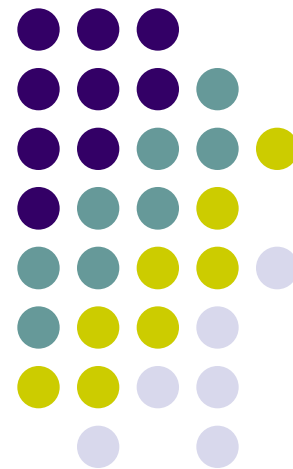
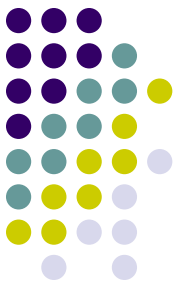


超大规模集成电路基础

Fundamental of VLSI

第八章 功能设计

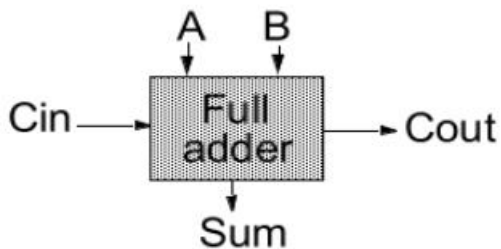




二进制加法器

- 时序电路的时间参数

$$\begin{aligned} S &= A \oplus B \oplus C_i \\ &= \bar{A}\bar{B}\bar{C}_i + \bar{A}B\bar{C}_i + \bar{A}BC_i + ABC_i \\ C_o &= AB + BC_i + AC_i \end{aligned}$$



A	B	C_i	S	C_o	<i>Carry status</i>
0	0	0	0	0	delete
0	0	1	1	0	delete
0	1	0	1	0	propagate
0	1	1	0	1	propagate
1	0	0	1	0	propagate
1	0	1	0	1	propagate
1	1	0	0	1	generate
1	1	1	1	1	generate



二进制加法器

- 产生，取消，传播

$$G = AB$$

$$D = \overline{A}\overline{B}$$

$$P = A \oplus B$$

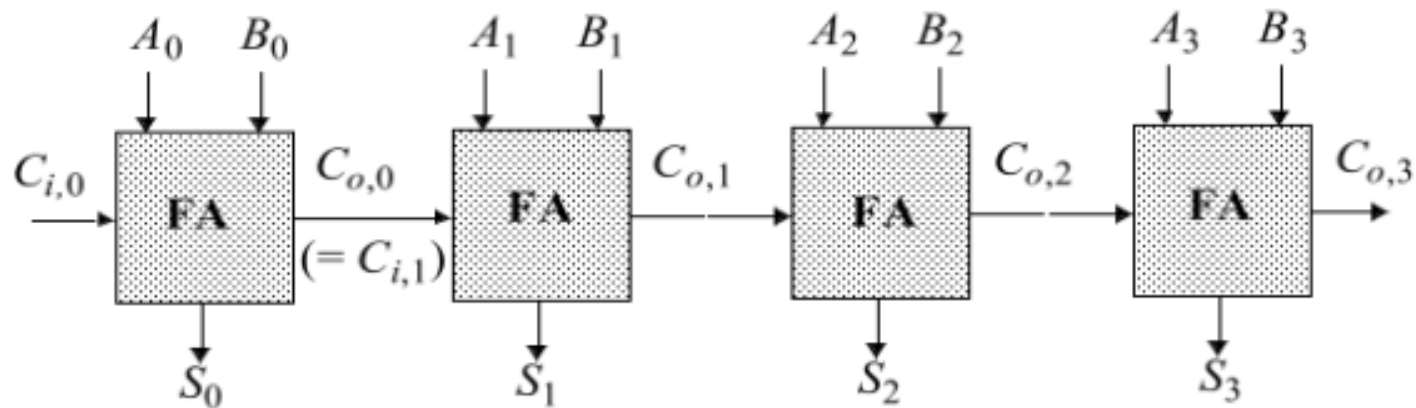
$$C_o(G, P) = G + PC_i$$

$$S(G, P) = P \oplus C_i$$



二进制加法器

- 行波进位加法器



$$t_{adder} \approx (N - 1)t_{carry} + t_{sum}$$

- 逐位进位加法器的传播延时与N成线性关系
- 加法器延时由进位延时决定

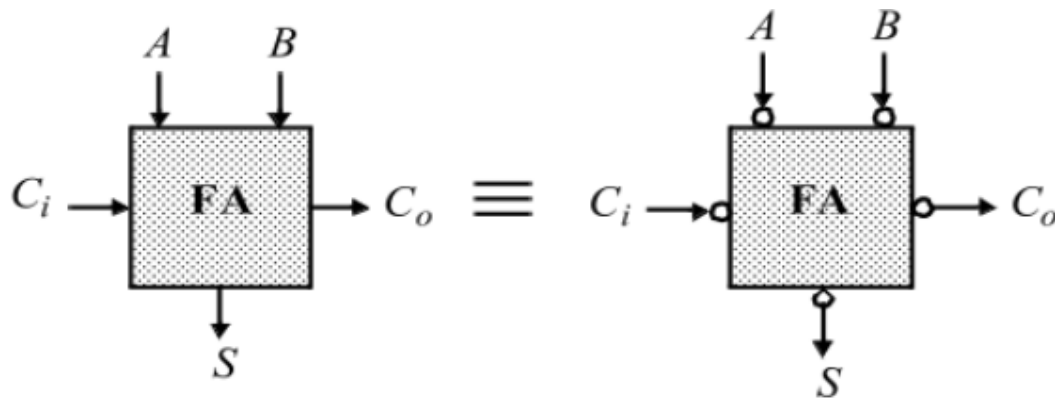
进位延迟最坏情况之一

$A: 00000001, B: 01111111$

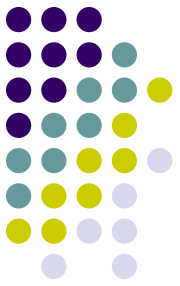


全加器电路设计考虑

- 加法器的反向特性
 - 把加法器的所有输入反向可以得到反向的输出

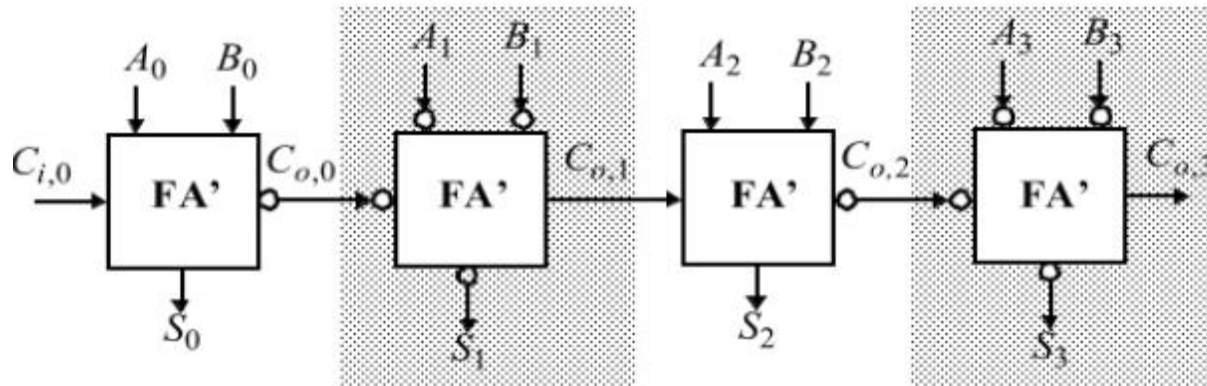


$$\begin{aligned}\bar{S}(A, B, C_i) &= S(\bar{A}, \bar{B}, \bar{C}_i) \\ \bar{C}_o(A, B, C_i) &= C_o(\bar{A}, \bar{B}, \bar{C}_i)\end{aligned}$$



全加器电路设计考虑

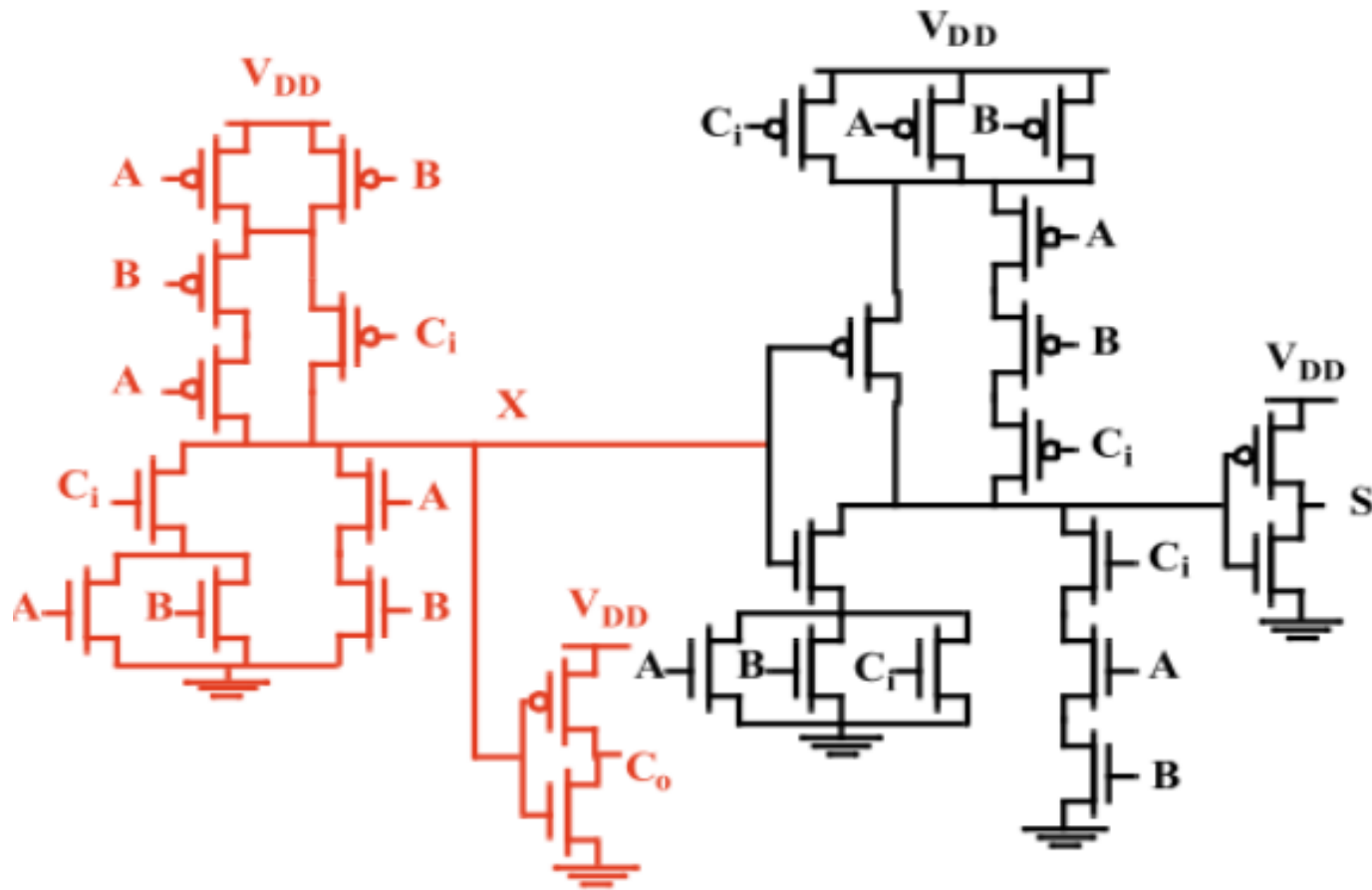
- 利用加法器反向特性设计的加法器结构





全加器电路设计考虑

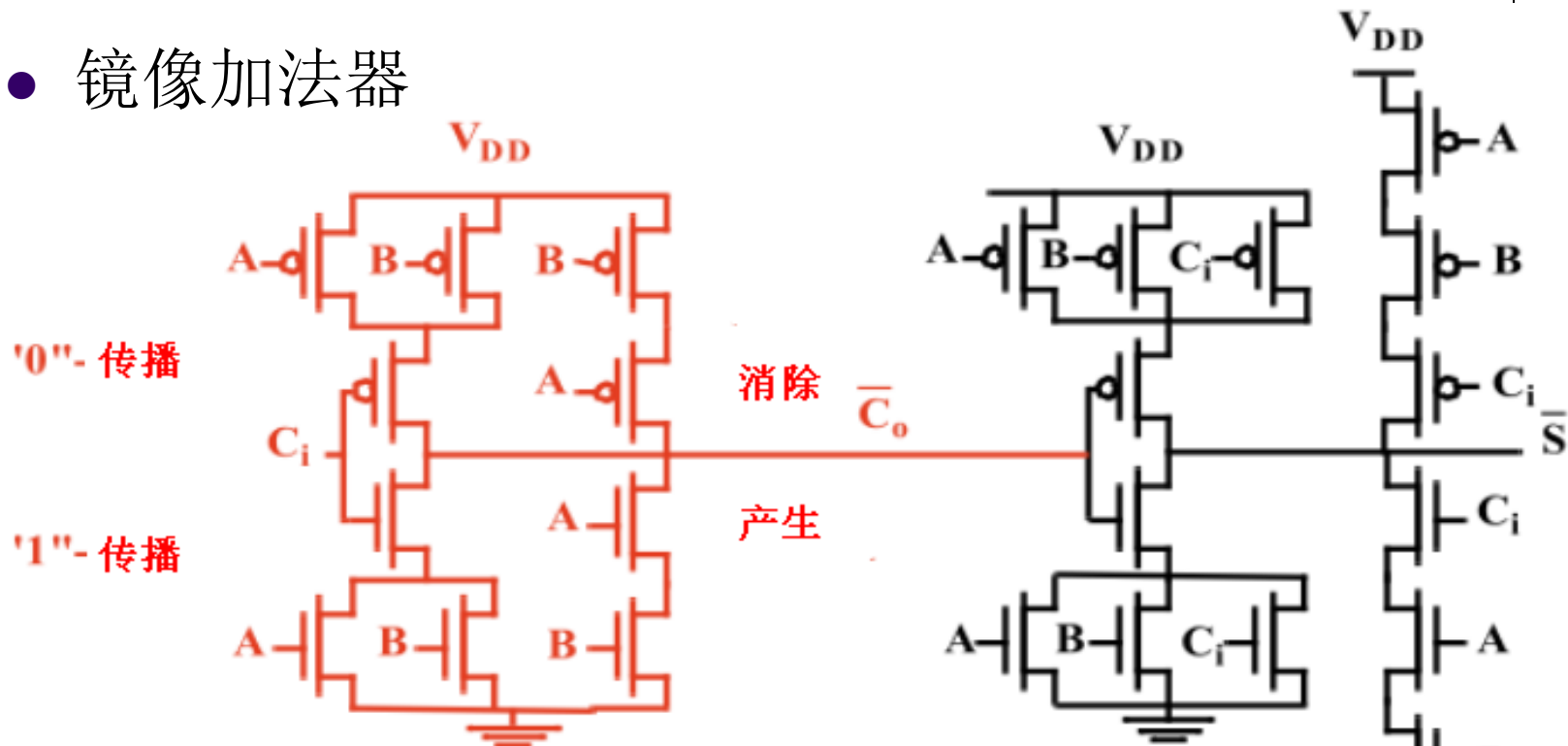
- 互补静态CMOS全加器





二进制加法器

- 镜像加法器



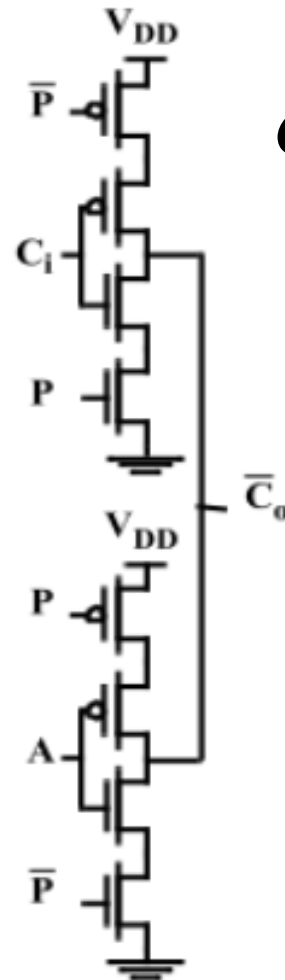
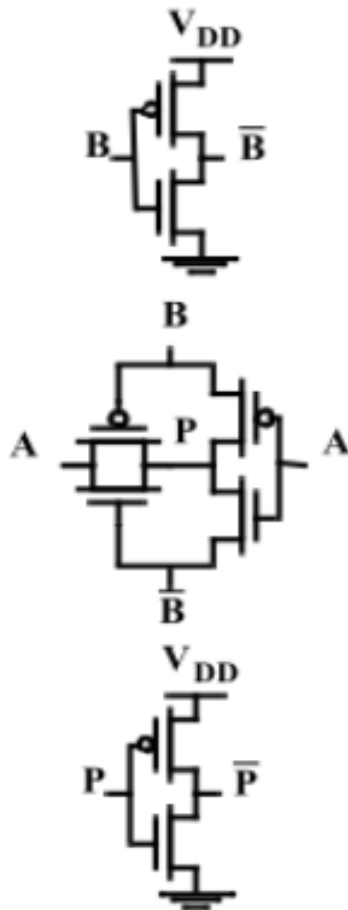
$$G = AB$$
$$D = \overline{A}\overline{B}$$
$$P = A \oplus B$$

$$C_o(G, P) = G + PC_i$$
$$\overline{S}(G, P) = P + C_i$$



二进制加法器

- 传输门型加法器



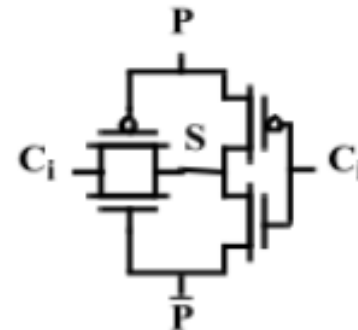
$$G = AB$$

$$D = \overline{A}\overline{B}$$

$$P = A \oplus B$$

$$C_o(G, P) = G + PC_i$$

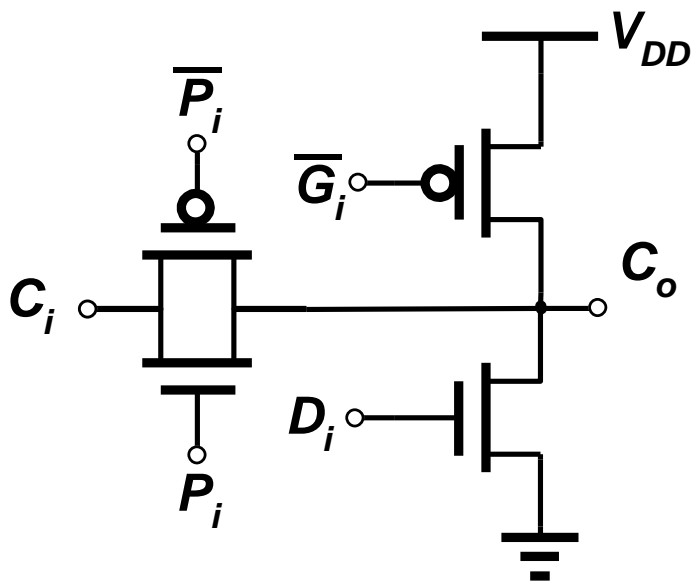
$$S(G, P) = P + C_i$$





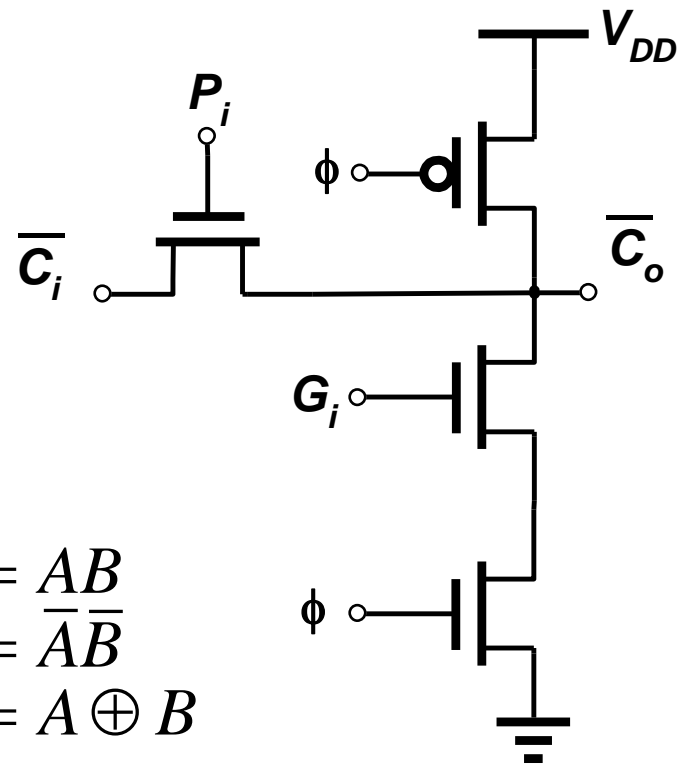
二进制加法器

- 曼彻斯特进位链加法器



$$\begin{aligned}G &= AB \\D &= \overline{A}\overline{B} \\P &= A \oplus B\end{aligned}$$

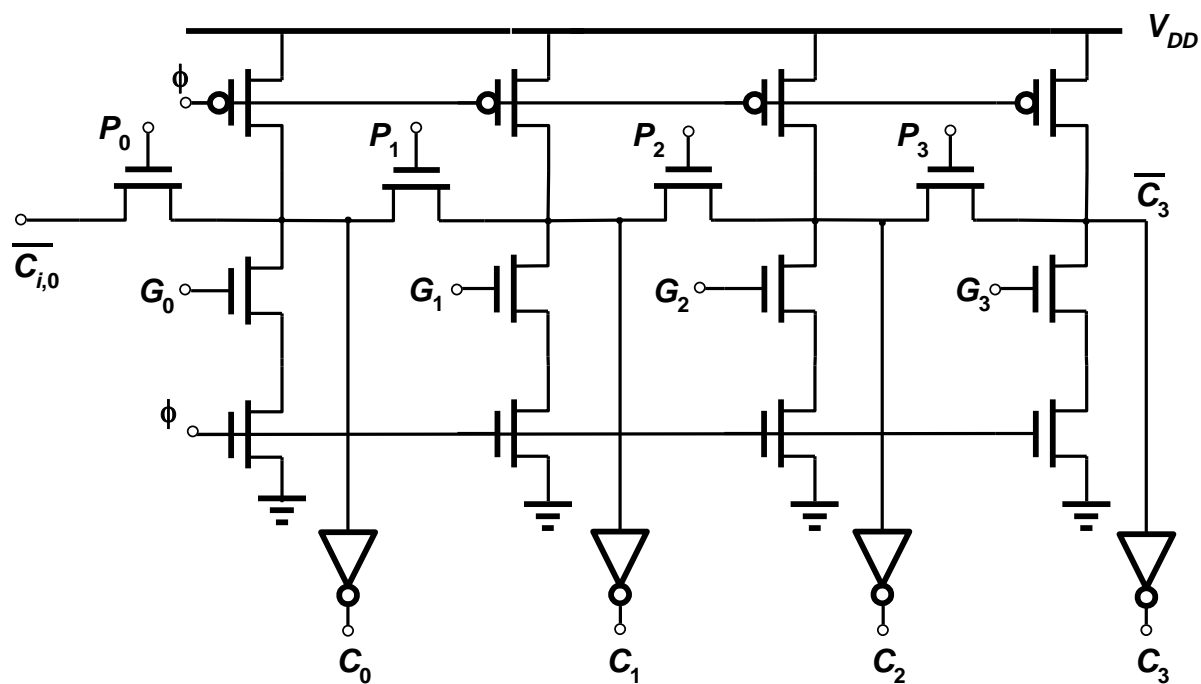
$$\begin{aligned}C_o(G, P) &= G + PC_i \\S(G, P) &= P + C_i\end{aligned}$$



二进制加法器



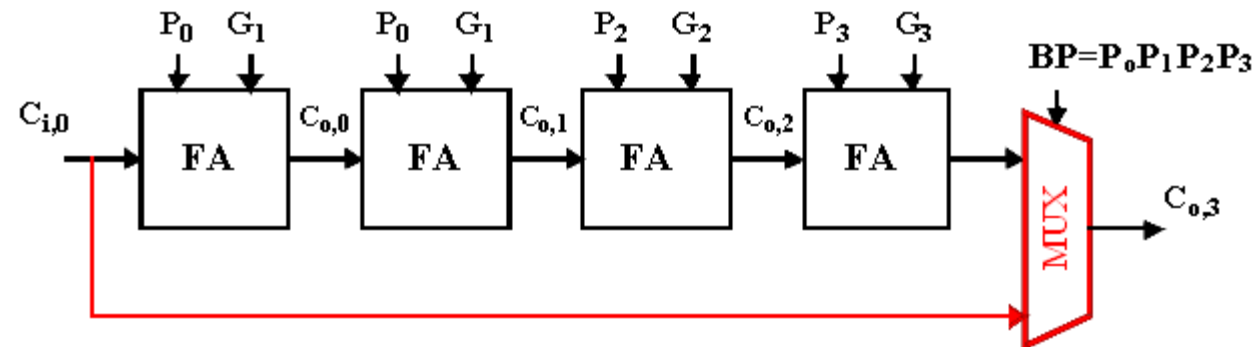
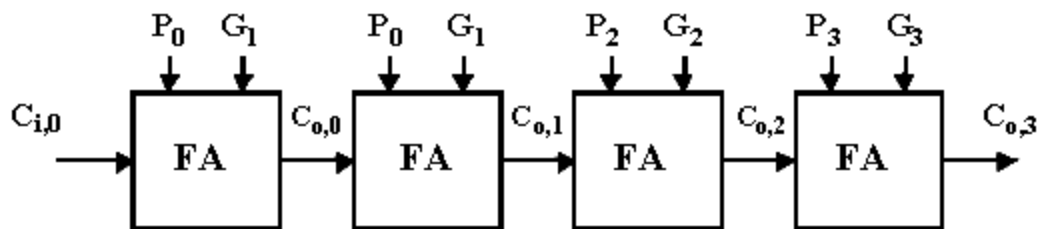
- 曼彻斯特进位链加法器





二进制加法器

- 旁路进位加法器



$$\begin{aligned} G &= AB \\ D &= \overline{A}\overline{B} \\ P &= A \oplus B \end{aligned}$$

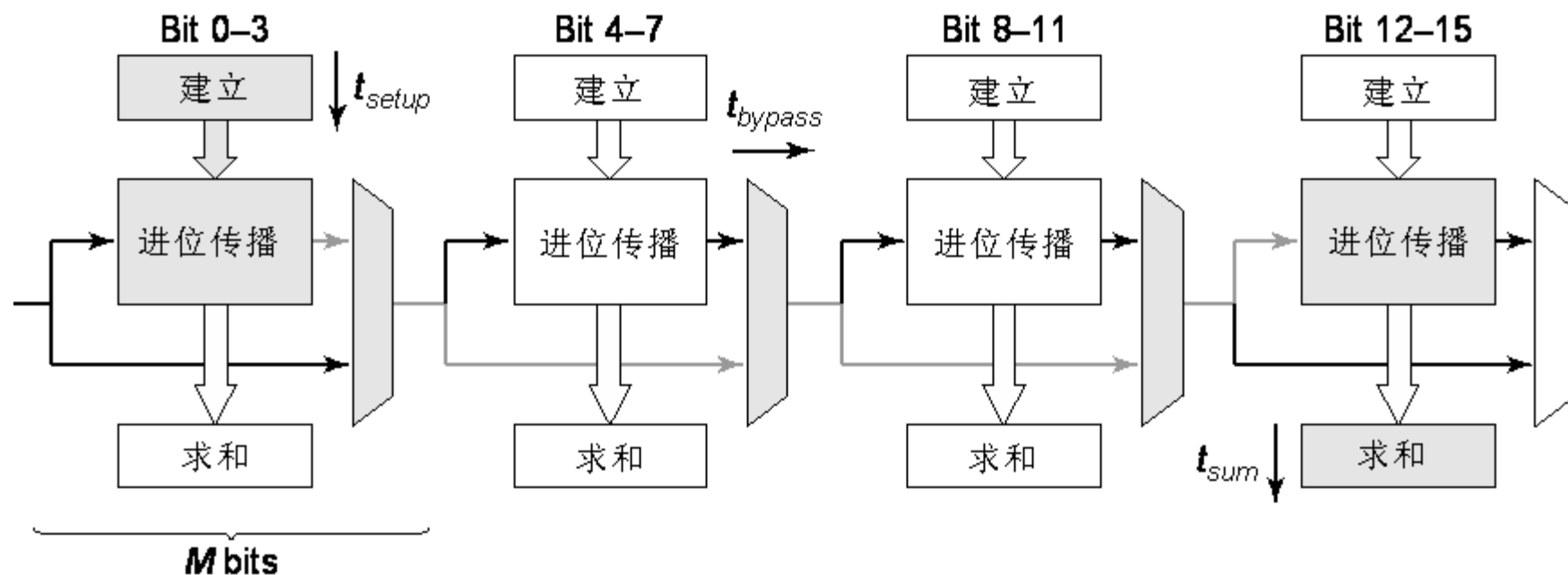
$$\begin{aligned} C_o(G, P) &= G + PC_i \\ S(G, P) &= P + C_i \end{aligned}$$

如果 $BP = P_0 P_1 P_2 P_3$, $C_{i,0} = C_{o,3}$



二进制加法器

- 旁路进位加法器
 - 关键路径

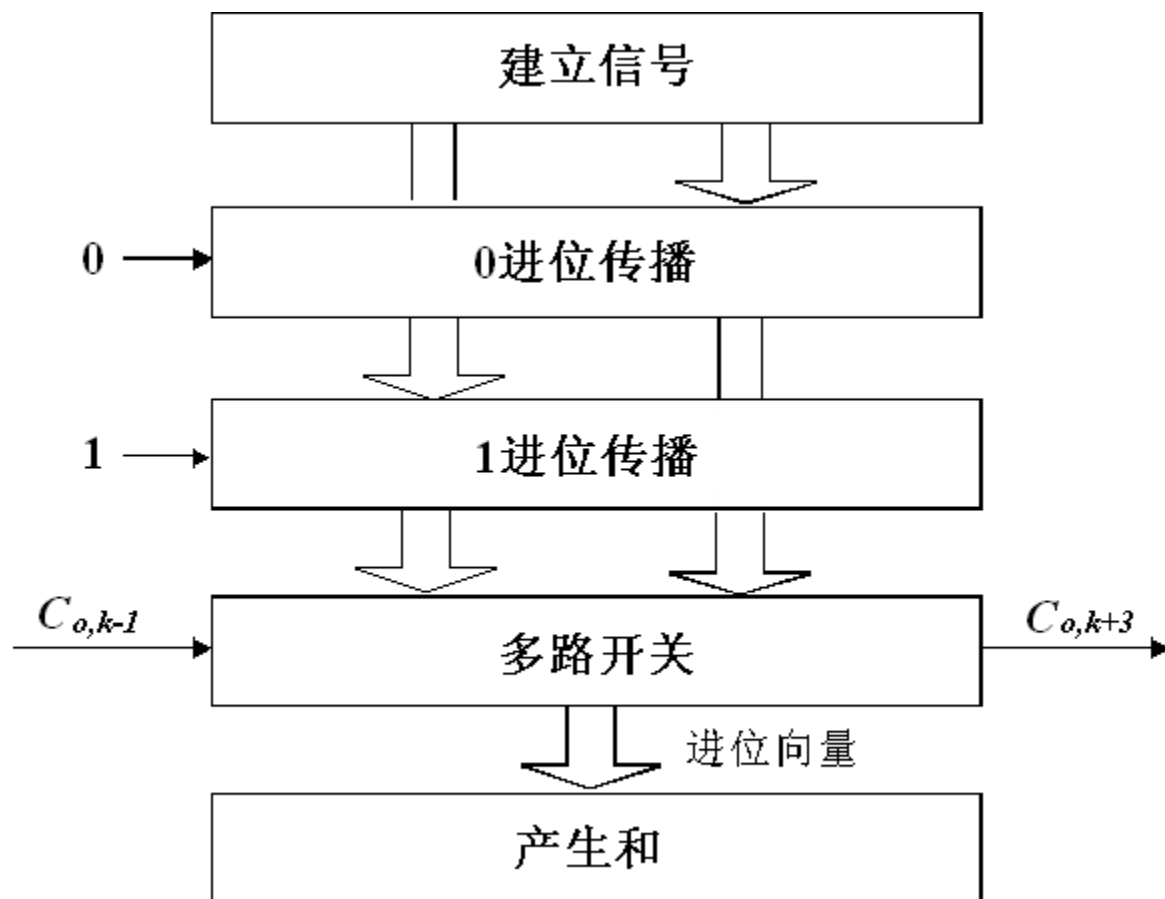


$$t_{adder} = t_{setup} + (N/M-1)t_{carry} + Mt_{bypass} + (M-1)t_{carry} + t_{sum}$$



二进制加法器

- 线性进位选择加法器

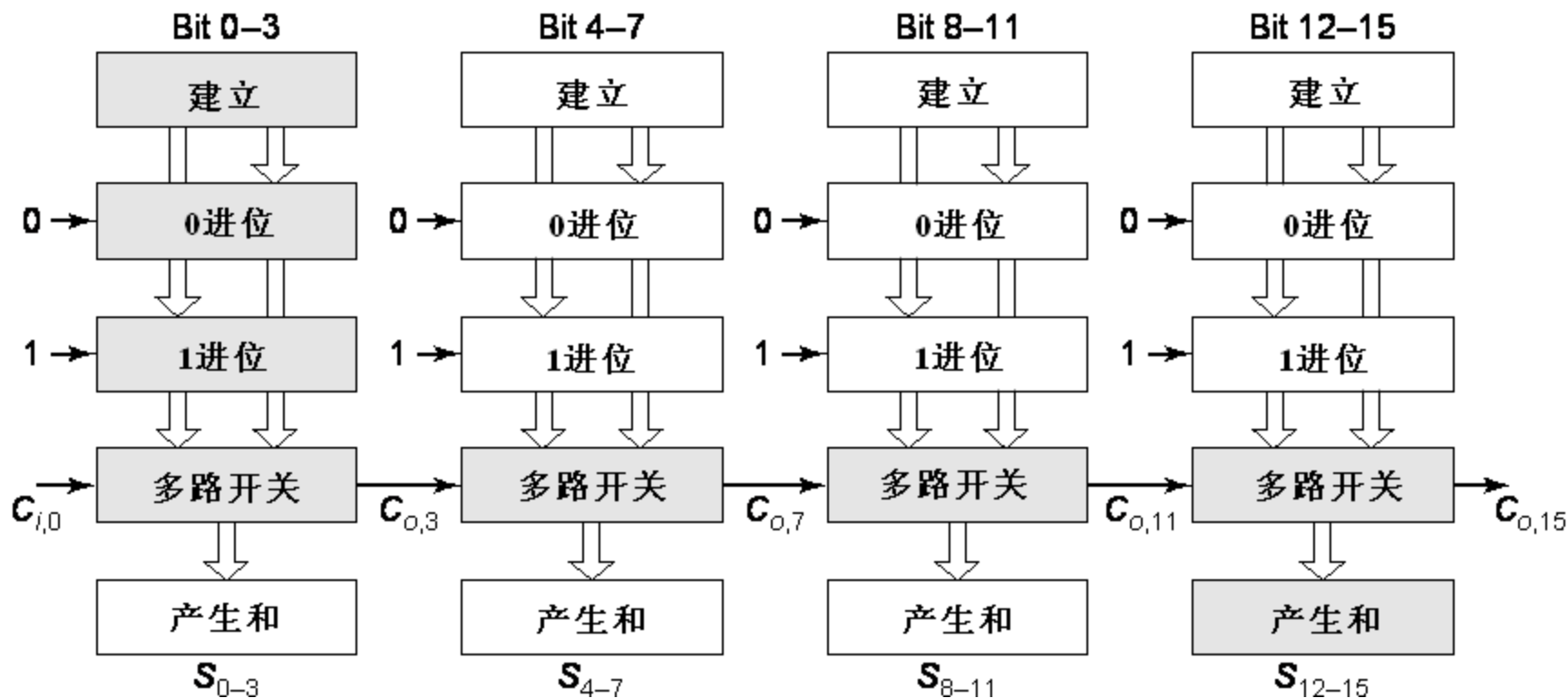




二进制加法器

- 线性进位选择加法器

- 关键路径

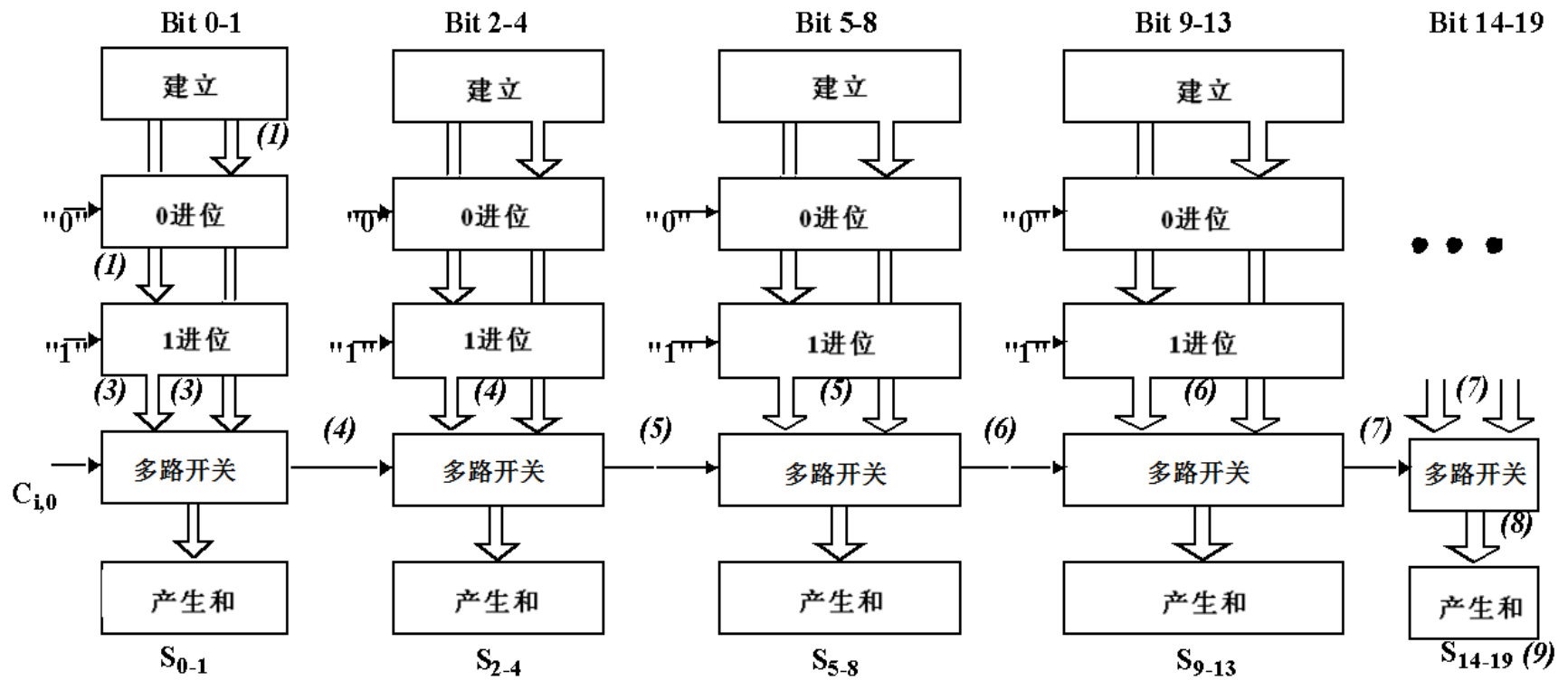


$$t_{adder} = t_{setup} + (N/M)t_{carry} + Mt_{mux} + t_{sum}$$



二进制加法器

- 平方根进位选择加法器

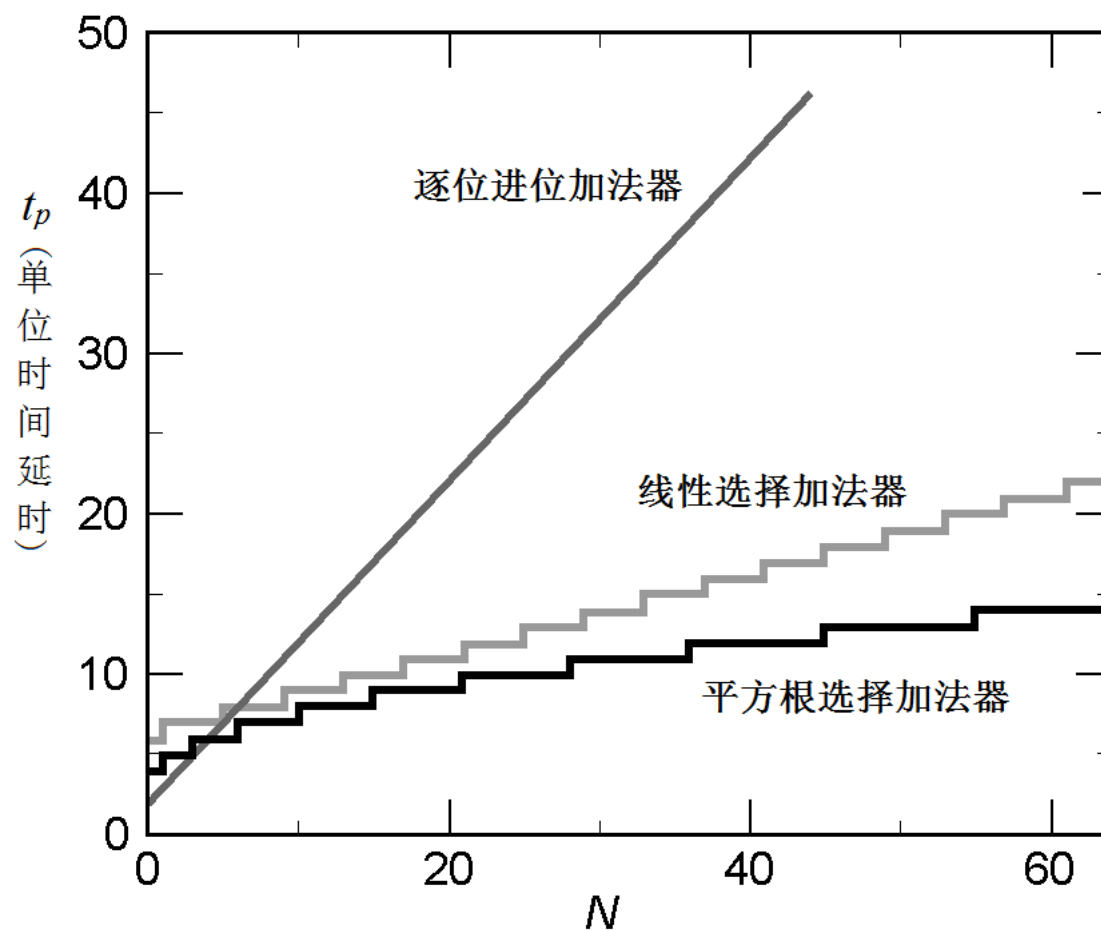


$$t_{adder} = t_{setup} + Pt_{carry} + (2N)^{1/2}t_{mux} + t_{sum}$$

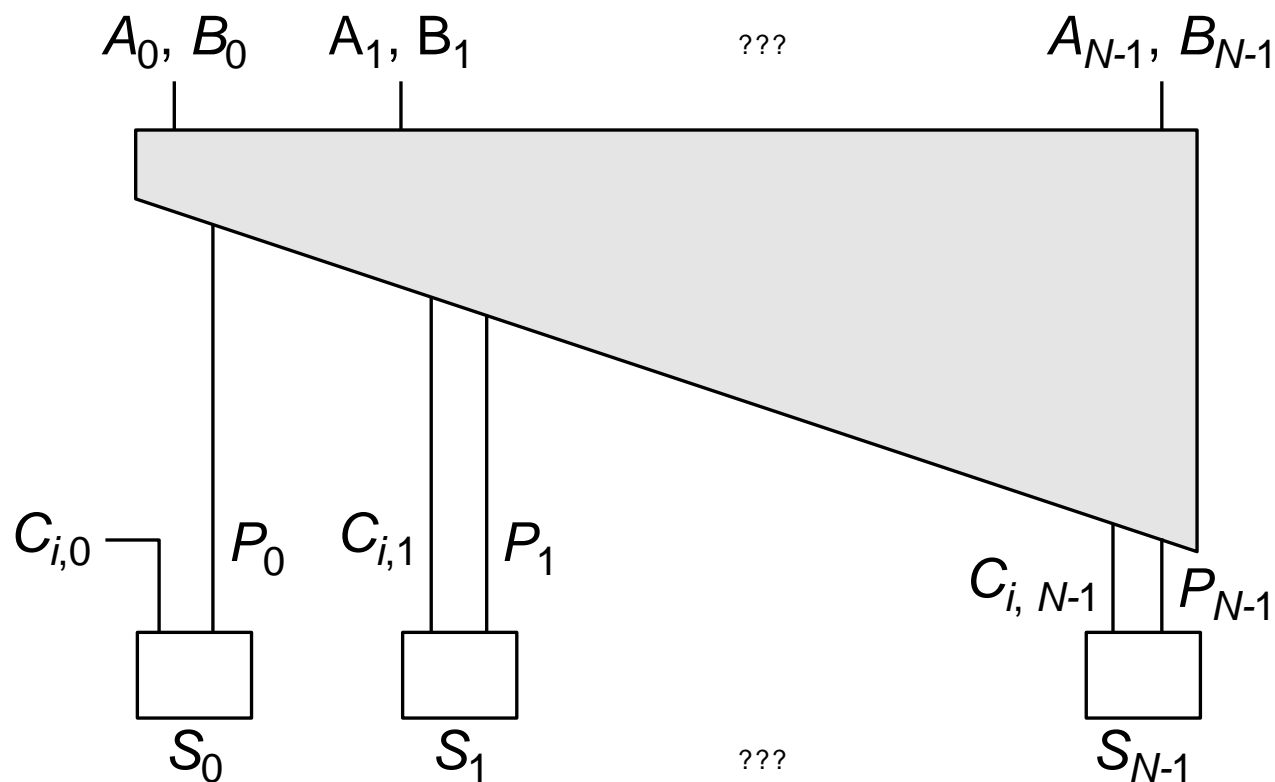


二进制加法器

- 几种进位选择加法器的传播延时比较

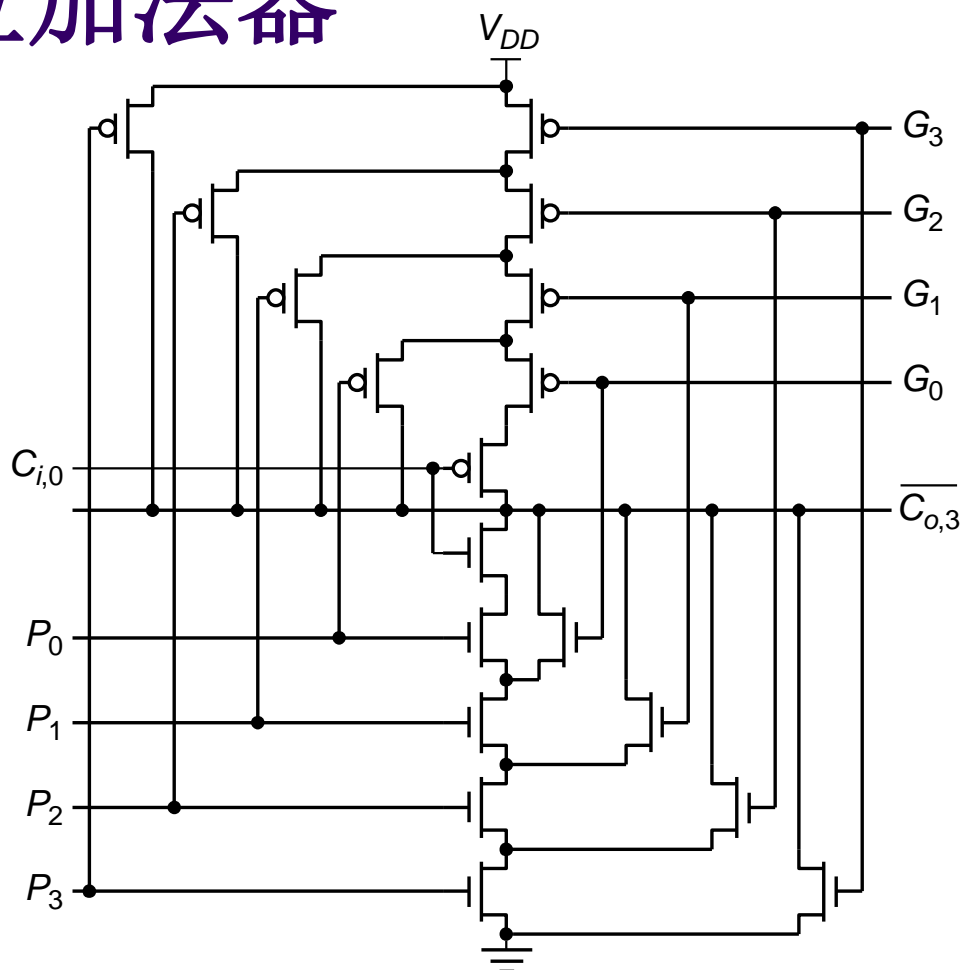


超前进位加法器



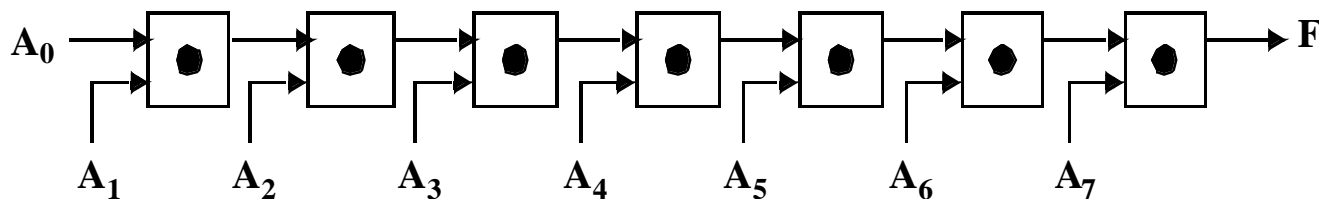
$$\begin{aligned}
 C_{o,k} &= G_k + P_k C_{o,k-1} = G_k + P_k (G_{k-1} + P_{k-1} C_{o,k-2}) \\
 &= G_k + P_k (G_{k-1} + P_{k-1} (\dots + P_1 (G_0 + P_0 C_{i,0})))
 \end{aligned}$$

超前进位加法器

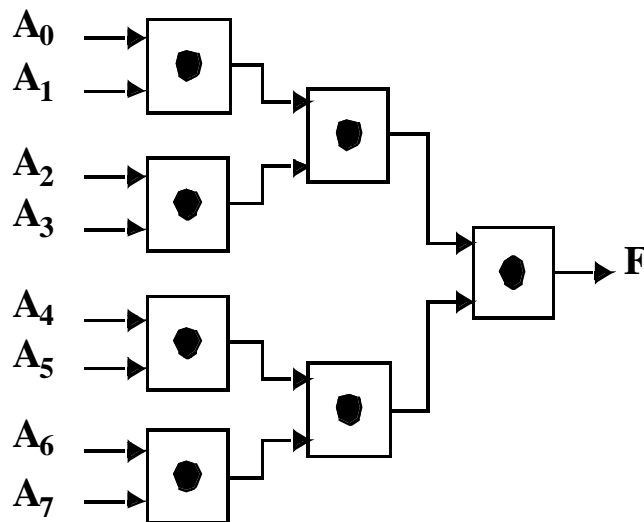


$$\begin{aligned}
 C_{o,k} &= G_k + P_k C_{o,k-1} = G_k + P_k (G_{k-1} + P_{k-1} C_{o,k-2}) \\
 &= G_k + P_k (G_{k-1} + P_{k-1} (\dots + P_1 (G_0 + P_0 C_{i,0})))
 \end{aligned}$$

对数超前进位加法器



$$t_p \sim N$$



$$t_p \sim \log_2(N)$$



对数超前进位加法器

$$C_{o,k} = G_k + P_k C_{o,k-1} = G_k + P_k (G_{k-1} + P_{k-1} C_{o,k-2})$$

$$= G_k + P_k (G_{k-1} + P_{k-1} (\dots + P_1 (G_0 + P_0 C_{i,0})))$$

$$(G_i + P_i G_j, P_i P_j) = (G_i, P_i) (G_j, P_j)$$

$$C_{o,0} = G_0 + P_0 C_{i,0} = (G_0, P_0) (C_{i,0}, 0)$$

$$C_{o,1} = G_1 + P_1 G_0 + P_1 P_0 C_{i,0} = (G_1, P_1) (G_0, P_0) (C_{i,0}, 0)$$

$$= (G_{1:0}, P_{1:0}) (C_{i,0}, 0)$$

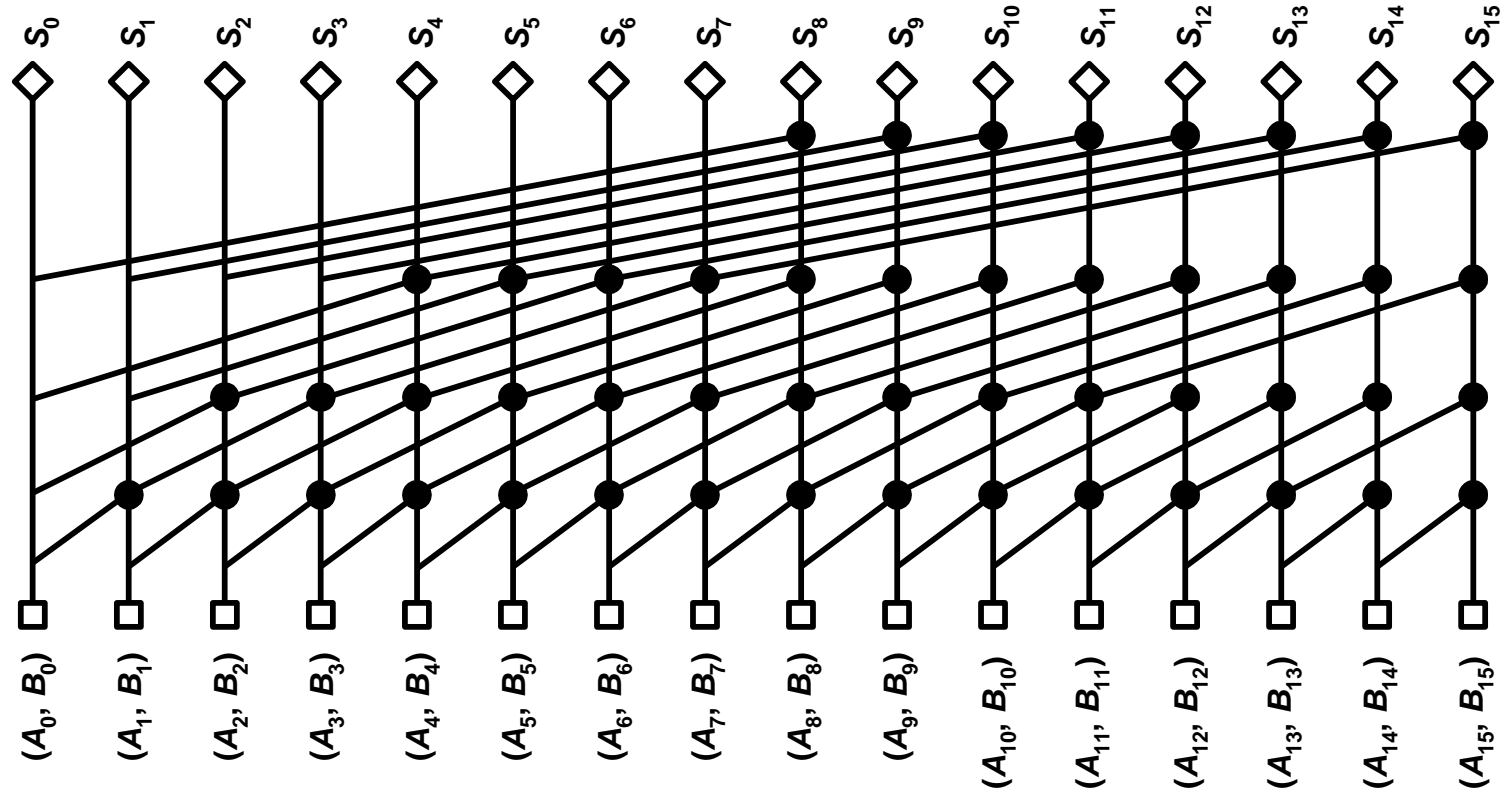
$$C_{o,2} = G_2 + P_2 G_1 + P_2 P_1 G_0 + P_2 P_1 P_0 C_{i,0}$$

$$= (G_2 + P_2 G_1) + (P_2 P_1) (G_0 + P_0 C_{i,0})$$

$$= (G_2, P_2) (G_1, P_1) (G_0, P_0) (C_{i,0}, 0)$$

$$= (G_{2:1}, P_{2:1}) (G_0, P_0) (C_{i,0}, 0)$$

对数超前进位加法器



乘法器

- 乘法器定义

$$X = \sum_{i=0}^{M-1} X_i 2^i \quad Y = \sum_{j=0}^{N-1} Y_j 2^j$$

$$\begin{aligned} Z = X \times Y &= \sum_{k=0}^{M+N-1} Z_k 2^k \\ &= \left(\sum_{i=0}^{M-1} X_i 2^i \right) \left(\sum_{j=0}^{N-1} Y_j 2^j \right) = \sum_{i=0}^{M-1} \left(\sum_{j=0}^{N-1} X_i Y_j 2^{i+j} \right) \end{aligned}$$

				1	0	1	0	1	0		被乘数	
x				1	0	1	1				乘数	
				<hr/>								
				1	0	1	0	1	0		} 部分积	
				1	0	1	0	1	0			
				0	0	0	0	0	0			
				0	0	0	0	0	0			
+				1	0	1	0	1	0		} 结果	
				<hr/>								
				1	1	1	0	0	1	1	1	0





乘法器

- 部分积的产生

- 减少非零行数

- 波兹编码

01111110

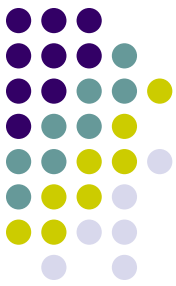
100000 $\bar{1}$ 0

减少“1”的数目

部分积最多情况：1010101010

$$Y = \sum_{j=0}^{(N-1)/2} Y_j 4^j \quad (Y_j \in \{-2, -1, 0, 1, 2\})$$

-2: $\bar{1}$ 0
-1: 0 $\bar{1}$
0: 00
1: 01
2: 10



乘法器

- 部分积的产生
 - 波兹编码方法

01111110

01(1), 11(1), 11(1), 10(0)

011 = 10

111 = 00

111 = 00

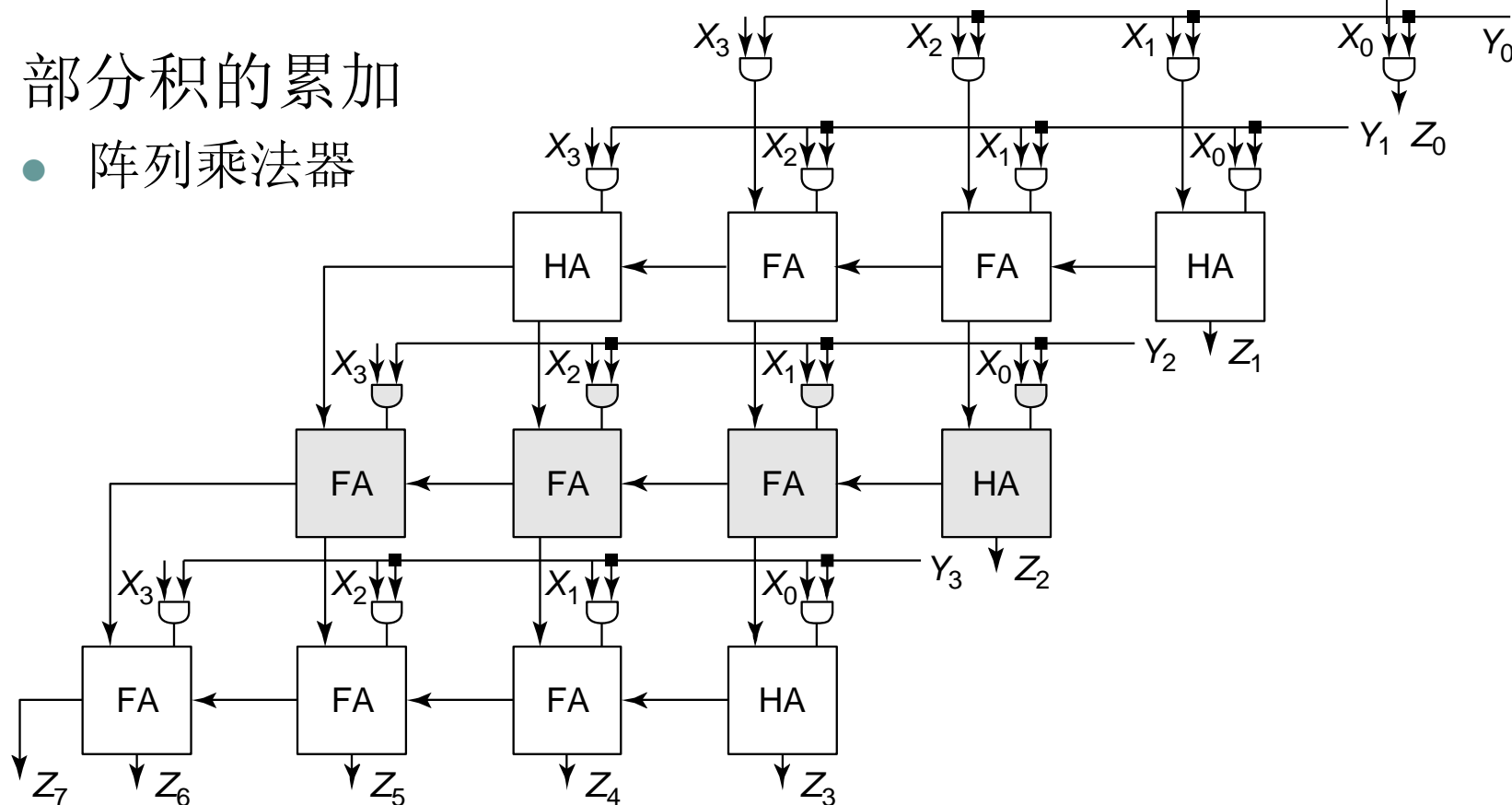
100 = $\overline{10}$

100000 $\overline{10}$

部分积选择表	
乘数位	编码位
000	0
001	+ 被乘数 $-2: \overline{10}$
010	+ 被乘数 $-1: 0\overline{1}$
011	+ 2 × 被乘数 0: 00
100	- 2 × 被乘数 1: 01
101	- 被乘数 2: 10
110	- 被乘数
111	0

乘法器

- 部分积的累加
 - 阵列乘法器

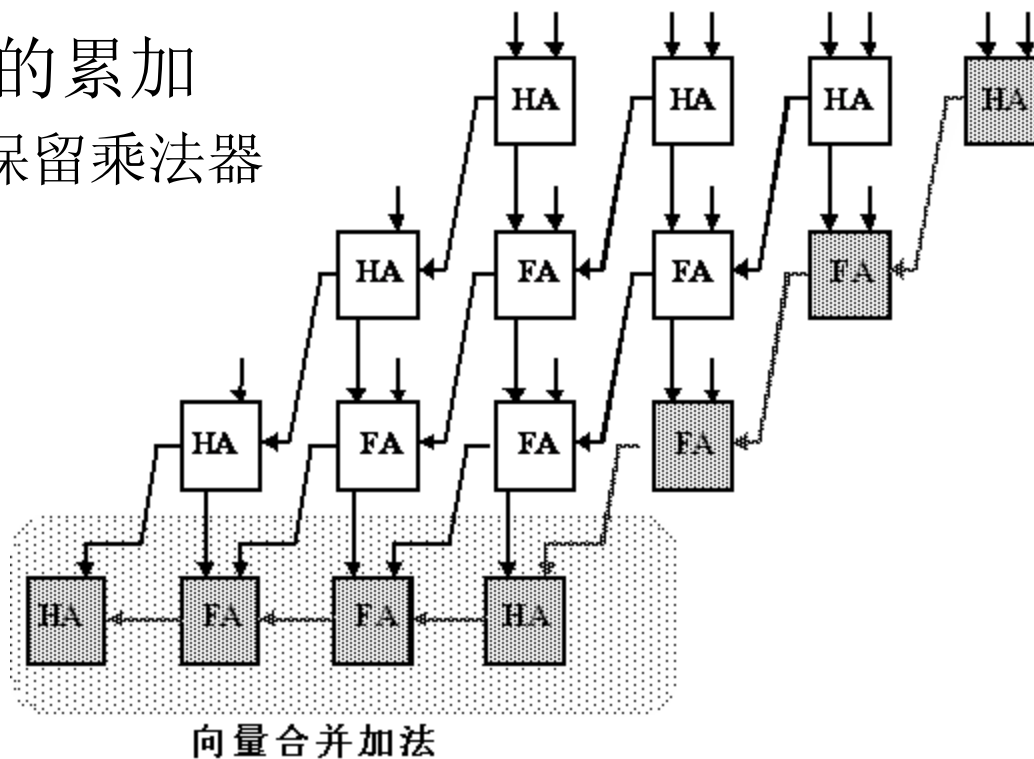


$$t_{mult} = [(M-1) + (N-2)] t_{carry} + (N-1) t_{sum} + t_{and}$$

乘法器



- 部分积的累加
 - 进位保留乘法器

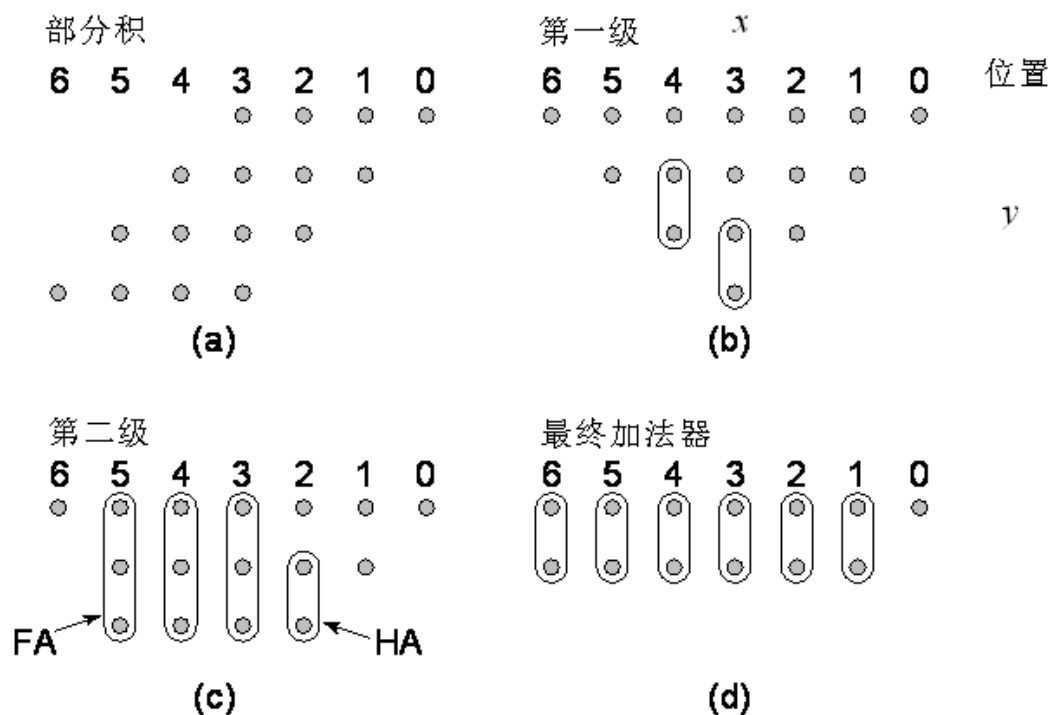


$$t_{mult} = t_{and} + (N-1) t_{carry} + t_{merge}$$

乘法器



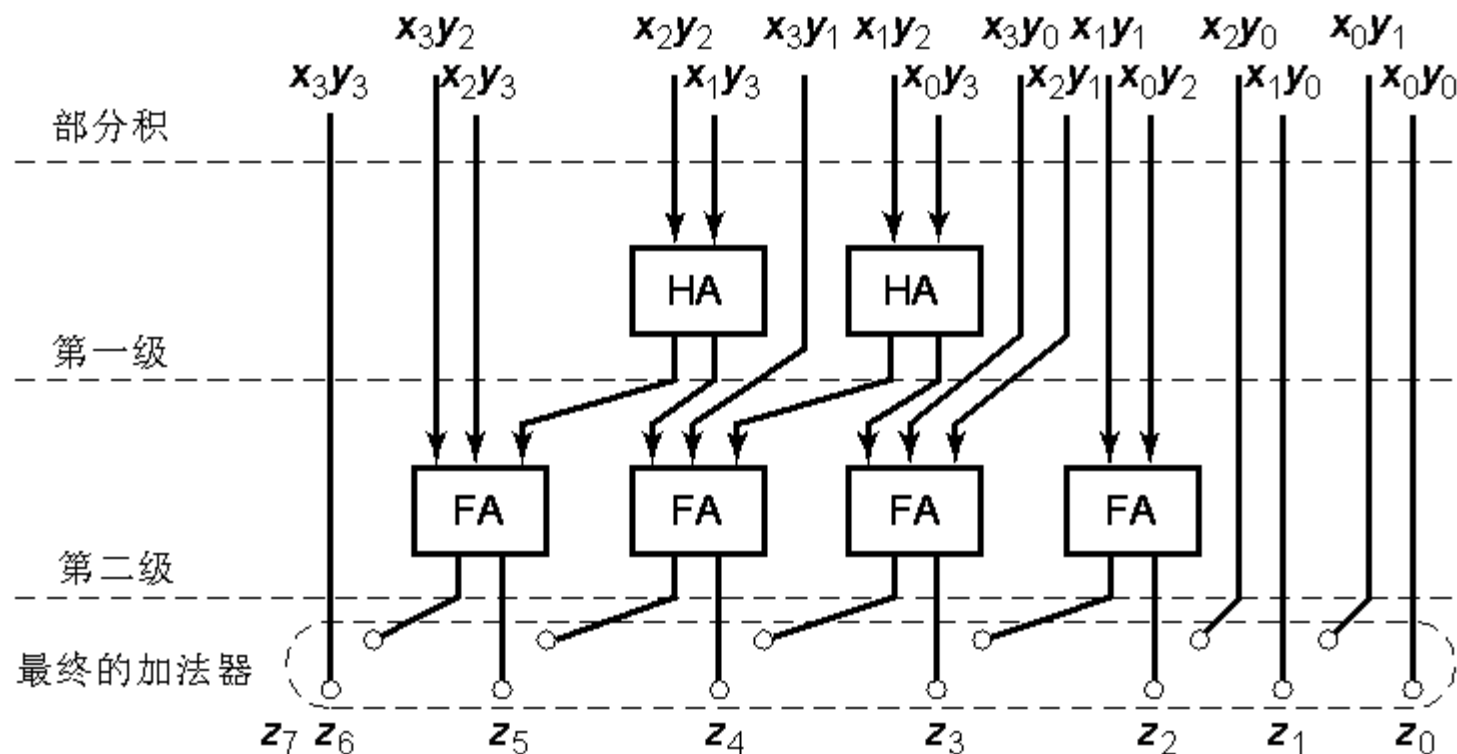
- 部分积的累加
 - 树形乘法器（华莱士树）
 - 减少关键路径和所需的加法器单元数目
 - 全加器：3-2压缩器





乘法器

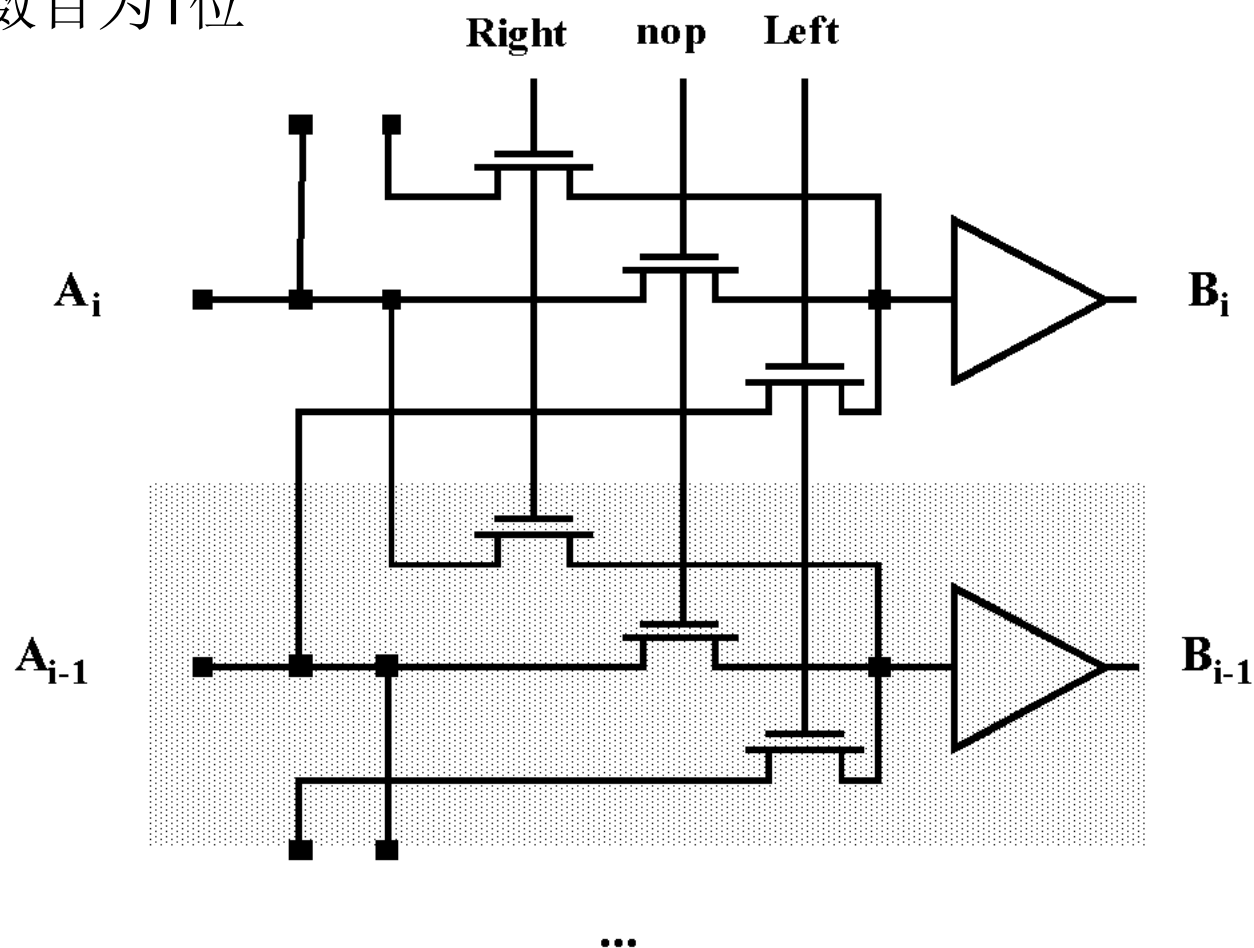
- 部分积的累加
 - 树形乘法器（华莱士树）



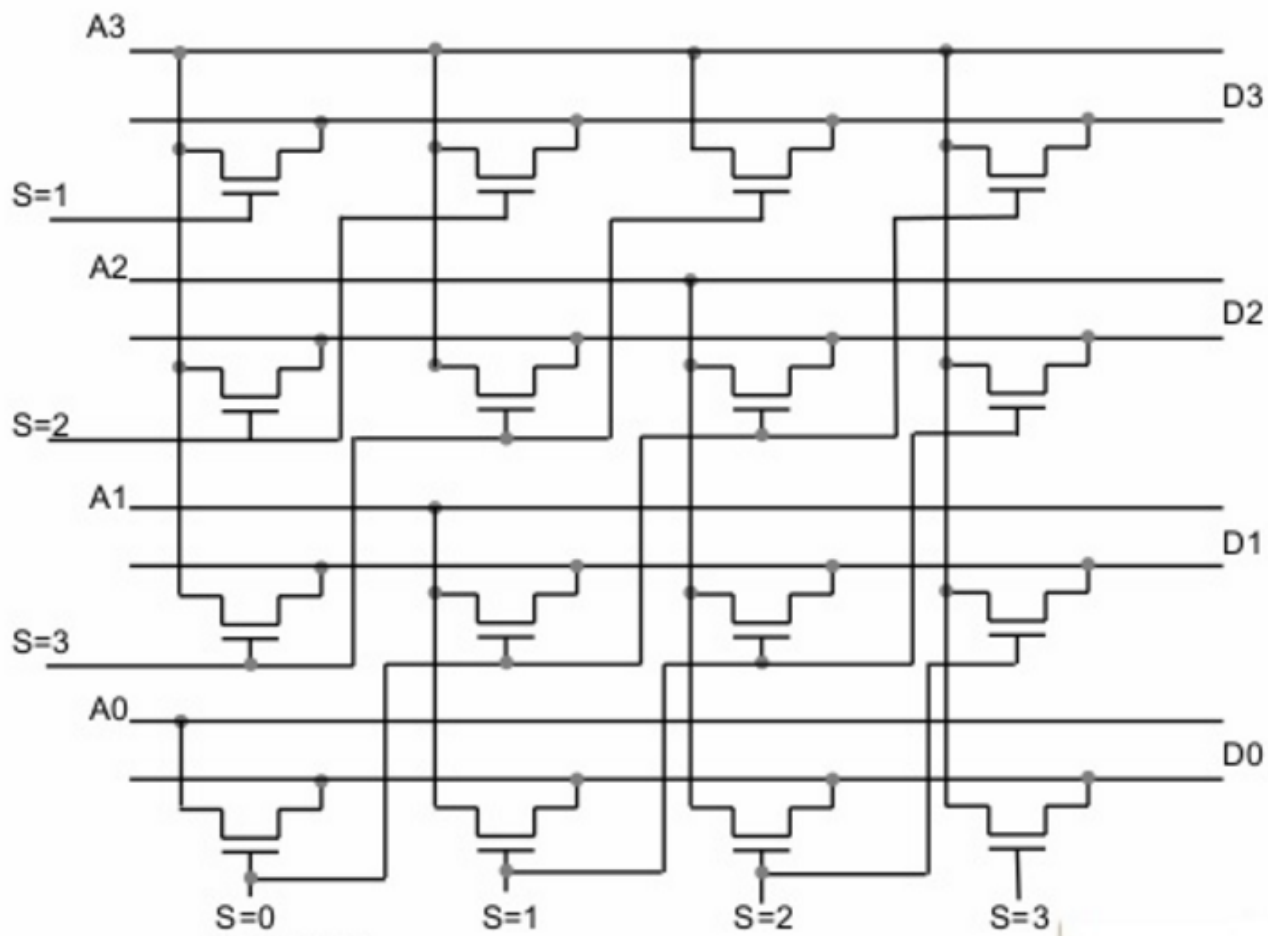
移位器



- 移位数目为1位



桶形移位器



对数移位器

