

Chapter 16 二端口网络

16.1 二端口概述

16.2 二端口网络的特性

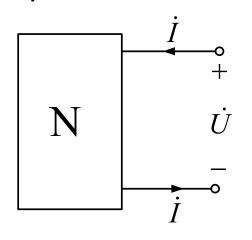
16.3 二端口网络的参数

16.4 二端口网络的电路模型

16.5 二端口网络的相互连接

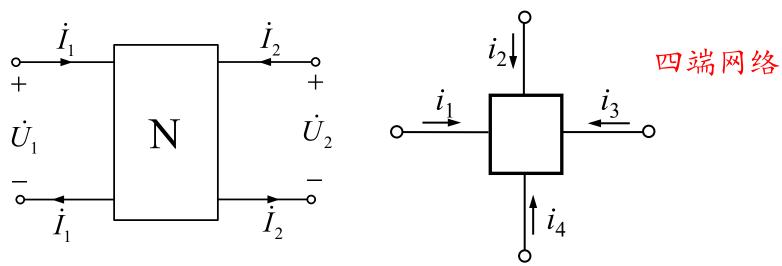
16.1 二端口概述

1. 端口

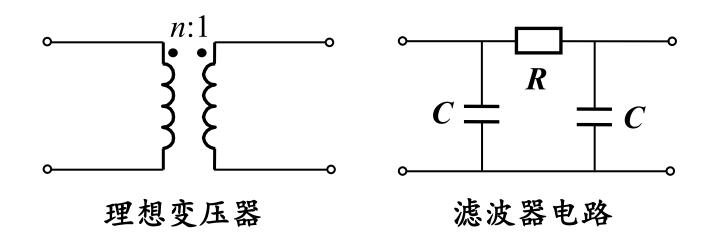


由一对端钮构成,且满足从一个端钮流入的电流等于从另一个端钮流出的电流。

2. 二端口网络 具有两个端口和外部相连的网络



典型二端口网络

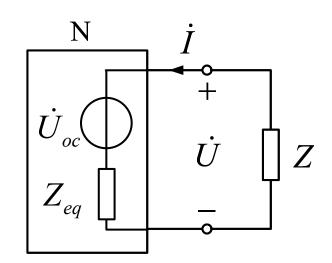


- ✓ 二端口网络大量出现在通信、控制系统和电力系统中, 信号(或能量)从一个端口输入,另一个端口输出。
- ✓ 很多情况下,我们只关心输入、输出端口的电压、电流,而并不关心二端口网络内部的情况,因此可以把二端口网络是为"黑盒子"来分析。
- ✓ 无需弄清网络内部的元件连接关系,只需了解二端口网络的端口特性即可。

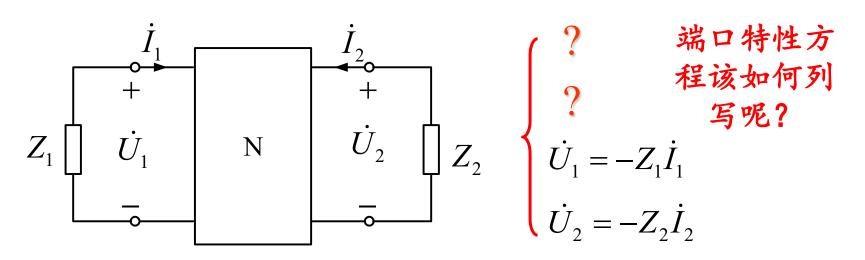
16.2 二端口网络的特性

1. 一端口网络

$$\begin{cases} \dot{U} = \dot{U}_{oc} + Z_{eq} \dot{I} \\ \dot{U} = Z \left(-\dot{I} \right) \end{cases}$$



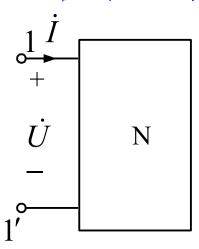
2. 二端口网络 求解4个端口变量需要列写4个方程



3. 含源网络

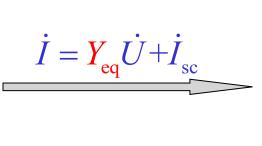
(1) 含源一端口网络

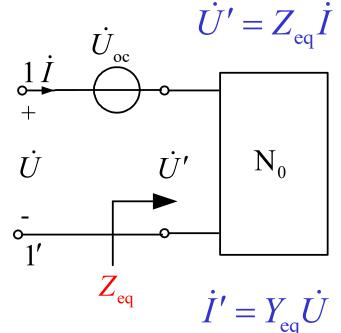
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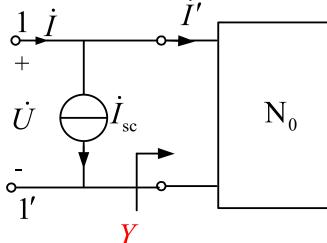


 \dot{U}

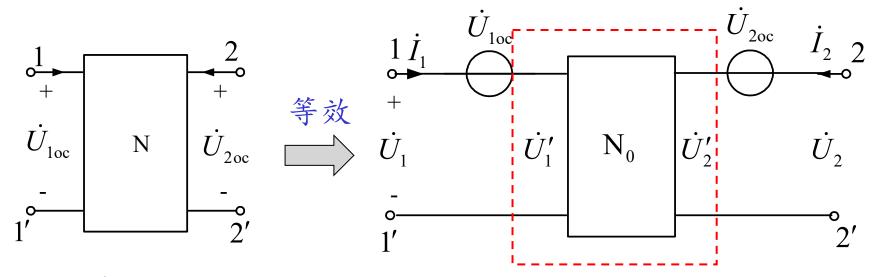
$$\dot{U} = \dot{U}_{\rm oc} + Z_{\rm eq} \dot{I}$$



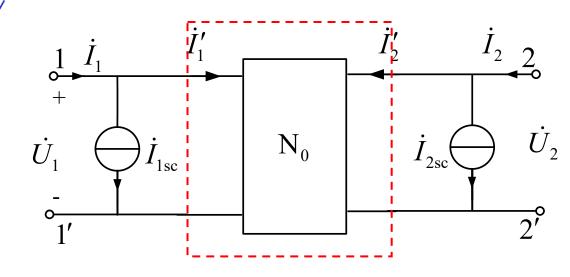




(2) 含源二端口网络 研究松弛二端口网络的端口特性



- ✓ 线性含源二端口网 络,可等效为电压/ 电流源与不含源二 端口网络的连接
- ✓ 不含独立源的二端 口网络称为松弛二 端口网络



六组松弛二端口网络的端口特性方程(___



两个端口 四个变量

阻抗参数方程(Z)

导纳参数方程(Y)

$$\begin{cases} \dot{U}_1 = Z_{11}\dot{I}_1 + Z_{12}\dot{I}_2 \\ \dot{U}_2 = Z_{21}\dot{I}_1 + Z_{22}\dot{I}_2 \end{cases}$$

$$\begin{cases} \dot{I}_1 = Y_{11}\dot{U}_1 + Y_{12}\dot{U}_2 \\ \dot{I}_2 = Y_{21}\dot{U}_1 + Y_{22}\dot{U}_2 \end{cases}$$

同一类型变量 位于不同端口

混合参数方程(H和G)

$$\begin{cases} \dot{U}_1 = h_{11}\dot{I}_1 + h_{12}\dot{U}_2 \\ \dot{I}_2 = h_{21}\dot{I}_1 + h_{22}\dot{U}_2 \end{cases}$$

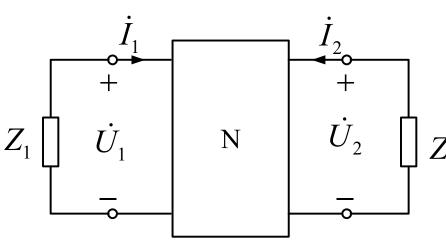
$$\begin{cases} \dot{U}_1 = h_{11}\dot{I}_1 + h_{12}\dot{U}_2 & \qquad \\ \dot{I}_2 = h_{21}\dot{I}_1 + h_{22}\dot{U}_2 & \qquad \\ \dot{U}_2 = g_{21}\dot{U}_1 + g_{22}\dot{I}_2 \end{cases}$$

不同类型变量 位于不同端口

传输参数方程(T和T')

$$\begin{cases} \dot{U}_1 = A\dot{U}_2 + B\left(-\dot{I}_2\right) & \begin{cases} \dot{U}_2 = A'\dot{U}_1 + B'\left(-\dot{I}_1\right) & \mathbf{不同类型变量} \\ \dot{I}_1 = C\dot{U}_2 + D\left(-\dot{I}_2\right) & \dot{I}_2 = C'\dot{U}_1 + D'\left(-\dot{I}_1\right) & \mathbf{位于同一端口} \end{cases}$$

16.3 二端口网络的参数



四个端口参数

 \dot{U}_1 \dot{I}_2 \dot{I}_1 \dot{U}_2

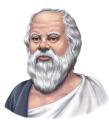
Z, 六组特性方程

Z参数、Y参数、H参数、

G参数、T参数、T'参数



灵魂三问???



我是谁?

参数是什么? 各种参数的物 理含义



我从哪里来?

参数如何获得? 参数计算、测 量方法



我要到哪里去?

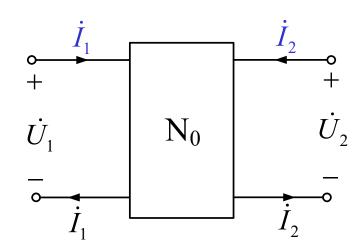
参数如何应用? 计算端口变量、 获得等效电路

1. 阻抗参数方程(Z参数)

$$\begin{cases} \dot{U}_1 = Z_{11}\dot{I}_1 + Z_{12}\dot{I}_2 \\ \dot{U}_2 = Z_{21}\dot{I}_1 + Z_{22}\dot{I}_2 \end{cases}$$

矩阵形式:

$$\begin{bmatrix} \dot{U}_1 \\ \dot{U}_2 \end{bmatrix} = \begin{bmatrix} Z_{11} & Z_{12} \\ Z_{21} & Z_{22} \end{bmatrix} \begin{bmatrix} \dot{I}_1 \\ \dot{I}_2 \end{bmatrix} \qquad Z = \begin{bmatrix} Z_{11} & Z_{12} \\ Z_{21} & Z_{22} \end{bmatrix}$$



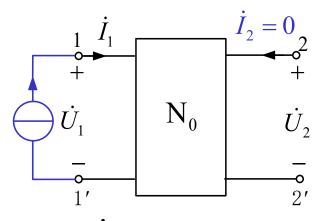
$$Z = egin{bmatrix} Z_{11} & Z_{12} \ Z_{21} & Z_{22} \end{bmatrix}$$

Q:如何测量Z参数?

如何测量
$$Z$$
参数 $\dot{I}_2=0$ 计算 Z_{11} 、 Z_{21} 用端口开路实验测 Z 参数 $\dot{I}_1=0$ 计算 Z_{12} 、 Z_{22}

阻抗参数测量 $\begin{cases} \dot{U}_1 = Z_{11}\dot{I}_1 + Z_{12}I_2 \\ \dot{U}_2 = Z_{21}\dot{I}_1 + Z_{22}\dot{I}_2 \end{cases}$

2-2′ 端口开路:

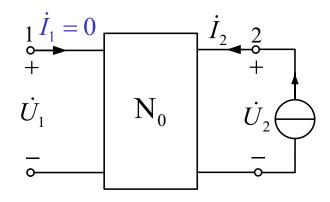


$$Z_{11} = \frac{\dot{U}_1}{\dot{I}_1} \Big|_{\dot{I}_2=0}$$
 输入阻抗

$$Z_{21} = \frac{\dot{U}_2}{\dot{I}_1}\Big|_{\dot{I}_2=0}$$
 转移阻抗

O: 两个转移阻抗值是否相等? 互易定理

1-1′ 端口开路:

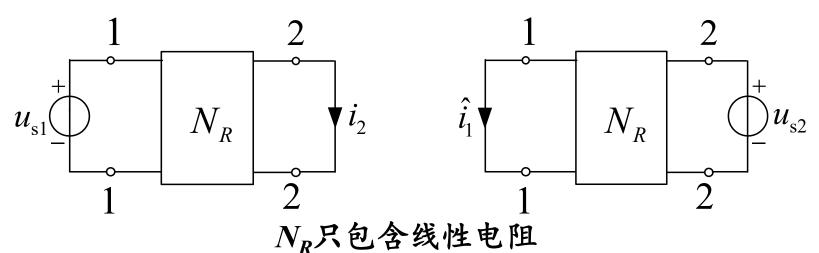


$$Z_{12} = \frac{U_1}{\dot{I}_2}\Big|_{\dot{I}_1=0}$$
 转移阻抗

$$Z_{22} = \frac{\dot{U}_2}{\dot{I}_2} \Big|_{\dot{I}_1=0}$$
 输入阻抗

互易定理(第四章内容回顾!)

第一种形式:激励是电压源,响应是电流。



$$u_{s1}\hat{i}_{1} + 0 + \sum_{k=3}^{b} u_{k}\hat{i}_{k} = 0$$

$$0 + u_{s2}i_{2} + \sum_{k=3}^{b} \hat{u}_{k}\hat{i}_{k} = 0$$

$$\hat{u}_{k}\hat{i}_{k} = R_{k}\hat{i}_{k}\hat{i}_{k} = u_{k}\hat{i}_{k}$$

$$\frac{i_{2}}{u_{s1}} = \frac{\hat{i}_{1}}{u_{s}}$$

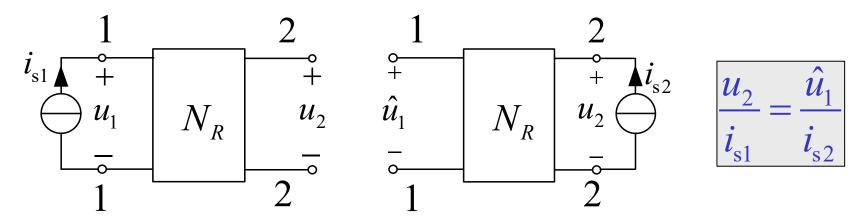
两电路的响应与激励之比相等



2022/6/9 电路理论

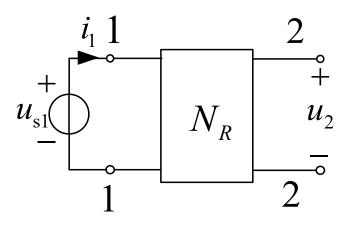
互易定理(第四章内容回顾!)

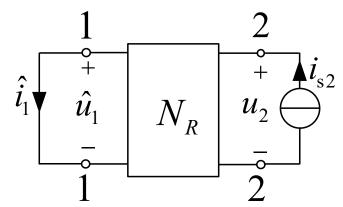
第二种形式:激励是电流源,响应是电压。



第三种形式: 电路1激励是电压源, 响应是电压;

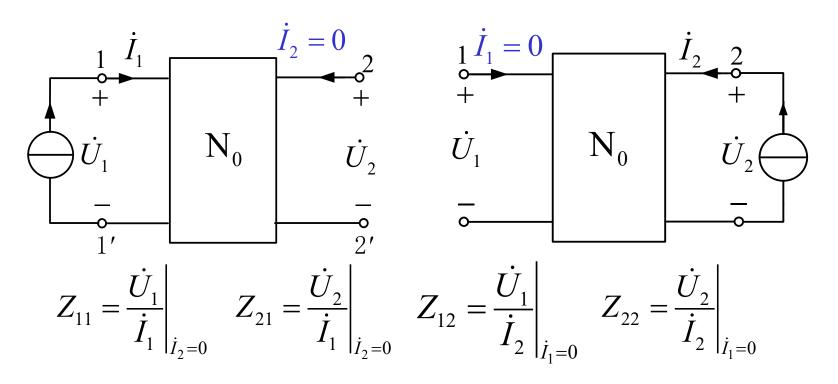
电路2激励是电流源,响应是电流。





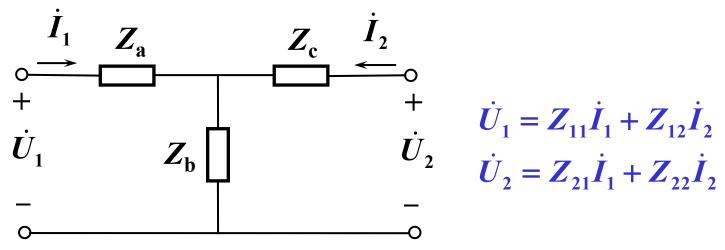
$$\frac{u_2}{u_{\rm s1}} = \frac{\hat{i}_1}{i_{\rm s2}}$$





若二端口网络内部无受控源, 电路满足互易定理2

例 求所示电路的Z参数。



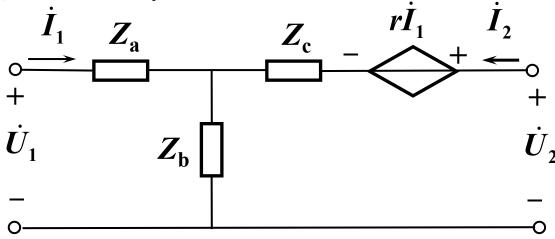
解 由实验测量法得到参数

$$Z_{11} = \frac{\dot{U}_1}{\dot{I}_1}\Big|_{\dot{I}_2=0} = Z_a + Z_b \qquad Z_{12} = \frac{\dot{U}_1}{\dot{I}_2}\Big|_{\dot{I}_1=0} = Z_b$$

$$Z_{21} = \frac{\dot{U}_2}{\dot{I}_1}\Big|_{\dot{I}_2=0} = Z_b \qquad Z_{22} = \frac{\dot{U}_2}{\dot{I}_2}\Big|_{\dot{I}_1=0} = Z_b + Z_c$$

互易二端口,且当 $Z_a=Z_c$ 时为对称二端口。





$$\dot{U}_1 = Z_a \dot{I}_1 + Z_b (\dot{I}_1 + \dot{I}_2)$$

$$\dot{U}_2 = r\dot{I}_1 + Z_c\dot{I}_2 + Z_b(\dot{I}_1 + \dot{I}_2)$$

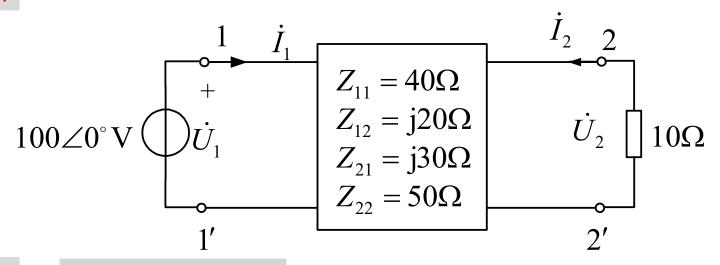
$$Z = \begin{bmatrix} Z_{a} + Z_{b} & Z_{b} \\ r + Z_{b} & Z_{b} + Z_{c} \end{bmatrix}$$
 4个独立参数

 $U_1 = Z_{11}I_1 + Z_{12}I_2$

 $\dot{U}_{2} = Z_{21}\dot{I}_{1} + Z_{22}\dot{I}_{2}$

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例 计算端口电流



解

已知参数,确

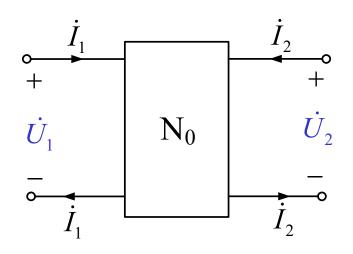
2. 导纳参数方程(Y参数)

$$\begin{cases} \dot{I}_1 = Y_{11}\dot{U}_1 + Y_{12}\dot{U}_2 \\ \dot{I}_2 = Y_{21}\dot{U}_1 + Y_{22}\dot{U}_2 \end{cases}$$

矩阵形式:

$$\begin{bmatrix} \dot{I}_1 \\ \dot{I}_2 \end{bmatrix} = \begin{bmatrix} Y_{11} & Y_{12} \\ Y_{21} & Y_{22} \end{bmatrix} \begin{bmatrix} \dot{U}_1 \\ \dot{U}_2 \end{bmatrix} \qquad Y = \begin{bmatrix} Y_{11} & Y_{12} \\ Y_{21} & Y_{22} \end{bmatrix}$$

$$\begin{bmatrix} \dot{U}_1 \\ \dot{U}_2 \end{bmatrix} = \begin{bmatrix} Z_{11} & Z_{12} \\ Z_{21} & Z_{22} \end{bmatrix} \begin{bmatrix} \dot{I}_1 \\ \dot{I}_2 \end{bmatrix} \qquad \Box$$



$$Y = \begin{bmatrix} Y_{11} & Y_{12} \\ Y_{21} & Y_{22} \end{bmatrix}$$

$$YZ = 1$$

同一网络的Z参数、 Y参数矩阵互逆

导纳参数测量

2-2' 端口短路:
$$\dot{U}_2 = 0$$

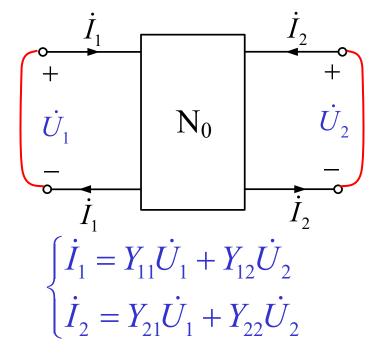
$$Y_{11} = \frac{\dot{I}_1}{\dot{U}_1} \Big|_{\dot{U}_2=0}$$
 输入导纳

$$Y_{21} = \frac{\dot{I}_2}{\dot{U}_1} \Big|_{\dot{U}_2=0}$$
 转移导纳

1-1' 端口短路: $\dot{U}_1 = 0$

$$Y_{12} = \frac{\dot{I}_1}{\dot{U}_2} \Big|_{\dot{U}_1=0}$$
 转移导纳

$$Y_{22} = \frac{\dot{I}_2}{\dot{U}_2} \Big|_{\dot{U}_1=0}$$
 输入导纳



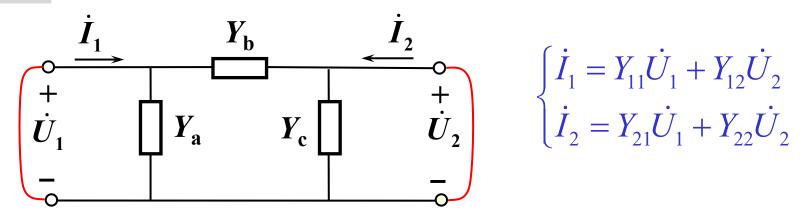
$$Y_{12} = Y_{21}$$

对称二端口:

$$Y_{12} = Y_{21}$$
 $Y_{11} = Y_{22}$

实验测量法是确定任何参数的通用方法!

例求图示二端口的Y参数。



解 由实验测量法得到参数

$$\begin{split} Y_{11} &= \frac{\dot{I}_1}{\dot{U}_1} \Big|_{\dot{U}_2 = 0} = Y_{\rm a} + Y_{\rm b} \qquad Y_{12} = \frac{\dot{I}_1}{\dot{U}_2} \Big|_{\dot{U}_1 = 0} = -Y_{\rm b} \\ Y_{21} &= \frac{\dot{I}_2}{\dot{U}_1} \Big|_{\dot{U}_2 = 0} = -Y_{\rm b} \qquad Y_{22} = \frac{\dot{I}_2}{\dot{U}_2} \Big|_{\dot{U}_2 = 0} = Y_{\rm b} + Y_{\rm c} \end{split}$$

互易二端口,且当 $Y_a=Y_c$ 时为对称二端口。

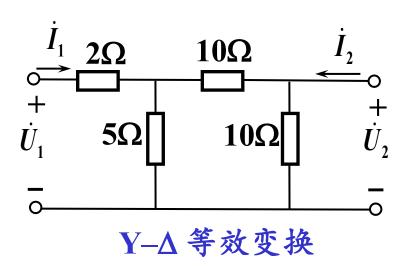
例求图示二端口的Y参数。

解 电阻网络, 互易

$$Y_{12} = Y_{21} = -\frac{1}{16} S$$

$$Y_{11} = \frac{1}{2+5/10} = \frac{3}{16} S$$

$$Y_{22} = \frac{1}{10/(10+2/5)} = \frac{3}{16} S$$



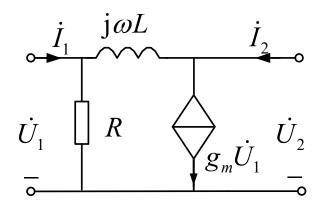
对称二端口网络
$$Y_{11} = Y_{22}$$
 $Y_{12} = Y_{21}$

对称二端口是指两个端口电气特性上对称。电路结构 左右对称的,端口电气特性对称;电路结构不对称的二端 口,其电气特性也可能是对称的。

解 方法一:实验测量法

方法二: 列写结点方程

$$\begin{cases} \left(\frac{1}{R} + \frac{1}{j\omega L}\right)\dot{U}_1 - \frac{1}{j\omega L}\dot{U}_2 = \dot{I}_1 \\ -\frac{1}{j\omega L}\dot{U}_1 + \frac{1}{j\omega L}\dot{U}_2 = \dot{I}_2 - g_m\dot{U}_1 \end{cases}$$



$$\begin{cases} \dot{I}_1 = Y_{11}\dot{U}_1 + Y_{12}\dot{U}_2 \\ \dot{I}_2 = Y_{21}\dot{U}_1 + Y_{22}\dot{U}_2 \end{cases}$$

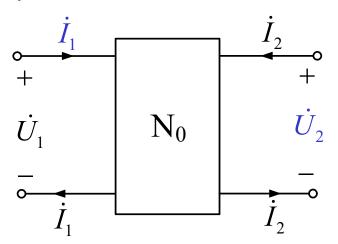
$$\begin{cases} \dot{I}_1 = \left(\frac{1}{R} + \frac{1}{j\omega L}\right)\dot{U}_1 - \frac{1}{j\omega L}\dot{U}_2 \\ \dot{I}_2 = \left(g_m - \frac{1}{j\omega L}\right)\dot{U}_1 + \frac{1}{j\omega L}\dot{U}_2 \end{cases} Y = \begin{bmatrix} \frac{1}{R} + \frac{1}{j\omega L} & -\frac{1}{j\omega L} \\ g_m - \frac{1}{j\omega L} & \frac{1}{j\omega L} \end{bmatrix}$$

$$Y = \begin{bmatrix} \frac{1}{R} + \frac{1}{j\omega L} & -\frac{1}{j\omega L} \\ g_m - \frac{1}{j\omega L} & \frac{1}{j\omega L} \end{bmatrix}$$

3. 混合参数方程(H参数和G参数)

H参数
$$\begin{cases} \dot{U}_1 = h_{11}\dot{I}_1 + h_{12}\dot{U}_2 & + \\ \dot{I}_2 = h_{21}\dot{I}_1 + h_{22}\dot{U}_2 & \dot{U}_1 \end{cases}$$
 N₀

G参数
$$\begin{cases} \dot{I}_1 = g_{11}\dot{U}_1 + g_{12}\dot{I}_2 \\ \dot{U}_2 = g_{21}\dot{U}_1 + g_{22}\dot{I}_2 \end{cases}$$



矩阵形式:

$$HG = 1$$

$$\begin{bmatrix} \dot{U}_1 \\ \dot{I}_2 \end{bmatrix} = \begin{bmatrix} h_{11} & h_{12} \\ h_{21} & h_{22} \end{bmatrix} \begin{bmatrix} \dot{I}_1 \\ \dot{U}_2 \end{bmatrix}$$

$$\begin{bmatrix} \dot{U}_1 \\ \dot{I}_2 \end{bmatrix} = \begin{bmatrix} h_{11} & h_{12} \\ h_{21} & h_{22} \end{bmatrix} \begin{bmatrix} \dot{I}_1 \\ \dot{U}_2 \end{bmatrix} \qquad \begin{bmatrix} \dot{I}_1 \\ \dot{U}_2 \end{bmatrix} = \begin{bmatrix} g_{11} & g_{12} \\ g_{21} & g_{22} \end{bmatrix} \begin{bmatrix} \dot{U}_1 \\ \dot{I}_2 \end{bmatrix}$$

混合参数测量(H参数)

2-2' 端口短路:
$$\dot{U}_2 = 0$$

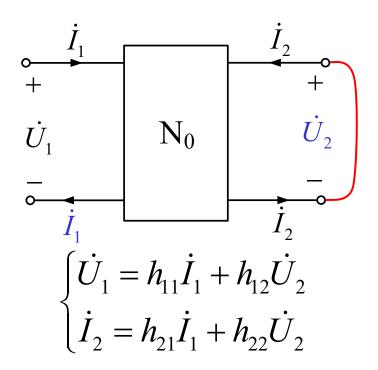
$$h_{11} = \frac{\dot{U}_1}{\dot{I}_1} \Big|_{\dot{U}_2=0}$$
 输入阻抗

$$h_{21} = \frac{I_2}{\dot{I}_1} \Big|_{\dot{U}_2=0}$$
 电流增益

1-1' 端口开路:
$$\dot{I}_1 = 0$$

$$h_{12} = \frac{U_1}{\dot{U}_2} \Big|_{\dot{I}_1 = 0}$$
 电压增益

$$h_{22} = \frac{\dot{I}_2}{\dot{U}_2} \Big|_{\dot{I}_1=0}$$
 输入导纳



互易二端口: 互易定理3
$$h_{12} = -h_{21}$$

对称二端口:
$$h_{12} = -h_{21}$$
$$h_{11}h_{22} - h_{12}h_{21} = 1$$

证明对称二端口H参数需满足: $h_{12} = -h_{21}$ $h_{11}h_{22} - h_{12}h_{21} = 1$

Y参数
$$\begin{cases} \dot{I}_1 = Y_{11}\dot{U}_1 + Y_{12}\dot{U}_2 \\ \dot{I}_2 = Y_{21}\dot{U}_1 + Y_{22}\dot{U}_2 \end{cases}$$
 对称二端口:
$$Y_{12} = Y_{21} \\ Y_{11} = Y_{22}$$

$$\vec{I}_{2} = h_{21}\dot{I}_{1} + h_{22}\dot{U}_{2}
\vec{I}_{2} = h_{21}\left(\frac{\dot{U}_{1}}{h_{11}} - \frac{h_{12}\dot{U}_{2}}{h_{11}}\right) + h_{22}\dot{U}_{2}
= \frac{h_{21}}{h_{11}}\dot{U}_{1} + \frac{h_{11}h_{22} - h_{21}h_{12}}{h_{11}}\dot{U}_{2}$$

$$Y_{12} = \frac{-h_{12}}{h_{11}} = \frac{h_{21}}{h_{11}} = Y_{21}
Y_{11} = \frac{1}{h_{11}} = \frac{h_{11}h_{22} - h_{21}h_{12}}{h_{11}} = Y_{22}$$

G参数测量方法以及互易、对称条件类似,课后自学

例求图示二端口的H参数。

解 含受控源,不满足互易定理

方法一:实验测量法

$$1-1'$$
 端口开路: $i_1=0$

$$2u_2 = u_1 + u_1 \implies h_{12} = 1$$

$$u_2 = 1 \cdot i_2 + 2u_2 \implies h_{22} = -1$$

$$\begin{cases} \dot{U}_1 = h_{11}\dot{I}_1 + h_{12}\dot{U}_2 \\ \dot{I}_2 = h_{21}\dot{I}_1 + h_{22}\dot{U}_2 \end{cases}$$

$$H = \begin{bmatrix} h_{11} & h_{12} \\ h_{21} & h_{22} \end{bmatrix} = \begin{bmatrix} 0.5 & 1 \\ 0 & -1 \end{bmatrix}$$

解 方法二: 列写网孔方程

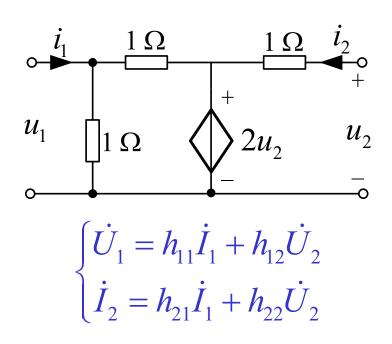
$$u_1 = \left(i_1 - \frac{u_1}{1}\right) \times 1 + 2u_2$$

$$\square > u_1 = \frac{1}{2}i_1 + u_2$$

$$i_2 = \frac{u_2 - 2u_2}{1} = -u_2$$

$$h_{11} = \frac{1}{2}$$
 $h_{12} = 1$

$$h_{21} = 0$$
 $h_{22} = -1$



$$H = \begin{bmatrix} h_{11} & h_{12} \\ h_{21} & h_{22} \end{bmatrix} = \begin{bmatrix} 0.5 & 1 \\ 0 & -1 \end{bmatrix}$$

已知 $U_s = 42\sqrt{2}\cos(5000t)$ V, $R = 6\Omega$,无源二端

口网络G参数为
$$G = \begin{bmatrix} \frac{1}{6} - j\frac{1}{6} & -0.5 + j0.5 \\ 0.5 - j0.5 & 1.5 + j2.5 \end{bmatrix}$$
, 求负

载 Z_L 的最大功率 P_L 及此时电源提供的功率 P_S 。

解

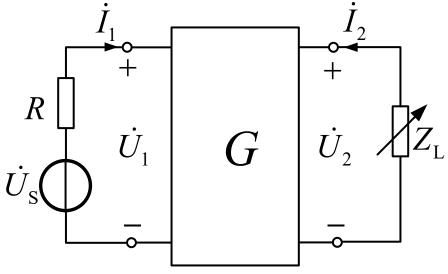
$$\dot{I}_{1} = g_{11}\dot{U}_{1} + g_{12}\dot{I}_{2}$$

$$\dot{U}_{2} = g_{21}\dot{U}_{1} + g_{22}\dot{I}_{2}$$

$$\dot{U}_{1} = 42 - 6\dot{I}_{1}$$

戴维南等效

端口2短路:
$$\dot{U}_2=0$$
 $\Longrightarrow \dot{I}_{\rm 2SC}=\dot{I}_2$



$$\dot{U}_2 = \dot{U}_{2OC} + Z_{eq} \dot{I}_{2SC}$$

例 已知 $U_s = 42\sqrt{2}\cos(5000t)$ V, $R = 6\Omega$,无源二端

口网络G参数为
$$G = \begin{bmatrix} \frac{1}{6} - j\frac{1}{6} & -0.5 + j0.5 \\ 0.5 - j0.5 & 1.5 + j2.5 \end{bmatrix}$$
, 求负

载 Z_L 的最大功率 P_L 及此时电源提供的功率 P_S 。

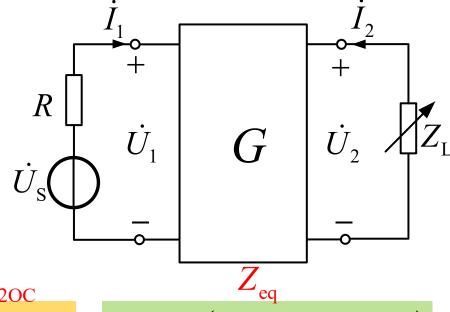
解 思路2: 戴维南等效

$$\begin{split} \dot{U}_2 &= \dot{U}_{\rm 2OC} + Z_{\rm eq} \dot{I}_{\rm 2SC} \\ &= \dot{U}_{\rm 2OC} + Z_{\rm eq} \dot{I}_2 \end{split}$$

只需找到Ü,和İ,之间的关系

$$\dot{I}_{1} = g_{11}\dot{U}_{1} + g_{12}\dot{I}_{2} \qquad \dot{U}_{2OC}$$

$$\dot{U}_{2} = g_{21}\dot{U}_{1} + g_{22}\dot{I}_{2} \rightarrow \dot{U}_{2} = 42\frac{g_{21}}{1 + 6g_{11}}$$



$$\frac{g_{22} + 6(g_{11}g_{22} - g_{12}g_{21})}{1 + 6g_{11}}\dot{I}_{2}$$

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$$\dot{U}_{2OC} = 42 \frac{g_{21}}{1 + 6g_{11}} = 13.28 \angle -18.43^{\circ} \text{ V}$$

$$Z_{\text{eq}} = \frac{g_{22} + 6(g_{11}g_{22} - g_{12}g_{21})}{1 + 6g_{11}} = 2.1 + \text{j}1.3 \ \Omega$$

(1) 最大功率
$$P_{\rm L}$$

$$Z_{\rm L} = 2.1 - \text{j}1.3 \ \Omega \qquad \to P_{\rm L} = \frac{U_{\rm OC}^2}{4 \, \text{Re}(Z_{\rm L})} = 21 \ \text{W}$$

(2) 电源功率 $P_{\rm S}$

$$\dot{I}_2 = -\frac{\dot{U}_{2OC}}{Z_{eq} + Z_L} = 3.16 \angle 161.57^{\circ} \text{ A}$$

$$\dot{I}_{1} = g_{11}\dot{U}_{1} + g_{12}\dot{I}_{2}
\dot{U}_{1} = 42 - 6\dot{I}_{1}$$

$$\rightarrow \dot{I}_{1} = \frac{g_{11}\dot{U}_{1} + g_{12}\dot{I}_{2}}{1 + 6g_{11}} = 5.38\angle - 21.8^{\circ} \text{ A}$$

$$\rightarrow P = 42 \times 5.38 \times \cos(21.8^{\circ}) = 210 \text{ W}$$

4. 传输参数方程(T参数和T'参数)

矩阵形式:

$$\begin{bmatrix} \dot{U}_1 \\ \dot{I}_1 \end{bmatrix} = \begin{bmatrix} A & B \\ C & D \end{bmatrix} \begin{bmatrix} \dot{U}_2 \\ -\dot{I}_2 \end{bmatrix} \qquad \begin{bmatrix} \dot{U}_2 \\ \dot{I}_2 \end{bmatrix} = \begin{bmatrix} A' & B' \\ C' & D' \end{bmatrix} \begin{bmatrix} \dot{U}_1 \\ -\dot{I}_1 \end{bmatrix}$$

T参数能方便的描述信号或能量从一个 端口向另一端口的传输特性。

$$\begin{bmatrix} \dot{U}_1 \\ \dot{I}_1 \end{bmatrix} = \begin{bmatrix} A & B \\ C & D \end{bmatrix} \begin{bmatrix} \dot{U}_2 \\ -\dot{I}_2 \end{bmatrix} \qquad \begin{bmatrix} \dot{U}_2 \\ \dot{I}_2 \end{bmatrix} = \begin{bmatrix} A' & B' \\ C' & D' \end{bmatrix} \begin{bmatrix} \dot{U}_1 \\ -\dot{I}_1 \end{bmatrix}$$

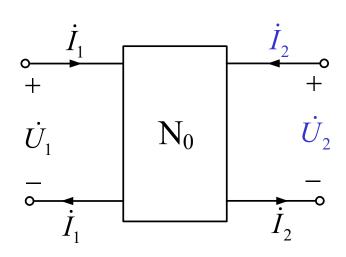
$$\begin{bmatrix} A & -B \\ C & -D \end{bmatrix} \begin{bmatrix} A' & -B' \\ C' & -D' \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}$$

传输参数测量 (T参数)

2-2' 端口开路:
$$\dot{I}_2 = 0$$

$$A = \frac{\dot{U}_1}{\dot{U}_2} \Big|_{\dot{I}_2 = 0}$$
 A^{-1} 电压增益

$$C = \frac{I_1}{\dot{U}_2}\Big|_{\dot{I}_2=0}$$
 C^{-1} 转移阻抗

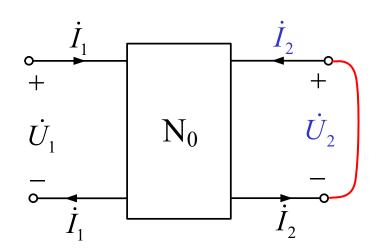


传输参数测量 (T参数)

2-2' 端口短路:
$$\dot{U}_2 = 0$$

$$D = \frac{\dot{I}_1}{-\dot{I}_2} \Big|_{\dot{U}_2 = 0} - D^{-1}$$
电流增益

$$B = \frac{\dot{U}_1}{-\dot{I}_2}\Big|_{\dot{U}_2=0} - B^{-1}$$
转移导纳



互易二端口:

$$AD - BC = 1$$

对称二端口:

$$AD - BC = 1$$
 $A = D$

T '参数的测量方法课后自学

解 方法一:实验测量法

2-2' 端口开路:
$$\dot{I}_2 = 0$$

$$\dot{I}_3 = 0$$
 $\dot{U}_2 = \frac{1}{n}\dot{U}_3$ $\dot{U}_3 = \dot{U}_1$

$$\dot{U}_2 = \frac{1}{n}\dot{U}_3 = \frac{1}{n}\dot{U}_1$$

$$\dot{I}_{1} = \frac{\dot{U}_{1}}{R_{1}} = \frac{n}{R_{1}}\dot{U}_{2}$$

$$A = \frac{\dot{U}_1}{\dot{U}_2} \Big|_{\dot{I}_2 = 0} = n$$

$$\begin{cases} \dot{I}_{1} = CU \\ \dot{I}_{1} = CU \end{cases}$$

$$= \frac{\dot{U}_{1}}{1 \cdot \cdot} = \frac{n}{R}$$

$$\begin{cases} \dot{U}_1 = A\dot{U}_2 + B(-\dot{I}_2) \\ \dot{I}_1 = C\dot{U}_2 + D(-\dot{I}_2) \end{cases}$$

例 求图示二端口的T参数。

解 方法一:实验测量法

2-2' 端口短路:
$$\dot{U}_2 = 0$$

$$\dot{U}_3 = 0 \quad \dot{I}_3 = \frac{\dot{U}_1}{R_2}$$

$$\dot{I}_1 = \frac{\dot{U}_1}{R_1} + \dot{I}_3 = (\frac{1}{R_1} + \frac{1}{R_2})\dot{U}_1$$

$$\dot{I}_2 = -n\dot{I}_3 = -\frac{n}{R_2}\dot{U}_1$$

$$\begin{cases} \dot{U}_1 = A\dot{U}_2 + B\left(-\dot{I}_2\right) \\ \dot{I}_1 = C\dot{U}_2 + D\left(-\dot{I}_2\right) \end{cases}$$

$$B = \frac{\dot{U}_1}{-\dot{I}_2}\Big|_{\dot{U}_2=0} = \frac{\dot{U}_1}{\frac{n}{R_2}\dot{U}_1} = \frac{R_2}{n} \quad D = \frac{\dot{I}_1}{-\dot{I}_2}\Big|_{\dot{U}_2=0} = \frac{(\frac{1}{R_1} + \frac{1}{R_2})\dot{U}_1}{\frac{n}{R_2}\dot{U}_1} = \frac{R_1 + R_2}{nR_1}$$

例 求图示二端口的T参数。

解 方法二: 列写端口特性方程

由理想变压器特性:

$$\dot{I}_3 = -\frac{1}{n}\dot{I}_2$$

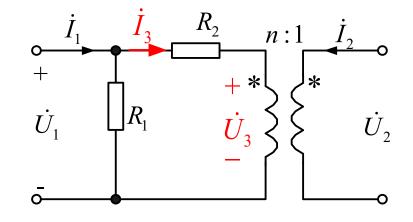
$$\dot{U}_3 = n\dot{U}_2$$

$$\dot{U}_1 = \dot{I}_3 R_2 + \dot{U}_3 = n \dot{U}_2 - \frac{R_2}{n} \dot{I}_2$$

$$\dot{I}_1 = \frac{U_1}{R_1} + \dot{I}_3 = \frac{1}{R_1} (n\dot{U}_2 - \frac{R_2}{n}\dot{I}_2) - \frac{1}{n}\dot{I}_2$$

$$\dot{U}_1 = n\dot{U}_2 + \frac{R_2}{n}(-\dot{I}_2)$$

$$\vec{I}_1 = \frac{n}{R_1} \dot{U}_2 + \left(\frac{R_2}{nR_1} + \frac{1}{n}\right) (-\dot{I}_2)$$



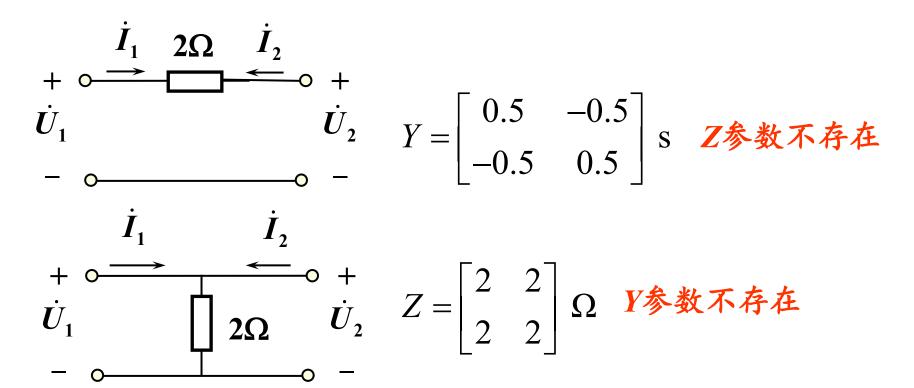
$$\begin{cases} \dot{U}_1 = A\dot{U}_2 + B\left(-\dot{I}_2\right) \\ \dot{I}_1 = C\dot{U}_2 + D\left(-\dot{I}_2\right) \end{cases}$$

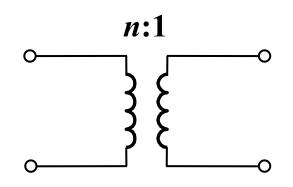
$$A = n \qquad B = \frac{R_2}{n}$$

$$C = \frac{n}{R_1} \quad D = \frac{R_1 + R_2}{nR_1}$$

小结

- (1) 为什么用这么多参数表示二端口网络特性?
 - > 为描述电路方便, 测量方便;
 - > 有些电路只存在某几种参数。





存在T参数,H参数。Z、Y均不存在。

- (2) 线性无源二端口为互易二端口网络;
- (3) 含有受控源的电路有四个独立参数;
- (4) 不同参数之间可以进行互换。 数学方程变换

H参数
$$\begin{cases} \dot{U}_{1} = h_{11}\dot{I}_{1} + h_{12}\dot{U}_{2} \Longrightarrow \dot{I}_{1} = \frac{1}{h_{11}}\dot{U}_{1} - \frac{h_{12}}{h_{11}}\dot{U}_{2} \\ \dot{I}_{2} = h_{21}\dot{I}_{1} + h_{22}\dot{U}_{2} \Longrightarrow \dot{I}_{2} = h_{21}\left(\frac{\dot{U}_{1}}{h_{11}} - \frac{h_{12}\dot{U}_{2}}{h_{11}}\right) + h_{22}\dot{U}_{2} \\ \dot{I}_{2} = Y_{21}\dot{U}_{1} + Y_{22}\dot{U}_{2} \Longrightarrow \dot{I}_{1} = \frac{h_{21}}{h_{11}}\dot{U}_{1} + \frac{h_{11}h_{22} - h_{21}h_{12}}{h_{11}}\dot{U}_{2} \end{cases}$$

互易二端口网络、对称二端口网络参数特点

	互易二端口	对称二端口
Z	$Z_{12} = Z_{21}$	$Z_{11} = Z_{22}$ $Z_{12} = Z_{21}$
Y	$Y_{12} = Y_{21}$	$Y_{11} = Y_{22}$ $Y_{12} = Y_{21}$
Н	$h_{12} = -h_{21}$	$h_{11}h_{22} - h_{12}h_{21} = 1$ $h_{12} = -h_{21}$
G	$g_{12} = -g_{21}$	$g_{11}g_{22} - g_{12}g_{21} = 1$ $g_{12} = -g_{21}$
T	AD - BC = 1	A = D AD - BC = 1
T'	A'D' - B'C' = 1	$A' = D' \qquad A'D' - B'C' = 1$

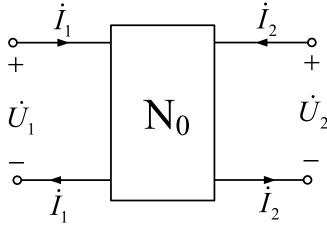
例 已知:端口2开路时, $U_1=10\,\mathrm{mV}$, $I_1=10\,\mu\mathrm{A}$, U_2 =-40V; 端口2短路时, U_1 =24mV, I_1 =20 μ A, $I_2=1$ mA。计算网络H参数。

解
$$\begin{cases} \dot{U}_1 = h_{11}\dot{I}_1 + h_{12}\dot{U}_2 \\ \dot{I}_2 = h_{21}\dot{I}_1 + h_{22}\dot{U}_2 \end{cases}$$

端口2短路时:

$$h_{11} = \frac{\dot{U}_1}{\dot{I}_1}\Big|_{\dot{U}_2=0} = \frac{24\text{mV}}{20\mu\text{A}} = 1.2 \text{ k}\Omega$$

$$h_{21} = \frac{\dot{I}_2}{\dot{I}_1}\Big|_{\dot{U}_2=0} = \frac{1\text{mA}}{20\mu\text{A}} = 50$$



端口1开路时:

$$h_{12} = \frac{\dot{U}_1}{\dot{U}_2}\Big|_{\dot{I}_1 = 0} = ?$$

$$h_{22} = \frac{\dot{I}_2}{\dot{U}_2}\Big|_{\dot{I}_1=0} = ?$$

例 已知:端口2开路时, $U_1=10\,\mathrm{mV}$, $I_1=10\,\mu\mathrm{A}$, U_2 =-40V; 端口2短路时, U_1 =24mV, I_1 =20 μ A, I_2 =1mA。计算网络H参数。

T参数方程
$$\begin{cases} \dot{U}_1 = A\dot{U}_2 + B\left(-\dot{I}_2\right) \\ \dot{I}_1 = C\dot{U}_2 + D\left(-\dot{I}_2\right) \end{cases}$$

$$A = \frac{\dot{U}_1}{\dot{U}_2} = \frac{10\text{mV}}{-40\text{V}} = -2.5 \times 10^4$$
 $B = \frac{\dot{U}_1}{-\dot{I}_2} = \frac{24\text{mV}}{-1\text{mA}} = -24 \Omega$

$$C = \frac{\dot{I}_1}{\dot{U}_2} = \frac{10\mu\text{A}}{-40\text{V}} = -2.5 \times 10^7 \text{ S}$$
 $D = \frac{\dot{I}_1}{-\dot{I}_2} = \frac{20\mu\text{A}}{-1\text{mA}} = -0.02$

$$\dot{I}_{1}$$
 \dot{I}_{2}
 \dot{V}_{1}
 \dot{V}_{1}
 \dot{I}_{2}
 \dot{I}_{2}

$$B = \frac{U_1}{-\dot{I}_2} = \frac{24 \text{mV}}{-1 \text{mA}} = -24 \Omega$$

$$D = \frac{\dot{I}_1}{-\dot{I}_2} = \frac{20\mu A}{-1\text{mA}} = -0.02$$

$$A = -2.5 \times 10^4$$

$$C = -2.5 \times 10^7 \text{ S}$$

$$B = -24 \Omega$$

$$D = -0.02$$

T参数方程

$$\begin{cases} \dot{U}_1 = A\dot{U}_2 + B\left(-\dot{I}_2\right) \\ \dot{I}_1 = C\dot{U}_2 + D\left(-\dot{I}_2\right) \end{cases}$$

$$\begin{cases} \dot{U}_1 = h_{11}\dot{I}_1 + h_{12}\dot{U}_2 \\ \dot{I}_2 = h_{21}\dot{I}_1 + h_{22}\dot{U}_2 \end{cases}$$

$$\dot{I}_{1} = C\dot{U}_{2} + D(-\dot{I}_{2})$$
 \Box $\dot{I}_{2} = -\frac{1}{D}\dot{I}_{1} + \frac{C}{D}\dot{U}_{2}$

$$\dot{I}_2 = -\frac{1}{D}\dot{I}_1 + \frac{C}{D}\dot{U}_2$$

$$h_{21} = -\frac{1}{D} = 50$$
 $h_{22} = \frac{C}{D} = 12.5 \text{ }\mu\text{S}$

$$\dot{U}_1 = A\dot{U}_2 + B\left(\frac{1}{D}\dot{I}_1 - \frac{C}{D}\dot{U}_2\right) \quad \Longrightarrow \quad$$

$$\dot{U}_{1} = A\dot{U}_{2} + B\left(\frac{1}{D}\dot{I}_{1} - \frac{C}{D}\dot{U}_{2}\right) \implies h_{11} = \frac{B}{D} = 1.2 \text{ k}\Omega$$

$$h_{12} = \frac{AD - BC}{D} = 5 \times 10^{-5}$$

例 已知:端口2开路时, U_1 =10mV, I_1 =10 μ A, U_2 =-40V;端口2短路时, U_1 =24mV, I_1 =20 μ A, I_2 =1mA。计算网络H参数。 i_1 i_2

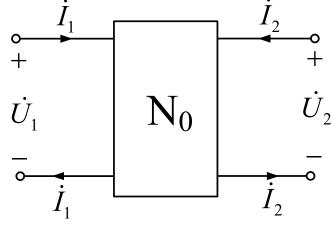
解 方法二:

$$\begin{cases} \dot{U}_1 = h_{11}\dot{I}_1 + h_{12}\dot{U}_2 \\ \dot{I}_2 = h_{21}\dot{I}_1 + h_{22}\dot{U}_2 \end{cases}$$

端口2短路时:

$$h_{11} = \frac{\dot{U}_1}{\dot{I}_1}\Big|_{\dot{U}_2=0} = \frac{24\text{mV}}{20\mu\text{A}} = 1.2 \text{ k}\Omega$$

$$h_{21} = \frac{\dot{I}_2}{\dot{I}_1}\Big|_{\dot{U}_2=0} = \frac{1\text{mA}}{20\mu\text{A}} = 50$$



端口2开路时: $\dot{I}_2=0$

$$h_{22} = \frac{-h_{21}\dot{I}_1}{\dot{U}_2} = 12.5 \text{ } \mu\text{S}$$

$$h_{12} = \frac{\dot{U}_1 - h_{11}\dot{I}_1}{\dot{U}_2} = 5 \times 10^{-5}$$

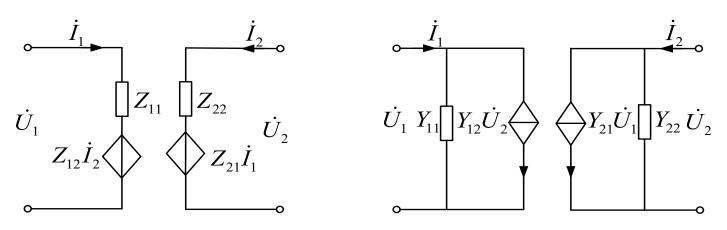
无论端口电量如何变化, 网络参数不变

16.4二端口网络的电路模型

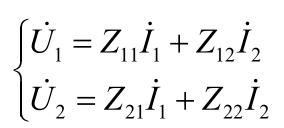
1. 等效二端口网络

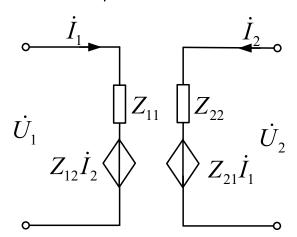
两个二端口等效是指对外电路而言,端口的电压、电流关系相同。

- > 可以用最简单的二端口网络来代替复杂的网络;
- > 等效二端口网络电路模型并不唯一;
- ▶ 最简单的二端口网络应包含4个独立的电路元件。

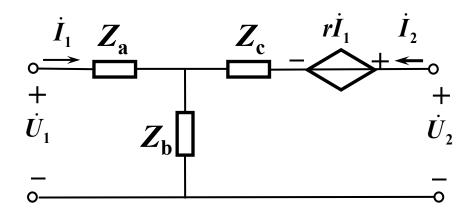


2. 由Z参数方程做等效电路 □ T形电路模型





回顾

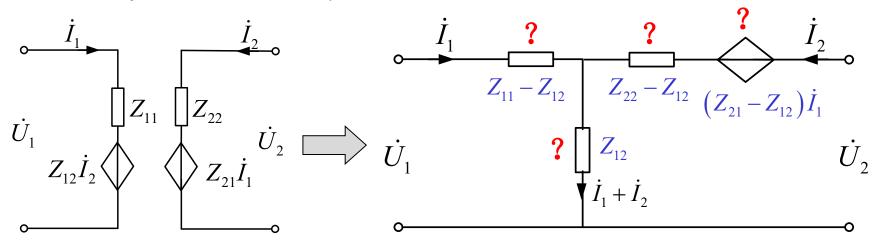


方法二: 列写网孔方程

$$\dot{U}_{1} = Z_{a}\dot{I}_{1} + Z_{b}(\dot{I}_{1} + \dot{I}_{2})$$

$$\dot{U}_{2} = r\dot{I}_{1} + Z_{c}\dot{I}_{2} + Z_{b}(\dot{I}_{1} + \dot{I}_{2})$$

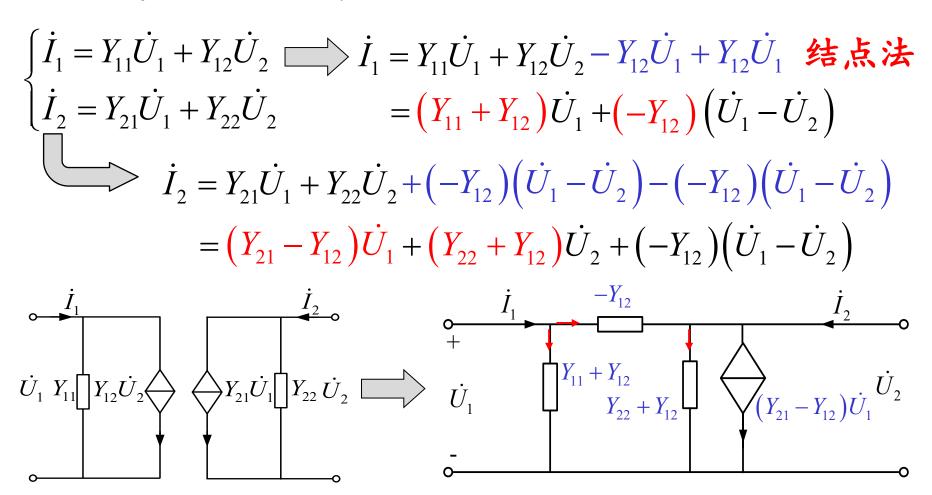
2. 由Z参数方程做等效电路 T形电路模型



$$\begin{cases} \dot{U}_{1} = Z_{11}\dot{I}_{1} + Z_{12}\dot{I}_{2} & \longrightarrow \dot{U}_{1} = Z_{11}\dot{I}_{1} + Z_{12}\dot{I}_{2} + Z_{12}\dot{I}_{1} - Z_{12}\dot{I}_{1} \\ \dot{U}_{2} = Z_{21}\dot{I}_{1} + Z_{22}\dot{I}_{2} & = \left(Z_{11} - Z_{12}\right)\dot{I}_{1} + Z_{12}\left(\dot{I}_{1} + \dot{I}_{2}\right) \\ & \longrightarrow \dot{U}_{2} = Z_{21}\dot{I}_{1} + Z_{22}\dot{I}_{2} + Z_{12}\left(\dot{I}_{1} + \dot{I}_{2}\right) - Z_{12}\left(\dot{I}_{1} + \dot{I}_{2}\right) \\ & = \left(Z_{21} - Z_{12}\right)\dot{I}_{1} + \left(Z_{22} - Z_{12}\right)\dot{I}_{2} + Z_{12}\left(\dot{I}_{1} + \dot{I}_{2}\right) \end{cases}$$

记住推导过程, 勿死记硬背公式

3. 由Y参数方程做等效电路 π形电路模型



通过已知参数,确定二端口网络等效电路模型

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例 已知二端口参数矩阵为 $Y = \begin{bmatrix} 5 & -2 \\ 0 & 3 \end{bmatrix}$ S, $Z = \begin{bmatrix} \frac{50}{9} & \frac{40}{9} \\ \frac{40}{9} & \frac{100}{9} \end{bmatrix}$ Ω 求它们的 π 形等效电路。

解 (1)
$$:: Y_{21} - Y_{12} \neq 0$$

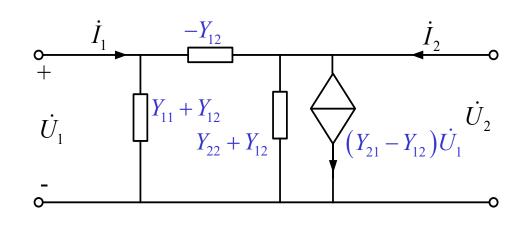
二. 非互易二端口网络, 含受控源

$$Y_{11} + Y_{12} = 3 \text{ S}$$

$$-Y_{12} = 2 \text{ S}$$

$$Y_{22} + Y_{12} = 1 \text{ S}$$

$$(Y_{21} - Y_{12})\dot{U}_1 = 2\dot{U}_1$$



例 已知二端口参数矩阵为
$$Y = \begin{bmatrix} 5 & -2 \\ 0 & 3 \end{bmatrix}$$
 S , $Z = \begin{bmatrix} \frac{60}{9} & \frac{40}{9} \\ \frac{40}{9} & \frac{100}{9} \end{bmatrix}$ Ω 求它们的π形等效电路。

解 (2) Z参数转换为Y参数矩阵

$$YZ = 1$$
 $Y = Z^{-1} = \begin{bmatrix} 0.2045 & -0.0818 \\ -0.0818 & 0.1227 \end{bmatrix} S$

$$Y_{11} + Y_{12} = 0.1227 \text{ S}$$

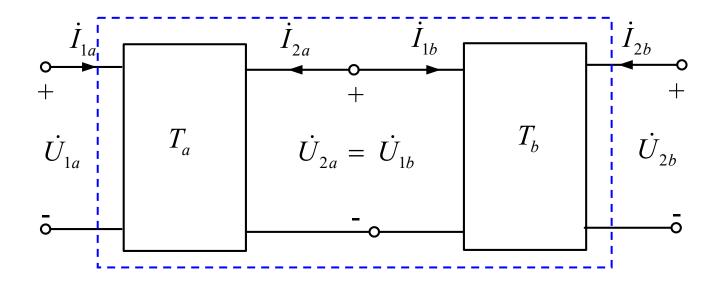
 $-Y_{12} = 0.0818 \text{ S}$
 $Y_{22} + Y_{12} = 0.0409 \text{ S}$

 $Y_{21} - Y_{12} = 0$

二端口网络等效电路模型不唯一,也可等效为T形电路模型

16.5 二端口网络的相互连接

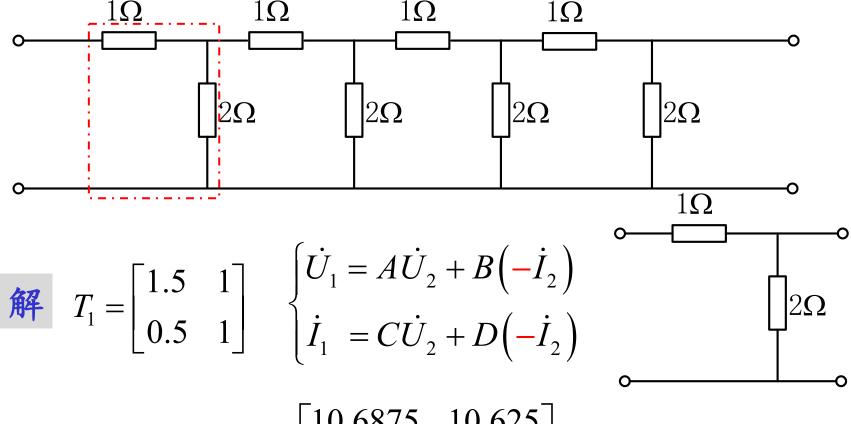
1. 级联 广泛用于电子信号滤波和传输中



$$\begin{bmatrix} \dot{U}_{1a} \\ \dot{I}_{1a} \end{bmatrix} = \boldsymbol{T}_a \begin{bmatrix} \dot{U}_{2a} \\ -\dot{I}_{2a} \end{bmatrix} = \boldsymbol{T}_a \begin{bmatrix} \dot{U}_{1b} \\ \dot{I}_{1b} \end{bmatrix} = \boldsymbol{T}_a \times \boldsymbol{T}_b \begin{bmatrix} \dot{U}_{2b} \\ -\dot{I}_{2b} \end{bmatrix}$$

$$T = T_a \times T_b$$

例求T参数矩阵。



$$T = T_1^4 = (T_1^2)^2 = \begin{bmatrix} 10.6875 & 10.625 \\ 5.3125 & 5.375 \end{bmatrix}$$

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