

$$\frac{1}{2} n \log_2 n$$

steps

Page No.

Date: 11

$$n = 1024 = 2^{10}$$

$$n^2 > 1000000 \rightarrow 1024 \times 1024 \text{ times}$$

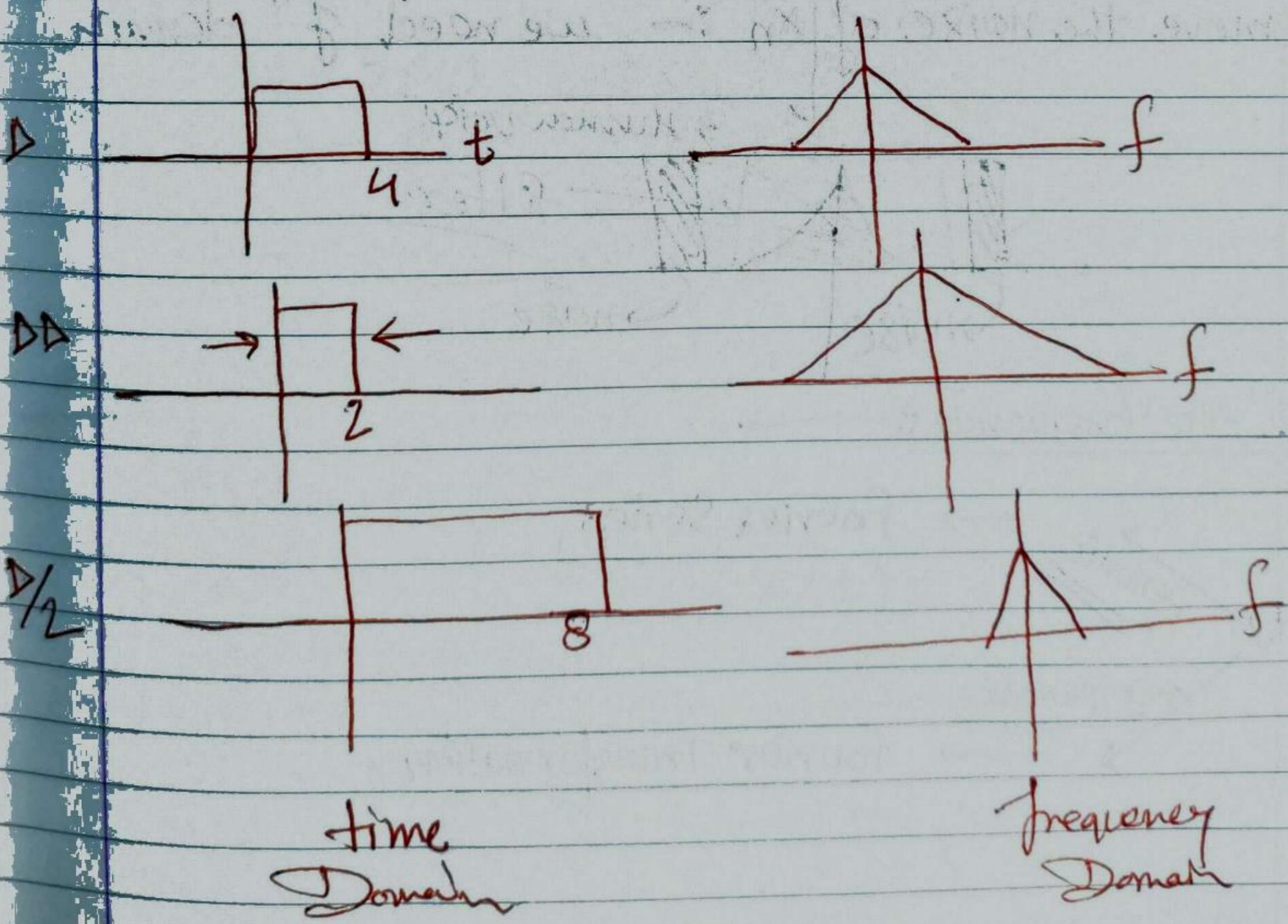
$$\frac{1}{2} n \log_2 n = (1024) \frac{10}{2} \checkmark [5 \times 1024] \checkmark \text{times}$$

Fatima

Applications

frequency means how fast a signal is changing

Audio) Hi, this is fatima, please call me back!



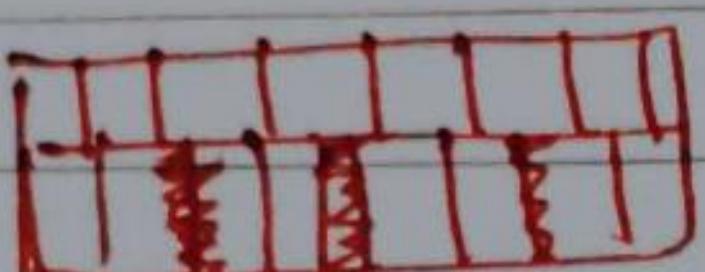
time
Domain

frequency
Domain

ex: $x_1(t)$

$x_n(t)$ (baby crying)

$x_2(t)$

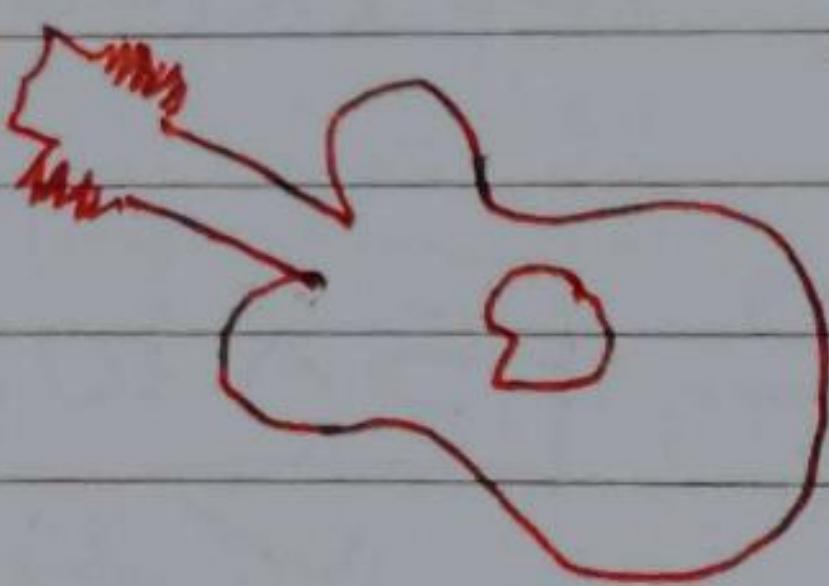


a_1

a_2

a_3

$u_3(t)$



\otimes

\otimes

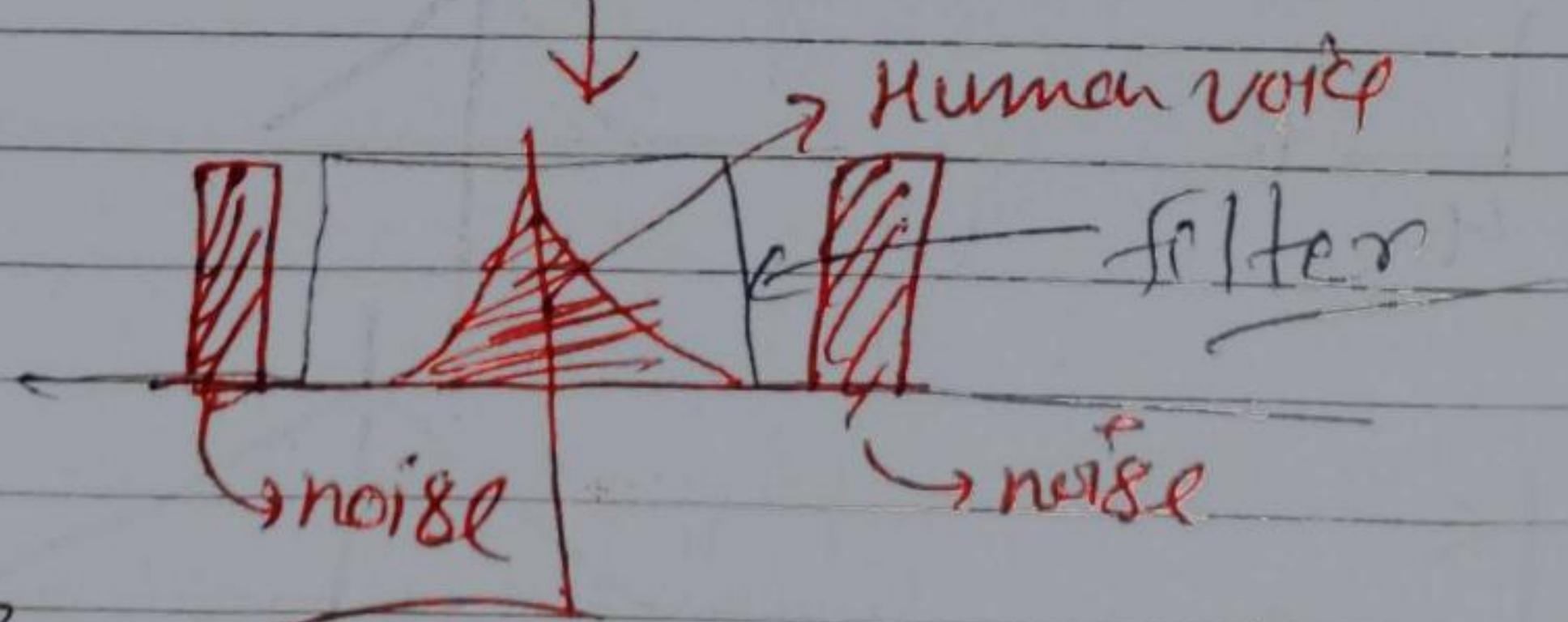
$$y(t) = \underline{a_1 u_1(t)} + \underline{a_2 u_2(t)} + \underline{a_3 u_3(t)}$$

1 1 1
0 | B
| | 3

with noise

$$y(t) = \underline{a_1(u_1(t) + x_n(t))} + \underline{a_2 u_2(t)} + \underline{a_3 u_3(t)}$$

to remove the noise of $x_n \rightarrow$ we need f'' domain



Time to Frequency

→ Fourier Series

Periodic

not periodic

→ Fourier Transformation

Complex Numbers

Cartesian

Polar

$$\text{Cartesian} \\ z = x + yj$$

$$z = 3j$$

$$z = -2$$

$$z = 4 - 3j$$

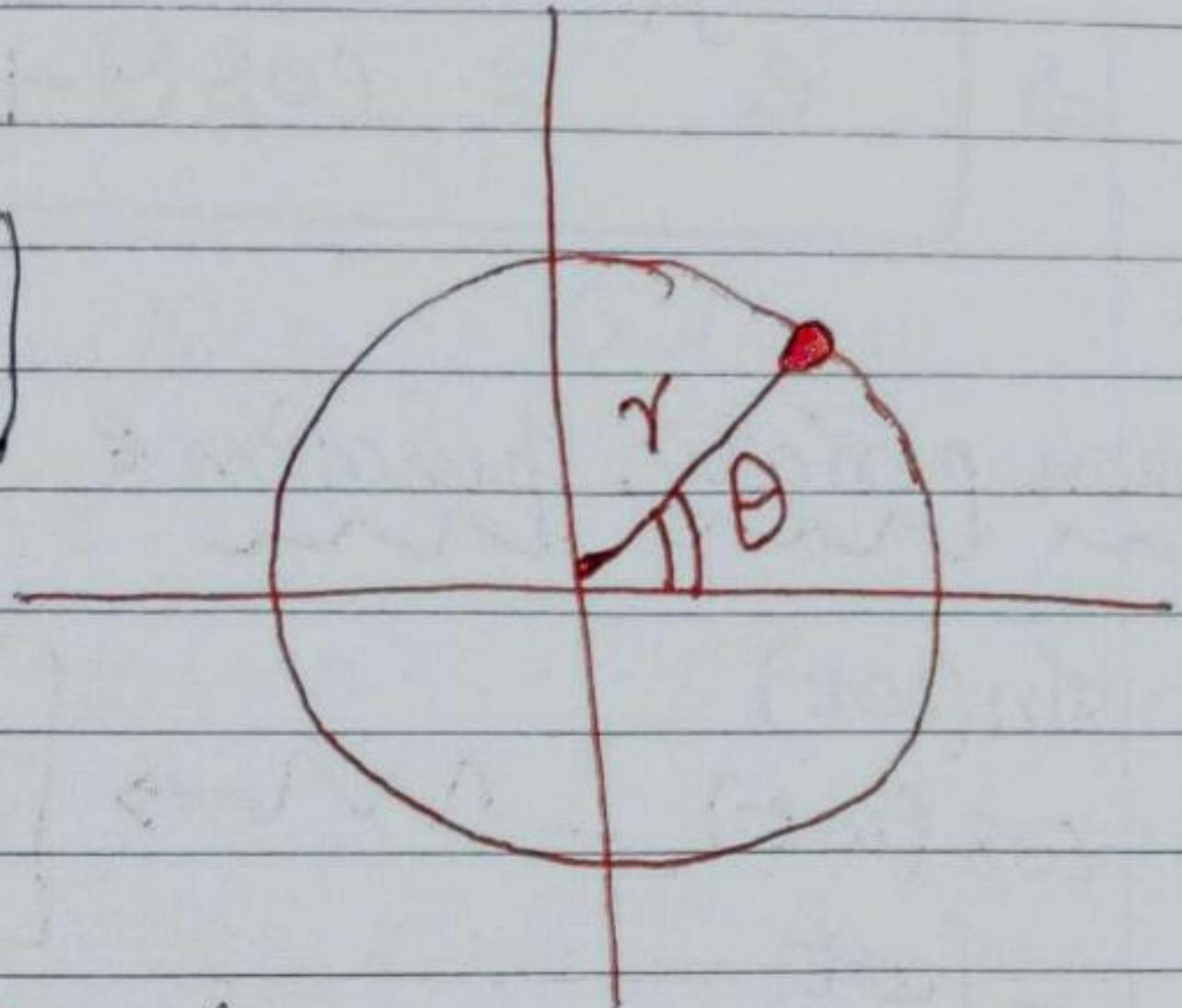
$$z = r e^{j\theta}$$

$r \rightarrow$ radius

$\theta \rightarrow$ angle

$$z = 2e^{j\pi/3}$$

= radius is 2; $\theta = \pi/3$

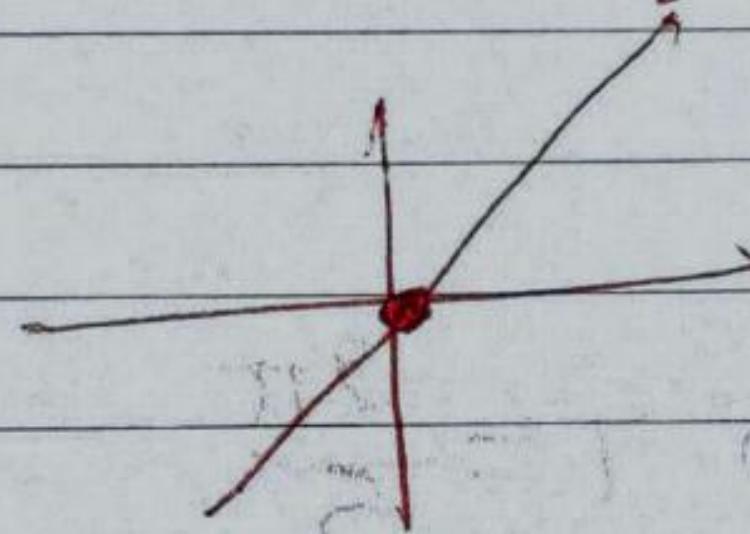


Signal is odd or even

$$\text{ex: } x(t) = 2t$$

Calculate

$$x(-t) = 2(-t) = -2t$$



symm around origin

$$x(t) = 2(t) \quad \text{mean}$$

$$-x(-t) = 2(t) \quad \underline{x(t) = x(-t)}$$

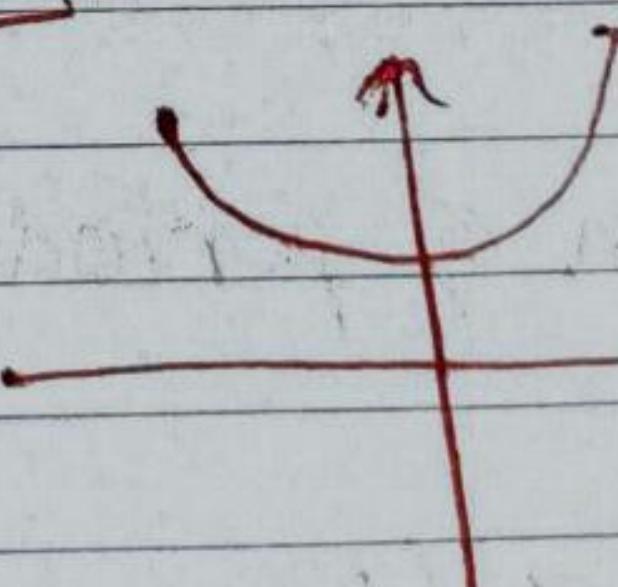
Odd

$$\text{ex: } x(t) = t^2 + 1$$

$$x(-t) = (-t)^2 + 1$$

$$x(t) = t^2 + 1 \Rightarrow x(t) = x(-t)$$

$$x(-t) = t^2 + 1$$



symm around

y-axis

Even

By Euler:

$$x(t) = e^{j\omega t}$$

18 $\Rightarrow e^{j\theta} = \cos\theta + j\sin\theta$

Page No.

Date: 11

common periodic function

$$\left. \begin{array}{l} \# \sin(\omega t) \\ \# \cos(\omega t) \\ \# e^{j\omega t} \end{array} \right\} \rightarrow T = \frac{2\pi}{\omega}$$

↳ period.

ex: $x(t) = \cos 5t$

(a) so period is $\Rightarrow T = \frac{2\pi}{5}$

(b) $x(t) = \sin(\pi t + \frac{\pi}{4})$

$$\omega = \pi \quad \begin{matrix} \leftarrow T \\ \uparrow \text{shift} \end{matrix}$$
$$T = \frac{2\pi}{\pi} = 2$$

(c) combination of periodic signals

$$x_3(t) = \underbrace{x_1(t)}_{T_1} + \underbrace{x_2(t)}_{T_2}$$

$x_2(t)$ is periodic, if $\frac{T_1}{T_2}$ is rational.

if (c) is true \Rightarrow

$$T_3 \rightarrow \text{lcm}(T_1, T_2)$$

↳ least common multiple

Exe $\sin(\pi t) + 8\sin(t)$ is periodic or not?

$T_1 = \frac{2\pi}{\pi} = 2; \quad T_2 = \frac{2\pi}{1} = 2\pi$

Page No. _____
Date: 11

$$\frac{T_1}{T_2} = \frac{2}{2\pi} = \cancel{\frac{1}{\pi}} \text{ Not periodic!}$$

Exe $\cos(6\pi t) + 8\sin(30\pi t)$

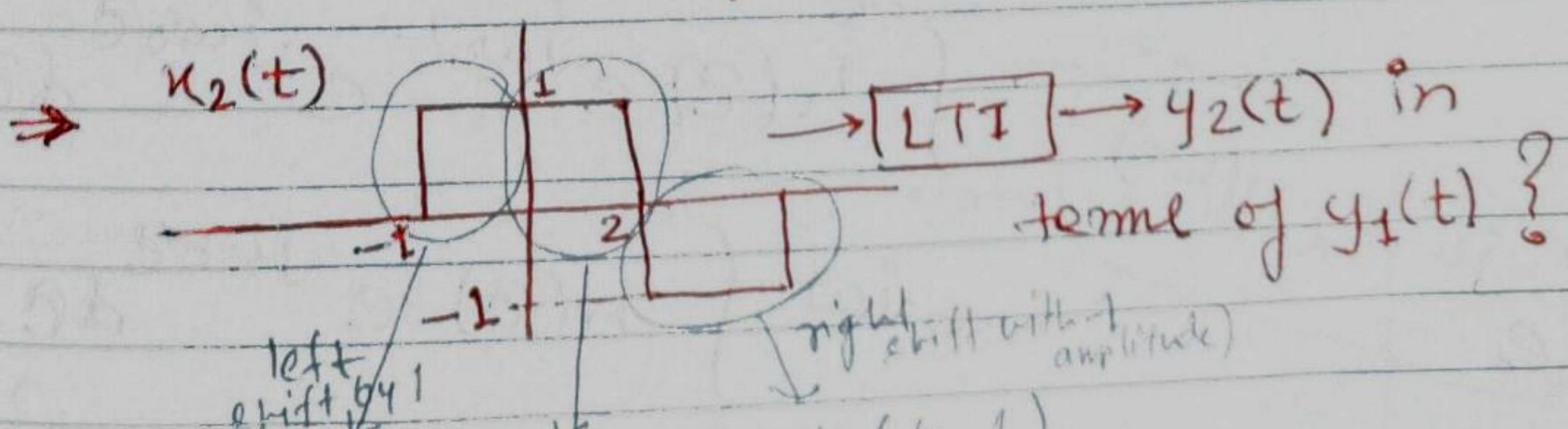
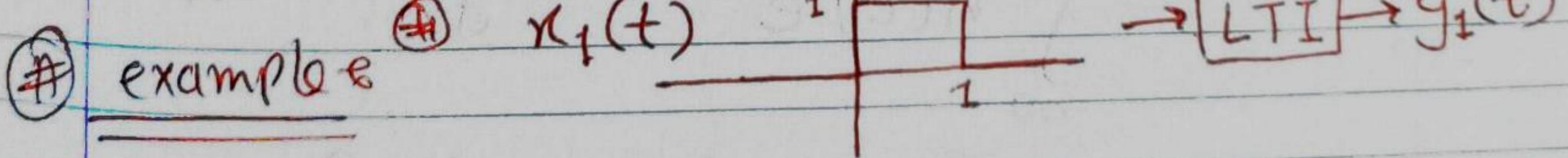
$$T_1 = \frac{2\pi}{6\pi} = \frac{1}{3}; \quad T_2 = \frac{2\pi}{30\pi} = \frac{1}{15}$$

$$\frac{T_1}{T_2} = \frac{1/3}{1/15} = 5;$$

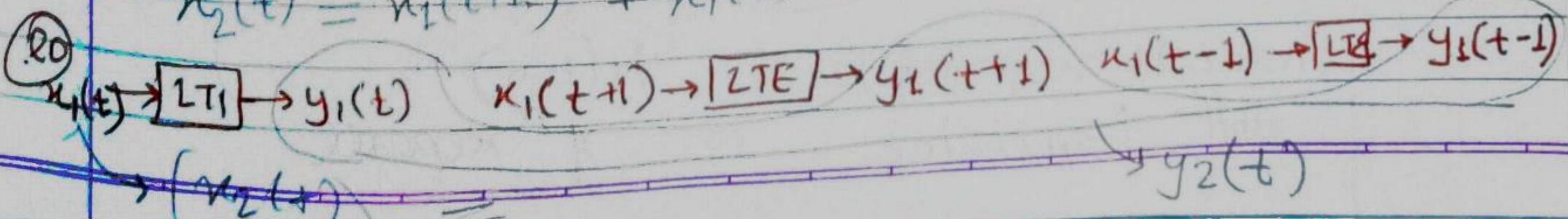
Elementary Signals

LTI : Linear & Time Invariant

System is time invariant if time shift in input ~~is~~ results in the same time shift in the o/p.



$$x_2(t) = x_1(t+1) + x_1(t) - x_1(t-1)$$



Frequency domain \rightarrow Fourier Series

Page No.

Date:

$$x(t) = \sum_{k=-\infty}^{\infty} c_k e^{jk\omega_0 t} \quad \text{FS synthesis equations}$$

$$c_k = \frac{1}{T} \int_T x(t) e^{-jk\omega_0 t} dt \quad \text{FS analysis equations}$$

fourier series coeff.

T : fundamental period of $x(t)$

$\omega_0 \Rightarrow$ fundamental frequency $\omega_0 = 2\pi/T$

$x(t) = e^{j\omega_0 t} \xrightarrow{\text{[LTI]}} y(t) = x(t) * h(t)$

$$y(t) = \int_{-\infty}^{\infty} h(\tau) x(t-\tau) d\tau \quad t \rightarrow t-\tau$$

$$= \int_{-\infty}^{\infty} h(\tau) e^{j\omega_0(t-\tau)} d\tau$$

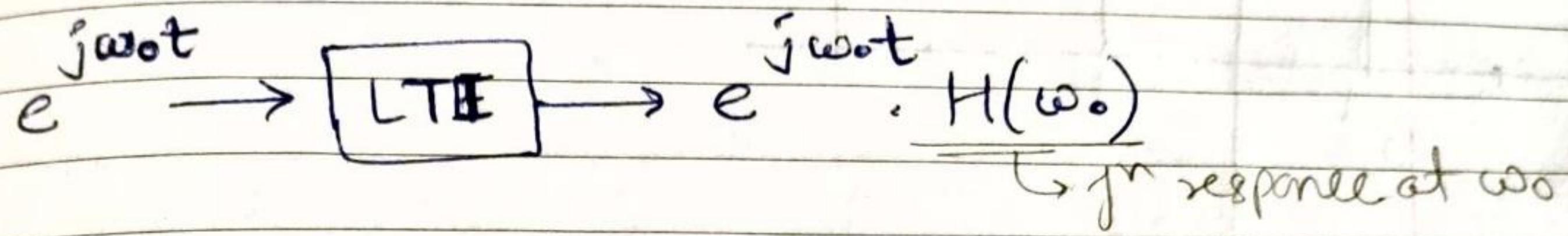
$$= \int_{-\infty}^{\infty} h(\tau) e^{j\omega_0 t} e^{-j\omega_0 \tau} d\tau$$

$$e^{j\omega_0 t} H(\omega_0) = \int_{-\infty}^{\infty} h(\tau) e^{-j\omega_0 \tau} d\tau$$

how LTI responded to
diff frequencies \leftarrow $H(\omega_0)$
 \downarrow \uparrow response

80 rewriting:

$$x(t) = \sum_{k=-\infty}^{\infty} c_k e^{jk\omega_0 t}$$



Conclusion:

- ① LTI system doesn't change the frequency.

Properties of LTI

$$\sum_{k=0}^{\infty} c_k e^{jk\omega_0 t} \xrightarrow{\text{LTI}} \sum_{k=-\infty}^{\infty} c_k e^{jk\omega_0 t} \cdot H(k\omega_0)$$

$x(t)$ ↳ f response at ω_0

$y(t)$

Fourier Series

$$x(t) = \sum_{k=-\infty}^{\infty} c_k e^{jk\omega_0 t}$$

$$c_k = \frac{1}{T} \int_{-T/2}^{T/2} x(t) e^{-jk\omega_0 t} dt$$

example:

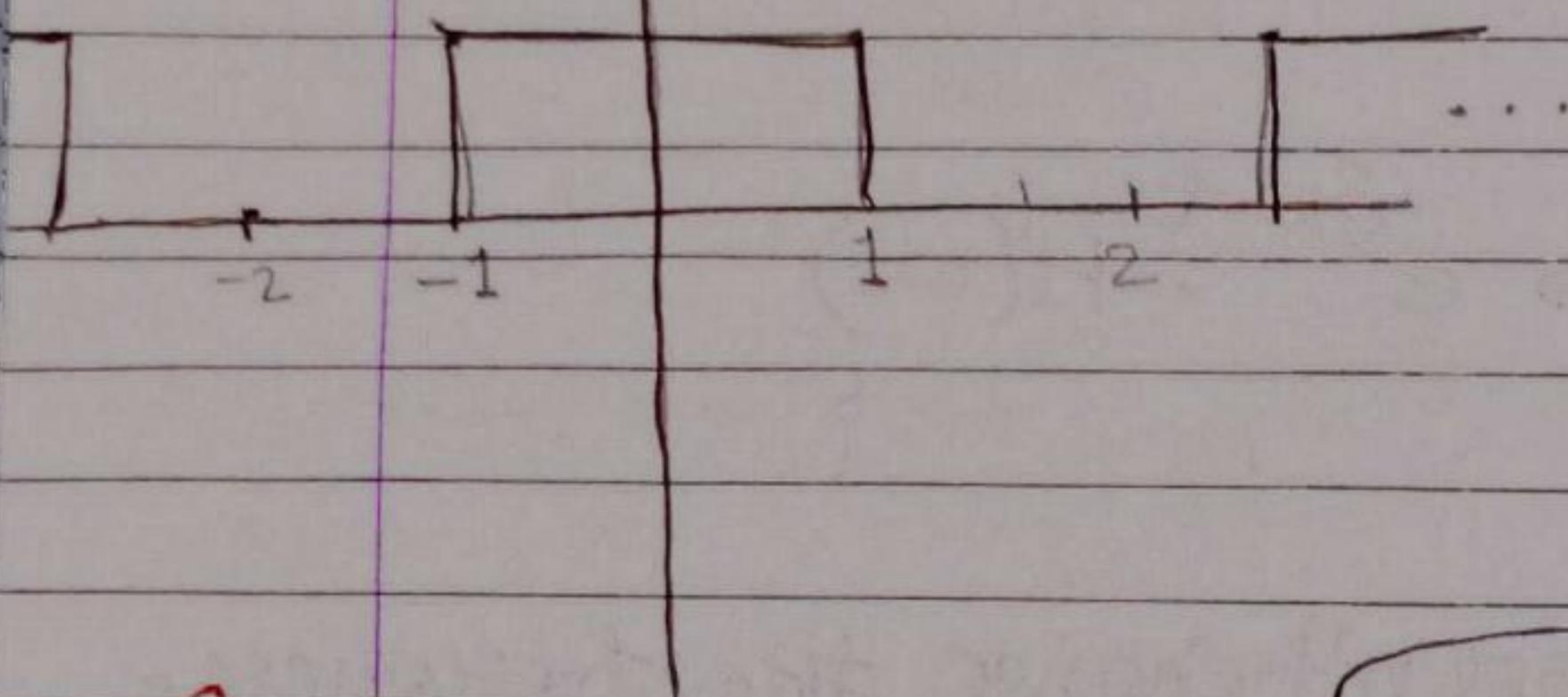
① find c_k for these periodic signals:-

Page No.

Date: 11

$x(t)$

$c_k = ?$



Soln

$$T=4 \rightarrow \omega_0 = \frac{2\pi}{T} = \frac{2\pi}{4} = \frac{\pi}{2}$$

(write)

$$c_k = \frac{1}{T} \int_T x(t) e^{-j k \omega_0 t} dt = \frac{1}{4} \int_{-2}^2 x(t) e^{-j k \omega_0 t} dt$$

Remark

$$\int e^{at} dt = \frac{1}{a} e^{at}$$

$$= \frac{1}{4} \int_{-1}^1 e^{-j k \omega_0 t} dt$$

($\because x(t)$ is 0 in this interval)
& $x(t)$ is 0 $-2 \rightarrow -1, 1 \rightarrow 2$)

Remarks

$$\begin{aligned}
 \sin a & c_k = \frac{1}{4} \left[\frac{1}{-j k \omega_0} e^{-j k \omega_0 t} \right]_{-1}^1 \\
 & = -\frac{1}{4 j k \omega_0} \left(e^{j k \omega_0} - e^{-j k \omega_0} \right) \\
 & = \frac{1}{4 j k \omega_0} \left(e^{j k \omega_0} - e^{-j k \omega_0} \right) \\
 & = \frac{1}{2 k \omega_0} \left(\frac{e^{j k \omega_0} - e^{-j k \omega_0}}{2j} \right) \\
 & = \frac{1}{2 k \omega_0} \sin(k \omega_0)
 \end{aligned}$$

Le

fin

$$c_k = \frac{\sin(k\omega_0)}{2k\omega_0} = \frac{\sin\left(k\frac{\pi}{2}\right)}{2k\cdot\frac{\pi}{2}} =$$

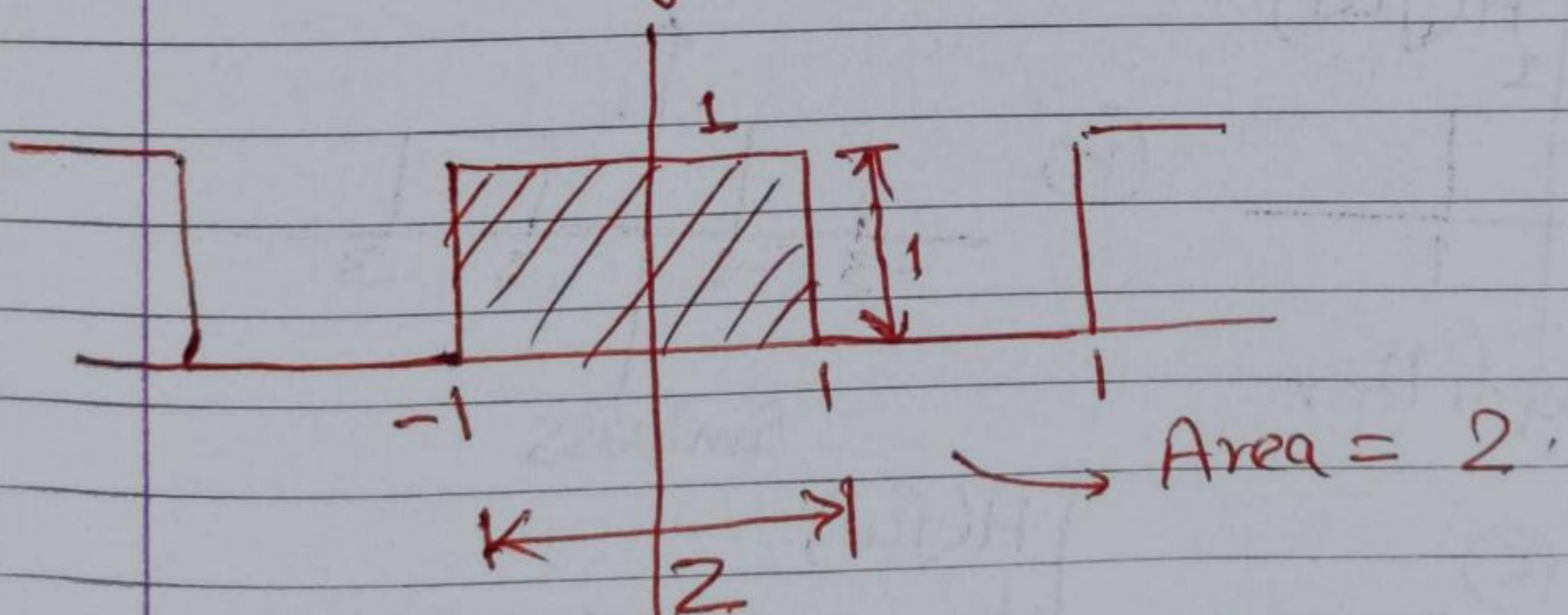
$$\boxed{c_k = \frac{\sin\left(k\frac{\pi}{2}\right)}{k\cdot\pi}} \quad \begin{matrix} \text{Valid only} \\ k \neq 0 \end{matrix} \quad \begin{matrix} \text{when } k \neq 0. \end{matrix}$$

$$c_k = \frac{1}{T} \int_{-T/2}^{T/2} x(t) e^{-jk\omega_0 t} dt \xrightarrow{k=0} c_0 = \frac{1}{T} \int_{-T/2}^{T/2} x(t) dt$$

$$c_0 = \frac{1}{T} \int_{-T/2}^{T/2} x(t) dt$$

denotes

area under
the curve over
one period



$$\text{so: } c_0 = \frac{1}{T} \cdot 2 = \frac{1}{2}.$$

find answer:

$$c_k = \begin{cases} \frac{\sin\left(k\frac{\pi}{2}\right)}{k\cdot\pi} & k \neq 0 \\ \frac{1}{2} & k = 0 \end{cases}$$

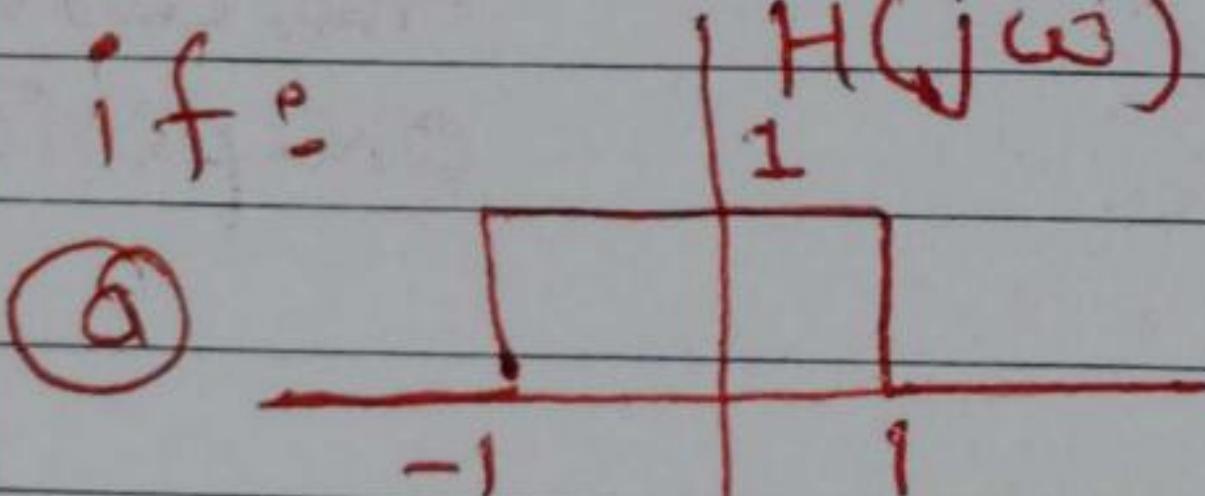
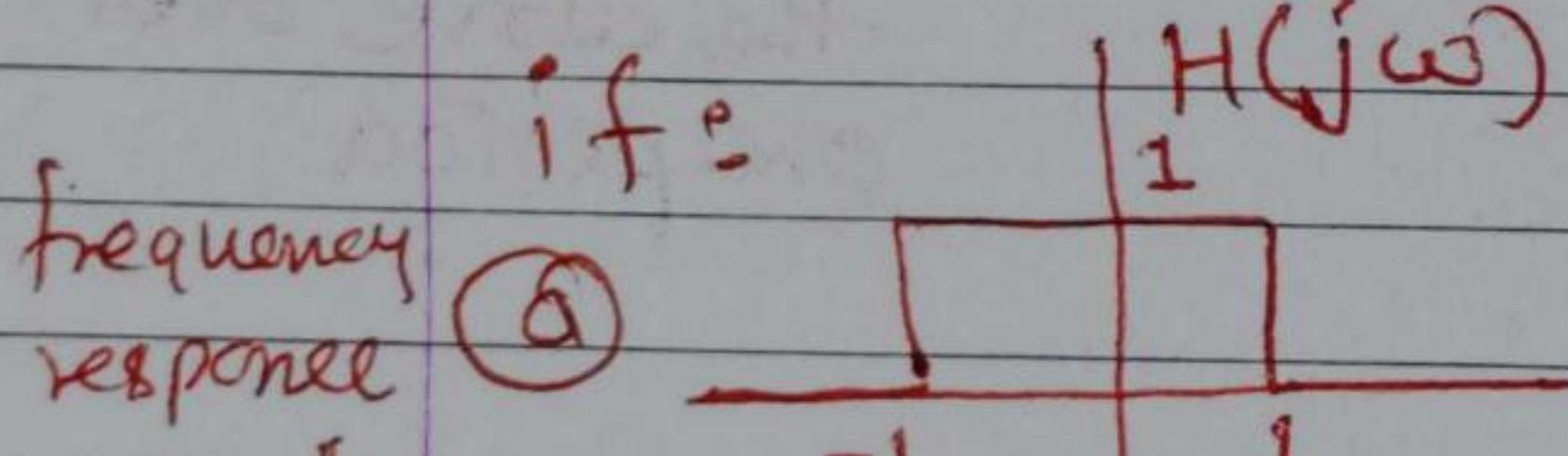
Note 6

- ① if $x(t)$ is even, c_k is even $\Rightarrow c_k = \bar{c}_k$
- ② if $x(t)$ is odd, c_k is odd $\Rightarrow c_k = -\bar{c}_{-k}$
- ③ if $x(t)$ is real, $\Rightarrow c_k = \bar{c}_{-k}^*$
- ④ if $x(t)$ is even and real $\Rightarrow c_k$ is real
- ⑤ if $x(t)$ is odd & real $\Rightarrow c_k$ is imaginary.

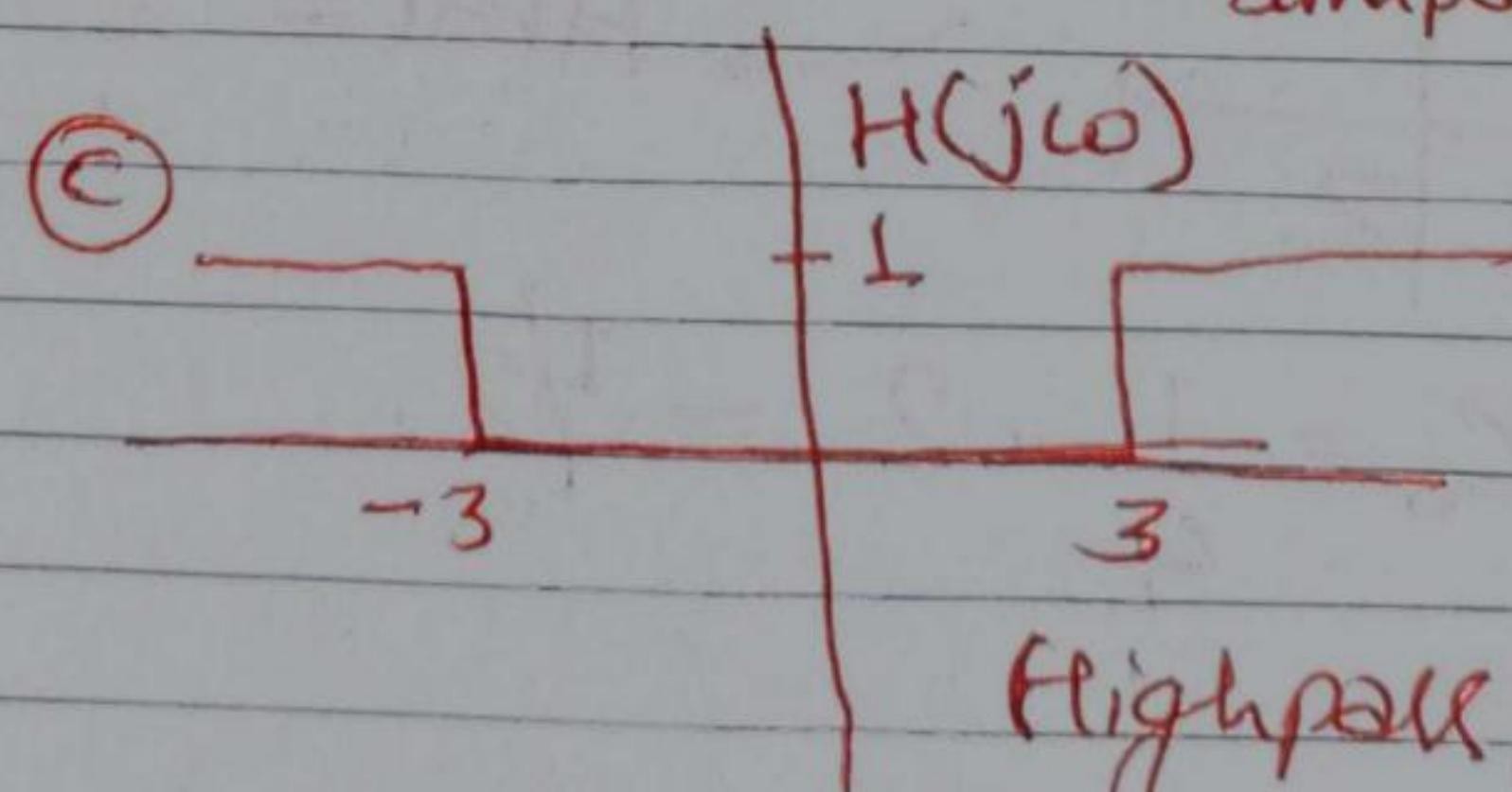
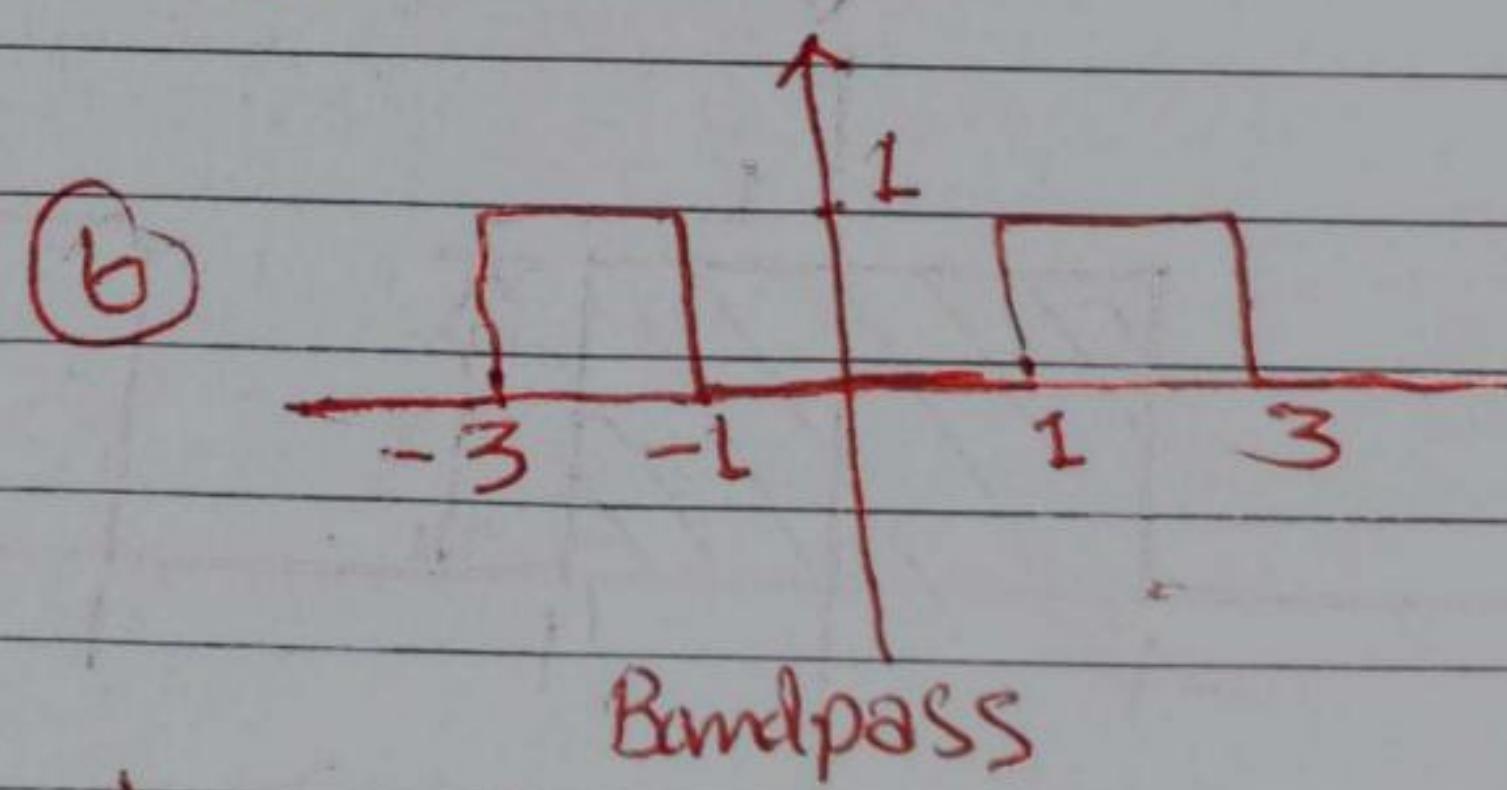
Filtering (Periodic Signals) :-

Example Signal $x(t) = 1 + 2\cos(2t) + 2\cos(4t)$

is applied to LTI system ;
what is the output ? $y(t)$?



lowpass filter



note :

Step ① Represent input into linear combination of complex exponentials.

$$x(t) = \sum_{k=-\infty}^{\infty} c_k e^{jk\omega_0 t}$$

Remark:

$$\cos(a) = \frac{e^{ja} + e^{-ja}}{2}$$

Page No. 79

Date: 11

mean \rightarrow low \rightarrow mid \rightarrow High \uparrow

$$\text{so: } x(t) = 1 + 2\cos(2t) + 2\cos(4t)$$

$$x(t) = 1 + 2\left(\frac{e^{j2t} + e^{-j2t}}{2}\right) + 2\left(\frac{e^{j4t} + e^{-j4t}}{2}\right)$$

$$\underline{x(t)} = e^{j0t} + e^{j2t} + e^{-j2t} + e^{j4t} + e^{-j4t} \rightarrow \boxed{\text{LTI}} \downarrow y(t)$$

As we know

$$\boxed{\text{LTI}} \quad e^{j\omega_0 t} \rightarrow \boxed{\text{LTI}} \rightarrow e^{j\omega_0 t} H(j\omega_0)$$

Op is same

~~$y(t) = e^{j0t} + e^{j2t}$~~

$$\underline{y(t)} = e^{j0t} H(j0) + e^{j2t} H(j2) + e^{-j2t} H(-j2) + e^{j4t} H(j4) + e^{-j4t} H(-j4)$$

frequency response

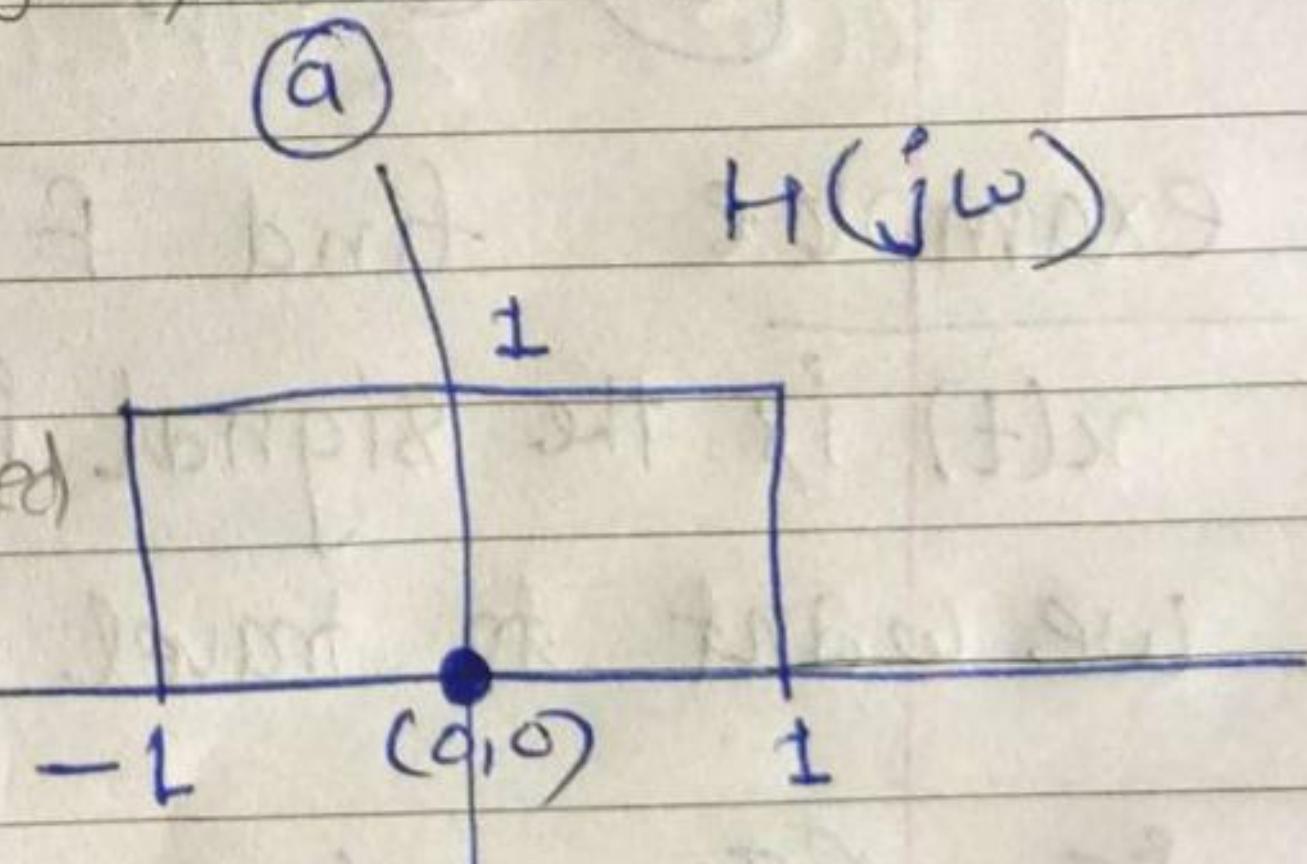
$$H(0) \Rightarrow (0,0) \rightarrow \text{is, 1}$$

so LTI passes this signal

similarly for $H(2) = 0$; \rightarrow signal is not passed

for $H(-2), H(4), H(-4) = 0$;

signal is not passed;



(a) Output: $y(t) = e^{j0t} = 1$

similarly

(b) $y(t) = 2\left(\frac{e^{j2t} + e^{-j2t}}{2}\right) \Rightarrow 2\cos(2t)$

(c) $y(t) = \frac{1}{2}\left(\frac{e^{j4t} + e^{-j4t}}{2}\right) = 2\cos(4t)$

Fourier Transform

non periodic

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{j\omega t} d\omega$$

$\xrightarrow{\text{IFT}}$

$$X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

$\xrightarrow{\text{FT}}$

fourier Series

$$x(t) = \sum_{k=-\infty}^{\infty} c_k e^{jk\omega_0 t}$$

$$c_k = \frac{1}{T} \int_{-\infty}^{\infty} x(t) e^{-jk\omega_0 t} dt$$

\boxed{T}
periodic

example OR + OG + OB

$X(\omega)$ → weighting system for diff colors

→ ω → colors (diff fundamental colors)

- Application of FT:-
- ① filtering the frequen
 - ② Object detection by Radar.
 - ③ Telecommunicat (modulation & Demodulation).

(4) MRI

(5) Quantum mechanics

example find FT for $x(t) = \delta(t-2)$

$x(t)$ is the signal in time domain by using FT

we want to travel to frequency domain.

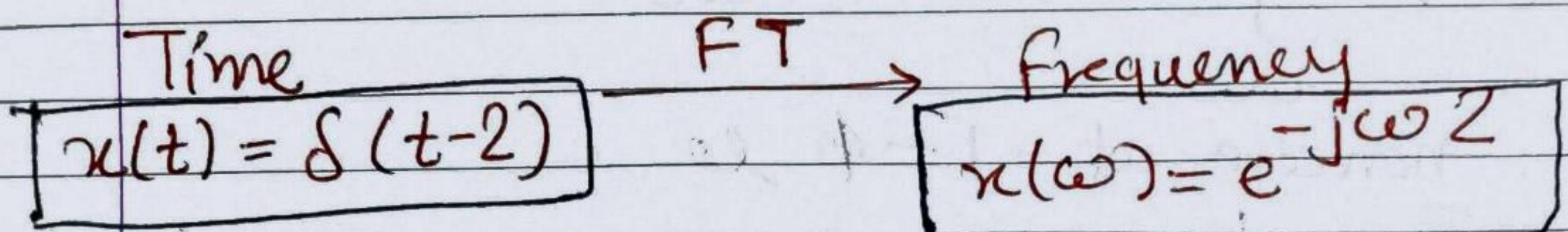
FT:
$$X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

$$= \int_{-\infty}^{\infty} \delta(t-2) e^{-j\omega t} dt$$

Note: $\delta(t-2)$ is non-zero at $t=2$.

sifting property:

$$\text{Ans} \rightarrow x(\omega) = e^{-j\omega 2} \quad \begin{matrix} \text{replace} \\ (t \text{ by } 2) \end{matrix}$$



example: Find IFT for $x(\omega) = 2\pi\delta(\omega-5)$ travel this signal is in frequency domain & we want to time domain using Inverse FT.

Solution:
$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} x(\omega) e^{j\omega t} d\omega$$

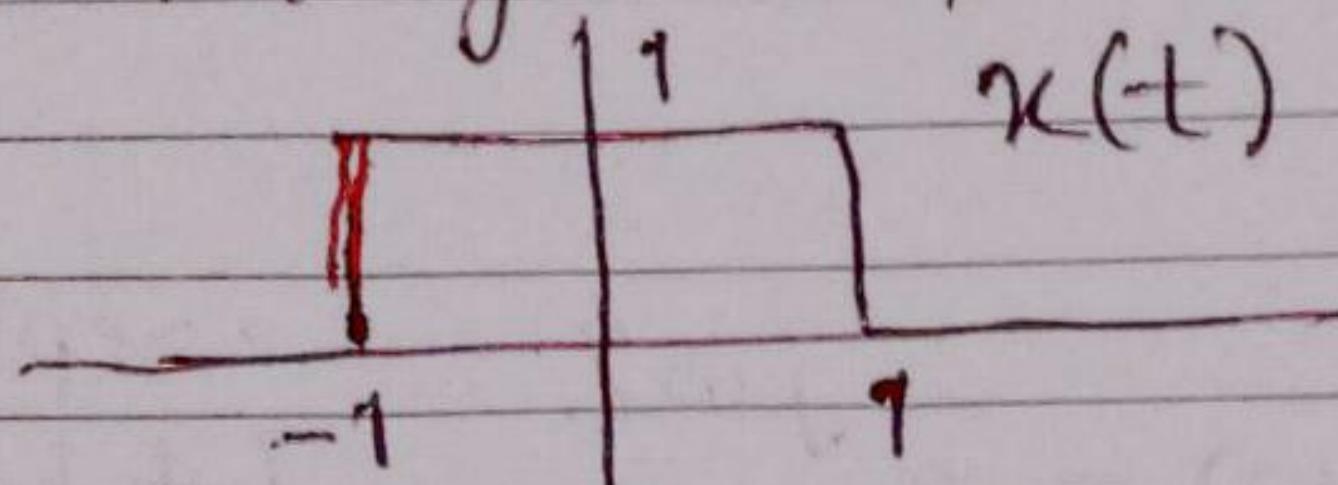
$$= \frac{1}{2\pi} \int_{-\infty}^{\infty} 2\pi\delta(\omega-5) e^{j\omega t} d\omega$$

$$= \int_{-\infty}^{\infty} \underbrace{\delta(\omega-5)}_{\text{non-zero at } \omega=5} e^{j\omega t} d\omega$$

Based on sifting property

$$\text{Ans} \boxed{x(t) = e^{j5t}} \quad (\text{replace } \omega \text{ by } 5)$$

example find FT for this window



$$\text{Sol: } X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

nonzero at $-1 < \omega < 1$

$$= \int_{-1}^1 e^{-j\omega t} dt \quad (x(t) = 1 \text{ for } -1 < t < 1)$$

$$X(\omega) = \frac{1}{-j\omega} [e^{j\omega} - e^{-j\omega}] = \frac{1}{j\omega} [e^{j\omega} - e^{-j\omega}]$$

$$\underline{X(\omega)} = \frac{1}{\omega} \left[\frac{e^{j\omega} - e^{-j\omega}}{2j} \right] = \underbrace{\frac{2}{\omega} \sin(\omega)}_{\text{sinc}(\omega)}$$

$$X(\omega) = 2 \operatorname{sinc}(\omega)$$

Fourier Transform tables

Page No.

Date: 11

Table of pairs

$$s(t) \longleftrightarrow 1$$

$$1 \longleftrightarrow 2\pi \delta(\omega)$$

$$\operatorname{rect}\left(\frac{t}{T}\right) \longleftrightarrow T \operatorname{sinc}\left(\frac{\omega T}{2}\right)$$

$$\frac{B}{\pi} \operatorname{sinc}(Bt) \longleftrightarrow \operatorname{rect}\left(\frac{\omega}{2B}\right)$$

$$e^{-at} u(t) \longleftrightarrow \frac{1}{j\omega + a}$$

$$e^{j\omega_0 t} \longleftrightarrow 2\pi \delta(\omega - \omega_0)$$

$$\cos \omega_0 t \longleftrightarrow \frac{\pi}{2} (\delta(\omega - \omega_0) + \delta(\omega + \omega_0))$$

$$\sin \omega_0 t \longleftrightarrow \frac{\pi}{j} (\delta(\omega - \omega_0) - \delta(\omega + \omega_0))$$

Time domain

frequency domain

Notes: Inverse relation b/w Time & frequency.