

# PWM and ADC

Lecture 4

# PWM and ADC

- Counters
- Timers and Alarms
- About Analog and Digital Signals
- Pulse Width Modulation (PWM)
- Analog to Digital Converters (ADC)

# Timers

# Bibliography

for this section

## **Raspberry Pi Ltd, *RP2040 Datasheet***

- Chapter 2 - *System Description*
  - Chapter 2.15 - *Clocks*
    - Subchapter 2.15.1
    - Subchapter 2.15.2
- Chapter 4 - *Peripherals*
  - Chapter 4.6 - *Timer*

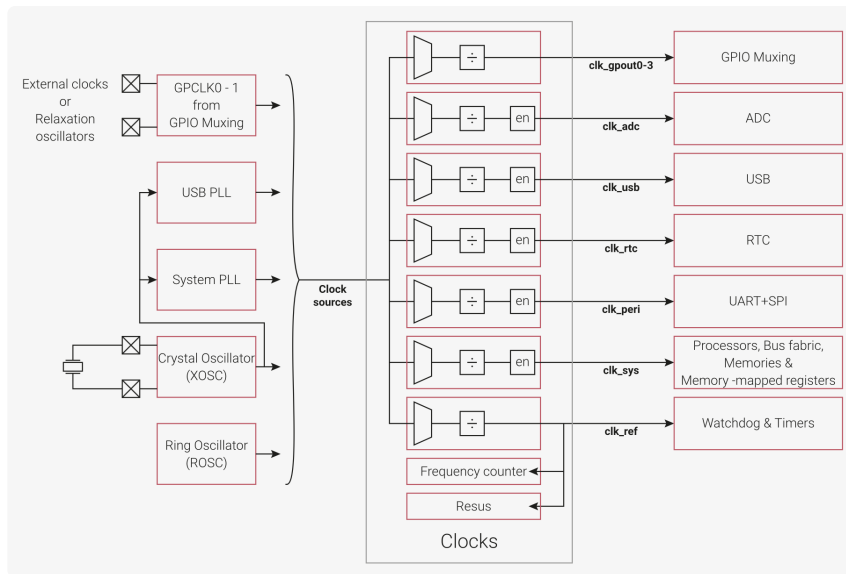
# Clocks

all peripherals and the MCU use a clock to execute at certain intervals

Source	Usage
<i>external crystal (XOSC)</i>	a stable frequency is required, for instance when using USB
<i>internal ring (ROSC)</i>	low frequency, in between 1.8 - 12 MHz (varies)

Embassy initializes the Raspberry Pi Pico with the clock source from the 12 MHz crystal.

```
1 let p = embassy_rp::init(Default::default());
```



# Frequency divider

stabilizing the signal and adjusting it

1. divides down the clock signals used for the timer, giving reduced overflow rates
2. allows the timer to be clocked at a user desires the rate



# Counter

increments a register at every clock cycle

Registers	Description
value	the current value of the counter
direction	set to count UP or DOWN
reset	UP: the value at which the counter resets to 0 DOWN: the value to which the counter resets after getting to 0



# SysTick

## ARM Cortex-M time counter

The ARM Cortex-M0+ registers start at a base address of `0xe0000000` (defined as `PPB_BASE` in SDK).

Offset	Name	Info
0xe010	<code>SYST_CSR</code>	SysTick Control and Status Register
0xe014	<code>SYST_RVR</code>	SysTick Reload Value Register
0xe018	<code>SYST_CVR</code>	SysTick Current Value Register
0xe01c	<code>SYST_CALIB</code>	SysTick Calibration Value Register

- decrements the value of `SYST_CVR` every  $\mu\text{s}$
- when `SYST_CVR` becomes 0 :
  - triggers the SysTick the exception
  - next clock cycle sets the value of `SYST_CVR` to `SYST_RVR`
- `SYST_CALIB` is the value of `SYST_RVR` for a 10ms interval (might not be available)

## SYST\_CSR register

Bits	Name	Description	Type	Reset
31:17	Reserved.	-	-	-
16	COUNTFLAG	Returns 1 if timer counted to 0 since last time this was read. Clears on read by application or debugger.	RO	0x0
15:3	Reserved.	-	-	-
2	CLKSOURCE	SysTick clock source. Always reads as one if SYST_CALIB reports NOREF. Selects the SysTick timer clock source: 0 = External reference clock. 1 = Processor clock.	RW	0x0
1	TICKINT	Enables SysTick exception request: 0 = Counting down to zero does not assert the SysTick exception request. 1 = Counting down to zero to asserts the SysTick exception request.	RW	0x0
0	ENABLE	Enable SysTick counter: 0 = Counter disabled. 1 = Counter enabled.	RW	0x0

$$f = \frac{1}{SYST\_RVR} * 1,000,000 [Hz]_{SI}$$



# SysTick

## ARM Cortex-M peripheral

The ARM Cortex-M0+ registers start at a base address of `0xe0000000` (defined as `PPB_BASE` in SDK).

Offset	Name	Info
0xe010	<a href="#">SYST_CSR</a>	SysTick Control and Status Register
0xe014	<a href="#">SYST_RVR</a>	SysTick Reload Value Register
0xe018	<a href="#">SYST_CVR</a>	SysTick Current Value Register
0xe01c	<a href="#">SYST_CALIB</a>	SysTick Calibration Value Register

```
1  const SYST_RVR: *mut u32 = 0xe000_e014 as *mut u32;
2  const SYST_CVR: *mut u32 = 0xe000_e018 as *mut u32;
3  const SYST_CSR: *mut u32 = 0xe000_e010 as *mut u32;
4
5  // fire systick every 5 seconds
6  let interval: u32 = 5_000_000;
7  unsafe {
8      write_volatile(SYST_RVR, interval);
9      write_volatile(SYST_CVR, 0);
10     // set fields `ENABLE` and `TICKINT`
11     write_volatile(SYST_CSR, 0b11);
12 }
```

## SYST\_CSR register

Bits	Name	Description	Type	Reset
31:17	Reserved.	-	-	-
16	COUNTFLAG	Returns 1 if timer counted to 0 since last time this was read. Clears on read by application or debugger.	RO	0x0
15:3	Reserved.	-	-	-
2	CLKSOURCE	SysTick clock source. Always reads as one if SYST_CALIB reports NOREF. Selects the SysTick timer clock source: 0 = External reference clock. 1 = Processor clock.	RW	0x0
1	TICKINT	Enables SysTick exception request: 0 = Counting down to zero does not assert the SysTick exception request. 1 = Counting down to zero to asserts the SysTick exception request.	RW	0x0
0	ENABLE	Enable SysTick counter: 0 = Counter disabled. 1 = Counter enabled.	RW	0x0

## Register SysTick handler

```
1  #[exception]
2  unsafe fn SysTick() {
3      /* systick fired */
4  }
```

# Alarm

counter that triggers interrupts after a time interval

Registers	Description
<code>value</code>	the current value of the counter
<code>direction</code>	set to count UP or DOWN
<code>reset</code>	UP: max value before 0 DOWN: value after 0
<code>alarm_x</code>	when <code>value == alarm_x</code> , triggers an interrupt, <code>x</code> in 1 ... n



# RP2040's Timer

- stores a 64 bit number ( `reset` is  $2^{64}-1$  )
- starts with `0` at (the peripheral's) reset
- increments the number every  $\mu\text{s}$
- in practice fully monotonic (cannot over flow)
- allows 4 alarms that trigger interrupts
  - `TIMER_IRQ_0`
  - `TIMER_IRQ_1`
  - `TIMER_IRQ_2`
  - `TIMER_IRQ_3`
- `alarm_0 ... alarm_3` registers are only 32 bits wide



# RP2040's Timer

read the number of elapsed  $\mu$ s since reset

The Timer registers start at a base address of `0x40054000` (defined as `TIMER_BASE` in SDK).

Offset	Name	Info
0x00	<a href="#">TIMEHW</a>	Write to bits 63:32 of time always write timelw before timehw
0x04	<a href="#">TIMELW</a>	Write to bits 31:0 of time writes do not get copied to time until timehw is written

## Reading the time elapsed since restart

```
1  const TIMERLR: *const u32 = 0x4005_400c;  
2  const TIMERHR: *const u32 = 0x4005_4008;  
3  
4  let time: u64 = unsafe {  
5      let low = read_volatile(TIMERLR);  
6      let high = read_volatile(TIMERHR);  
7      high as u64 << 32 | low  
8  }
```

The **reading order matters** as reading `TIMERLR` latches the value in `TIMERHR` (stops being updated) until `TIMERHR` is read. Works only in **single core**.

Offset	Name	Info
0x08	<a href="#">TIMEHR</a>	Read from bits 63:32 of time always read timelr before timehr
0x0c	<a href="#">TIMELR</a>	Read from bits 31:0 of time
0x10	<a href="#">ALARM0</a>	Arm alarm 0, and configure the time it will fire. Once armed, the alarm fires when <code>TIMER_ALARM0 == TIMELR</code> . The alarm will disarm itself once it fires, and can be disarmed early using the ARMED status register.
0x14	<a href="#">ALARM1</a>	Arm alarm 1, and configure the time it will fire. Once armed, the alarm fires when <code>TIMER_ALARM1 == TIMELR</code> . The alarm will disarm itself once it fires, and can be disarmed early using the ARMED status register.
0x18	<a href="#">ALARM2</a>	Arm alarm 2, and configure the time it will fire. Once armed, the alarm fires when <code>TIMER_ALARM2 == TIMELR</code> . The alarm will disarm itself once it fires, and can be disarmed early using the ARMED status register.
0x1c	<a href="#">ALARM3</a>	Arm alarm 3, and configure the time it will fire. Once armed, the alarm fires when <code>TIMER_ALARM3 == TIMELR</code> . The alarm will disarm itself once it fires, and can be disarmed early using the ARMED status register.
0x20	<a href="#">ARMED</a>	Indicates the armed/disarmed status of each alarm. A write to the corresponding ALARMx register arms the alarm. Alarms automatically disarm upon firing, but writing ones here will disarm immediately without waiting to fire.
0x24	<a href="#">TIMERAWH</a>	Raw read from bits 63:32 of time (no side effects)
0x28	<a href="#">TIMERAWL</a>	Raw read from bits 31:0 of time (no side effects)
0x2c	<a href="#">DBGPAUSE</a>	Set bits high to enable pause when the corresponding debug ports are active
0x30	<a href="#">PAUSE</a>	Set high to pause the timer
0x34	<a href="#">INTR</a>	Raw Interrupts
0x38	<a href="#">INTE</a>	Interrupt Enable
0x3c	<a href="#">INTF</a>	Interrupt Force
0x40	<a href="#">INTS</a>	Interrupt status after masking & forcing

# Alarm

triggering an interrupt at an interval

```
1  #[interrupt]
2  unsafe fn TIMER_IRQ_0() { /* alarm fired */ }

1  const TIMERLR: *const u32 = 0x4005_400c;
2  const ALARM0: *mut u32 = 0x4005_4010;
3  // + 0x2000 is bitwise set
4  const INTE_SET: *mut u32 = 0x4005_4038 + 0x2000;
5
6  // set an alarm after 3 seconds
7  let us = 3_0000_0000;
8
9  unsafe {
10     let time = read_volatile(TIMERLR);
11     // use `wrapping_add` as overflowing may panic
12     write_volatile(ALARM0, time.wrapping_add(us));
13     write_volatile(INTE_SET, 1 << 0);
14 };
```

- the alarm can be set only for the lower 32 bits
- maximum 72 minutes (use *RTC* for longer alarms)

Offset	Name	Info
0x08	<a href="#">TIMEHR</a>	Read from bits 63:32 of time always read timelr before timehr
0x0c	<a href="#">TIMELR</a>	Read from bits 31:0 of time
0x10	<a href="#">ALARM0</a>	Arm alarm 0, and configure the time it will fire. Once armed, the alarm fires when <code>TIMER_ALARM0 == TIMELR</code> . The alarm will disarm itself once it fires, and can be disarmed early using the ARMED status register.
0x14	<a href="#">ALARM1</a>	Arm alarm 1, and configure the time it will fire. Once armed, the alarm fires when <code>TIMER_ALARM1 == TIMELR</code> . The alarm will disarm itself once it fires, and can be disarmed early using the ARMED status register.
0x18	<a href="#">ALARM2</a>	Arm alarm 2, and configure the time it will fire. Once armed, the alarm fires when <code>TIMER_ALARM2 == TIMELR</code> . The alarm will disarm itself once it fires, and can be disarmed early using the ARMED status register.
0x1c	<a href="#">ALARM3</a>	Arm alarm 3, and configure the time it will fire. Once armed, the alarm fires when <code>TIMER_ALARM3 == TIMELR</code> . The alarm will disarm itself once it fires, and can be disarmed early using the ARMED status register.
0x20	<a href="#">ARMED</a>	Indicates the armed/disarmed status of each alarm. A write to the corresponding ALARMx register arms the alarm. Alarms automatically disarm upon firing, but writing ones here will disarm immediately without waiting to fire.
0x24	<a href="#">TIMERAWH</a>	Raw read from bits 63:32 of time (no side effects)
0x28	<a href="#">TIMERAWL</a>	Raw read from bits 31:0 of time (no side effects)
0x2c	<a href="#">DBGPAUSE</a>	Set bits high to enable pause when the corresponding debug ports are active
0x30	<a href="#">PAUSE</a>	Set high to pause the timer
0x34	<a href="#">INTR</a>	Raw Interrupts
0x38	<a href="#">INTE</a>	Interrupt Enable
0x3c	<a href="#">INTF</a>	Interrupt Force
0x40	<a href="#">INTS</a>	Interrupt status after masking & forcing

# Signals

Digital Signals - Recap

# Signals

## Analog vs Digital

- *analog signals* are *real* signals
- *digital signals* are a *numerical representation* of an analog signal (software level)
- hardware usually works with two-level digital signals (hardware level)

### Exceptions

- in wireless and in high-speed cable communication things get more complicated

for PCB level / between integrated circuits on the same board / inside the same chip - things are a "a little simpler" - as detailed in the following



# Why use digital in computing?

Signal that we *want* to generate with an output pin



Signal that what we actually generate



Why we still use it? Because after passing through an IC or a gate inside an IC - the signal is "rebuilt" and if the "digital discipline" described in the following is respected - we can preserve the information after numerous "passes". Thus, each element can behave with a large margin for error, yet the final result is correct.



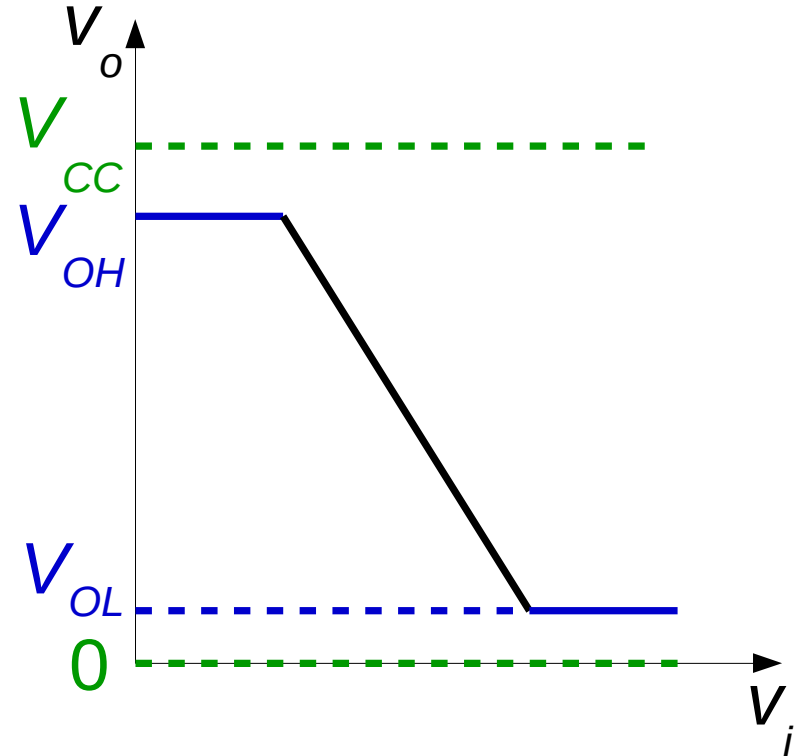
# Noise Margin



# Why is the output not ideal?

The two corresponding voltage output levels are affected by:

- power supply voltage
- output current
- temperature
- variations in the manufacturing process



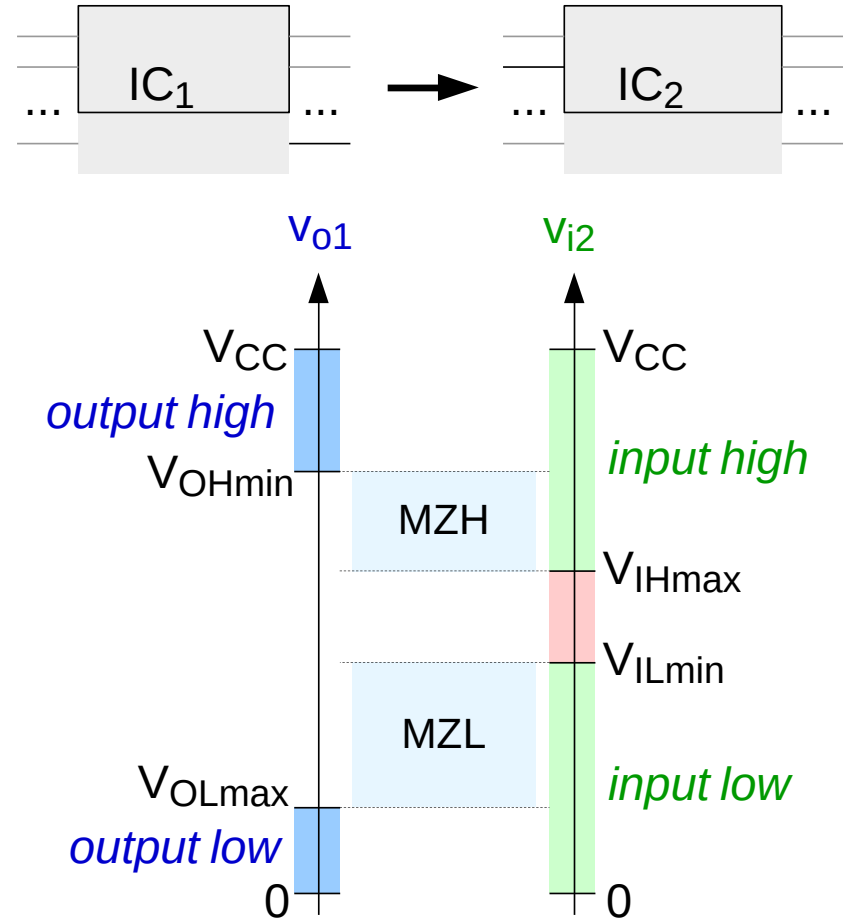
# ICs same voltage

Usually will work as is

- usually, they will be compatible
- conditions:

$$V_{OH\_transmitter} > V_{IH\_receiver}$$

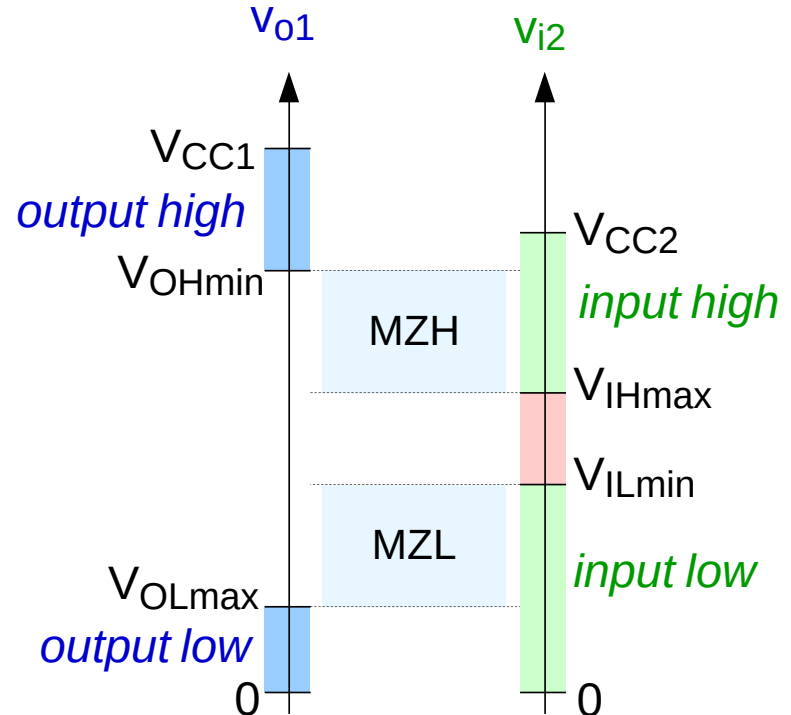
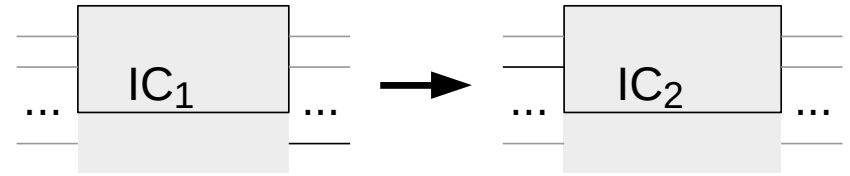
$$V_{OL\_transmitter} < V_{IL\_receiver}$$



# $V_{CC1} > V_{CC2}$

Might work, might produce magic smock

$$V_{OH\_transmitter} > V_{CC\_receiver}$$



## PROBLEM

Solutions:

- level shifter
- resistor divider / voltage limiter

Examples:

- Bi-Directional Level Shifter with 4 Channels
- Level Shifter Multi-Channel
- 8 Channels Level Shifter

# $V_{CC1} < V_{CC2}$

Might work

$$V_{CC\_transmitter} \lesssim V_{IH\_receiver}$$

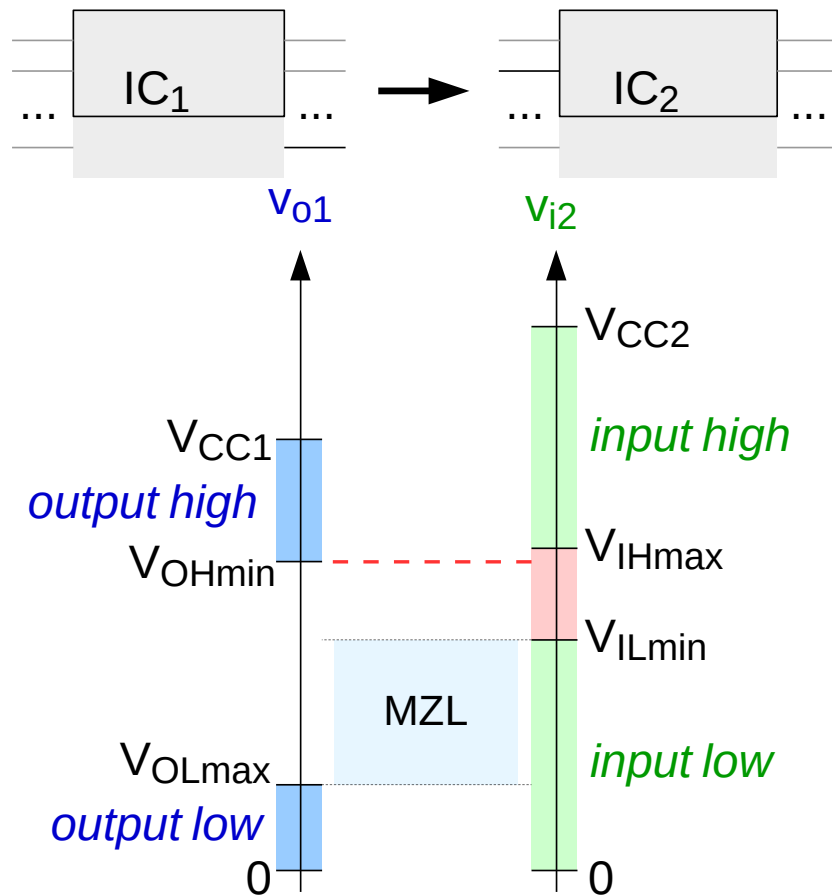
**Might work in an intermittent mode - hard to debug!**

Solutions:

- level shifter
- resistor divider / voltage limiter

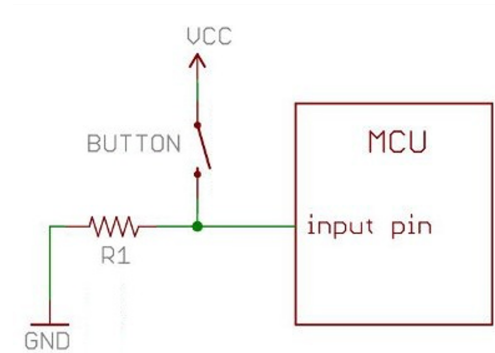
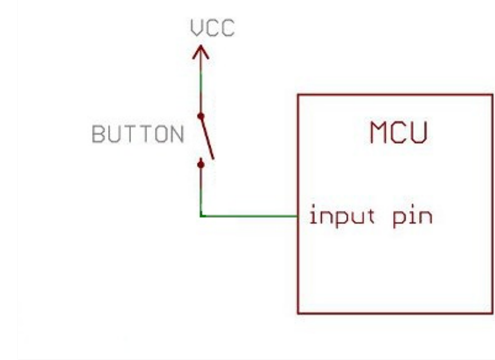
Examples:

- Bi-Directional Level Shifter with 4 Channels
- Level Shifter Multi-Channel
- 8 Channels Level Shifter



# Why Pull-Down R

- Without pull-down – when the button is not pressed, it leaves the input pin floating.
- The second design ensures that the voltage level has a well-defined state, regardless of the button's state.
- R1 is called a "pull-down" resistor.

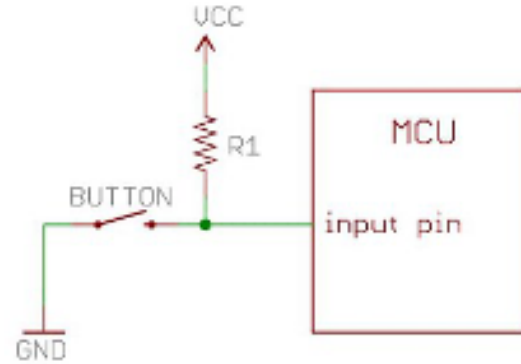


# Why Pull-Up R

- Same reasoning
- R1 is called a "pull-up" resistor.

##Obs:

- most microcontrollers have at least a pull-up resistor incorporated on GPIOs - that can be activated in software
- some have both pull-up and pull-down
- typically, these are sized for a 50 - 10 nA current consumption



# Notes on output pins

- most microcontrollers have a limit of around 10mA per output PIN
- ! do not connect an LED without a resistor in series (to limit the current)
- ! do not connect a motor / any type of inductive load

## Solutions:

- use a transistor
- use an IC with incorporated Darlingtontons (eg: ULN2003)



# PWM

Pulse Width Modulation

# Bibliography

for this section

## 1. **Raspberry Pi Ltd**, *RP2040 Datasheet*

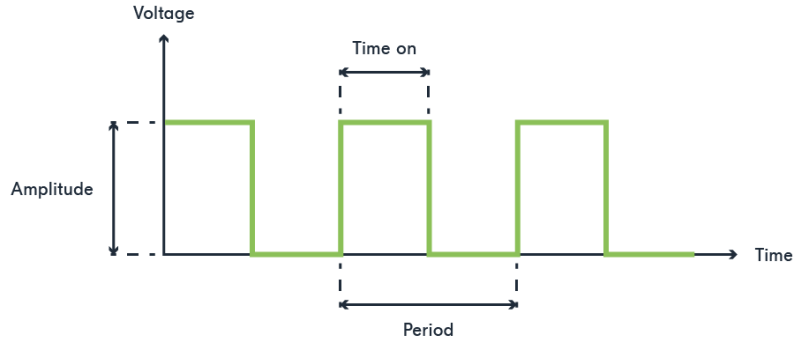
- Chapter 4 - *Peripherals*
  - Chapter 4.5 - *PWM*

## 2. **Paul Denisowski**, *Understanding PWM*

# PWM

simulates an *analog* signal (using integration)

- generates a square signal
- if integrated (averaged), it looks like an analog signal

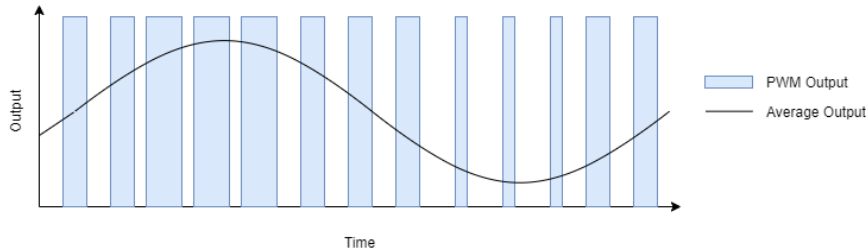


*frequency*    Hz    The number of repeats per s

*duty\_cycle*    %    The percentage of the time when the signal is High

$$f = \frac{1}{\text{period}} \left[ \frac{1}{s} = 1\text{Hz} \right]_{SI}$$

$$\text{duty\_cycle} = \frac{\text{time\_on}}{\text{period}} \%$$



# PWM

generic device

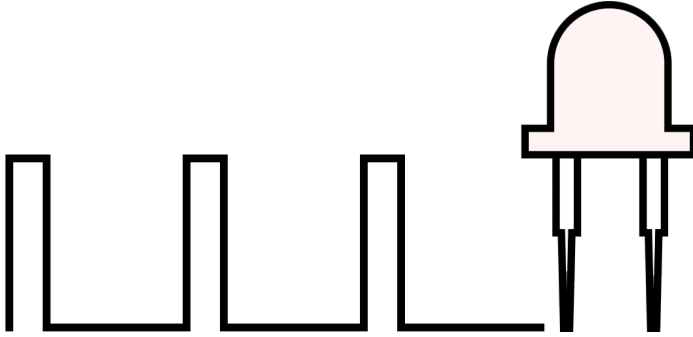
$$f = \begin{cases} \frac{f_{clock}}{divider \times (top+1)} & correction = 0 \\ \frac{f_{clock}}{divider \times 2 \times (top+1)} & correction = 1 \end{cases}$$

$$pin_{a,b} = \begin{cases} 0 & compare_{a,b} \geq value \\ 1 & compare_{a,b} < value \end{cases}$$

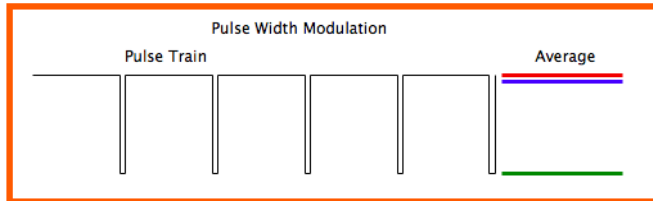


# Usage examples

- dimming an LED

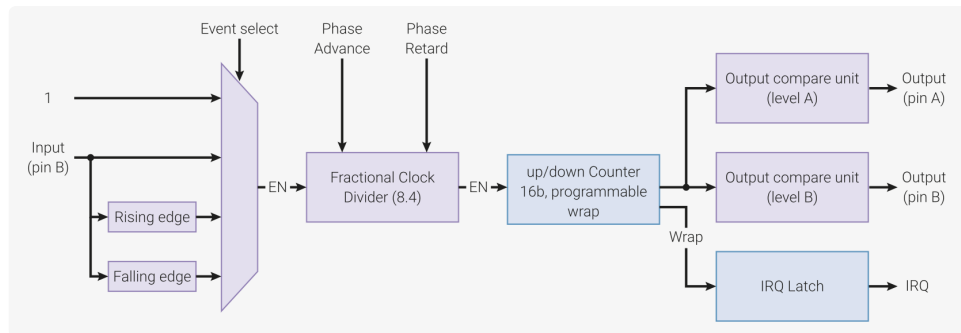


- controlling motors
  - controlling the angle of a stepper motor
  - controlling the RPM of a motor



# RP2040's PWM

- generates square signals
- counts the pulse with of input signals
- 8 PWM units, each with 2 channels (A and B)
- each PWM channel is connected to a certain pin
- some channels are connected to two pins



All 30 GPIO pins on RP2040 can be used for PWM:

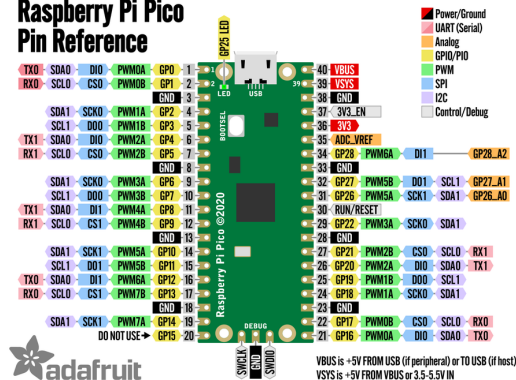
GPIO	0	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15
PWM Channel	0A	0B	1A	1B	2A	2B	3A	3B	4A	4B	5A	5B	6A	6B	7A	7B
GPIO	16	17	18	19	20	21	22	23	24	25	26	27	28	29		
PWM Channel	0A	0B	1A	1B	2A	2B	3A	3B	4A	4B	5A	5B	6A	6B		

## Registers

The PWM registers start at a base address of `0x40050000` (defined as `PWM_BASE` in SDK).

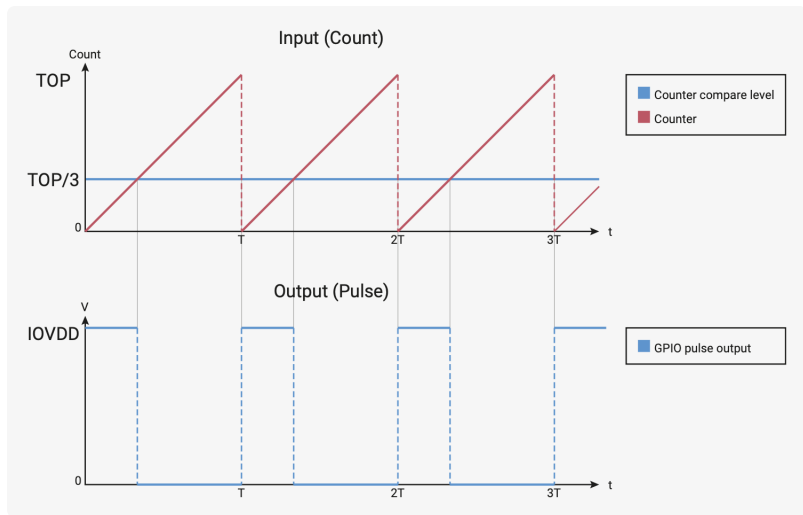
Offset	Name	Info
0x00	CH0_CSR	Control and status register
0x04	CH0_DIV	INT and FRAC form a fixed-point fractional number. Counting rate is system clock frequency divided by this number. Fractional division uses simple 1st-order sigma-delta.
0x08	CH0_CTR	Direct access to the PWM counter
0x0c	CH0_CC	Counter compare values
0x10	CH0_TOP	Counter wrap value

## Raspberry Pi Pico Pin Reference

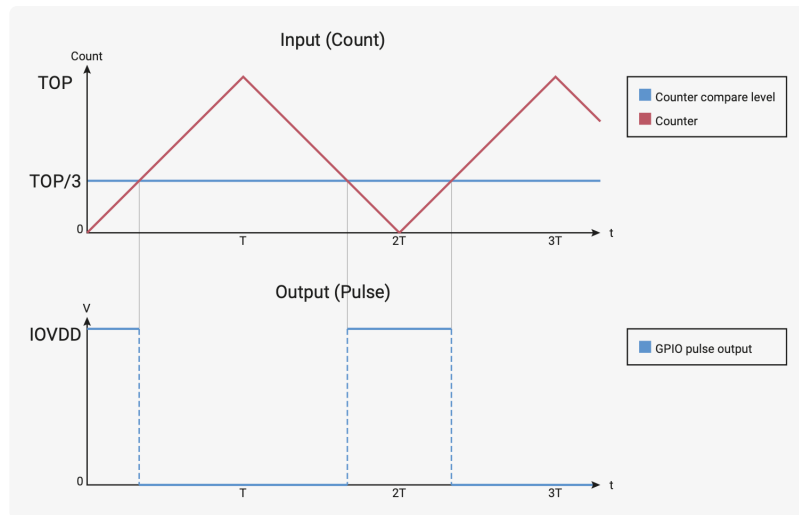


# RP2040's PWM Modes

standard mode



phase-correct mode



$$period = (TOP + 1) \times (PH\_CORRECT + 1) \times \left( DIV\_INT + \frac{DIV\_FRAC}{16} \right) [s]_{SI}$$

$$f = \frac{f_{sys}}{period} [Hz]_{SI}$$

# Example

using Embassy

```
1 use embassy_rp::pwm::{Config, Pwm};
2
3 let p = embassy_rp::init(Default::default());
4
5 let mut c: Config = Default::default();
6 c.top = 0x8000;
7 c.compare_b = 8;
8
9 let mut pwm = Pwm::new_output_b(
10     p.PWM_CH4,
11     p.PIN_25,
12     c.clone()
13 );
14
15 loop {
16     info!("LED duty cycle: {}/32768", c.compare_b);
17     Timer::after_secs(1).await;
18     c.compare_b += 10;
19     pwm.set_config(&c);
20 }
```

```
pub struct Config {
    /// Inverts the PWM output signal on channel A.
    pub invert_a: bool,
    /// Inverts the PWM output signal on channel B.
    pub invert_b: bool,
    /// Enables phase-correct mode for PWM operation.
    pub phase_correct: bool,
    /// Enables the PWM slice, allowing it to generate an out
    pub enable: bool,
    /// A fractional clock divider, represented as a fixed-po
    /// 8 integer bits and 4 fractional bits. It allows preci
    /// the PWM output frequency by gating the PWM counter in
    /// A higher value will result in a slower output frequen
    pub divider: fixed::FixedU16<fixed::types::extra::U4>,
    /// The output on channel A goes high when `compare_a` is
    /// counter. A compare of 0 will produce an always low ou
    pub compare_a: u16,
    /// The output on channel B goes high when `compare_b` is
    /// counter.
    pub compare_b: u16,
    /// The point at which the counter wraps, representing th
    /// period. The counter will either wrap to 0 or reverse
    /// setting of `phase_correct`.
    pub top: u16,
}
```



# ADC

Analog to Digital Converter

# Bibliography

for this section

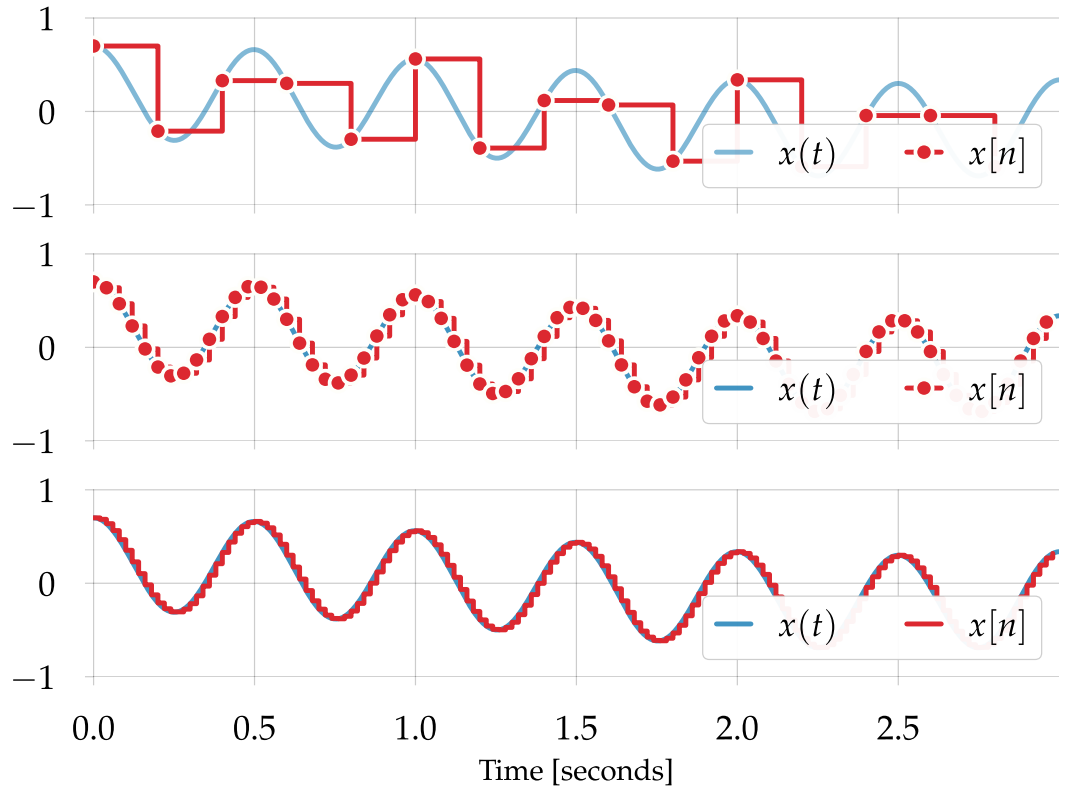
## **Raspberry Pi Ltd, *RP2040 Datasheet***

- Chapter 4 - *Peripherals*
  - Chapter 4.9 - *ADC and Temperature Sensor*
    - Subchapter 4.9.1
    - Subchapter 4.9.2
    - Subchapter 4.9.5

# ADC

sampling an analog signal to an array of values

<i>sampling rate</i>	Hz	the frequency at which a new sample is read
<i>resolution</i>	bits	the number of bits used to store a sampled value

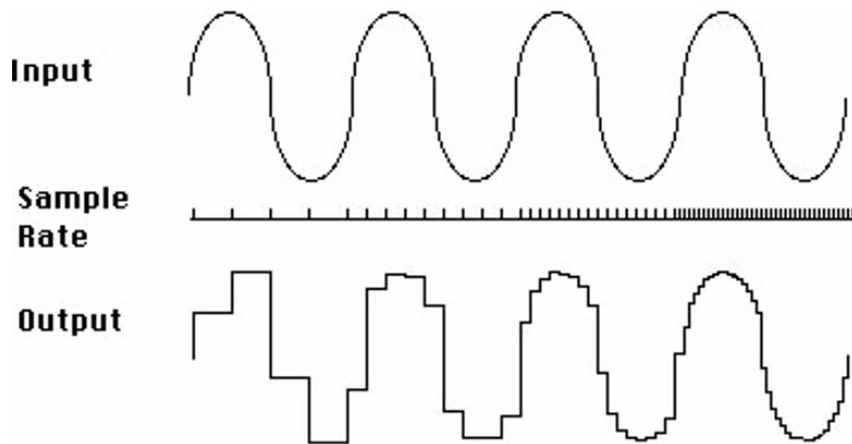


Lower sample rates yield the *aliasing effect*.

# Nyquist–Shannon Sampling Theorem

$$sampling_f \geq 2 \times max_f$$

The **sampling frequency** has to be at least **two times higher** than the **maximum frequency** of the signal to avoid frequency aliasing<sup>[1]</sup>.

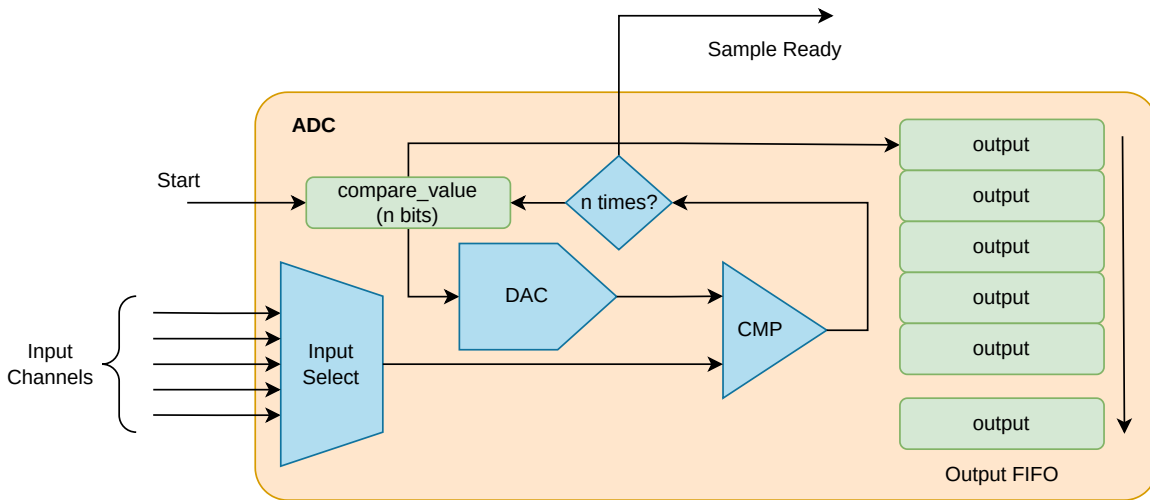


- 
1. Aliasing is the overlapping of frequency components. This overlap results in distortion or artifacts when the signal is reconstructed from samples which causes the **reconstructed signal to differ from the original** continuous signal. ↩

# Sampling

how the ADC works

- assumes  $\text{bit}_{n-1}$  of `compare_value` is 1
- compares the input signal with a generated analog signal from `compare_value`
  - if input is lower,  $\text{bit}_{n-1}$  is 0
  - if input is higher,  $\text{bit}_{n-1}$  is 1
- repeats for  $\text{bit}_{n-2}$ ,  $\text{bit}_{n-3} \dots \text{bit}_0$



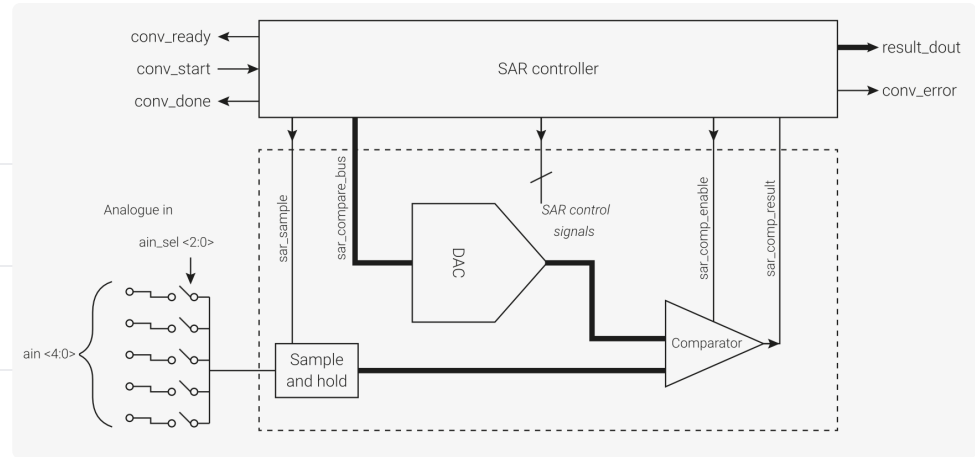
There are different types of ADCs depending on the architecture. The most common used is SAR (Successive Approximation Register) ADC, also integrated in RP2040.

# RP2040's ADC

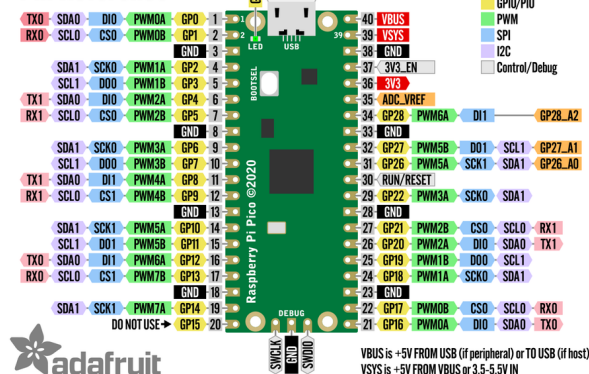
<i>channels</i>	5
<i>sampling rate</i>	500 kHz
<i>resolution</i>	12 bits
<i>V<sub>max</sub></i>	3.3 V

- requires a 48 MHz clock signal
- channel 4 is connected to the internal temperature sensor

$$t = 27 - \frac{(V_{input\_4} - 0.706)}{0.001721} [^{\circ}C]_{SI}$$



## Raspberry Pi Pico Pin Reference



# ADC

in Embassy

```
1 use embassy_rp::adc::{Adc, Channel, Config, InterruptHandler};
2
3 bind_interrupts!(struct Irqs {
4     ADC_IRQ_FIFO => InterruptHandler;
5 });
6
7 let p = embassy_rp::init(Default::default());
8 let mut adc = Adc::new(p.ADC, Irqs, Config::default());
9
10 let mut p26 = Channel::new_pin(p.PIN_26, Pull::None);
11
12 loop {
13     let level = adc.read(&mut p26).await.unwrap();
14     info!("Pin 26 ADC: {}", level);
15     let voltage = 3300 * level / 4095;
16     info!("Pin 26 voltage: {:.{}V", voltage / 1000, voltage % 1000);
17     Timer::after_secs(1).await;
18 }
```

# Conclusion

we talked about

- Counters
- SysTick
- Timers and Alarms
- PWM
- Analog and Digital
- ADC