulate like a priority encoder but will infer non-priority encoder logic when synthesized, as shown in Figure 11.

Example 6 - Non-parallel case statement with parallel\_case directive

Warning: You are using the parallel_case directive with a case statement in which some case-items may overlap				
Statistics for case statements in always block at line 6 in file '/intctllb.v'				
	Line	full/ parallel		
	9	no/user		
======	==========			

Figure 10 - Case statement report for Example 6

In Example 6, the parallel\_case directive has "worked" and now the synthesized logic does not match the simulated Verilog functional model.

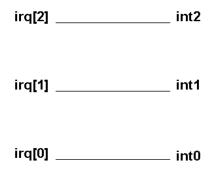


Figure 11 - Incorrectly implemented interrupt control logic

This is an example that demonstrates that adding the parallel\_case directive makes the design smaller and faster, but in the process it also adversely changes the functionality of the design.

## 5.4 A parallel case statement with parallel\_case directive

The casez statement in Example 7 is parallel! If a parallel\_case directive is added to the casez statement, it will make no difference. The design will synthesize the same with or without the parallel\_case directive.

The point is, the parallel\_case directive is always most dangerous when it works! When it does not work, it is just extra characters at the end of the case header.

Example 7 - Parallel case statement with parallel\_case directive

# 5.5 Verilog & VHDL case statements

VHDL case statements are required to have no overlap in any of the case items, and therefore are parallel and cannot infer priority encoders. VHDL case items are constants that are used to compare against the VHDL case expression. For this reason, it is also easy to parse multiple VHDL case item statements without the need to include begin and end keywords for case item statements.

Verilog case statements are permitted to have overlapping case items. Verilog case items can be separate and distinct Boolean expressions where one or more of the expressions can evaluate to "true" or "false." In those instances where more than one case item can match a "true" or "false" case expression, the first matching case item has priority over subsequent matching case items, therefore priority logic will be inferred by synthesis tools.

Verilog casez and casex statements can also include case items with constant vector expressions that include "don't-cares" that would permit a case expression to match multiple case items in the casez or casex statements, also inferring a priority encoder.

If all goes well, full\_case parallel\_case will do nothing to your design, and your design will work fine. The problem happens when full\_case parallel\_case DO work to change the functionality of your design or increase the size and area of your design.

There is one other style of Verilog case statement that frequently infers a priority encoder. The style is frequently referred to as the "case if true" or "reverse case" statement coding style.

This case statement style evaluates expressions for each case item and then tests to see of they are "true" (equal to 1'b1). This coding style is used to infer very efficient one-hot finite state machines, but is otherwise a somewhat dangerous coding practice[1][3][4].

## 5.6 Synopsys case statement reports - parallel\_case

For each case statement that is read by Synopsys DC, a case statement report is generated that indicates one of the following conditions with respect to the parallel nature of each case statement:

• parallel / no (Figure 12) - The case statement was not recognized to be parallel by Synopsys Design Compiler (DC) tools.

Figure 12 - parallel / no - Case statement not parallel

• parallel / auto (Figure 13) - Synopsys DC has determined that the case statement as coded is parallel.

Figure 13 - parallel / auto - Case statement is parallel

• parallel / user (Figure 14) - A Synopsys parallel\_case directive was added to the case statement header by the user.

Figure 14 - parallel / user - "// synopsys parallel\_case" added to the case header

# 5.7 Coding priority encoders

Non-parallel case statements infer priority encoders; however, it is a poor coding practice to code priority encoders using case statements. It is better to code priority encoders using if-else-if statements.

Guideline: Code all intentional priority encoders using if-else-if statements. It is easier for the typical design engineer to recognize a priority encoder when it is coded as an if-else-if statement.

**Guideline:** Case statements can be used to create tabular coded parallel logic. Coding with case statements is recommended when a truth-table-like structure makes the Verilog code more concise and readable.

Although good priority encoders can be inferred from case statements, following the above coding guidelines will help to prevent mistakes and mismatches between pre-synthesis and post-synthesis simulations.

# 6.0 SystemVerilog unique case

SystemVerilog adds the new case statement modifier called "unique."

Any Verilog case statement is permitted to have overlapping case items, which means that a case expression could match more than one case item. When case items overlap, priority encoders are simulated and synthesized.

The unique keyword shall cause the simulator to report a run-time error if a case expression is ever found to match more than one of the case items; hence, the unique keyword is a both a case statement modifier and an assertion. The engineer asserts that the case items are unique and that it is impossible to match a case expression to more than one of the case items.

The unique keyword informs the synthesis tool that no priority encoders are needed to build the logic defined by the unique case statement. Eliminating unnecessary priority encoders typically guarantees smaller and faster logic.

The unique keyword also does the same runtime existence checking that is performed by the simulator. If a case expression does not match any of the unique case items, a runtime error is reported. As with the SystemVerilog priority case statement, adding a case default statement nullifies the testing for nonexistent matches between the case expression and the case items; however, adding the case default to a unique case statement does not remove the uniqueness testing.

The biggest difference between a parallel\_case directive and a unique case statement modifier is that the unique keyword is part of the SystemVerilog syntax that will be interpreted the same by simulators, synthesis tools and formal verification tools. In essence, the unique case statement is a "safe" parallel\_case case statement.

# 7.0 Mapping full\_case & parallel\_case to priority & unique

From the preceding sections, we can now determine which combinations of priority and unique will infer the exact same synthesis functionality as full\_case and parallel\_case. These equivalents are shown in Table 1.

full_case parallel_case version	SystemVerilog version
case ()	case ()
•••	•••
endcase	endcase
case () // full_case	priority case ()
•••	•••
endcase	endcase
<pre>case () // parallel_case</pre>	unique case ()
• • •	•••
endcase	default:
	endcase
<pre>case () // full_case parallel_case</pre>	unique case ()
•••	•••
endcase	endcase

Table 1 - full, parallel, priority, unique case mapping

From Table 1, it is important to note that unique case without case default is equivalent to adding both of the older full\_case parallel\_case directives, while using unique case with case default is equivalent to only adding the older parallel\_case directive.

Of course, the advantage to using the SystemVerilog case modifiers is that not only do the modifiers affect synthesis, but they are also recognized as design assertions and SystemVerilog simulators will report errors whenever the assertions are violated during simulation.

# 8.0 SystemVerilog priority & unique if

SystemVerilog adds the new if statement modifiers, priority and unique and only one of these keywords can be used before an if-else-if statement. A nested if-else-if statement could include another priority or unique modifier, but any of the else-branches of an if-else-if statement cannot be modified with priority or unique.

```
priority if (...) ...
else if (...) ...
else if (...) ...
```

Editorial comment: A useful enhancement to the SystemVerilog language would be a binary else, one that does not default-trigger when the if-tests fail due to x's or z's.

From a purely coding style perspective, if-else-if statements suggest priority encoders, while case statements suggest parallel logic, even though a case statement can infer the exact same priority encoder.

So what does priority if mean? When used in the context of an if statement, the priority modifier means that the if-else-if must be true for at least one of the if-tests. If during simulation the if expression ever becomes a value that cannot be matched to any of the if-else-if tests, a runtime error shall be reported. Of course a final else-statement will disable any priority if testing.

So what does unique if mean? The unique keyword shall cause the simulator to report a runtime error if an if-else-if statement is ever found that matches more than one of the if-else-if tests during the current execution of the if-else-if statement. Just like the case statement, the unique if testing also does the equivalent priority if run-time testing.

# 9.0 Synthesis coding styles

Sunburst Design Assumption: it is generally a bad coding practice to give the synthesis tool different information about the functionality of a design than is given to the simulator.

Whenever either full\_case or parallel\_case directives are added to the Verilog source code, more information is potentially being given about the design to the synthesis tool than is being given to the simulator.

Guideline: In general, do not use full\_case parallel\_case directives with any Verilog case statements.

Guideline: Replace the older full\_case parallel\_case directives with new and safer SystemVerilog priority and unique directives.

Guideline: Educate (or fire) any employee or consultant who routinely adds full\_case parallel\_case to all case statements in their Verilog code, especially if the project involves the design of medical diagnostic equipment, medical implants, or detonation logic for thermonuclear devices!

Guideline: Do not use the priority and unique directives with all SystemVerilog case statements and if-else-if statements. There are still some efficient coding styles that do not define all possible testing conditions within conditional statement and these efficient coding styles would trigger run-time simulation errors if priority and unique directives were added to all case statements and if-else-if statements.

Other exceptions and guidelines will surely be discovered as engineers use the new SystemVerilog modifiers. These exceptions and guidelines will be added to later versions of this paper and will be freely downloadable from the www.sunburst-design.com/papers web site.

# 10.0 Latch example using full\_case

Myth: // synopsys full\_case removes all latches that would otherwise be inferred from a case statement.

Truth: The full\_case directive and now the SystemVerilog priority and unique case-statement and if-statement modifiers only remove latches from a case statement for missing case items or incomplete if-else-if statements. One of the most common ways to infer a latch is to make assignments to multiple outputs from a single case statement but neglect to assign all outputs for each case item. Even adding the full\_case directive or SystemVerilog priority or unique modifiers to this type of case statement will not eliminate latches[5].

Example 8 shows Verilog code for a simple address decoder that will infer a latch for the mce0\_n mce1\_n and rce\_n outputs (as reported in Figure 15 and shown in Figure 16), despite the fact that the full\_case directive was used with the case statement. In this example, the case statement is full but not all outputs are assigned for each case item; therefore, latches were inferred for all three outputs.

Example 8 - **full** case directive with latched outputs

Figure 15 - Case statement report and latch report for **full\_case** latched example

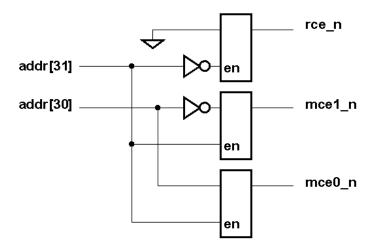


Figure 16 - Address decoder with unwanted latches

The easiest way to eliminate latches is to make initial default value assignments to all outputs immediately beneath the sensitivity list, before executing the case statement, as shown in Example 9. The correctly synthesized logic is shown in Figure 18.

Example 9 - Initial default value assignments to remove latches

Figure 17 - Case statement report for Example 9

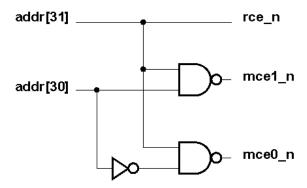


Figure 18 - Correctly implemented address decoder

# 11.0 Synopsys warnings

When Verilog files are read by DC, Synopsys issues warnings when the full\_case directive is used with a case statement that was not full (see Synopsys full\_case description in section 3.4). The same warnings are now issued when the new SystemVerilog priority and unique keywords are used in equivalent full\_case coding styles with a case statement that is not full.

Example 10 shows a non-full case statement with full\_case directive. Figure 19 shows the warning that is reported when the full\_case directive is used with a non-full case statement.

```
module fcasewarn1b
  (output reg y,
    input    d, en);

always @*
    case (en) // synopsys full_case
        1'b1: y = d;
    endcase
endmodule
```

Example 10 - Non-full case statement with full\_case directive

Figure 19 - Synopsys full\_case warning

The warning in Figure 19 should really say, "watch out! the full\_case directive might work and cause your design to break!!" Unfortunately this warning is easy to miss when running a synthesis script and the design might be adversely affected by the full\_case directive.

Similarly, when Verilog files are read by DC, Synopsys issues warnings when the parallel\_case directive (or new SystemVerilog unique case modifier) is used with a case statement that was not parallel. (see Synopsys parallel\_case description in section 5.3).

Example 11 shows a non-parallel case statement with parallel\_case directive. Figure 20 shows the warning that is reported when the parallel\_case directive is used with a non-parallel case statement.

```
module pcasewarn1b
  (output reg y, z,
    input     a, b, c, d);

always @* begin
    {y,z} = 2'b00;
    casez ({a,b,c,d}) // synopsys parallel_case
        4'b11??: y = 1'b1;
        4'b??11: z = 1'b1;
    endcase
    end
endmodule
```

Example 11 - Non-parallel case statement with parallel\_case directive

Figure 20 - Synopsys parallel\_case warning

The warning in Figure 20 should really say, "watch out! the parallel\_case directive might work and cause your design to break!!" Unfortunately this warning, like the full\_case warning, is also easy to miss when running a synthesis script and the design might be adversely affected by the parallel\_case directive.

# 12.0 Actual full\_case design problem

The 2-to-4 decoder with enable in Example 12, uses a case statement that is coded without using any synthesis directives. The resultant design was a decoder built from 3-input and gates and inverters as shown in Figure 22. No latches were inferred because all outputs are given a default

assignment before the case statement. For this example, the pre-synthesis and post-synthesis designs and simulations matched.

Example 12 - Decoder example with no **full\_case** directive

Figure 21 - Case statement report for Example 12

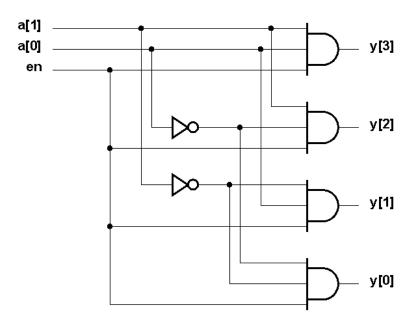


Figure 22 - Correctly synthesized 2-to-4 decoder with output enable

The 2-to-4 decoder with enable in Example 13, uses a case statement with the full\_case synthesis directive. Because of this synthesis directive, the enable input (en) was optimized away during synthesis and left as a dangling input as shown in Figure 24. The pre-synthesis simulation results of modules code4a and code4b matched the post-synthesis simulation results of module code4b [2].

```
// full case example
// Decoder built from four 2-input nor gates
     and two inverters
// The enable input is dangling (has been optimized away)
module code4b
  (output reg [3:0] y,
   input
              [1:0] a,
   input
                    en);
  always @* begin
    y = 4'h0;
    case ({en,a}) // synopsys full_case
      3'b1_00: y[a] = 1'b1;
      3'b1_01: y[a] = 1'b1;
      3'b1_10: y[a] = 1'b1;
      3'b1 11: y[a] = 1'b1;
    endcase
  end
endmodule
```

Example 13 - Decoder example with **full\_case** directive

Figure 23 - Case statement report for Example 13

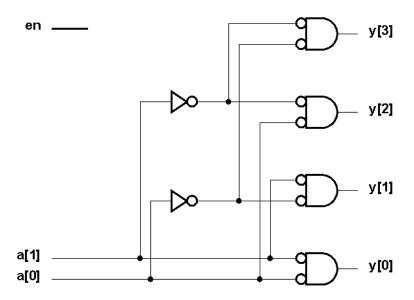


Figure 24 - Incorrectly synthesized 2-to-4 decoder with output enable

This same problem occurs when either the **priority** or **unique** modifiers are used with this **case** statement. The difference is that during simulation, a run-time error would be reported anytime the **en** input goes to **0**. This would be a hint that the **priority** and **unique** keywords are going to cause problems when this coding style is synthesized.

# 12.1 Should you always use priority and unique case modifiers?

Should SystemVerilog case and if-else-if statements always add the priority and unique modifiers? The answer is NO! It can be seen from Example 13 that priority or unique would break this efficient RTL coding style.

Remember that the priority modifier asserts that all possible cases have been defined within the case statement, and that if a case expression does not match any of the defined case items, then an error shall be reported. This causes problems with some commonly recommended and efficient RTL coding styles.

One of the easiest ways to avoid unwanted latches is to make a default assignment at the top of an always block to all of the outputs that are assigned within the always block as was shown in the 2-to-4 decoder code of Example 12.

The equivalent coding style using priority case is shown below. This coding style actually causes two problems: (1) the synthesis results will be the same faulty design shown in Figure 24 due to the full\_case directive that was added to the Example 13 code, and (2) the pre-synthesis simulation will cause errors whenever the enable is 0. because the en=0 case was covered by the default assignment at the top of the always block and not defined within the priority case statement.

```
// priority case example
// Decoder built from four 2-input nor gates
    and two inverters
// The enable input is dangling (has been optimized away)
module code4c
  (output reg [3:0] y,
   input [1:0] a,
   input
                  en);
  always_comb begin
   y = '0;
    priority case ({en,a})
      3'b1_00: y[a] = 1'b1;
      3'b1_01: y[a] = 1'b1;
      3'b1_10: y[a] = 1'b1;
      3'b1_1: y[a] = 1'b1;
    endcase
  end
endmodule
```

Figure 25 - Decoder example with **priority case** modifier

The problem in this example could be solved by adding a case default assignment to set all of the y-bits to 0, but as discussed earlier, adding a case default nullifies the effects of the priority modifier, so there is no point in adding the extraneous priority keyword if you are also going to cancel the modifier with a case default.

Similarly, if you are going to design a priority encoder using if-else statements, almost all priority encoder designs depend on overlapping if-else-if tests, and most RTL-coded priority encoders end with an else statement, again nullifying the priority modifier. If the priority modifier is replaced with a unique modifier, simulation with overlapping case items would cause run-time errors.

At the time this paper was written, the author was of the opinion that most case statements can be reasonably and safely coded using the unique modifier and adding the case default. This is the simulation equivalent of a parallel\_case assertion but not a full\_case assertion. This opinion may change over time as more designs and experiments are conducted and the author welcomes feedback from other users on their experiences using the unique and priority case-statement modifiers.

# 13.0 Actual parallel\_case design problem

One consultant shared the experience where parallel\_case was added to the Verilog code for a large ASIC design to remove stray priority encoders and infer a smaller and faster design. The Verilog case statement was coded as a priority encoder and all RTL simulations worked correctly. Unfortunately, the gate-level design without priority encoder did not function correctly and the gate-level simulations did not catch the problem. This ASIC had to be re-designed,

costing \$100,000's of actual dollars, delayed product release, and unknown lost dollars for being months late to market.

# 14.0 Summary of guidelines and conclusions

Guideline: Code all intentional priority encoders using if-else-if statements. It is easier for the typical design engineer to recognize a priority encoder when it is coded as an if-else-if statement.

**Guideline:** Coding with case statements is recommended when a truth-table-like structure makes the Verilog code more concise and readable.

Guideline: In general, do not use full\_case parallel\_case directives with any Verilog case statements.

Guideline: Replace the older full\_case parallel\_case directives with new and safer SystemVerilog priority and unique directives.

Guideline: Educate (or fire) any employee or consultant who routinely adds full\_case parallel\_case to all case statements in their Verilog code.

Guideline: Do not use the priority and unique directives with all SystemVerilog case statements and if-else-if statements.

Other exceptions and guidelines will surely be discovered as engineers use the new SystemVerilog modifiers. These exceptions and guidelines will be added to later versions of this paper and will be freely downloadable from the <a href="www.sunburst-design.com/papers">www.sunburst-design.com/papers</a> web site. Please email the author if you believe you have other additional useful coding guidelines.

Conclusion: full\_case and parallel\_case directives are most dangerous when they work!

Conclusion: the SystemVerilog priority and unique case statement modifiers are safe replacements for the full\_case parallel\_case evil twins, but an engineer still needs to know how they work and when their usage is most advantageous. The SystemVerilog priority and unique keywords should not be indiscriminately used with all case statements.

The SystemVerilog priority and unique keywords are not only valuable case statement modifiers, they are also valuable simulation, synthesis and formal verification *assertions*.

It is exciting to see so many SystemVerilog features already implemented by both VCS and Design Compiler. VCS now recognizes and reports run-time errors when the priority and unique design assertions are violated, and DC recognizes the new priority and unique modifiers, allowing for safe full\_case parallel\_case equivalent synthesis optimization.

## 15.0 Case statement definitions

To fully understand how the "full\_case parallel\_case" directives work, a common set of terms is needed to describe the different parts of a case statement. This section defines a common set of terms that are used to describe case statement functionality throughout the rest of the paper.

#### 15.1 Case statement

In Verilog, a case statement includes all of the code between the Verilog keywords, case (casez, casex) and endcase.

A case statement is a select-one-of-many construct that is roughly equivalent to an if-else-if statement. The general case statement in Figure 26 is equivalent to the general if-else-if statement shown in Figure 27.

```
case (case_expression)
  case_item1 : case_item_statement1;
  case_item2 : case_item_statement2;
  case_item3 : case_item_statement3;
  case_item4 : case_item_statement4;
  default : case_item_statement5;
endcase
```

Figure 26 - Case Statement - General Form

```
if (case_expression === case_item1) case_item_statement1;
else if (case_expression === case_item2) case_item_statement2;
else if (case_expression === case_item3) case_item_statement3;
else if (case_expression === case_item4) case_item_statement4;
else case_item_statement5;
```

Figure 27 - If-else-if Statement - General Form

As can be seen from the coding syntax shown in Figure 26 and Figure 27, when testing against a common expression, case statements offer a very nice and concise shorthand for describing the same problem.

#### 15.2 Case statement header

A case statement header consists of the case (casez, casex) keyword followed by the case expression, usually all on one line of code.

Typically, when adding full\_case or parallel\_case directives to a case statement, the directives are added as a comment immediately following the case expression at the end of the case statement header and before any of the case items on subsequent lines of code.

```
// Synopsys comment
case (case_expression) // synopsys full_case parallel_case
```

```
// Synplicity comment
case (case_expression) // synthesis full_case parallel_case
...
```

The IEEE Verilog-2002 RTL Synthesis Standard[8] modified the way these directives should be added to a case statement. Instead of using comments to express a synthesis directive, the directives were added to attributes prior to the case statement.

```
(* synthesis, full_case, parallel_case *) case (case_expression) ...
```

The attribute method for adding these directives meant that tools did not have to parse Verilog comments to find synthesis directives. A properly coded design would allow tools to treat all comments as comments and not as potential commands.

Even though an IEEE Verilog-2002 RTL Synthesis-compliant tool now reads synthesis directives within attributes instead of directives within comments, the evil and flawed nature of the full\_case parallel\_case "evil twins" still exist. The attribute method for adding directives did not correct the fundamental problems related to these directives.

# 15.3 Case expression

A Verilog case expression is the expression enclosed between parentheses immediately following the case keyword. In Verilog, a case expression can either be a constant, such as 1'b1 (one bit of 1, or "true"), it can be an expression that evaluates to a constant value, or most often it is a bit or vector of bits that are used to compare against case items.

#### 15.4 Case item

The case item is the bit, vector or Verilog expression that is used to compare against the case expression.

Unlike other high-level programming languages such as 'C', the Verilog case statement includes an implied break statement. The first case item that matches the current case expression causes the corresponding case item statement to be executed and then all of the rest of the case items are skipped (ignored) for the current pass through the case statement.

## 15.5 Case item statement

A case item statement is one or more Verilog statements that are executed if the corresponding case item matches the current case expression.

Unlike VHDL, Verilog case items can themselves be expressions. To simplify parsing of Verilog source code, Verilog case item statements must be enclosed between the keywords begin and end if more than one statement is to be executed for a selected case item. This is one

of the few places were Verilog syntax requirements are considered by VHDL-literate engineers to be too verbose.

#### 15.6 Case default

An optional case default can be included in the case statement to indicate what actions to perform if none of the defined case items matches the current case expression. It is good coding style to place the case default last, even though the Verilog standard does not require it.

#### **15.7** Casez

In Verilog there is a casez statement, a variation of the case statement that permits z and? values to be treated during case-comparison as "don't care" values. z and? are treated as a don't care if they are in the case expression *and/or* if they are in the case item.

More information on the precautions that should be taken when using casez for RTL modeling and synthesis are detailed in a paper authored by myself and Don Mills in 1999[5].

Guideline: Exercise caution when coding synthesizable models using the Verilog casez statement.

Coding Style Guideline: When coding a case statement with "don't cares," use a casez statement and use? characters instead of z characters in the case items to indicate "don't care" bits.

#### 15.8 Casex

In Verilog there is a casex statement, a variation of the case statement that permits z,? and x values to be treated during comparison as "don't care" values. x, z and? are treated as a don't care if they are in the case expression and/or if they are in the case item.

More information on the dangers of using casex for RTL modeling and synthesis are detailed in a paper authored by myself and Don Mills in 1999[5]. In our 1999 paper, Don and I gave a guideline to not use casex in RTL models. To date, I have not found a good safe usage for the casex statement in either RTL, behavioral or testbench coding so my current guideline is:

Guideline: Do not use the casex statement.

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#### **Author & Contact Information**

Cliff Cummings, President of Sunburst Design, Inc., is an independent EDA consultant and trainer with 23 years of ASIC, FPGA and system design experience and 13 years of Verilog, synthesis and methodology training experience.

Mr. Cummings, a member of the IEEE 1364 Verilog Standards Group (VSG) since 1994, is the only Verilog and SystemVerilog trainer to co-develop and co-author the IEEE 1364-1995 & IEEE 1364-2001 Verilog Standards, the IEEE 1364.1-2002 Verilog RTL Synthesis Standard and the Accellera SystemVerilog 3.0, 3.1 & 3.1a Standards. Mr. Cummings is also an active member of the IEEE P1364-2005 Verilog and IEEE P1800-2005 SystemVerilog Standards Groups.

Mr. Cummings holds a BSEE from Brigham Young University and an MSEE from Oregon State University.

Sunburst Design, Inc. offers Verilog, Verilog Synthesis and SystemVerilog training courses. For more information, visit the www.sunburst-design.com web site.

Email address: cliffc@sunburst-design.com

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