

Lecture 4b. Transfer Functions

... in which we begin our descent into the ...

FREQUENCY DOMAIN !!

I. Transfer Functions for Linear ODEs.

Consider an ODE of the form

$$a_2 \ddot{y} + a_1 \dot{y} + a_0 y = b_1 \dot{u} + b_0 u$$

where y is the output and u is the input. The transfer function of the ODE is defined by

$$\frac{Y(s)}{U(s)} = \frac{\text{Laplace Transform of the output}}{\text{Laplace Transform of the Input}}$$

Assuming $y(0) = \dot{y}(0) = u(0) = 0$.

Recall the Laplace Transform:

$$\mathcal{L}[f(t)] = \int_0^\infty e^{-st} f(t) dt$$

where $s \in \mathbb{C}$.

The most important consequence of this definition is the following:

Say $F(s)$ is the Laplace Transform of $f(t)$. Then

$$\begin{aligned}\mathcal{L}[f'(t)] &= \int_0^\infty e^{-st} f'(t) dt \\ &= e^{-st} f(t) \Big|_0^\infty + s \int_0^\infty e^{-st} f(t) dt \quad \boxed{\int u dv = uv - \int v du} \\ &= -f(0) + sF(s).\end{aligned}$$

Also, $f''(t) \xrightarrow{\mathcal{L}} s^2 F(s)$ etc.

Example:

$$2\ddot{y} - \dot{y} + y = u - u$$

$$\overset{\uparrow}{L}$$

$$2s^2Y(s) - sY(s) + Y(s) = sU(s) - u(s)$$

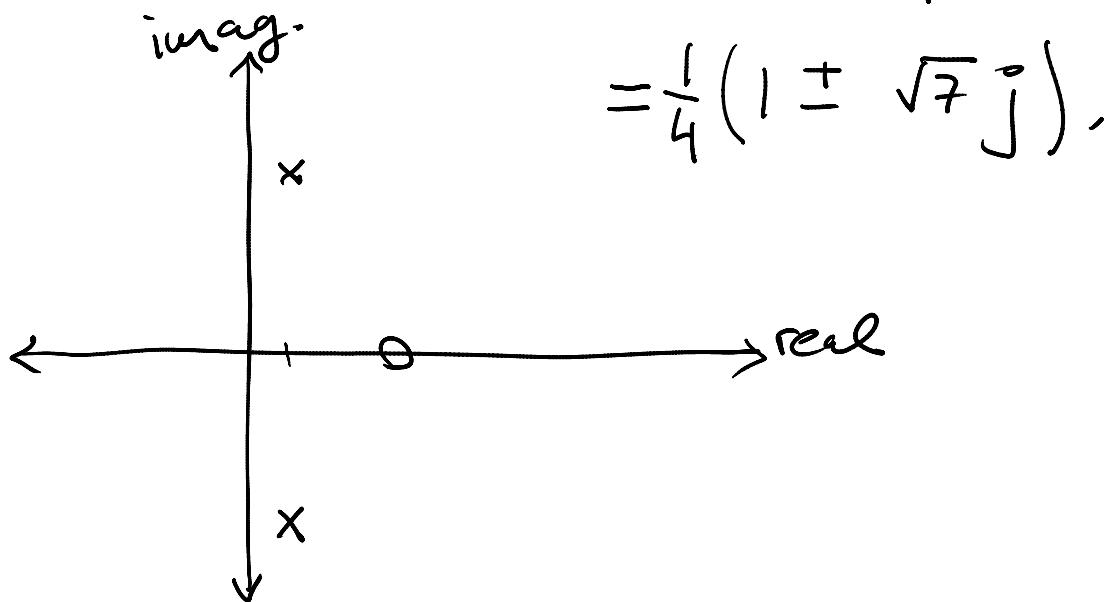
$$\Leftrightarrow Y(s)(2s^2 - s + 1) = (s-1)U(s)$$

$$T(s) = \frac{Y(s)}{U(s)} = \boxed{\frac{s-1}{2s^2 - s + 1}}$$

- $2s^2 - s + 1$ is the characteristic equation. Its roots are the poles or eigenvalues.
- The roots of $s-1$ are called the zeros of $T(s)$.

Pole zero plots:

$$\begin{array}{l|l} 0 = \text{zero} & 2s^2 - s + 1 = 0 \\ x = \text{pole} & \Rightarrow s = \frac{1 \pm \sqrt{1-8}}{4} \end{array}$$



We will get back to these. The point is that a pole-zero plot tells you pretty much everything you need to know about $T(s)$.

II. T.F.s from State Space

Consider the system

$$\dot{\vec{x}} = A\vec{x} + Bu$$

$$y = C\vec{x},$$

Our goal is to write the T.F. for this system. Taking the Laplace Transform:

$$sX(s) = AX(s) + BU(s)$$

$$Y(s) = \underbrace{CX(s)}_{\substack{\text{vector of} \\ \text{L.P.T. of } x(t)}} \quad$$

$$\Leftrightarrow sX(s) - AX(s) = BU(s)$$

$$\Leftrightarrow (sI - A)X(s) = BU(s)$$

$$X(s) = (sI - A)^{-1}BU(s)$$

Since $\Psi(s) = CX(s)$, we get

$$\begin{aligned} T(s) &= \frac{\Psi(s)}{U(s)} = \frac{CX(s)}{U(s)} \\ &= \frac{C(SI-A)^{-1}B\cancel{U(s)}}{\cancel{U(s)}} \\ &= C(SI-A)^{-1}B \end{aligned}$$

The quantity $(SI-A)^{-1}$ is usually denoted $\Phi(s)$.

[Notice the similarity between the eigenvalue formula and $\Phi(s)$.]

We get

$$T(s) = C\Phi(s)B$$

Example:

$$\dot{\vec{x}} = \begin{pmatrix} 0 & 1 \\ -\frac{1}{2} & \frac{1}{2} \end{pmatrix} \vec{x} + \begin{pmatrix} -1 \\ \frac{1}{2} \end{pmatrix} u$$

$$y = \left(\begin{smallmatrix} -\frac{1}{2} & 0 \end{smallmatrix} \right) \vec{x}$$

$$\Phi(s) = (sI - A)^{-1} = \begin{pmatrix} s & -1 \\ \frac{1}{2} & s - \frac{1}{2} \end{pmatrix}^{-1}$$

$$= \frac{1}{s(s - \frac{1}{2}) + \frac{1}{2}} \begin{pmatrix} s - \frac{1}{2} & 1 \\ -\frac{1}{2} & s \end{pmatrix}$$

$$= \frac{2}{s^2 - s + 1} \begin{pmatrix} s - \frac{1}{2} & 1 \\ -\frac{1}{2} & s \end{pmatrix}.$$

S_0

$$T(s) = C \dot{\phi}(s) B$$

$$= \frac{2}{2s^2 - s + 1} \begin{pmatrix} \frac{1}{2} & 0 \\ -\frac{1}{2} & 0 \end{pmatrix} \begin{pmatrix} s - \frac{1}{2} & 1 \\ -\frac{1}{2} & s \end{pmatrix} \begin{pmatrix} -1 \\ \frac{1}{2} \end{pmatrix}$$

$$= \frac{1}{2s^2 - s + 1} \begin{pmatrix} \frac{1}{2} - s & -1 \\ \frac{1}{2} & \end{pmatrix} \begin{pmatrix} -1 \\ \frac{1}{2} \end{pmatrix}$$

$$= \frac{1}{2s^2 - s + 1} \begin{pmatrix} s - \frac{1}{2} & -\frac{1}{2} \\ \frac{1}{2} & \end{pmatrix} = \boxed{\frac{s-1}{2s^2 - s + 1}}$$

Note: This is the same as for

$$2\ddot{y} - \dot{y} + y = \ddot{u} - u. \quad \boxed{\text{Why?}}$$

III. Final Value Theorem

If $F(s) \longleftrightarrow f(t)$ then

$$\mathcal{L}[f'(t)] = sF(s) - f(0)$$

$$= \int_0^\infty f'(t) e^{-st} dt$$

Take the limit as $s \rightarrow 0$:

$$\int_0^\infty f'(t) dt = \lim_{s \rightarrow 0} (sF(s) - f(0))$$

$$f(\infty) - f(0) = \lim_{s \rightarrow 0} sF(s) - f(0)$$

$$\boxed{\lim_{t \rightarrow \infty} f(t) = \lim_{s \rightarrow 0} sF(s)}.$$

Note:

- ① Can't have more than one pole at 0
- ② Must be stable (no poles in RHP).

Example:

$$Y(s) = \frac{s+4}{s(s+1)}$$

$$\lim_{s \rightarrow 0} s Y(s) = \lim_{s \rightarrow 0} \frac{s+4}{s+1} = 4.$$

Check: $Y(s) = \frac{4}{s} - \frac{3}{s+1}$



$$y(t) = 4 - 3e^{-t} \rightarrow 4.$$

This gives a way to check the response of a transfer function to various inputs.

STEP INPUTS:

A unit step is the function

$$r(t) = 1 \quad \xleftarrow{\mathcal{L}} \quad R(s) = \frac{1}{s} .$$

($t \geq 0$).

$$\begin{aligned} \text{So } T(s) &= \frac{Y(s)}{R(s)} \Rightarrow Y(s) = R(s)T(s) \\ &= \frac{1}{s} T(s) . \end{aligned}$$

for example, say

$$T(s) = \frac{Y(s)}{R(s)} = \frac{s+2}{(s+1)^2}$$

$$Y(s) = \frac{1}{s} T(s) = \frac{s+2}{s(s+1)^2}$$

$$\lim_{s \rightarrow 0} s Y(s) = \lim_{s \rightarrow 0} \frac{s+2}{(s+1)^2} \\ = 2.$$